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Details

Product Status	Active
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, LINbus, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	53
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	1.5K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 16x12b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LQFP
Supplier Device Package	64-LQFP (10x10)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mc9s08dn32aclh

MC9S08DN60 Series Features

8-Bit HCS08 Central Processor Unit (CPU)

- 40-MHz HCS08 CPU (20-MHz bus)
- HC08 instruction set with added BGND instruction
- Support for up to 32 interrupt/reset sources

On-Chip Memory

- Flash read/program/erase over full operating voltage and temperature
 - MC9S08DN60 = 60K
 - MC9S08DN48 = 48K
 - MC9S08DN32 = 32K
 - MC9S08DN16 = 16K
- Up to 2K EEPROM in-circuit programmable memory; 8-byte single-page or 4-byte dual-page erase sector; Program and Erase while executing Flash; Erase abort
- Up to 2K random-access memory (RAM)

Power-Saving Modes

- Two very low power stop modes
- Reduced power wait mode
- Very low power real time interrupt for use in run, wait, and stop

Clock Source Options

- Oscillator (XOSC) — Loop-control Pierce oscillator; Crystal or ceramic resonator range of 31.25 kHz to 38.4 kHz or 1 MHz to 16 MHz
- Multi-purpose Clock Generator (MCG) — PLL and FLL modes (FLL capable of 1.5% deviation using internal temperature compensation); Internal reference clock with trim adjustment (trimmed at factory, with trim value stored in flash); External reference with oscillator/resonator options

System Protection

- Watchdog computer operating properly (COP) reset with option to run from backup dedicated 1-kHz internal clock source or bus clock
- Low-voltage detection with reset or interrupt; selectable trip points
- Illegal opcode detection with reset
- Illegal address detection with reset
- Flash block protect
- Loss-of-lock protection

Development Support

- Single-wire background debug interface
- On-chip, in-circuit emulation (ICE) with real-time bus capture

Peripherals

- **ADC** — 16-channel, 12-bit resolution, 2.5 μ s conversion time, automatic compare function, temperature sensor, internal bandgap reference channel
- **ACMPx** — Two analog comparators with selectable interrupt on rising, falling, or either edge of comparator output; compare option to fixed internal bandgap reference voltage
- **SCI** — One SCI supporting LIN 2.0 Protocol and SAE J2602 protocols; Full duplex non-return to zero (NRZ); Master extended break generation; Slave extended break detection; Wakeup on active edge
- **SPI** — Full-duplex or single-wire bidirectional; Double-buffered transmit and receive; Master or Slave mode; MSB-first or LSB-first shifting
- **IIC** — Up to 100 kbps with maximum bus loading; Multi-master operation; Programmable slave address; General Call Address; Interrupt driven byte-by-byte data transfer
- **TPMx** — One 6-channel (TPM1) and one 2-channel (TPM2); Selectable input capture, output compare, or buffered edge-aligned PWM on each channel
- **RTC** — (Real-time counter) 8-bit modulus counter with binary or decimal based prescaler; Real-time clock capabilities using external crystal and RTC for precise time base, time-of-day, calendar or task scheduling functions; Free running on-chip low power oscillator (1 kHz) for cyclic wake-up without external components

Input/Output

- 53 general-purpose input/output (I/O) pins and 1 input-only pin
- 24 interrupt pins with selectable polarity on each pin
- Hysteresis and configurable pull device on all input pins.
- Configurable slew rate and drive strength on all output pins.

Package Options

- 64-pin low-profile quad flat-pack (LQFP) — 10x10 mm
- 48-pin low-profile quad flat-pack (LQFP) — 7x7 mm
- 32-pin low-profile quad flat-pack (LQFP) — 7x7 mm

NOTE

To avoid extra current drain from floating input pins, the reset initialization routine in the application program should either enable on-chip pull-up devices or change the direction of unused or non-bonded pins to outputs so they do not float.

Table 4-2. Direct-Page Register Summary (Sheet 2 of 3)

Address	Register Name	Bit 7	6	5	4	3	2	1	Bit 0
0x0028	TPM1C1SC	CH1F	CH1IE	MS1B	MS1A	ELS1B	ELS1A	0	0
0x0029	TPM1C1VH	Bit 15	14	13	12	11	10	9	Bit 8
0x002A	TPM1C1VL	Bit 7	6	5	4	3	2	1	Bit 0
0x002B	TPM1C2SC	CH2F	CH2IE	MS2B	MS2A	ELS2B	ELS2A	0	0
0x002C	TPM1C2VH	Bit 15	14	13	12	11	10	9	Bit 8
0x002D	TPM1C2VL	Bit 7	6	5	4	3	2	1	Bit 0
0x002E	TPM1C3SC	CH3F	CH3IE	MS3B	MS3A	ELS3B	ELS3A	0	0
0x002F	TPM1C3VH	Bit 15	14	13	12	11	10	9	Bit 8
0x0030	TPM1C3VL	Bit 7	6	5	4	3	2	1	Bit 0
0x0031	TPM1C4SC	CH4F	CH4IE	MS4B	MS4A	ELS4B	ELS4A	0	0
0x0032	TPM1C4VH	Bit 15	14	13	12	11	10	9	Bit 8
0x0033	TPM1C4VL	Bit 7	6	5	4	3	2	1	Bit 0
0x0034	TPM1C5SC	CH5F	CH5IE	MS5B	MS5A	ELS5B	ELS5A	0	0
0x0035	TPM1C5VH	Bit 15	14	13	12	11	10	9	Bit 8
0x0036	TPM1C5VL	Bit 7	6	5	4	3	2	1	Bit 0
0x0037	Reserved	—	—	—	—	—	—	—	—
0x0038	SCI1BDH	LBKDIE	RXEDGIE	0	SBR12	SBR11	SBR10	SBR9	SBR8
0x0039	SCI1BDL	SBR7	SBR6	SBR5	SBR4	SBR3	SBR2	SBR1	SBR0
0x003A	SCI1C1	LOOPS	SCISWAI	RSRC	M	WAKE	ILT	PE	PT
0x003B	SCI1C2	TIE	TCIE	RIE	ILIE	TE	RE	RWU	SBK
0x003C	SCI1S1	TDRE	TC	RDRF	IDLE	OR	NF	FE	PF
0x003D	SCI1S2	LBKDIF	RXEDGIF	0	RXINV	RWUID	BRK13	LBKDE	RAF
0x003E	SCI1C3	R8	T8	TXDIR	TXINV	ORIE	NEIE	FEIE	PEIE
0x003F	SCI1D	Bit 7	6	5	4	3	2	1	Bit 0
0x0040– 0x0047	Reserved	—	—	—	—	—	—	—	—
0x0048	MCGC1	CLKS		RDIV			IREFS	IRCLKEN	IREFSTEN
0x0049	MCGC2	BDIV		RANGE	HGO	LP	EREFS	ERCLKEN	EREFSTEN
0x004A	MCGTRM	TRIM							
0x004B	MCGSC	LOLS	LOCK	PLLST	IREFST	CLKST		OSCINIT	FTRIM
0x004C	MCGC3	LOLIE	PLLS	CME	0	VDIV			
0x004D– 0x004F	Reserved	—	—	—	—	—	—	—	—
0x0050	SPIC1	SPIE	SPE	SPTIE	MSTR	CPOL	CPHA	SSOE	LSBFE
0x0051	SPIC2	0	0	0	MODFEN	BIDIROE	0	SPISWAI	SPC0
0x0052	SPIBR	0	SPPR2	SPPR1	SPPR0	0	SPR2	SPR1	SPR0
0x0053	SPIS	SPRF	0	SPTEF	MODF	0	0	0	0
0x0054	Reserved	0	0	0	0	0	0	0	0
0x0055	SPID	Bit 7	6	5	4	3	2	1	Bit 0

An output pin can be selected to have high output drive strength by setting the corresponding bit in the drive strength select register (PTxDSn). When high drive is selected, a pin is capable of sourcing and sinking greater current. Even though every I/O pin can be selected as high drive, the user must ensure that the total current source and sink limits for the MCU are not exceeded. Drive strength selection is intended to affect the DC behavior of I/O pins. However, the AC behavior is also affected. High drive allows a pin to drive a greater load with the same switching speed as a low drive enabled pin into a smaller load. Because of this, the EMC emissions may be affected by enabling pins as high drive.

6.3 Pin Interrupts

Port A, port B, and port D pins can be configured as external interrupt inputs and as an external means of waking the MCU from stop or wait low-power modes.

The block diagram for each port interrupt logic is shown Figure 6-2.

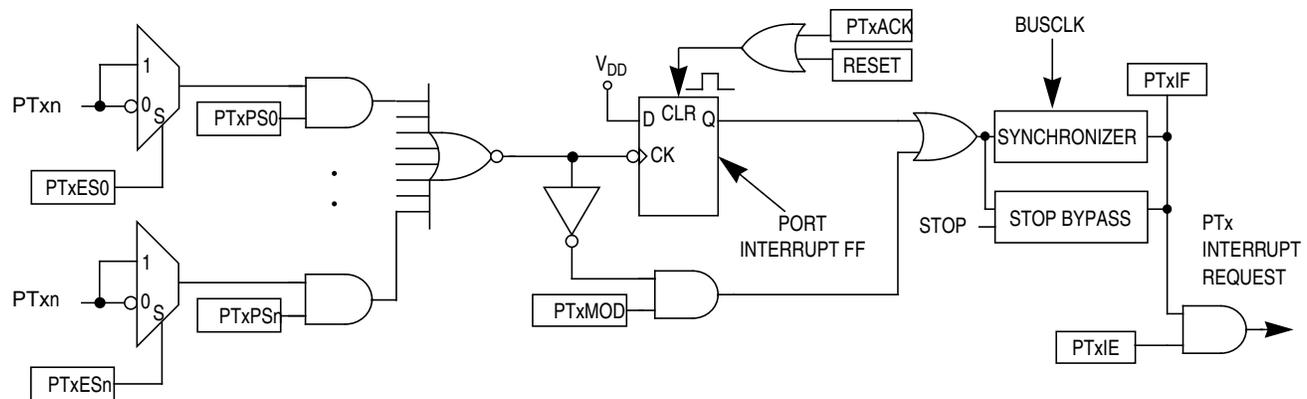


Figure 6-2. Port Interrupt Block Diagram

Writing to the PTxPSn bits in the port interrupt pin select register (PTxPS) independently enables or disables each port pin. Each port can be configured as edge sensitive or edge and level sensitive based on the PTxMOD bit in the port interrupt status and control register (PTxSC). Edge sensitivity can be software programmed to be either falling or rising; the level can be either low or high. The polarity of the edge or edge and level sensitivity is selected using the PTxESn bits in the port interrupt edge select register (PTxES).

Synchronous logic is used to detect edges. Prior to detecting an edge, enabled port inputs must be at the deasserted logic level. A falling edge is detected when an enabled port input signal is seen as a logic 1 (the deasserted level) during one bus cycle and then a logic 0 (the asserted level) during the next cycle. A rising edge is detected when the input signal is seen as a logic 0 during one bus cycle and then a logic 1 during the next cycle.

6.3.1 Edge Only Sensitivity

A valid edge on an enabled port pin will set PTxIF in PTxSC. If PTxIE in PTxSC is set, an interrupt request will be presented to the CPU. Clearing of PTxIF is accomplished by writing a 1 to PTxACK in PTxSC.

6.5.3.5 Port C Drive Strength Selection Register (PTCDS)

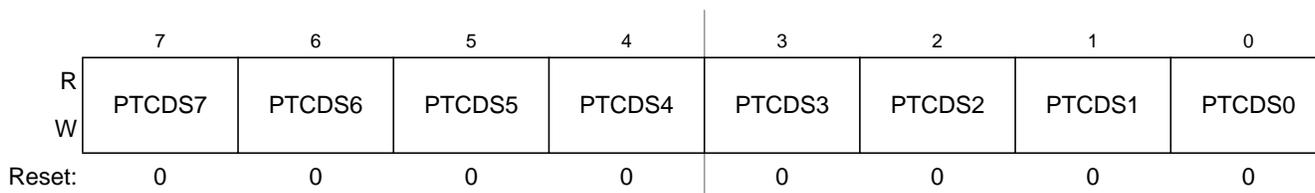


Figure 6-23. Drive Strength Selection for Port C Register (PTCDS)

Table 6-21. PTCDS Register Field Descriptions

Field	Description
7:0 PTCDS[7:0]	<p>Output Drive Strength Selection for Port C Bits — Each of these control bits selects between low and high output drive for the associated PTC pin. For port C pins that are configured as inputs, these bits have no effect.</p> <p>0 Low output drive strength selected for port C bit n.</p> <p>1 High output drive strength selected for port C bit n.</p>

8.1.2 Modes of Operation

There are nine modes of operation for the MCG:

- FLL Engaged Internal (FEI)
- FLL Engaged External (FEE)
- FLL Bypassed Internal (FBI)
- FLL Bypassed External (FBE)
- PLL Engaged External (PEE)
- PLL Bypassed External (PBE)
- Bypassed Low Power Internal (BLPI)
- Bypassed Low Power External (BLPE)
- Stop

For details see [Section 8.4.1, “Operational Modes.](#)

8.2 External Signal Description

There are no MCG signals that connect off chip.

8.4.7 Fixed Frequency Clock

The MCG presents the divided reference clock as MCGFFCLK for use as an additional clock source. The MCGFFCLK frequency must be no more than 1/4 of the MCGOUT frequency to be valid. Because of this requirement, the MCGFFCLK is not valid in bypass modes for the following combinations of BDIV and RDIV values:

- BDIV=00 (divide by 1), RDIV < 010
- BDIV=01 (divide by 2), RDIV < 011

When MCGFFCLK is valid then MCGFFCLKVALID is set to 1. When MCGFFCLK is not valid then MCGFFCLKVALID is set to 0.

8.5 Initialization / Application Information

This section describes how to initialize and configure the MCG module in application. The following sections include examples on how to initialize the MCG and properly switch between the various available modes.

8.5.1 MCG Module Initialization Sequence

The MCG comes out of reset configured for FEI mode with the BDIV set for divide-by-2. The internal reference will stabilize in t_{refst} microseconds before the FLL can acquire lock. As soon as the internal reference is stable, the FLL will acquire lock in $t_{\text{fl_lock}}$ milliseconds.

Upon POR, the internal reference will require trimming to guarantee an accurate clock. Freescale recommends using FLASH location 0xFFAE for storing the fine trim bit, FTRIM in the MCGSC register, and 0xFFAF for storing the 8-bit trim value in the MCGTRM register. The MCU will not automatically copy the values in these FLASH locations to the respective registers. Therefore, user code must copy these values from FLASH to the registers.

NOTE

The BDIV value should not be changed to divide-by-1 without first trimming the internal reference. Failure to do so could result in the MCU running out of specification.

8.5.1.1 Initializing the MCG

Because the MCG comes out of reset in FEI mode, the only MCG modes which can be directly switched to upon reset are FEE, FBE, and FBI modes (see [Figure 8-8](#)). Reaching any of the other modes requires first configuring the MCG for one of these three initial modes. Care must be taken to check relevant status bits in the MCGSC register reflecting all configuration changes within each mode.

To change from FEI mode to FEE or FBE modes, follow this procedure:

1. Enable the external clock source by setting the appropriate bits in MCGC2.
2. Write to MCGC1 to select the clock mode.

- If entering FEE, set RDIV appropriately, clear the IREFS bit to switch to the external reference, and leave the CLKS bits at %00 so that the output of the FLL is selected as the system clock source.
 - If entering FBE, clear the IREFS bit to switch to the external reference and change the CLKS bits to %10 so that the external reference clock is selected as the system clock source. The RDIV bits should also be set appropriately here according to the external reference frequency because although the FLL is bypassed, it is still on in FBE mode.
 - The internal reference can optionally be kept running by setting the IRCLKEN bit. This is useful if the application will switch back and forth between internal and external modes. For minimum power consumption, leave the internal reference disabled while in an external clock mode.
3. After the proper configuration bits have been set, wait for the affected bits in the MCGSC register to be changed appropriately, reflecting that the MCG has moved into the proper mode.
 - If ERCLKEN was set in step 1 or the MCG is in FEE, FBE, PEE, PBE, or BLPE mode, and EREFS was also set in step 1, wait here for the OSCINIT bit to become set indicating that the external clock source has finished its initialization cycles and stabilized. Typical crystal startup times are given in Appendix A, “Electrical Characteristics”.
 - If in FEE mode, check to make sure the IREFST bit is cleared and the LOCK bit is set before moving on.
 - If in FBE mode, check to make sure the IREFST bit is cleared, the LOCK bit is set, and the CLKST bits have changed to %10 indicating the external reference clock has been appropriately selected. Although the FLL is bypassed in FBE mode, it is still on and will lock in FBE mode.

To change from FEI clock mode to FBI clock mode, follow this procedure:

1. Change the CLKS bits to %01 so that the internal reference clock is selected as the system clock source.
2. Wait for the CLKST bits in the MCGSC register to change to %01, indicating that the internal reference clock has been appropriately selected.

8.5.2 MCG Mode Switching

When switching between operational modes of the MCG, certain configuration bits must be changed in order to properly move from one mode to another. Each time any of these bits are changed (PLLS, IREFS, CLKS, or EREFS), the corresponding bits in the MCGSC register (PLLST, IREFST, CLKST, or OSCINIT) must be checked before moving on in the application software.

Additionally, care must be taken to ensure that the reference clock divider (RDIV) is set properly for the mode being switched to. For instance, in PEE mode, if using a 4 MHz crystal, RDIV must be set to %001 (divide-by-2) or %010 (divide-by-4) in order to divide the external reference down to the required frequency between 1 and 2 MHz.

The RDIV and IREFS bits should always be set properly before changing the PLLS bit so that the FLL or PLL clock has an appropriate reference clock frequency to switch to.

The table below shows MCGOUT frequency calculations using RDIV, BDIV, and VDIV settings for each clock mode. The bus frequency is equal to MCGOUT divided by 2.

Table 8-6. MCGOUT Frequency Calculation Options

Clock Mode	f_{MCGOUT}^1	Note
FEI (FLL engaged internal)	$(f_{int} * 1024) / B$	Typical $f_{MCGOUT} = 16$ MHz immediately after reset. RDIV bits set to %000.
FEE (FLL engaged external)	$(f_{ext} / R * 1024) / B$	f_{ext} / R must be in the range of 31.25 kHz to 39.0625 kHz
FBE (FLL bypassed external)	f_{ext} / B	f_{ext} / R must be in the range of 31.25 kHz to 39.0625 kHz
FBI (FLL bypassed internal)	f_{int} / B	Typical $f_{int} = 32$ kHz
PEE (PLL engaged external)	$[(f_{ext} / R) * M] / B$	f_{ext} / R must be in the range of 1 MHz to 2 MHz
PBE (PLL bypassed external)	f_{ext} / B	f_{ext} / R must be in the range of 1 MHz to 2 MHz
BLPI (Bypassed low power internal)	f_{int} / B	
BLPE (Bypassed low power external)	f_{ext} / B	

¹ R is the reference divider selected by the RDIV bits, B is the bus frequency divider selected by the BDIV bits, and M is the multiplier selected by the VDIV bits.

This section will include 3 mode switching examples using a 4 MHz external crystal. If using an external clock source less than 1 MHz, the MCG should not be configured for any of the PLL modes (PEE and PBE).

8.5.2.1 Example # 1: Moving from FEI to PEE Mode: External Crystal = 4 MHz, Bus Frequency = 8 MHz

In this example, the MCG will move through the proper operational modes from FEI to PEE mode until the 4 MHz crystal reference frequency is set to achieve a bus frequency of 8 MHz. Because the MCG is in FEI mode out of reset, this example also shows how to initialize the MCG for PEE mode out of reset. First, the code sequence will be described. Then a flowchart will be included which illustrates the sequence.

1. First, FEI must transition to FBE mode:
 - a) MCGC2 = 0x36 (%00110110)
 - BDIV (bits 7 and 6) set to %00, or divide-by-1
 - RANGE (bit 5) set to 1 because the frequency of 4 MHz is within the high frequency range
 - HGO (bit 4) set to 1 to configure external oscillator for high gain operation
 - EREFS (bit 2) set to 1, because a crystal is being used
 - ERCLKEN (bit 1) set to 1 to ensure the external reference clock is active
 - b) Loop until OSCINIT (bit 1) in MCGSC is 1, indicating the crystal selected by the EREFS bit has been initialized.

Table 10-6. ADCCFG Register Field Descriptions (continued)

Field	Description
3:2 MODE	Conversion Mode Selection. MODE bits are used to select between 12-, 10-, or 8-bit operation. See Table 10-8 .
1:0 ADICLK	Input Clock Select. ADICLK bits select the input clock source to generate the internal clock ADCK. See Table 10-9 .

Table 10-7. Clock Divide Select

ADIV	Divide Ratio	Clock Rate
00	1	Input clock
01	2	Input clock ÷ 2
10	4	Input clock ÷ 4
11	8	Input clock ÷ 8

Table 10-8. Conversion Modes

MODE	Mode Description
00	8-bit conversion (N=8)
01	12-bit conversion (N=12)
10	10-bit conversion (N=10)
11	Reserved

Table 10-9. Input Clock Select

ADICLK	Selected Clock Source
00	Bus clock
01	Bus clock divided by 2
10	Alternate clock (ALTCLK)
11	Asynchronous clock (ADACK)

10.3.8 Pin Control 1 Register (APCTL1)

The pin control registers disable the I/O port control of MCU pins used as analog inputs. APCTL1 is

Table 10-12. APCTL3 Register Field Descriptions

Field	Description
7 ADPC23	ADC Pin Control 23. ADPC23 controls the pin associated with channel AD23. 0 AD23 pin I/O control enabled 1 AD23 pin I/O control disabled
6 ADPC22	ADC Pin Control 22. ADPC22 controls the pin associated with channel AD22. 0 AD22 pin I/O control enabled 1 AD22 pin I/O control disabled
5 ADPC21	ADC Pin Control 21. ADPC21 controls the pin associated with channel AD21. 0 AD21 pin I/O control enabled 1 AD21 pin I/O control disabled
4 ADPC20	ADC Pin Control 20. ADPC20 controls the pin associated with channel AD20. 0 AD20 pin I/O control enabled 1 AD20 pin I/O control disabled
3 ADPC19	ADC Pin Control 19. ADPC19 controls the pin associated with channel AD19. 0 AD19 pin I/O control enabled 1 AD19 pin I/O control disabled
2 ADPC18	ADC Pin Control 18. ADPC18 controls the pin associated with channel AD18. 0 AD18 pin I/O control enabled 1 AD18 pin I/O control disabled
1 ADPC17	ADC Pin Control 17. ADPC17 controls the pin associated with channel AD17. 0 AD17 pin I/O control enabled 1 AD17 pin I/O control disabled
0 ADPC16	ADC Pin Control 16. ADPC16 controls the pin associated with channel AD16. 0 AD16 pin I/O control enabled 1 AD16 pin I/O control disabled

10.4 Functional Description

The ADC module is disabled during reset or when the ADCH bits are all high. The module is idle when a conversion has completed and another conversion has not been initiated. When idle, the module is in its lowest power state.

The ADC can perform an analog-to-digital conversion on any of the software selectable channels. In 12-bit and 10-bit mode, the selected channel voltage is converted by a successive approximation algorithm into a 12-bit digital result. In 8-bit mode, the selected channel voltage is converted by a successive approximation algorithm into a 9-bit digital result.

When the conversion is completed, the result is placed in the data registers (ADCRH and ADCRL). In 10-bit mode, the result is rounded to 10 bits and placed in the data registers (ADCRH and ADCRL). In 8-bit mode, the result is rounded to 8 bits and placed in ADCRL. The conversion complete flag (COCO) is then set and an interrupt is generated if the conversion complete interrupt has been enabled (AIEN = 1).

The ADC module has the capability of automatically comparing the result of a conversion with the contents of its compare registers. The compare function is enabled by setting the ACFE bit and operates with any of the conversion modes and configurations.

Table 11-2. IICF Field Descriptions

Field	Description
7–6 MULT	<p>IIC Multiplier Factor. The MULT bits define the multiplier factor, mul. This factor, along with the SCL divider, generates the IIC baud rate. The multiplier factor mul as defined by the MULT bits is provided below.</p> <p>00 mul = 01 01 mul = 02 10 mul = 04 11 Reserved</p>
5–0 ICR	<p>IIC Clock Rate. The ICR bits are used to prescale the bus clock for bit rate selection. These bits and the MULT bits determine the IIC baud rate, the SDA hold time, the SCL Start hold time, and the SCL Stop hold time. Table 11-4 provides the SCL divider and hold values for corresponding values of the ICR.</p> <p>The SCL divider multiplied by multiplier factor mul generates IIC baud rate.</p> $\text{IIC baud rate} = \frac{\text{bus speed (Hz)}}{\text{mul} \times \text{SCLdivider}} \quad \text{Eqn. 11-1}$ <p>SDA hold time is the delay from the falling edge of SCL (IIC clock) to the changing of SDA (IIC data).</p> $\text{SDA hold time} = \text{bus period (s)} \times \text{mul} \times \text{SDA hold value} \quad \text{Eqn. 11-2}$ <p>SCL start hold time is the delay from the falling edge of SDA (IIC data) while SCL is high (Start condition) to the falling edge of SCL (IIC clock).</p> $\text{SCL Start hold time} = \text{bus period (s)} \times \text{mul} \times \text{SCL Start hold value} \quad \text{Eqn. 11-3}$ <p>SCL stop hold time is the delay from the rising edge of SCL (IIC clock) to the rising edge of SDA (IIC data) while SCL is high (Stop condition).</p> $\text{SCL Stop hold time} = \text{bus period (s)} \times \text{mul} \times \text{SCL Stop hold value} \quad \text{Eqn. 11-4}$

For example, if the bus speed is 8 MHz, the table below shows the possible hold time values with different ICR and MULT selections to achieve an IIC baud rate of 100kbps.

Table 11-3. Hold Time Values for 8 MHz Bus Speed

MULT	ICR	Hold Times (μs)		
		SDA	SCL Start	SCL Stop
0x2	0x00	3.500	3.000	5.500
0x1	0x07	2.500	4.000	5.250
0x1	0x0B	2.250	4.000	5.250
0x0	0x14	2.125	4.250	5.125
0x0	0x18	1.125	4.750	5.125

Chapter 12

Serial Peripheral Interface (S08SPIV3)

12.1 Introduction

The serial peripheral interface (SPI) module provides for full-duplex, synchronous, serial communication between the MCU and peripheral devices. These peripheral devices can include other microcontrollers, analog-to-digital converters, shift registers, sensors, memories, etc.

The SPI runs at a baud rate up to the bus clock divided by two in master mode and bus clock divided by four in slave mode.

All devices in the MC9S08DN60 Series MCUs contain one SPI module, as shown in the following block diagram.

NOTE

Ensure that the SPI should not be disabled ($SPE=0$) at the same time as a bit change to the CPHA bit. These changes should be performed as separate operations or unexpected behavior may occur.

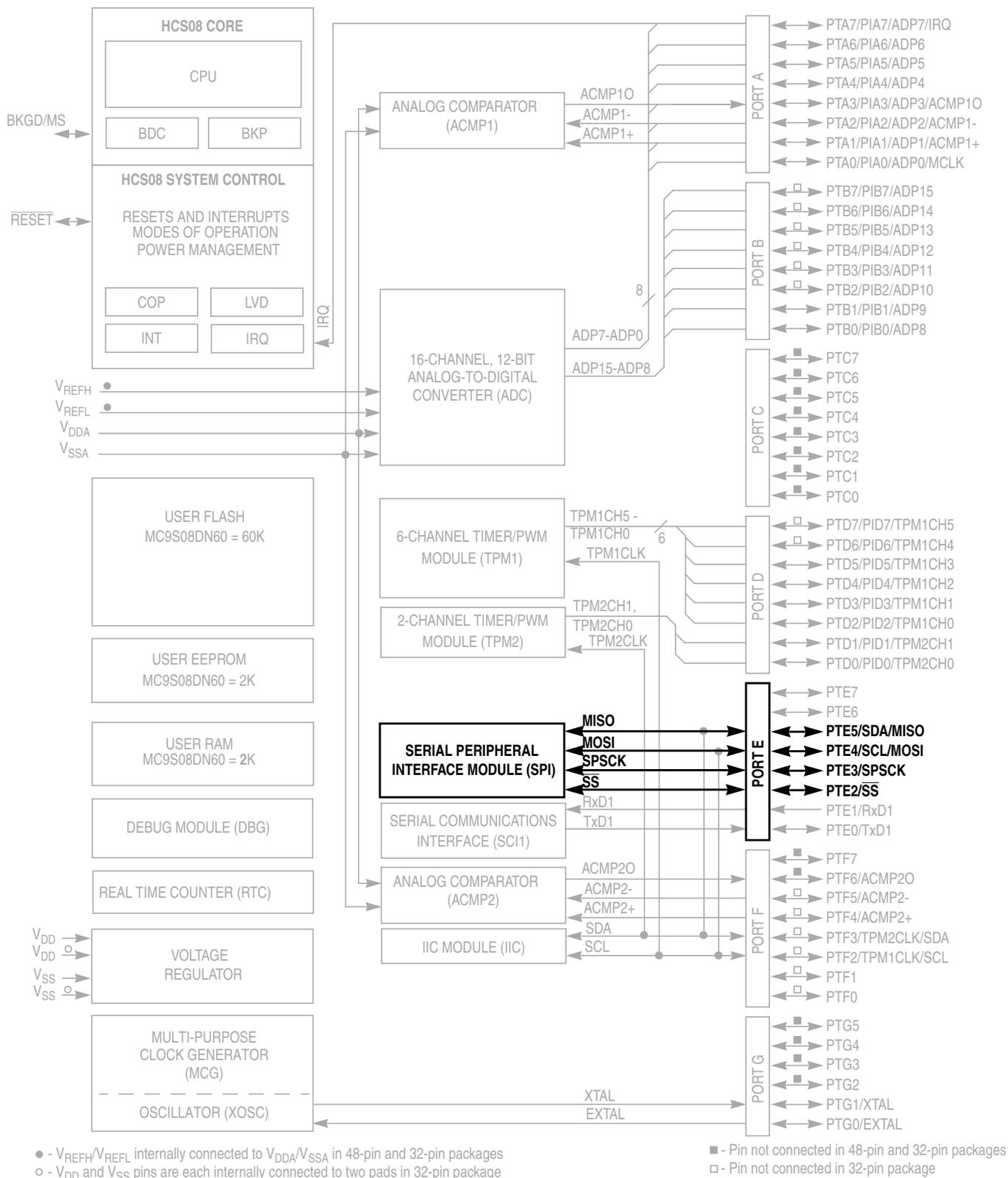


Figure 12-1. MC9S08DN60 Block Diagram

12.5 Functional Description

An SPI transfer is initiated by checking for the SPI transmit buffer empty flag (SPTEF = 1) and then writing a byte of data to the SPI data register (SPID) in the master SPI device. When the SPI shift register is available, this byte of data is moved from the transmit data buffer to the shifter, SPTEF is set to indicate there is room in the buffer to queue another transmit character if desired, and the SPI serial transfer starts.

During the SPI transfer, data is sampled (read) on the MISO pin at one SPSCCK edge and shifted, changing the bit value on the MOSI pin, one-half SPSCCK cycle later. After eight SPSCCK cycles, the data that was in the shift register of the master has been shifted out the MOSI pin to the slave while eight bits of data were shifted in the MISO pin into the master's shift register. At the end of this transfer, the received data byte is moved from the shifter into the receive data buffer and SPRF is set to indicate the data can be read by reading SPID. If another byte of data is waiting in the transmit buffer at the end of a transfer, it is moved into the shifter, SPTEF is set, and a new transfer is started.

Normally, SPI data is transferred most significant bit (MSB) first. If the least significant bit first enable (LSBFE) bit is set, SPI data is shifted LSB first.

When the SPI is configured as a slave, its \overline{SS} pin must be driven low before a transfer starts and \overline{SS} must stay low throughout the transfer. If a clock format where CPHA = 0 is selected, \overline{SS} must be driven to a logic 1 between successive transfers. If CPHA = 1, \overline{SS} may remain low between successive transfers. See [Section 12.5.1, "SPI Clock Formats"](#) for more details.

Because the transmitter and receiver are double buffered, a second byte, in addition to the byte currently being shifted out, can be queued into the transmit data buffer, and a previously received character can be in the receive data buffer while a new character is being shifted in. The SPTEF flag indicates when the transmit buffer has room for a new character. The SPRF flag indicates when a received character is available in the receive data buffer. The received character must be read out of the receive buffer (read SPID) before the next transfer is finished or a receive overrun error results.

In the case of a receive overrun, the new data is lost because the receive buffer still held the previous character and was not ready to accept the new data. There is no indication for such an overrun condition so the application system designer must ensure that previous data has been read from the receive buffer before a new transfer is initiated.

12.5.1 SPI Clock Formats

To accommodate a wide variety of synchronous serial peripherals from different manufacturers, the SPI system has a clock polarity (CPOL) bit and a clock phase (CPHA) control bit to select one of four clock formats for data transfers. CPOL selectively inserts an inverter in series with the clock. CPHA chooses between two different clock phase relationships between the clock and data.

[Figure 12-10](#) shows the clock formats when CPHA = 1. At the top of the figure, the eight bit times are shown for reference with bit 1 starting at the first SPSCCK edge and bit 8 ending one-half SPSCCK cycle after the sixteenth SPSCCK edge. The MSB first and LSB first lines show the order of SPI data bits depending on the setting in LSBFE. Both variations of SPSCCK polarity are shown, but only one of these waveforms applies for a specific transfer, depending on the value in CPOL. The SAMPLE IN waveform applies to the MOSI input of a slave or the MISO input of a master. The MOSI waveform applies to the MOSI output

15.4.1.3 Counting Modes

The main timer counter has two counting modes. When center-aligned PWM is selected (CPWMS=1), the counter operates in up/down counting mode. Otherwise, the counter operates as a simple up counter. As an up counter, the timer counter counts from 0x0000 through its terminal count and then continues with 0x0000. The terminal count is 0xFFFF or a modulus value in TPMxMODH:TPMxMODL.

When center-aligned PWM operation is specified, the counter counts up from 0x0000 through its terminal count and then down to 0x0000 where it changes back to up counting. Both 0x0000 and the terminal count value are normal length counts (one timer clock period long). In this mode, the timer overflow flag (TOF) becomes set at the end of the terminal-count period (as the count changes to the next lower count value).

15.4.1.4 Manual Counter Reset

The main timer counter can be manually reset at any time by writing any value to either half of TPMxCNTH or TPMxCNTL. Resetting the counter in this manner also resets the coherency mechanism in case only half of the counter was read before resetting the count.

15.4.2 Channel Mode Selection

Provided CPWMS=0, the MSnB and MSnA control bits in the channel n status and control registers determine the basic mode of operation for the corresponding channel. Choices include input capture, output compare, and edge-aligned PWM.

15.4.2.1 Input Capture Mode

With the input-capture function, the TPM can capture the time at which an external event occurs. When an active edge occurs on the pin of an input-capture channel, the TPM latches the contents of the TPM counter into the channel-value registers (TPMxCnVH:TPMxCnVL). Rising edges, falling edges, or any edge may be chosen as the active edge that triggers an input capture.

In input capture mode, the TPMxCnVH and TPMxCnVL registers are read only.

When either half of the 16-bit capture register is read, the other half is latched into a buffer to support coherent 16-bit accesses in big-endian or little-endian order. The coherency sequence can be manually reset by writing to the channel status/control register (TPMxCnSC).

An input capture event sets a flag bit (CHnF) which may optionally generate a CPU interrupt request.

While in BDM, the input capture function works as configured by the user. When an external event occurs, the TPM latches the contents of the TPM counter (which is frozen because of the BDM mode) into the channel value registers and sets the flag bit.

15.4.2.2 Output Compare Mode

With the output-compare function, the TPM can generate timed pulses with programmable position, polarity, duration, and frequency. When the counter reaches the value in the channel-value registers of an output-compare channel, the TPM can set, clear, or toggle the channel pin.

15.6.2.1.2 Center-Aligned PWM Case

When CPWMS=1, TOF gets set when the timer counter changes direction from up-counting to down-counting at the end of the terminal count (the value in the modulo register). In this case the TOF corresponds to the end of a PWM period.

15.6.2.2 Channel Event Interrupt Description

The meaning of channel interrupts depends on the channel's current mode (input-capture, output-compare, edge-aligned PWM, or center-aligned PWM).

15.6.2.2.1 Input Capture Events

When a channel is configured as an input capture channel, the ELSnB:ELSnA control bits select no edge (off), rising edges, falling edges or any edge as the edge which triggers an input capture event. When the selected edge is detected, the interrupt flag is set. The flag is cleared by the two-step sequence described in Section 15.6.2, "Description of Interrupt Operation."

15.6.2.2.2 Output Compare Events

When a channel is configured as an output compare channel, the interrupt flag is set each time the main timer counter matches the 16-bit value in the channel value register. The flag is cleared by the two-step sequence described Section 15.6.2, "Description of Interrupt Operation."

15.6.2.2.3 PWM End-of-Duty-Cycle Events

For channels configured for PWM operation there are two possibilities. When the channel is configured for edge-aligned PWM, the channel flag gets set when the timer counter matches the channel value register which marks the end of the active duty cycle period. When the channel is configured for center-aligned PWM, the timer count matches the channel value register twice during each PWM cycle. In this CPWM case, the channel flag is set at the start and at the end of the active duty cycle period which are the times when the timer counter matches the channel value register. The flag is cleared by the two-step sequence described Section 15.6.2, "Description of Interrupt Operation."

15.7 The Differences from TPM v2 to TPM v3

1. Write to TPMxCNTH:L registers (Section 15.3.2, "TPM-Counter Registers (TPMxCNTH:TPMxCNTL)") [SE110-TPM case 7]

Any write to TPMxCNTH or TPMxCNTL registers in TPM v3 clears the TPM counter (TPMxCNTH:L) and the prescaler counter. Instead, in the TPM v2 only the TPM counter is cleared in this case.

2. Read of TPMxCNTH:L registers (Section 15.3.2, "TPM-Counter Registers (TPMxCNTH:TPMxCNTL)")

— In TPM v3, any read of TPMxCNTH:L registers during BDM mode returns the value of the TPM counter that is frozen. In TPM v2, if only one byte of the TPMxCNTH:L registers was read before the BDM mode became active, then any read of TPMxCNTH:L registers during

16.4.3.9 Debug Status Register (DBGS)

This is a read-only status register.

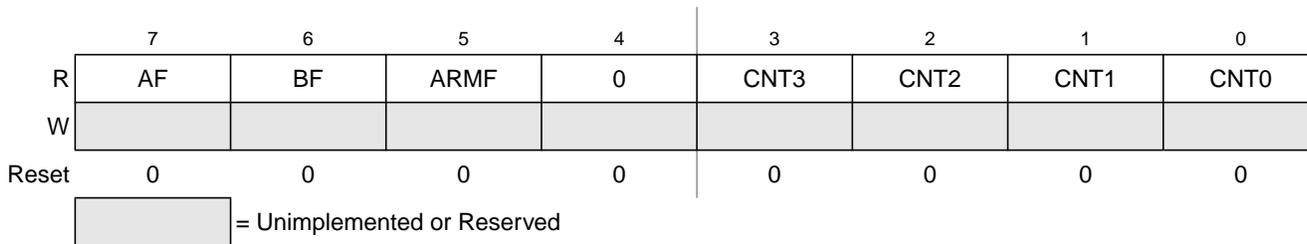


Figure 16-9. Debug Status Register (DBGS)

Table 16-6. DBGS Register Field Descriptions

Field	Description
7 AF	Trigger Match A Flag — AF is cleared at the start of a debug run and indicates whether a trigger match A condition was met since arming. 0 Comparator A has not matched 1 Comparator A match
6 BF	Trigger Match B Flag — BF is cleared at the start of a debug run and indicates whether a trigger match B condition was met since arming. 0 Comparator B has not matched 1 Comparator B match
5 ARMF	Arm Flag — While DBGEN = 1, this status bit is a read-only image of ARM in DBG. This bit is set by writing 1 to the ARM control bit in DBG (while DBGEN = 1) and is automatically cleared at the end of a debug run. A debug run is completed when the FIFO is full (begin trace) or when a trigger event is detected (end trace). A debug run can also be ended manually by writing 0 to ARM or DBGEN in DBG. 0 Debugger not armed 1 Debugger armed
3:0 CNT[3:0]	FIFO Valid Count — These bits are cleared at the start of a debug run and indicate the number of words of valid data in the FIFO at the end of a debug run. The value in CNT does not decrement as data is read out of the FIFO. The external debug host is responsible for keeping track of the count as information is read out of the FIFO. 0000 Number of valid words in FIFO = No valid data 0001 Number of valid words in FIFO = 1 0010 Number of valid words in FIFO = 2 0011 Number of valid words in FIFO = 3 0100 Number of valid words in FIFO = 4 0101 Number of valid words in FIFO = 5 0110 Number of valid words in FIFO = 6 0111 Number of valid words in FIFO = 7 1000 Number of valid words in FIFO = 8

B.2.4 Timer Channel n Status and Control Register (TPMxCnSC)

TPMxCnSC contains the channel interrupt status flag and control bits that are used to configure the interrupt enable, channel configuration, and pin function.

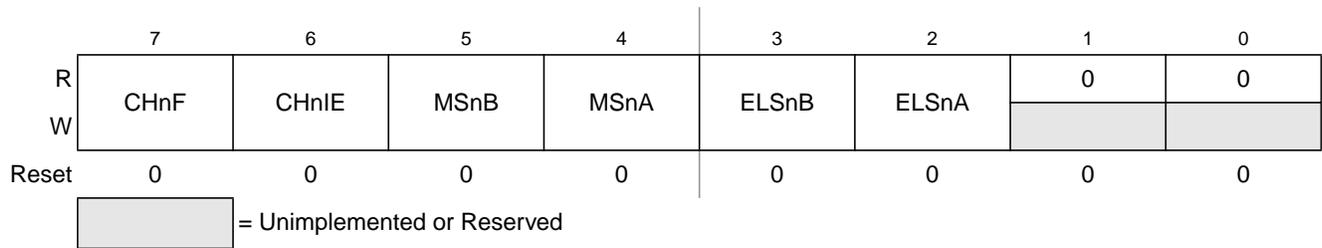


Figure B-7. Timer Channel n Status and Control Register (TPMxCnSC)

Table B-4. TPMxCnSC Register Field Descriptions

Field	Description
7 CHnF	<p>Channel n Flag — When channel n is configured for input capture, this flag bit is set when an active edge occurs on the channel n pin. When channel n is an output compare or edge-aligned PWM channel, CHnF is set when the value in the TPM counter registers matches the value in the TPM channel n value registers. This flag is seldom used with center-aligned PWMs because it is set every time the counter matches the channel value register, which correspond to both edges of the active duty cycle period.</p> <p>A corresponding interrupt is requested when CHnF is set and interrupts are enabled (CHnIE = 1). Clear CHnF by reading TPMxCnSC while CHnF is set and then writing a 0 to CHnF. If another interrupt request occurs before the clearing sequence is complete, the sequence is reset so CHnF would remain set after the clear sequence was completed for the earlier CHnF. This is done so a CHnF interrupt request cannot be lost by clearing a previous CHnF. Reset clears CHnF. Writing a 1 to CHnF has no effect.</p> <p>0 No input capture or output compare event occurred on channel n 1 Input capture or output compare event occurred on channel n</p>
6 CHnIE	<p>Channel n Interrupt Enable — This read/write bit enables interrupts from channel n. Reset clears CHnIE.</p> <p>0 Channel n interrupt requests disabled (use software polling) 1 Channel n interrupt requests enabled</p>
5 MSnB	<p>Mode Select B for TPM Channel n — When CPWMS = 0, MSnB = 1 configures TPM channel n for edge-aligned PWM mode. For a summary of channel mode and setup controls, refer to Table B-5.</p>
4 MSnA	<p>Mode Select A for TPM Channel n — When CPWMS = 0 and MSnB = 0, MSnA configures TPM channel n for input capture mode or output compare mode. Refer to Table B-5 for a summary of channel mode and setup controls.</p>
3:2 ELSn[B:A]	<p>Edge/Level Select Bits — Depending on the operating mode for the timer channel as set by CPWMS:MSnB:MSnA and shown in Table B-5, these bits select the polarity of the input edge that triggers an input capture event, select the level that will be driven in response to an output compare match, or select the polarity of the PWM output.</p> <p>Setting ELSnB:ELSnA to 0:0 configures the related timer pin as a general-purpose I/O pin unrelated to any timer channel functions. This function is typically used to temporarily disable an input capture channel or to make the timer pin available as a general-purpose I/O pin when the associated timer channel is set up as a software timer that does not require the use of a pin.</p>

the value in the TPMxCNTH:TPMxCNTL counter is 0x0000. (The new duty cycle does not take effect until the next full period.)

B.3.3 Center-Aligned PWM Mode

This type of PWM output uses the up-/down-counting mode of the timer counter (CPWMS = 1). The output compare value in TPMxCnVH:TPMxCnVL determines the pulse width (duty cycle) of the PWM signal and the period is determined by the value in TPMxMODH:TPMxMODL. TPMxMODH:TPMxMODL should be kept in the range of 0x0001 to 0x7FFF because values outside this range can produce ambiguous results. ELSnA will determine the polarity of the CPWM output.

$$\text{pulse width} = 2 \times (\text{TPMxCnVH:TPMxCnVL}) \quad \text{Eqn. 16-1}$$

$$\begin{aligned} \text{period} &= 2 \times (\text{TPMxMODH:TPMxMODL}); \\ &\text{for TPMxMODH:TPMxMODL} = 0\text{x}0001\text{--}0\text{x}7\text{FFF} \end{aligned} \quad \text{Eqn. 16-2}$$

If the channel value register TPMxCnVH:TPMxCnVL is zero or negative (bit 15 set), the duty cycle will be 0%. If TPMxCnVH:TPMxCnVL is a positive value (bit 15 clear) and is greater than the (nonzero) modulus setting, the duty cycle will be 100% because the duty cycle compare will never occur. This implies the usable range of periods set by the modulus register is 0x0001 through 0x7FFE (0x7FFF if generation of 100% duty cycle is not necessary). This is not a significant limitation because the resulting period is much longer than required for normal applications.

TPMxMODH:TPMxMODL = 0x0000 is a special case that should not be used with center-aligned PWM mode. When CPWMS = 0, this case corresponds to the counter running free from 0x0000 through 0xFFFF, but when CPWMS = 1 the counter needs a valid match to the modulus register somewhere other than at 0x0000 in order to change directions from up-counting to down-counting.

Figure B-11 shows the output compare value in the TPM channel registers (multiplied by 2), which determines the pulse width (duty cycle) of the CPWM signal. If ELSnA = 0, the compare match while counting up forces the CPWM output signal low and a compare match while counting down forces the output high. The counter counts up until it reaches the modulo setting in TPMxMODH:TPMxMODL, then counts down until it reaches zero. This sets the period equal to two times TPMxMODH:TPMxMODL.

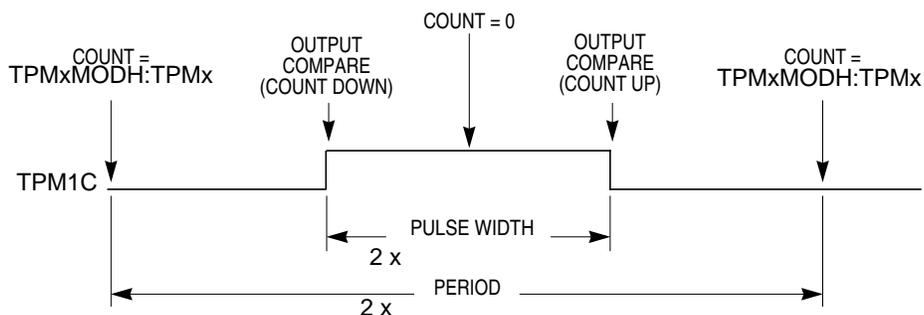


Figure B-11. CPWM Period and Pulse Width (ELSnA = 0)

Center-aligned PWM outputs typically produce less noise than edge-aligned PWMs because fewer I/O pin transitions are lined up at the same system clock edge. This type of PWM is also required for some types of motor drives.