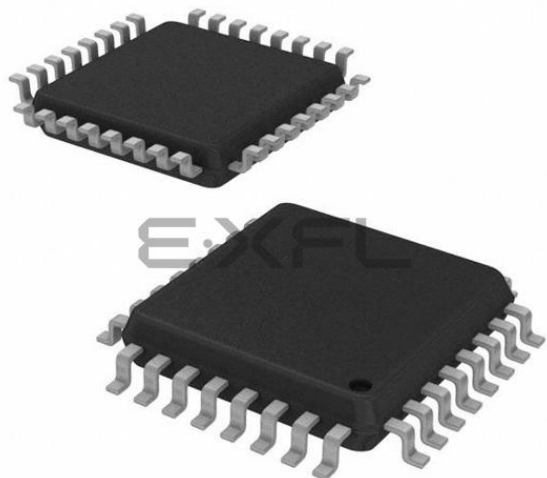




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Details

Product Status	Obsolete
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, LINbus, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	16KB (16K x 8)
Program Memory Type	FLASH
EEPROM Size	512 x 8
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 10x12b
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	32-LQFP
Supplier Device Package	32-LQFP (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc9s08dn16mlc

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Chapter 7

Central Processor Unit (S08CPUV3)

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Chapter 8

Multi-Purpose Clock Generator (S08MCGV1)

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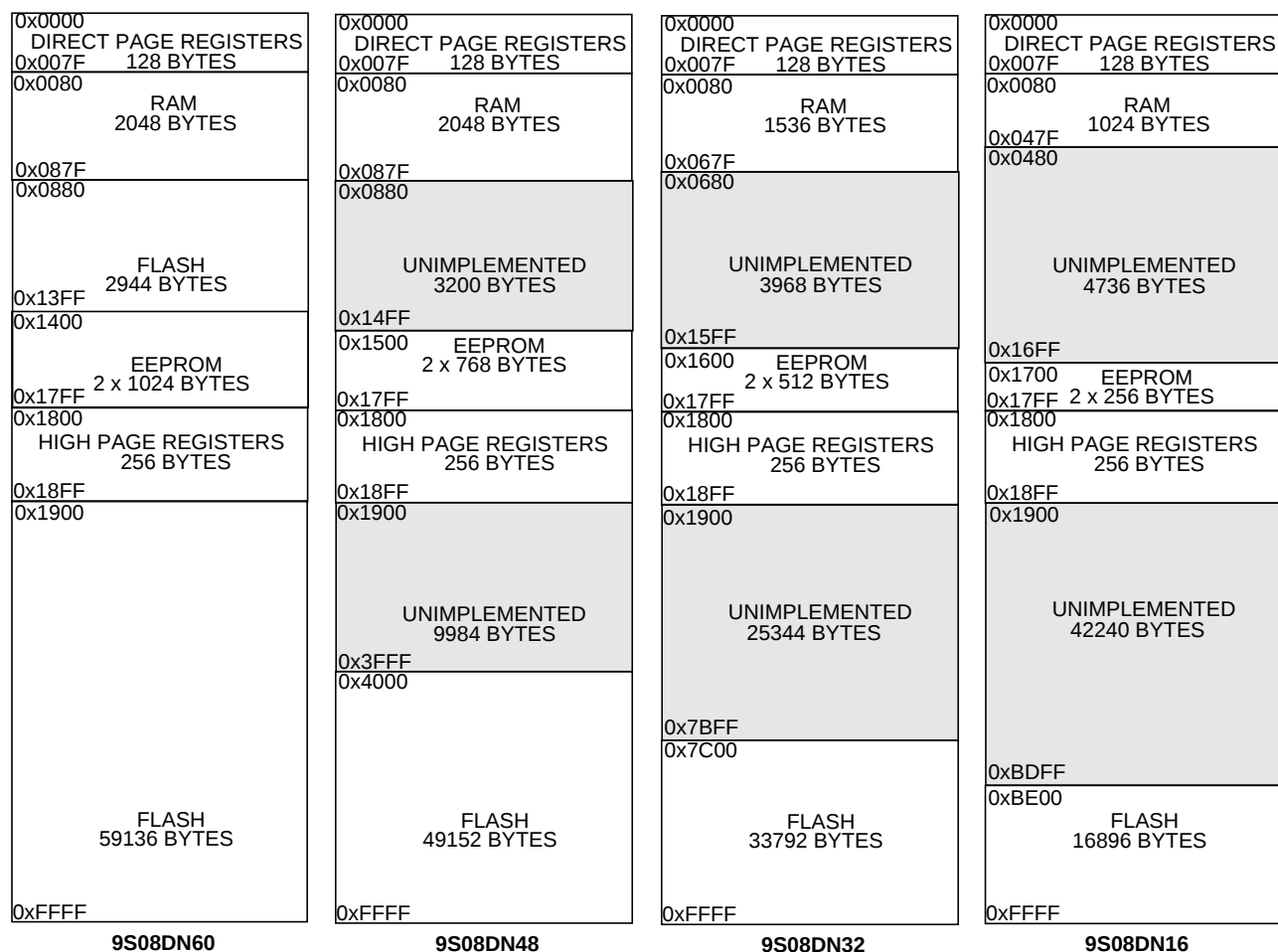


Figure 4-1. MC9S08DN60 Series Memory Map

4.2 Reset and Interrupt Vector Assignments

Table 4-1 shows address assignments for reset and interrupt vectors. The vector names shown in this table are the labels used in the MC9S08DN60 Series equate file provided by Freescale Semiconductor.

Table 4-1. Reset and Interrupt Vectors

Address (High/Low)	Vector	Vector Name
0xFFC0:0xFFC1	ACMP2	Vacmp2
0xFFC2:0xFFC3	ACMP1	Vacmp1
0xFFC4:0xFFCB	Reserved	—
0xFFCC:0xFFCD	RTC	Vrtc
0xFFCE:0xFFCF	IIC	Viic
0xFFD0:0xFFD1	ADC Conversion	Vadc

Table 4-1. Reset and Interrupt Vectors

Address (High/Low)	Vector	Vector Name
0xFFD2:0xFFD3	Port A, Port B, Port D	Vport
0xFFD4:0xFFD9	Reserved	—
0xFFDA:0xFFDB	SCI1 Transmit	Vsci1tx
0xFFDC:0xFFDD	SCI1 Receive	Vsci1rx
0xFFDE:0xFFDF	SCI1 Error	Vsci1err
0xFFE0:0xFFE1	SPI	Vspi
0xFFE2:0xFFE3	TPM2 Overflow	Vtpm2ovf
0xFFE4:0xFFE5	TPM2 Channel 1	Vtpm2ch1
0xFFE6:0xFFE7	TPM2 Channel 0	Vtpm2ch0
0xFFE8:0xFFE9	TPM1 Overflow	Vtpm1ovf
0xFFEA:0xFFEB	TPM1 Channel 5	Vtpm1ch5
0xFFEC:0xFFED	TPM1 Channel 4	Vtpm1ch4
0xFFEE:0xFFEF	TPM1 Channel 3	Vtpm1ch3
0xFFFF0:0xFFFF1	TPM1 Channel 2	Vtpm1ch2
0xFFFF2:0xFFFF3	TPM1 Channel 1	Vtpm1ch1
0xFFFF4:0xFFFF5	TPM1 Channel 0	Vtpm1ch0
0xFFFF6:0xFFFF7	MCG Loss of lock	Vlol
0xFFFF8:0xFFFF9	Low-Voltage Detect	Vlvd
0xFFFFA:0xFFFFB	IRQ	Virq
0xFFFFC:0xFFFFD	SWI	Vswi
0xFFFFE:0xFFFFF	Reset	Vreset

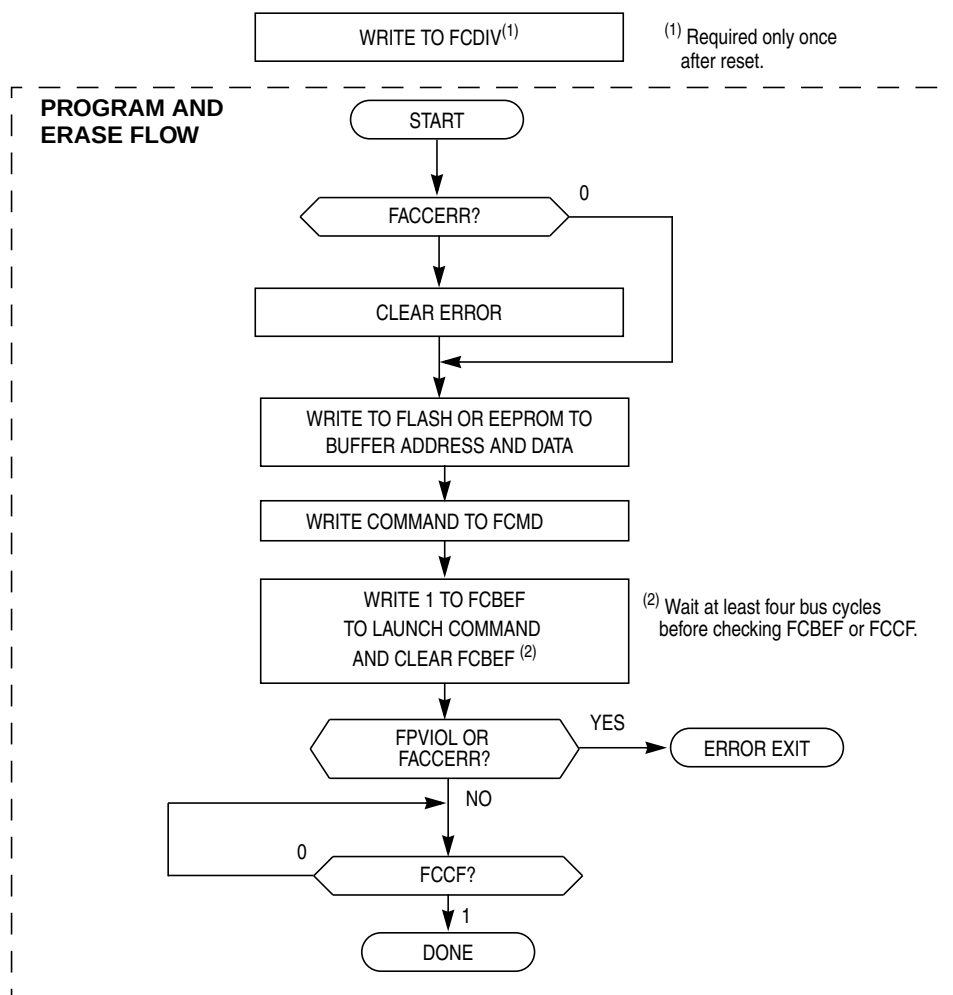


Figure 4-2. Program and Erase Flowchart

4.5.4 Burst Program Execution

The burst program command is used to program sequential bytes of data in less time than would be required using the standard program command. This is possible because the high voltage to the Flash array does not need to be disabled between program operations. Ordinarily, when a program or erase command is issued, an internal charge pump associated with the Flash memory must be enabled to supply high voltage to the array. Upon completion of the command, the charge pump is turned off. When a burst program command is issued, the charge pump is enabled and remains enabled after completion of the burst program operation if these two conditions are met:

- The next burst program command sequence has begun before the FCCF bit is set.
- The next sequential address selects a byte on the same burst block as the current byte being programmed. A burst block in this Flash memory consists of 32 bytes. A new burst block begins at each 32-byte address boundary.

The first byte of a series of sequential bytes being programmed in burst mode will take the same amount of time to program as a byte programmed in standard mode. Subsequent bytes will program in the burst

Table 5-1. Vector Summary¹

Vector No.	Address (High/Low)	Vector Name	Module	Source	Enable	Description
31	0xFFC0/0xFFC1	Vacmp2	ACMP2	ACF	ACIE	Analog comparator 2
30	0xFFC2/0xFFC3	Vacmp1	ACMP1	ACF	ACIE	Analog comparator 1
29–26	0xFFC4/0xFFC5– 0xFFCA/0xFFCB					(Reserved)
25	0xFFCC/0xFFCD	Vrtc	RTC	RTIF	RTIE	Real-time interrupt
24	0xFFCE/0xFFCF	Viiic	IIC	IICIS	IICIE	IIC control
23	0xFFD0/0xFFD1	Vadc	ADC	COCO	AIEN	ADC
22	0xFFD2/0xFFD3	Vport	Port A,B,D	PTAIF, PTBIF, PTDIF	PTAIE, PTBIE, PTDIE	Port Pins
21–19	0xFFD4/0xFFD5– 0xFFD8/0xFFD9					(Reserved)
18	0xFFDA/0xFFDB	Vsci1tx	SCI1	TDRE, TC	TIE, TCIE	SCI1 transmit
17	0xFFDC/0xFFDD	Vsci1rx	SCI1	IDLE, LBKDIF, RDRF, RXEDGIF	ILIE, LBKDIE, RIE, RXEDGIE	SCI1 receive
16	0xFFDE/0xFFDF	Vsci1err	SCI1	OR, NF, FE, PF	ORIE, NFIE, FEIE, PFIE	SCI1 error
15	0xFFE0/0xFFE1	Vspi	SPI	SPIF, MODF, SPTEF	SPIE, SPIE, SPTIE	SPI
14	0xFFE2/0xFFE3	Vtpm2ovf	TPM2	TOF	TOIE	TPM2 overflow
13	0xFFE4/0xFFE5	Vtpm2ch1	TPM2	CH1F	CH1IE	TPM2 channel 1
12	0xFFE6/0xFFE7	Vtpm2ch0	TPM2	CH0F	CH0IE	TPM2 channel 0
11	0xFFE8/0xFFE9	Vtpm1ovf	TPM1	TOF	TOIE	TPM1 overflow
10	0xFFEA/0xFFEB	Vtpm1ch5	TPM1	CH5F	CH5IE	TPM1 channel 5
9	0xFFEC/0xFFED	Vtpm1ch4	TPM1	CH4F	CH4IE	TPM1 channel 4
8	0xFFEE/0xFFEF	Vtpm1ch3	TPM1	CH3F	CH3IE	TPM1 channel 3
7	0xFFFF0/0xFFFF1	Vtpm1ch2	TPM1	CH2F	CH2IE	TPM1 channel 2
6	0xFFFF2/0xFFFF3	Vtpm1ch1	TPM1	CH1F	CH1IE	TPM1 channel 1
5	0xFFFF4/0xFFFF5	Vtpm1ch0	TPM1	CH0F	CH0IE	TPM1 channel 0
4	0xFFFF6/0xFFFF7	Vlol	MCG	LOLS	LOLIE	MCG loss of lock
3	0xFFFF8/0xFFFF9	Vlvd	System control	LVWF	LVWIE	Low-voltage warning
2	0xFFFFA/0xFFFFB	Virq	IRQ	IRQF	IRQIE	IRQ pin
1	0xFFFFC/0xFFFFD	Vswi	Core	SWI Instruction	—	Software interrupt
0	0xFFFFE/0xFFFFF	Vreset	System control	COP, LOC, LVD, RESET, ILOP, ILAD, POR, BDFR	COPE CME LVDRE — — — — —	Watchdog timer Loss-of-clock Low-voltage detect External pin Illegal opcode Illegal address Power-on-reset BDM-forced reset

¹ Vector priority is shown from lowest (first row) to highest (last row). For example, Vreset is the highest priority vector.

5.6 Low-Voltage Detect (LVD) System

The MC9S08DN60 Series includes a system to protect against low-voltage conditions in order to protect memory contents and control MCU system states during supply voltage variations. The system is

6.5.1.7 Port A Interrupt Pin Select Register (PTAPS)

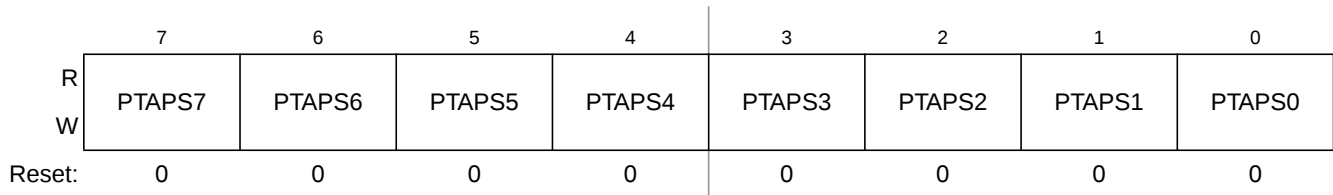


Figure 6-9. Port A Interrupt Pin Select Register (PTAPS)

Table 6-7. PTAPS Register Field Descriptions

Field	Description
7:0 PTAPS[7:0]	Port A Interrupt Pin Selects — Each of the PTAPSn bits enable the corresponding port A interrupt pin. 0 Pin not enabled as interrupt. 1 Pin enabled as interrupt.

6.5.1.8 Port A Interrupt Edge Select Register (PTAES)

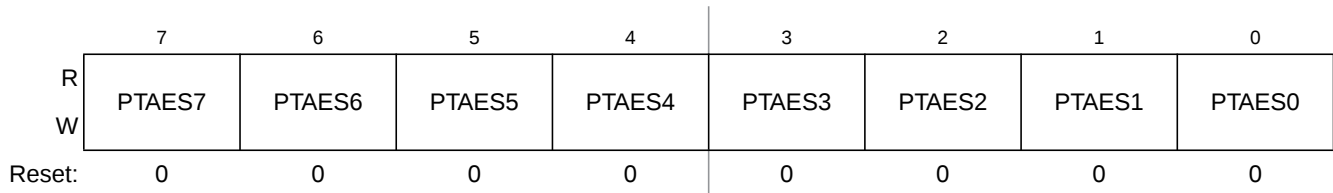


Figure 6-10. Port A Edge Select Register (PTAES)

Table 6-8. PTAES Register Field Descriptions

Field	Description
7:0 PTAES[7:0]	Port A Edge Selects — Each of the PTAESn bits serves a dual purpose by selecting the polarity of the active interrupt edge as well as selecting a pull-up or pull-down device if enabled. 0 A pull-up device is connected to the associated pin and detects falling edge/low level for interrupt generation. 1 A pull-down device is connected to the associated pin and detects rising edge/high level for interrupt generation.

6.5.5.5 Port E Drive Strength Selection Register (PTEDS)

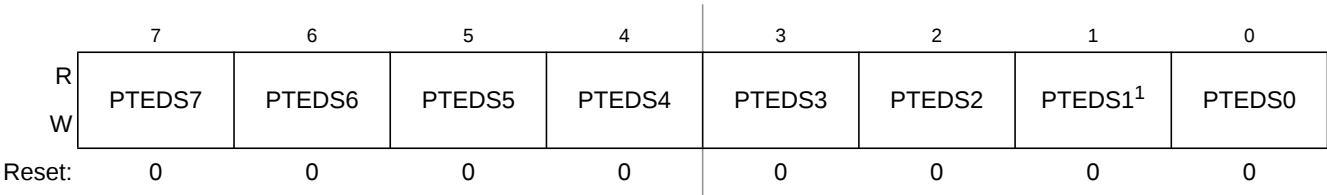


Figure 6-36. Drive Strength Selection for Port E Register (PTEDS)

¹ PTEDS1 has no effect on the input-only PTE1 pin.

Table 6-34. PTEDS Register Field Descriptions

Field	Description
7:0 PTEDS[7:0]	Output Drive Strength Selection for Port E Bits — Each of these control bits selects between low and high output drive for the associated PTE pin. For port E pins that are configured as inputs, these bits have no effect. 0 Low output drive strength selected for port E bit n. 1 High output drive strength selected for port E bit n.



- LP bit is written to 0

In FLL bypassed internal mode, the MCGOUT clock is derived from the internal reference clock. The FLL clock is controlled by the internal reference clock, and the FLL clock frequency locks to 1024 times the reference frequency, as selected by the RDIV bits. The MCGLCLK is derived from the FLL and the PLL is disabled in a low power state.

8.4.1.4 FLL Bypassed External (FBE)

In FLL bypassed external (FBE) mode, the MCGOUT clock is derived from the external reference clock and the FLL is operational but its output clock is not used. This mode is useful to allow the FLL to acquire its target frequency while the MCGOUT clock is driven from the external reference clock.

The FLL bypassed external mode is entered when all the following conditions occur:

- CLKS bits are written to 10
- IREFS bit is written to 0
- PLLS bit is written to 0
- RDIV bits are written to divide reference clock to be within the range of 31.25 kHz to 39.0625 kHz
- LP bit is written to 0

In FLL bypassed external mode, the MCGOUT clock is derived from the external reference clock. The external reference clock which is enabled can be an external crystal/resonator or it can be another external clock source. The FLL clock is controlled by the external reference clock, and the FLL clock frequency locks to 1024 times the reference frequency, as selected by the RDIV bits. The MCGLCLK is derived from the FLL and the PLL is disabled in a low power state.

NOTE

It is possible to briefly operate in FBE mode with an FLL reference clock frequency that is greater than the specified maximum frequency. This can be necessary in applications that operate in PEE mode using an external crystal with a frequency above 5 MHz. Please see [8.5.2.4, “Example # 4: Moving from FEI to PEE Mode: External Crystal = 8 MHz, Bus Frequency = 8 MHz”](#) for a detailed example.

8.4.1.5 PLL Engaged External (PEE)

The PLL engaged external (PEE) mode is entered when all the following conditions occur:

- CLKS bits are written to 00
- IREFS bit is written to 0
- PLLS bit is written to 1
- RDIV bits are written to divide reference clock to be within the range of 1 MHz to 2 MHz

In PLL engaged external mode, the MCGOUT clock is derived from the PLL clock which is controlled by the external reference clock. The external reference clock which is enabled can be an external crystal/resonator or it can be another external clock source. The PLL clock frequency locks to a

multiplication factor, as selected by the VDIV bits, times the reference frequency, as selected by the RDIV bits. If BDM is enabled then the MCGLCLK is derived from the DCO (open-loop mode) divided by two. If BDM is not enabled then the FLL is disabled in a low power state.

8.4.1.6 PLL Bypassed External (PBE)

In PLL bypassed external (PBE) mode, the MCGOUT clock is derived from the external reference clock and the PLL is operational but its output clock is not used. This mode is useful to allow the PLL to acquire its target frequency while the MCGOUT clock is driven from the external reference clock.

The PLL bypassed external mode is entered when all the following conditions occur:

- CLKS bits are written to 10
- IREFS bit is written to 0
- PLLS bit is written to 1
- RDIV bits are written to divide reference clock to be within the range of 1 MHz to 2 MHz
- LP bit is written to 0

In PLL bypassed external mode, the MCGOUT clock is derived from the external reference clock. The external reference clock which is enabled can be an external crystal/resonator or it can be another external clock source. The PLL clock frequency locks to a multiplication factor, as selected by the VDIV bits, times the reference frequency, as selected by the RDIV bits. If BDM is enabled then the MCGLCLK is derived from the DCO (open-loop mode) divided by two. If BDM is not enabled then the FLL is disabled in a low power state.

8.4.1.7 Bypassed Low Power Internal (BLPI)

The bypassed low power internal (BLPI) mode is entered when all the following conditions occur:

- CLKS bits are written to 01
- IREFS bit is written to 1
- PLLS bit is written to 0
- LP bit is written to 1
- BDM mode is not active

In bypassed low power internal mode, the MCGOUT clock is derived from the internal reference clock.

The PLL and the FLL are disabled at all times in BLPI mode and the MCGLCLK will not be available for BDC communications. If the BDM becomes active the mode will switch to FLL bypassed internal (FBI) mode.

8.4.1.8 Bypassed Low Power External (BLPE)

The bypassed low power external (BLPE) mode is entered when all the following conditions occur:

- CLKS bits are written to 10
- IREFS bit is written to 0
- PLLS bit is written to 0 or 1

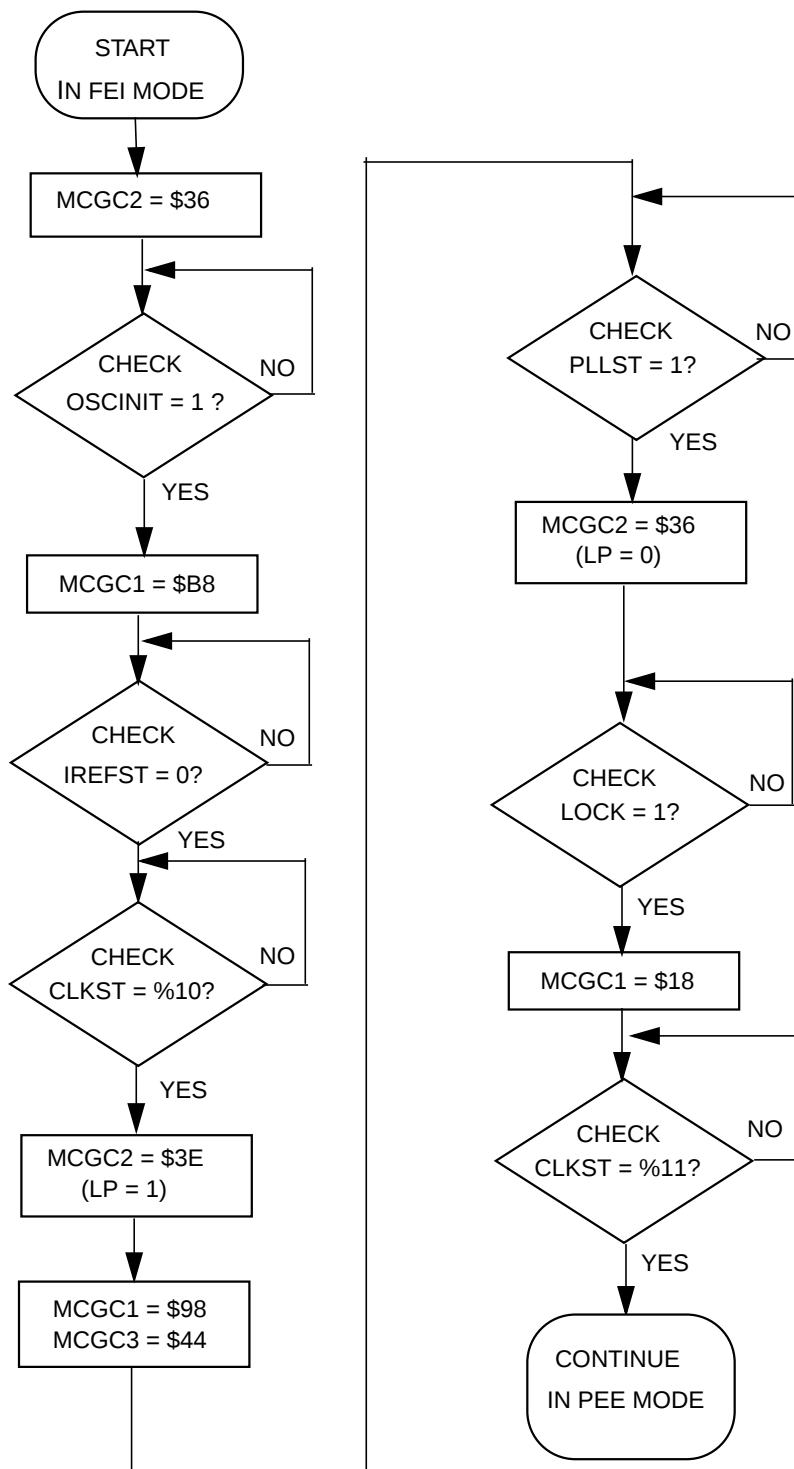


Figure 8-12. Flowchart of FEI to PEE Mode Transition using a 8 MHz crystal

Chapter 10

Analog-to-Digital Converter (S08ADC12V1)

10.1 Introduction

The 12-bit analog-to-digital converter (ADC) is a successive approximation ADC designed for operation within an integrated microcontroller system-on-chip.

NOTE

MC9S08DN60 Series devices operate at a higher voltage range (2.7 V to 5.5 V) and do not include stop1 mode. Please ignore references to stop1.

10.1.1 Analog Power and Ground Signal Names

References to V_{DDAD} and V_{SSAD} in this chapter correspond to signals V_{DDA} and V_{SSA} , respectively.

10.1.2 Channel Assignments

NOTE

The ADC channel assignments for the MC9S08DN60 Series devices are shown in [Table 10-1](#). Reserved channels convert to an unknown value.

This chapter shows bits for all S08ADC12V1 channels. MC9S08DN60 Series MCUs do not use all of these channels. All bits corresponding to channels that are not available on a device are reserved.

Table 10-13. Total Conversion Time vs. Control Conditions

Conversion Type	ADICLK	ADLSMP	Max Total Conversion Time
Subsequent continuous 10-bit or 12-bit; $f_{\text{BUS}} \geq f_{\text{ADCK}}/11$	xx	1	40 ADCK cycles

The maximum total conversion time is determined by the clock source chosen and the divide ratio selected. The clock source is selectable by the ADICLK bits, and the divide ratio is specified by the ADIV bits. For example, in 10-bit mode, with the bus clock selected as the input clock source, the input clock divide-by-1 ratio selected, and a bus frequency of 8 MHz, then the conversion time for a single conversion is:

$$\text{Conversion time} = \frac{23 \text{ ADCK Cyc}}{8 \text{ MHz}/1} + \frac{5 \text{ bus Cyc}}{8 \text{ MHz}} = 3.5 \text{ ms}$$

$$\text{Number of bus cycles} = 3.5 \text{ ms} \times 8 \text{ MHz} = 28 \text{ cycles}$$

NOTE

The ADCK frequency must be between f_{ADCK} minimum and f_{ADCK} maximum to meet ADC specifications.

10.4.5 Automatic Compare Function

The compare function can be configured to check for an upper or lower limit. After the input is sampled and converted, the result is added to the two's complement of the compare value (ADCCVH and ADCCVL). When comparing to an upper limit (ACFGT = 1), if the result is greater-than or equal-to the compare value, COCO is set. When comparing to a lower limit (ACFGT = 0), if the result is less than the compare value, COCO is set. The value generated by the addition of the conversion result and the two's complement of the compare value is transferred to ADCRH and ADCRL.

Upon completion of a conversion while the compare function is enabled, if the compare condition is not true, COCO is not set and no data is transferred to the result registers. An ADC interrupt is generated upon the setting of COCO if the ADC interrupt is enabled (AIEN = 1).

NOTE

The compare function can monitor the voltage on a channel while the MCU is in wait or stop3 mode. The ADC interrupt wakes the MCU when the compare condition is met.

10.4.6 MCU Wait Mode Operation

Wait mode is a lower power-consumption standby mode from which recovery is fast because the clock sources remain active. If a conversion is in progress when the MCU enters wait mode, it continues until completion. Conversions can be initiated while the MCU is in wait mode by means of the hardware trigger or if continuous conversions are enabled.

The bus clock, bus clock divided by two, and ADACK are available as conversion clock sources while in wait mode. The use of ALTCLK as the conversion clock source in wait is dependent on the definition of

Refer to the direct-page register summary in the [memory](#) chapter of this document for the absolute address assignments for all IIC registers. This section refers to registers and control bits only by their names. A Freescale-provided equate or header file is used to translate these names into the appropriate absolute addresses.

11.3.1 IIC Address Register (IICA)



Figure 11-3. IIC Address Register (IICA)

Table 11-1. IICA Field Descriptions

Field	Description
7–1 AD[7:1]	Slave Address. The AD[7:1] field contains the slave address to be used by the IIC module. This field is used on the 7-bit address scheme and the lower seven bits of the 10-bit address scheme.

11.3.2 IIC Frequency Divider Register (IICF)

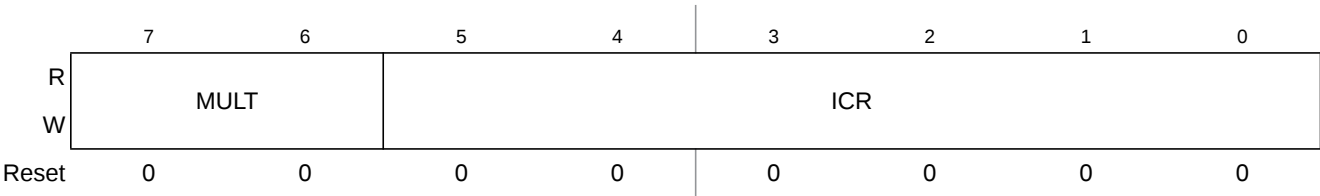


Figure 11-4. IIC Frequency Divider Register (IICF)

pin from a master and the MISO waveform applies to the MISO output from a slave. The $\overline{SS}$ OUT waveform applies to the slave select output from a master (provided MODFEN and SSOE = 1). The master $\overline{SS}$ output goes to active low one-half SPSCK cycle before the start of the transfer and goes back high at the end of the eighth bit time of the transfer. The $\overline{SS}$ IN waveform applies to the slave select input of a slave.

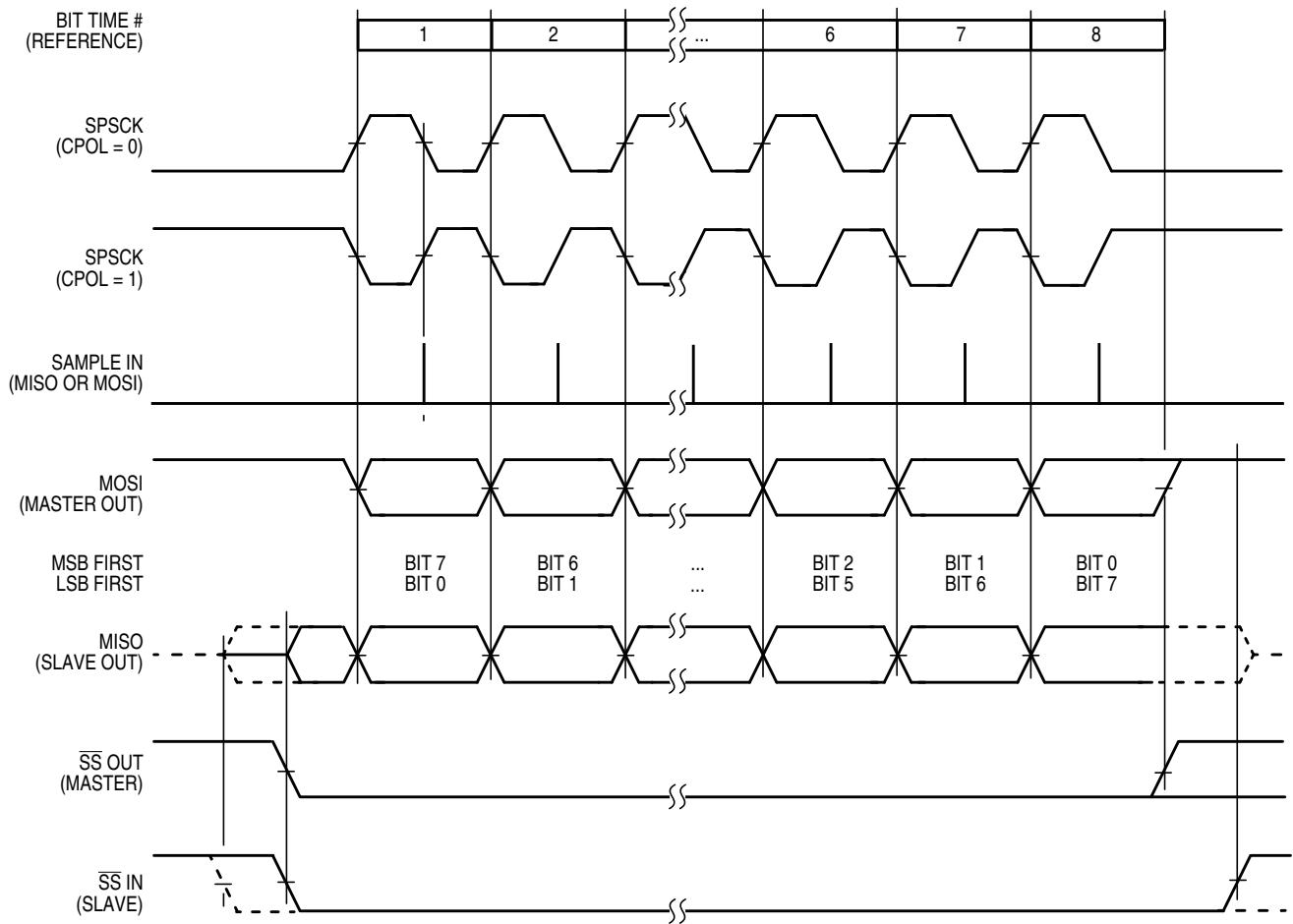


Figure 12-10. SPI Clock Formats (CPHA = 1)

When CPHA = 1, the slave begins to drive its MISO output when $\overline{SS}$ goes to active low, but the data is not defined until the first SPSCK edge. The first SPSCK edge shifts the first bit of data from the shifter onto the MOSI output of the master and the MISO output of the slave. The next SPSCK edge causes both the master and the slave to sample the data bit values on their MISO and MOSI inputs, respectively. At the third SPSCK edge, the SPI shifter shifts one bit position which shifts in the bit value that was just sampled, and shifts the second data bit value out the other end of the shifter to the MOSI and MISO outputs of the master and slave, respectively. When CHPA = 1, the slave's $\overline{SS}$ input is not required to go to its inactive high level between transfers.

Figure 12-11 shows the clock formats when CPHA = 0. At the top of the figure, the eight bit times are shown for reference with bit 1 starting as the slave is selected ($\overline{SS}$ IN goes low), and bit 8 ends at the last SPSCK edge. The MSB first and LSB first lines show the order of SPI data bits depending on the setting

Appendix A

Electrical Characteristics

A.1 Introduction

This section contains the most accurate electrical and timing information for the MC9S08DN60 Series of microcontrollers available at the time of publication.

A.2 Parameter Classification

The electrical parameters shown in this supplement are guaranteed by various methods. To give the customer a better understanding the following classification is used and the parameters are tagged accordingly in the tables where appropriate:

Table A-1. Parameter Classifications

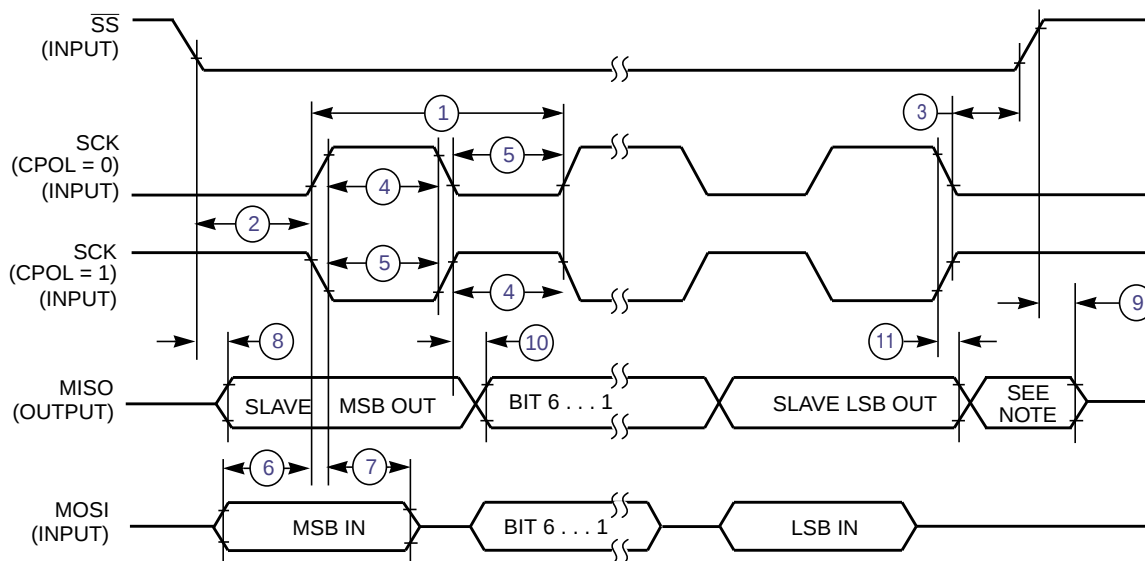
P	Those parameters are guaranteed during production testing on each individual device.
C	Those parameters are achieved by the design characterization by measuring a statistically relevant sample size across process variations.
T	Those parameters are achieved by design characterization on a small sample size from typical devices under typical conditions unless otherwise noted. All values shown in the typical column are within this category.
D	Those parameters are derived mainly from simulations.

NOTE

The classification is shown in the column labeled “C” in the parameter tables where appropriate.

A.3 Absolute Maximum Ratings

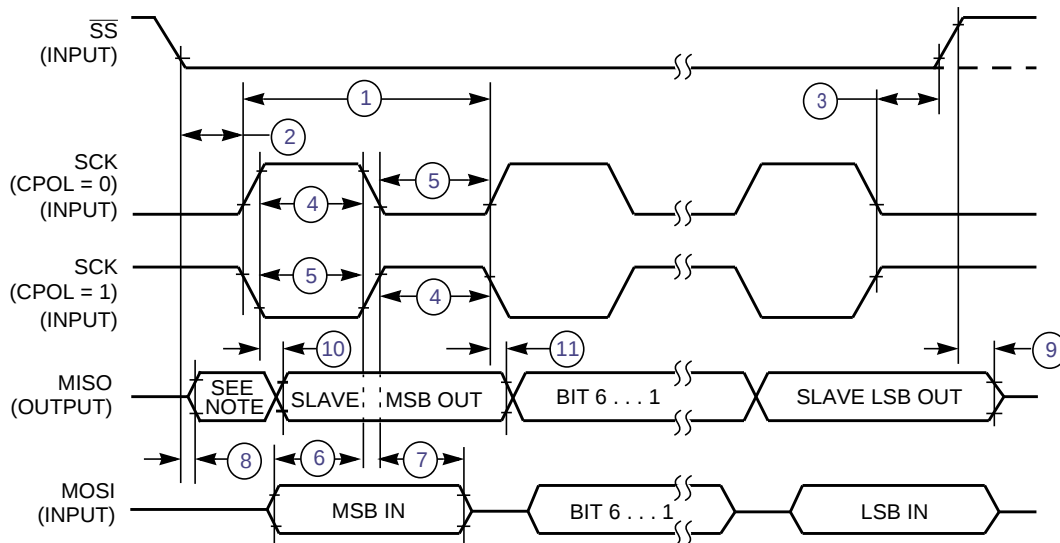
Absolute maximum ratings are stress ratings only, and functional operation at the maxima is not guaranteed. Stress beyond the limits specified in [Table A-2](#) may affect device reliability or cause permanent damage to the device. For functional operating conditions, refer to the remaining tables in this section.



NOTE:

1. Not defined but normally MSB of character just received

Figure A-9. SPI Slave Timing (CPHA = 0)



NOTE:

1. Not defined but normally LSB of character just received

Figure A-10. SPI Slave Timing (CPHA = 1)

at the transition from the value set in the modulus register and the next lower count value. This corresponds to the end of a PWM period. (The 0x0000 count value corresponds to the center of a period.)

B.4.3 Channel Event Interrupt Description

The meaning of channel interrupts depends on the current mode of the channel (input capture, output compare, edge-aligned PWM, or center-aligned PWM).

When a channel is configured as an input capture channel, the ELSnB:ELSnA control bits select rising edges, falling edges, any edge, or no edge (off) as the edge that triggers an input capture event. When the selected edge is detected, the interrupt flag is set. The flag is cleared by the 2-step sequence described in [Section B.4.1, “Clearing Timer Interrupt Flags.”](#)

When a channel is configured as an output compare channel, the interrupt flag is set each time the main timer counter matches the 16-bit value in the channel value register. The flag is cleared by the 2-step sequence described in [Section B.4.1, “Clearing Timer Interrupt Flags.”](#)

B.4.4 PWM End-of-Duty-Cycle Events

For channels that are configured for PWM operation, there are two possibilities:

- When the channel is configured for edge-aligned PWM, the channel flag is set when the timer counter matches the channel value register that marks the end of the active duty cycle period.
- When the channel is configured for center-aligned PWM, the timer count matches the channel value register twice during each PWM cycle. In this CPWM case, the channel flag is set at the start and at the end of the active duty cycle, which are the times when the timer counter matches the channel value register.

The flag is cleared by the 2-step sequence described in [Section B.4.1, “Clearing Timer Interrupt Flags.”](#)

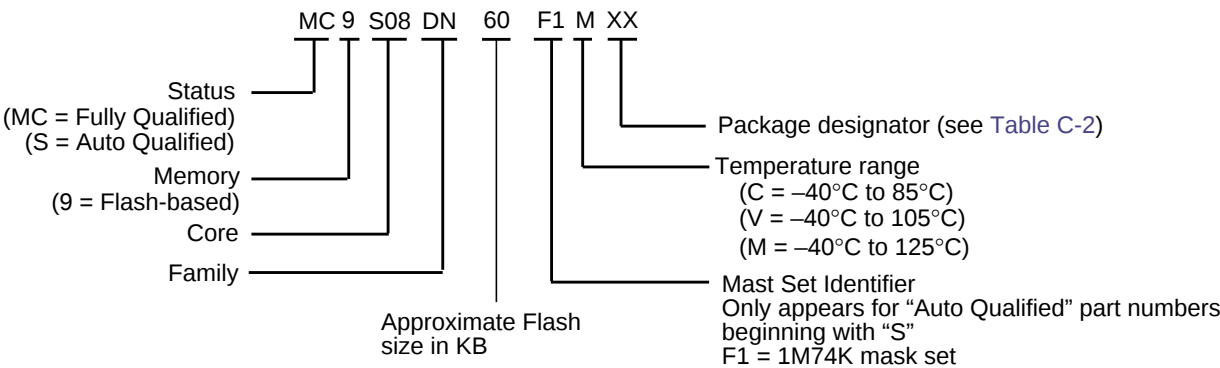
Appendix C

Ordering Information and Mechanical Drawings

C.1 Ordering Information

This section contains ordering information for MC9S08DN60 Series devices.

Example of the device numbering system:



C.1.1 MC9S08DN60 Series Devices

Table C-1. Devices in the MC9S08DN60 Series

Device Number	Memory			Available Packages ¹
	FLASH	RAM	EEPROM	
MC9S08DN60	62,080	2048	2048	64-LQFP, 48-LQFP, 32-LQFP
MC9S08DN48	49,152	2048	1536	
MC9S08DN32	33,792	1536	1024	
MC9S08DN16	16,896	1024	512	48-LQFP, 32-LQFP

¹ See Table C-2 for package information.

C.2 Mechanical Drawings

The following pages are mechanical drawings for the packages described in the following table:



NOTES:

1. DIMENSIONS AND TOLERANCING PER ASME Y14.5M—1994.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DATUM PLANE AB IS LOCATED AT BOTTOM OF LEAD AND IS COINCIDENT WITH THE LEAD WHERE THE LEAD EXITS THE PLASTIC BODY AT THE BOTTOM OF THE PARTING LINE.
4. DATUMS T, U, AND Z TO BE DETERMINED AT DATUM PLANE AB.



5. DIMENSIONS TO BE DETERMINED AT SEATING PLANE AC.



6. DIMENSIONS DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS 0.250 PER SIDE. DIMENSIONS DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT DATUM PLANE AB.



7. THIS DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION. DAMBAR PROTRUSION SHALL NOT CAUSE THE LEAD WIDTH TO EXCEED 0.350.

8. MINIMUM SOLDER PLATE THICKNESS SHALL BE 0.0076.



9. EXACT SHAPE OF EACH CORNER IS OPTIONAL.

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