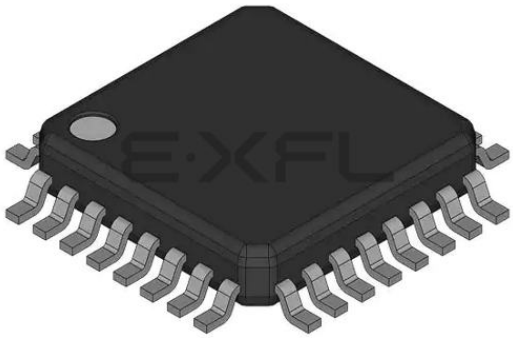


Welcome to E-XFL.COM

What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	I²C, LINbus, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	1.5K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 10x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	32-LQFP
Supplier Device Package	32-LQFP (7x7)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=s9s08dn32f2vlc

Table 4-3. High-Page Register Summary (Sheet 2 of 3)

Address	Register Name	Bit 7	6	5	4	3	2	1	Bit 0
0x1806	SDIDH	—	—	—	—	ID11	ID10	ID9	ID8
0x1807	SDIDL	ID7	ID6	ID5	ID4	ID3	ID2	ID1	ID0
0x1808	Reserved	—	—	—	—	—	—	—	—
0x1809	SPMSC1	LVWF	LVWACK	LVWIE	LVDRE	LVDSE	LVDE	0	BGBE
0x180A	SPMSC2	0	0	LVDV	LVWV	PPDF	PPDACK	0	PPDC
0x180B–0x180F	Reserved	— —	— —	— —	— —	— —	— —	— —	— —
0x1810	DBGCAH	Bit 15	14	13	12	11	10	9	Bit 8
0x1811	DBGCAL	Bit 7	6	5	4	3	2	1	Bit 0
0x1812	DBGCBH	Bit 15	14	13	12	11	10	9	Bit 8
0x1813	DBGCBL	Bit 7	6	5	4	3	2	1	Bit 0
0x1814	DBGFH	Bit 15	14	13	12	11	10	9	Bit 8
0x1815	DBGFL	Bit 7	6	5	4	3	2	1	Bit 0
0x1816	DBGC	DBGEN	ARM	TAG	BRKEN	RWA	RWAEN	RWB	RWBEN
0x1817	DBGT	TRGSEL	BEGIN	0	0	TRG3	TRG2	TRG1	TRG0
0x1818	DBGS	AF	BF	ARMF	0	CNT3	CNT2	CNT1	CNT0
0x1819–0x181F	Reserved	— —	— —	— —	— —	— —	— —	— —	— —
0x1820	FCDIV	DIVLD	PRDIV8	DIV					
0x1821	FOPT	KEYEN	FNORED	EPGMOD	0	0	0	SEC	
0x1822	Reserved	—	—	—	—	—	—	—	—
0x1823	FCNFG	0	EPGSEL	KEYACC	Reserved[1]	0	0	0	1
0x1824	FPROT	EPS		FPS					
0x1825	FSTAT	FCBEF	FCCF	FPVIOL	FACCERR	0	FBLANK	0	0
0x1826	FCMD	FCMD							
0x1827–0x183F	Reserved	— —	— —	— —	— —	— —	— —	— —	— —
0x1840	PTAPE	PTAPE7	PTAPE6	PTAPE5	PTAPE4	PTAPE3	PTAPE2	PTAPE1	PTAPE0
0x1841	PTASE	PTASE7	PTASE6	PTASE5	PTASE4	PTASE3	PTASE2	PTASE1	PTASE0
0x1842	PTADS	PTADS7	PTADS6	PTADS5	PTADS4	PTADS3	PTADS2	PTADS1	PTADS0
0x1843	Reserved	—	—	—	—	—	—	—	—
0x1844	PTASC	0	0	0	0	PTAIF	PTAACK	PTAIE	PTAMOD
0x1845	PTAPS	PTAPS7	PTAPS6	PTAPS5	PTAPS4	PTAPS3	PTAPS2	PTAPS1	PTAPS0
0x1846	PTAES	PTAES7	PTAES6	PTAES5	PTAES4	PTAES3	PTAES2	PTAES1	PTAES0
0x1847	Reserved	—	—	—	—	—	—	—	—
0x1848	PTBPE	PTBPE7	PTBPE6	PTBPE5	PTBPE4	PTBPE3	PTBPE2	PTBPE1	PTBPE0
0x1849	PTBSE	PTBSE7	PTBSE6	PTBSE5	PTBSE4	PTBSE3	PTBSE2	PTBSE1	PTBSE0
0x184A	PTBDS	PTBDS7	PTBDS6	PTBDS5	PTBDS4	PTBDS3	PTBDS2	PTBDS1	PTBDS0
0x184B	Reserved	—	—	—	—	—	—	—	—
0x184C	PTBSC	0	0	0	0	PTBIF	PTBACK	PTBIE	PTBMOD

4.5.5 Sector Erase Abort

The sector erase abort operation will terminate the active sector erase operation so that other sectors are available for read and program operations without waiting for the sector erase operation to complete.

The sector erase abort command write sequence is as follows:

1. Write to any Flash or EEPROM address to start the command write sequence for the sector erase abort command. The address and data written are ignored.
2. Write the sector erase abort command, 0x47, to the FCMD register.
3. Clear the FCBEF flag in the FSTAT register by writing a 1 to FCBEF to launch the sector erase abort command.

If the sector erase abort command is launched resulting in the early termination of an active sector erase operation, the FACCERR flag will set once the operation completes as indicated by the FCCF flag being set. The FACCERR flag sets to inform the user that the Flash sector may not be fully erased and a new sector erase command must be launched before programming any location in that specific sector.

If the sector erase abort command is launched but the active sector erase operation completes normally, the FACCERR flag will not set upon completion of the operation as indicated by the FCCF flag being set. Therefore, if the FACCERR flag is not set after the sector erase abort command has completed, a sector being erased when the abort command was launched will be fully erased.

A flowchart to execute the sector erase abort operation is shown in Figure 4-4.

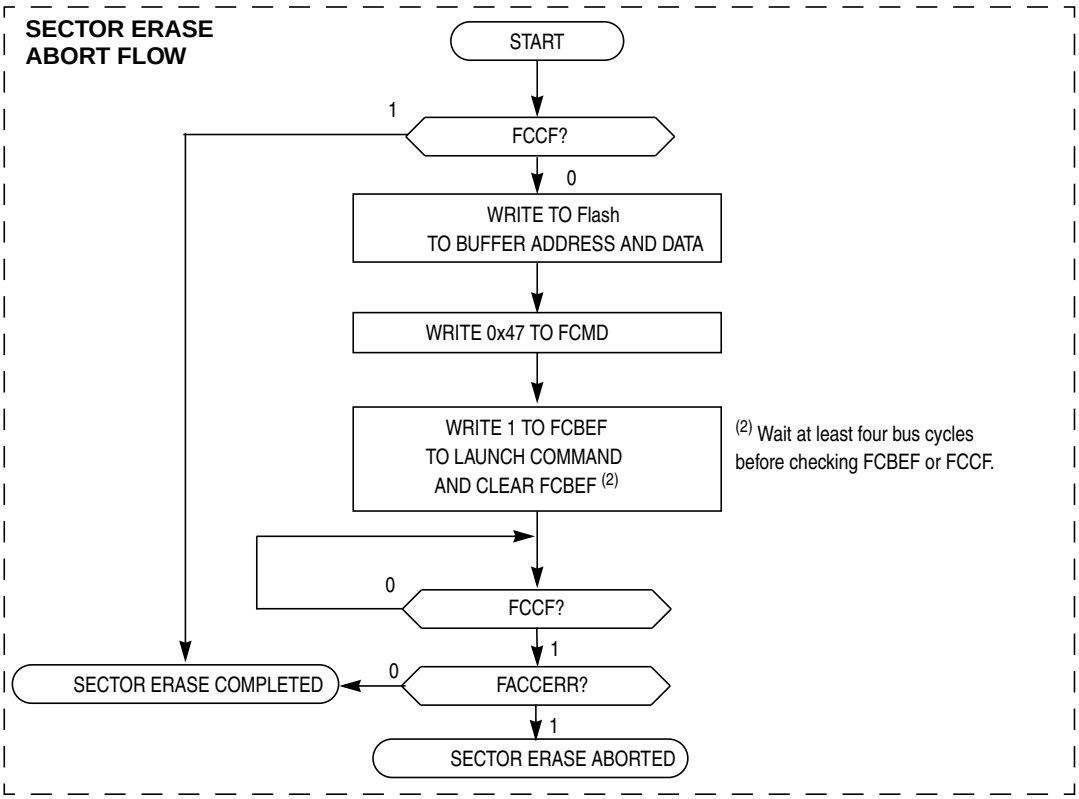


Figure 4-4. Sector Erase Abort Flowchart

Table 4-11. FPROT Register Field Descriptions

Field	Description
7:6 EPS	**EEPROM Protect Select Bits** — This 2-bit field determines the protected EEPROM locations that cannot be erased or programmed. See Table 4-12.
5:0 FPS	**Flash Protect Select Bits** — This 6-bit field determines the protected Flash locations that cannot be erased or programmed. SeeTable 4-13.

Table 4-12. EEPROM Block Protection

EPS	Address Area Protected	Memory Size Protected (bytes)	Number of Sectors Protected
0x3	N/A	0	0
0x2	0x17F0 - 0x17FF	32	4
0x1	0x17E0 - 0x17FF	64	8
0x0	0x17C0–0x17FF	128	16

Table 4-13. Flash Block Protection

FPS	Address Area Protected	Memory Size Protected (bytes)	Number of Sectors Protected
0x3F	N/A	0	0
0x3E	0xFA00–0xFFFF	1.5K	2
0x3D	0xF400–0xFFFF	3K	4
0x3C	0xEE00–0xFFFF	4.5K	6
0x3B	0xE800–0xFFFF	6K	8
...	...	...	...
0x37	0xD000–0xFFFF	12K	16
0x36	0xCA00–0xFFFF	13.5K	18
0x35	0xC400–0xFFFF	15K	20
0x34	0xBE00–0xFFFF	16.5K	22
...	...	...	...
0x2C	0x8E00–0xFFFF	28.5K	38
0x2B	0x8800–0xFFFF	30K	40
0x2A	0x8200–0xFFFF	31.5K	42
0x29	0x7C00–0xFFFF	33K	44
...	...	...	...
0x22	0x5200–0xFFFF	43.5K	58
0x21	0x4C00–0xFFFF	45K	60
0x20	0x4600–0xFFFF	46.5K	62
0x1F	0x4000–0xFFFF	48K	64
...	...	...	...

Table 4-13. Flash Block Protection (continued)

FPS	Address Area Protected	Memory Size Protected (bytes)	Number of Sectors Protected
0x1B	0x2800–0xFFFF	54K	72
0x1A	0x2200–0xFFFF	55.5K	74
0x19	0x1C00–0xFFFF	57K	76
0x18–0x00	0x0000–0xFFFF	64K	86

4.5.11.5 Flash and EEPROM Status Register (FSTAT)

	7	6	5	4	3	2	1	0
R	FCBEF	FCCF	FPVIOL	FACCERR	0	FBLANK	0	0
W								
Reset	1	1	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 4-9. Flash and EEPROM Status Register (FSTAT)

Table 4-14. FSTAT Register Field Descriptions

Field	Description
7 FCBEF	**Command Buffer Empty Flag** — The FCBEF bit is used to launch commands. It also indicates that the command buffer is empty so that a new command sequence can be executed when performing burst programming. The FCBEF bit is cleared by writing a 1 to it or when a burst program command is transferred to the array for programming. Only burst program commands can be buffered. 0 Command buffer is full (not ready for additional commands). 1 A new burst program command can be written to the command buffer.
6 FCCF	**Command Complete Flag** — FCCF is set automatically when the command buffer is empty and no command is being processed. FCCF is cleared automatically when a new command is started (by writing 1 to FCBEF to register a command). Writing to FCCF has no meaning or effect. 0 Command in progress 1 All commands complete
5 FPVIOL	**Protection Violation Flag** — FPVIOL is set automatically when a command that attempts to erase or program a location in a protected block is launched (the erroneous command is ignored). FPVIOL is cleared by writing a 1 to FPVIOL. 0 No protection violation. 1 An attempt was made to erase or program a protected location.

5.5.1 Interrupt Stack Frame

Figure 5-1 shows the contents and organization of a stack frame. Before the interrupt, the stack pointer (SP) points at the next available byte location on the stack. The current values of CPU registers are stored on the stack starting with the low-order byte of the program counter (PCL) and ending with the CCR. After stacking, the SP points at the next available location on the stack which is the address that is one less than the address where the CCR was saved. The PC value that is stacked is the address of the instruction in the main program that would have executed next if the interrupt had not occurred.

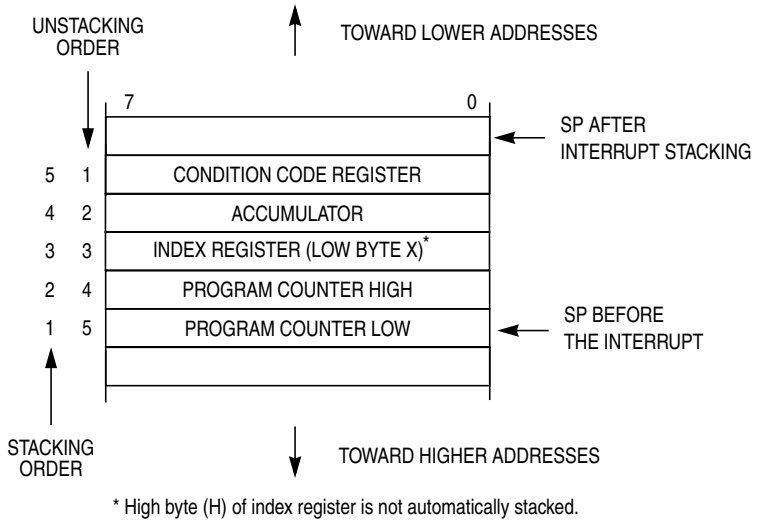


Figure 5-1. Interrupt Stack Frame

When an RTI instruction is executed, these values are recovered from the stack in reverse order. As part of the RTI sequence, the CPU fills the instruction pipeline by reading three bytes of program information, starting from the PC address recovered from the stack.

The status flag corresponding to the interrupt source must be acknowledged (cleared) before returning from the ISR. Typically, the flag is cleared at the beginning of the ISR so that if another interrupt is generated by this same source, it will be registered so it can be serviced after completion of the current ISR.

5.5.2 External Interrupt Request (IRQ) Pin

External interrupts are managed by the IRQ status and control register, IRQSC. When the IRQ function is enabled, synchronous logic monitors the pin for edge-only or edge-and-level events. When the MCU is in stop mode and system clocks are shut down, a separate asynchronous path is used so the IRQ (if enabled) can wake the MCU.

5.5.2.1 Pin Configuration Options

The IRQ pin enable (IRQPE) control bit in IRQSC must be 1 in order for the IRQ pin to act as the interrupt request (IRQ) input. As an IRQ input, the user can choose the polarity of edges or levels detected (IRQEDG), whether the pin detects edges-only or edges and levels (IRQMOD), and whether an event causes an interrupt or only sets the IRQF flag which can be polled by software.

5.8.4 System Options Register 1 (SOPT1)

This high page register is a write-once register so only the first write after reset is honored. It can be read at any time. Any subsequent attempt to write to SOPT1 (intentionally or unintentionally) is ignored to avoid accidental changes to these sensitive settings. This register should be written during the user's reset initialization program to set the desired controls even if the desired settings are the same as the reset settings.

	7	6	5	4	3	2	1	0
R	COPT		STOPE	Reserved	IICPS	0	0	0
W								
Reset:	1	1	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 5-5. System Options Register 1 (SOPT1)

Table 5-5. SOPT1 Register Field Descriptions

Field	Description
7:6 COPT[1:0]	**COP Watchdog Timeout** — These write-once bits select the timeout period of the COP. COPT along with COPCLKS in SOPT2 defines the COP timeout period. See Table 5-6.
5 STOPE	**Stop Mode Enable** — This write-once bit is used to enable stop mode. If stop mode is disabled and a user program attempts to execute a STOP instruction, an illegal opcode reset is forced. 0 Stop mode disabled. 1 Stop mode enabled.
3 IICPS	**IIC Pin Select**— This write-once bit selects the location of the SCL and SDA pins of the IIC module. 0 SCL on PTF2, SDA on PTF3. 1 SCL on PTE4, SDA on PTE5.

Table 5-6. COP Configuration Options

Control Bits		Clock Source	COP Window[1] Opens (COPW = 1)	COP Overflow Count
COPCLKS	COPT[1:0]			
N/A	0:0	N/A	N/A	COP is disabled
0	0:1	1 kHz	N/A	2^5 cycles (32 ms[2])
0	1:0	1 kHz	N/A	2^8 cycles (256 ms[1])
0	1:1	1 kHz	N/A	2^{10} cycles (1.024 s[1])
1	0:1	Bus	6144 cycles	2^{13} cycles
1	1:0	Bus	49,152 cycles	2^{16} cycles
1	1:1	Bus	196,608 cycles	2^{18} cycles

1 Windowed COP operation requires the user to clear the COP timer in the last 25% of the selected timeout period. This column displays the minimum number of clock counts required before the COP timer can be reset when in windowed COP mode (COPW = 1).

2 Values shown in milliseconds based on t_{LPO} = 1 ms. See t_{LPO} in the appendix Section A.12.1, "Control Timing," for the tolerance of this value.

6.5.3.5 Port C Drive Strength Selection Register (PTCDS)

	7	6	5	4	3	2	1	0
R W	PTCDS7	PTCDS6	PTCDS5	PTCDS4	PTCDS3	PTCDS2	PTCDS1	PTCDS0
Reset:	0	0	0	0	0	0	0	0

Figure 6-23. Drive Strength Selection for Port C Register (PTCDS)

Table 6-21. PTCDS Register Field Descriptions

Field	Description
7:0 PTCDS[7:0]	**Output Drive Strength Selection for Port C Bits** — Each of these control bits selects between low and high output drive for the associated PTC pin. For port C pins that are configured as inputs, these bits have no effect. 0 Low output drive strength selected for port C bit n. 1 High output drive strength selected for port C bit n.

8.1.1 Features

Key features of the MCG module are:

- Frequency-locked loop (FLL)
  - 0.2% resolution using internal 32-kHz reference
  - 2% deviation over voltage and temperature using internal 32-kHz reference
  - Internal or external reference can be used to control the FLL
- Phase-locked loop (PLL)
  - Voltage-controlled oscillator (VCO)
  - Modulo VCO frequency divider
  - Phase/Frequency detector
  - Integrated loop filter
  - Lock detector with interrupt capability
- Internal reference clock
  - Nine trim bits for accuracy
  - Can be selected as the clock source for the MCU
- External reference clock
  - Control for external oscillator
  - Clock monitor with reset capability
  - Can be selected as the clock source for the MCU
- Reference divider is provided
- Clock source selected can be divided down by 1, 2, 4, or 8
- BDC clock (MCGLCLK) is provided as a constant divide by 2 of the DCO output whether in an FLL or PLL mode.

multiplication factor, as selected by the VDIV bits, times the reference frequency, as selected by the RDIV bits. If BDM is enabled then the MCGLCLK is derived from the DCO (open-loop mode) divided by two. If BDM is not enabled then the FLL is disabled in a low power state.

8.4.1.6 PLL Bypassed External (PBE)

In PLL bypassed external (PBE) mode, the MCGOUT clock is derived from the external reference clock and the PLL is operational but its output clock is not used. This mode is useful to allow the PLL to acquire its target frequency while the MCGOUT clock is driven from the external reference clock.

The PLL bypassed external mode is entered when all the following conditions occur:

- CLKS bits are written to 10
- IREFS bit is written to 0
- PLLS bit is written to 1
- RDIV bits are written to divide reference clock to be within the range of 1 MHz to 2 MHz
- LP bit is written to 0

In PLL bypassed external mode, the MCGOUT clock is derived from the external reference clock. The external reference clock which is enabled can be an external crystal/resonator or it can be another external clock source. The PLL clock frequency locks to a multiplication factor, as selected by the VDIV bits, times the reference frequency, as selected by the RDIV bits. If BDM is enabled then the MCGLCLK is derived from the DCO (open-loop mode) divided by two. If BDM is not enabled then the FLL is disabled in a low power state.

8.4.1.7 Bypassed Low Power Internal (BLPI)

The bypassed low power internal (BLPI) mode is entered when all the following conditions occur:

- CLKS bits are written to 01
- IREFS bit is written to 1
- PLLS bit is written to 0
- LP bit is written to 1
- BDM mode is not active

In bypassed low power internal mode, the MCGOUT clock is derived from the internal reference clock.

The PLL and the FLL are disabled at all times in BLPI mode and the MCGLCLK will not be available for BDC communications If the BDM becomes active the mode will switch to FLL bypassed internal (FBI) mode.

8.4.1.8 Bypassed Low Power External (BLPE)

The bypassed low power external (BLPE) mode is entered when all the following conditions occur:

- CLKS bits are written to 10
- IREFS bit is written to 0
- PLLS bit is written to 0 or 1

Table 10-6. ADCCFG Register Field Descriptions (continued)

Field	Description
3:2 MODE	Conversion Mode Selection. MODE bits are used to select between 12-, 10-, or 8-bit operation. See Table 10-8.
1:0 ADICLK	Input Clock Select. ADICLK bits select the input clock source to generate the internal clock ADCK. See Table 10-9.

Table 10-7. Clock Divide Select

ADIV	Divide Ratio	Clock Rate
00	1	Input clock
01	2	Input clock ÷ 2
10	4	Input clock ÷ 4
11	8	Input clock ÷ 8

Table 10-8. Conversion Modes

MODE	Mode Description
00	8-bit conversion (N=8)
01	12-bit conversion (N=12)
10	10-bit conversion (N=10)
11	Reserved

Table 10-9. Input Clock Select

ADICLK	Selected Clock Source
00	Bus clock
01	Bus clock divided by 2
10	Alternate clock (ALTCLK)
11	Asynchronous clock (ADACK)

10.3.8 Pin Control 1 Register (APCTL1)

The pin control registers disable the I/O port control of MCU pins used as analog inputs. APCTL1 is

configured for low power operation, long sample time, continuous conversion, and automatic compare of the conversion result to a software determined compare value.

10.4.4.1 Initiating Conversions

A conversion is initiated:

- Following a write to ADCSC1 (with ADCH bits not all 1s) if software triggered operation is selected.
- Following a hardware trigger (ADHWT) event if hardware triggered operation is selected.
- Following the transfer of the result to the data registers when continuous conversion is enabled.

If continuous conversions are enabled, a new conversion is automatically initiated after the completion of the current conversion. In software triggered operation, continuous conversions begin after ADCSC1 is written and continue until aborted. In hardware triggered operation, continuous conversions begin after a hardware trigger event and continue until aborted.

10.4.4.2 Completing Conversions

A conversion is completed when the result of the conversion is transferred into the data result registers, ADCRH and ADCRL. This is indicated by the setting of COCO. An interrupt is generated if AIEN is high at the time that COCO is set.

A blocking mechanism prevents a new result from overwriting previous data in ADCRH and ADCRL if the previous data is in the process of being read while in 12-bit or 10-bit MODE (the ADCRH register has been read but the ADCRL register has not). When blocking is active, the data transfer is blocked, COCO is not set, and the new result is lost. In the case of single conversions with the compare function enabled and the compare condition false, blocking has no effect and ADC operation is terminated. In all other cases of operation, when a data transfer is blocked, another conversion is initiated regardless of the state of ADCO (single or continuous conversions enabled).

If single conversions are enabled, the blocking mechanism could result in several discarded conversions and excess power consumption. To avoid this issue, the data registers must not be read after initiating a single conversion until the conversion completes.

10.4.4.3 Aborting Conversions

Any conversion in progress is aborted when:

- A write to ADCSC1 occurs (the current conversion will be aborted and a new conversion will be initiated, if ADCH are not all 1s).
- A write to ADCSC2, ADCCFG, ADCCVH, or ADCCVL occurs. This indicates a mode of operation change has occurred and the current conversion is therefore invalid.
- The MCU is reset.
- The MCU enters stop mode with ADACK not enabled.

Chapter 11
Inter-Integrated Circuit (S08IICV2)

11.1 Introduction

The inter-integrated circuit (IIC) provides a method of communication between a number of devices. The interface is designed to operate up to 100 kbps with maximum bus loading and timing. The device is capable of operating at higher baud rates, up to a maximum of clock/20, with reduced bus loading. The maximum communication length and the number of devices that can be connected are limited by a maximum bus capacitance of 400 pF.

All MC9S08DN60 Series MCUs feature the IIC, as shown in the following block diagram.

NOTE

Drive strength must be disabled (DSE=0) for the IIC pins when using the IIC module for correct operation.

in LSBFE. Both variations of SPSCK polarity are shown, but only one of these waveforms applies for a specific transfer, depending on the value in CPOL. The SAMPLE IN waveform applies to the MOSI input of a slave or the MISO input of a master. The MOSI waveform applies to the MOSI output pin from a master and the MISO waveform applies to the MISO output from a slave. The $\overline{SS}$ OUT waveform applies to the slave select output from a master (provided MODFEN and SSOE = 1). The master $\overline{SS}$ output goes to active low at the start of the first bit time of the transfer and goes back high one-half SPSCK cycle after the end of the eighth bit time of the transfer. The $\overline{SS}$ IN waveform applies to the slave select input of a slave.

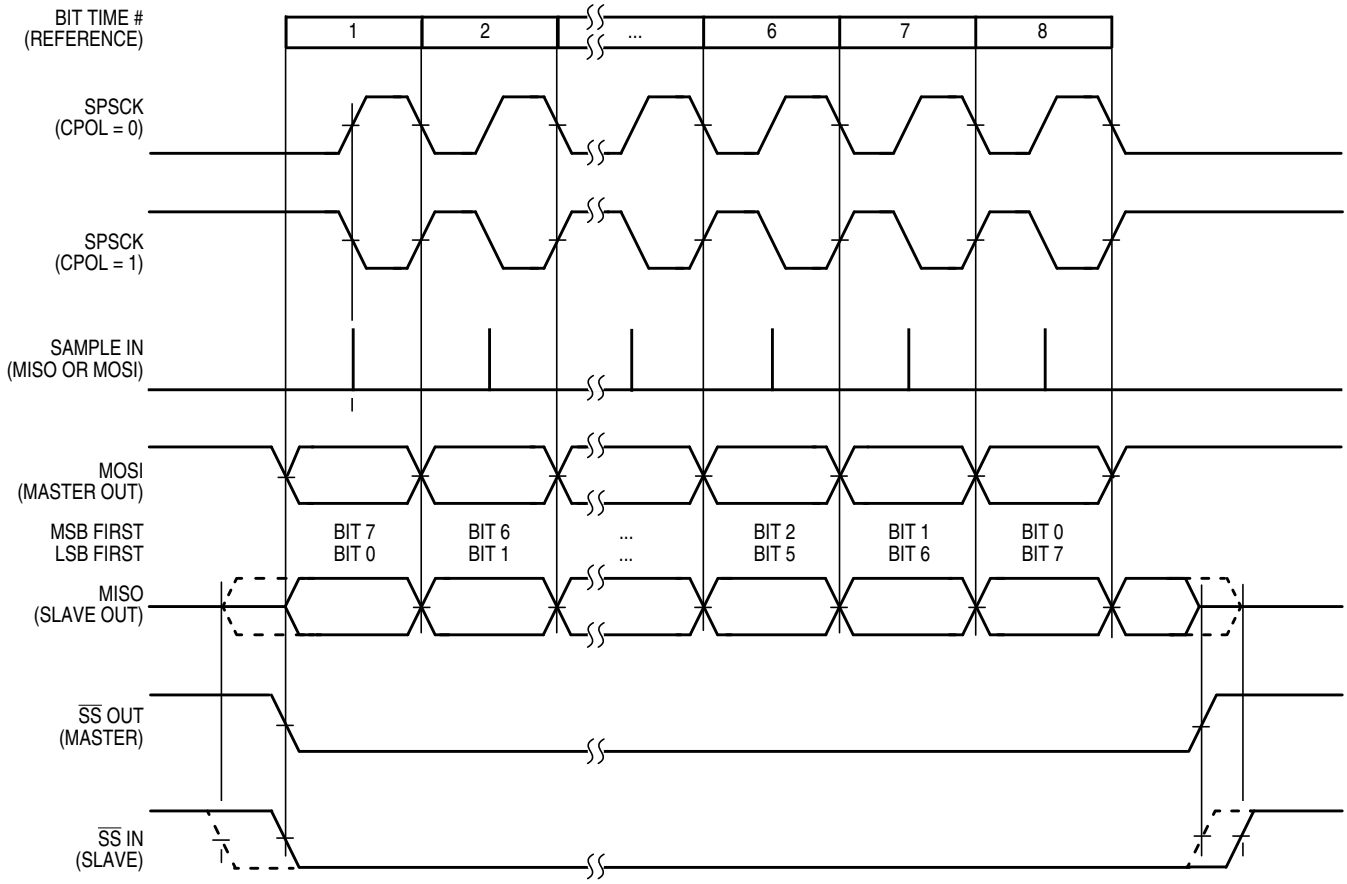


Figure 12-11. SPI Clock Formats (CPHA = 0)

When CPHA = 0, the slave begins to drive its MISO output with the first data bit value (MSB or LSB depending on LSBFE) when $\overline{SS}$ goes to active low. The first SPSCK edge causes both the master and the slave to sample the data bit values on their MISO and MOSI inputs, respectively. At the second SPSCK edge, the SPI shifter shifts one bit position which shifts in the bit value that was just sampled and shifts the second data bit value out the other end of the shifter to the MOSI and MISO outputs of the master and slave, respectively. When CPHA = 0, the slave's $\overline{SS}$ input must go to its inactive high level between transfers.

Figure 16-2 shows an external host transmitting a logic 1 or 0 to the BKGD pin of a target HCS08 MCU. The host is asynchronous to the target so there is a 0-to-1 cycle delay from the host-generated falling edge to where the target perceives the beginning of the bit time. Ten target BDC clock cycles later, the target senses the bit level on the BKGD pin. Typically, the host actively drives the pseudo-open-drain BKGD pin during host-to-target transmissions to speed up rising edges. Because the target does not drive the BKGD pin during the host-to-target transmission period, there is no need to treat the line as an open-drain signal during this period.

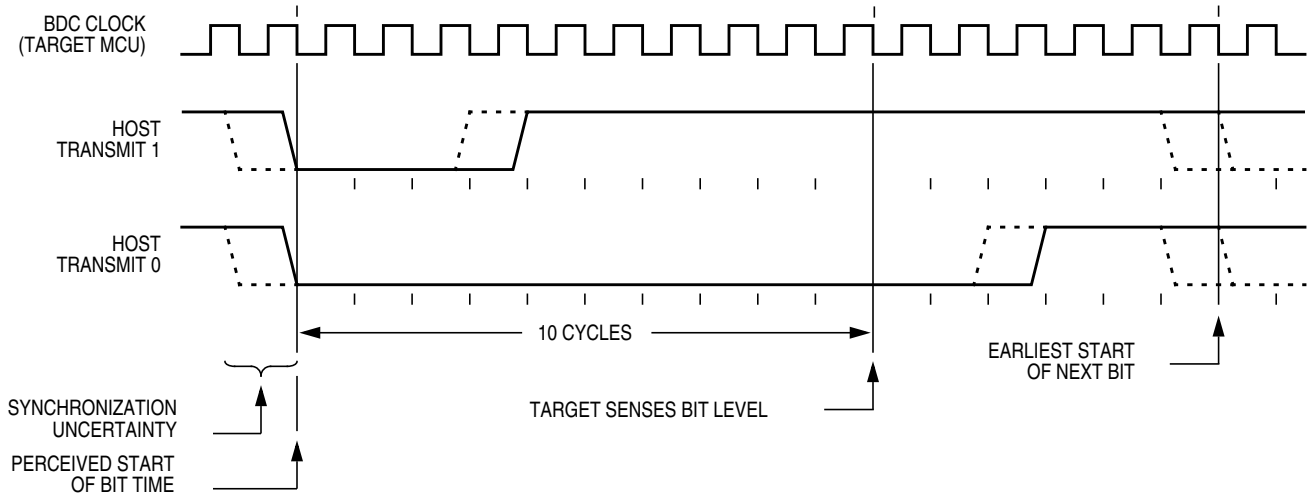


Figure 16-2. BDC Host-to-Target Serial Bit Timing

Table A-3. Thermal Characteristics

Num	C	Rating	Symbol	Value	Unit	Temp. Code
1	D	Operating temperature range (packaged)	T_A	–40 to 125 –40 to 105 –40 to 85	°C	M V C
2	T	Maximum Junction Temperature[1]	T_J	135	°C	—
3	D	Thermal resistance[2]				
		Single-layer board				
		64-pin LQFP	θ_{JA}	69	°C/W	
		48-pin LQFP	θ_{JA}	75	°C/W	
		32-pin LQFP	θ_{JA}	80	°C/W	
		Four-Layer board				
		64-pin LQFP	θ_{JA}	51	°C/W	
		48-pin LQFP	θ_{JA}	51	°C/W	
		32-pin LQFP	θ_{JA}	52	°C/W	

1 Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.

2 Junction to Ambient Natural Convection

The average chip-junction temperature (T_J) in °C can be obtained from:

$$T_J = T_A + (P_D \times \theta_{JA}) \qquad \textit{Eqn. A-1}$$

where:

T_A = Ambient temperature, °C

θ_{JA} = Package thermal resistance, junction-to-ambient, °C/W

$P_D = P_{int} + P_{I/O}$

$P_{int} = I_{DD} \times V_{DD}$, Watts — chip internal power

$P_{I/O}$ = Power dissipation on input and output pins — user determined

For most applications, $P_{I/O} << P_{int}$ and can be neglected. An approximate relationship between P_D and T_J (if $P_{I/O}$ is neglected) is:

$$P_D = K \div (T_J + 273°C) \qquad \textit{Eqn. A-2}$$

Solving equations 1 and 2 for K gives:

$$K = P_D \times (T_A + 273°C) + \theta_{JA} \times (P_D)^2 \qquad \textit{Eqn. A-3}$$

3 Monotonicity and No-Missing-Codes guaranteed in 10 bit and 8 bit modes

4 Based on input pad leakage current. Refer to pad electricals.

A.10 External Oscillator (XOSC) Characteristics

Table A-11. Oscillator Electrical Specifications (Temperature Range = –40 to 125°C Ambient)

Num	C	Rating	Symbol	Min	Typ[1]	Max	Unit
1	C	Oscillator crystal or resonator (EREFS = 1, ERCLKEN = 1					
		Low range (RANGE = 0)	f_{lo}	32	—	38.4	kHz
		High range (RANGE = 1) FEE or FBE mode [2]	f_{hi-fll}	1	—	5	MHz
		High range (RANGE = 1) PEE or PBE mode[3]	f_{hi-pll}	1	—	16	MHz
		High range (RANGE = 1, HGO = 1) BLPE mode	f_{hi-hgo}	1	—	16	MHz
		High range (RANGE = 1, HGO = 0) BLPE mode	f_{hi-lp}	1	—	8	MHz
2	—	Load capacitors	C_1 C_2	See crystal or resonator manufacturer's recommendation.			
3	—	Feedback resistor					
		Low range (32 kHz to 100 kHz)	R_F	—	10	—	MΩ
		High range (1 MHz to 16 MHz)		—	1	—	MΩ
4	—	Series resistor					
		Low range, low gain (RANGE = 0, HGO = 0)	R_S	—	0	—	kΩ
		Low range, high gain (RANGE = 0, HGO = 1)		—	100	—	
		High range, low gain (RANGE = 1, HGO = 0)		—	0	—	
		High range, high gain (RANGE = 1, HGO = 1) ≥ 8 MHz		—	0	0	
		4 MHz		—	0	10	
		1 MHz		—	0	20	
5	T	Crystal start-up time [4]					
		Low range, low gain (RANGE = 0, HGO = 0)	$t_{CSTL-LP}$	—	200	—	ms
		Low range, high gain (RANGE = 0, HGO = 1)	$t_{CSTL-HGO}$	—	400	—	
		High range, low gain (RANGE = 1, HGO = 0)[5]	$t_{CSTH-LP}$	—	5	—	
		High range, high gain (RANGE = 1, HGO = 1)[4]	$t_{CSTH-HGO}$	—	15	—	
6	T	Square wave input clock frequency (EREFS = 0, ERCLKEN = 1)					
		FEE or FBE mode [2]	f_{extal}	0.03125	—	5	MHz
		PEE or PBE mode[3]		1	—	16	
		BLPE mode		0	—	40	

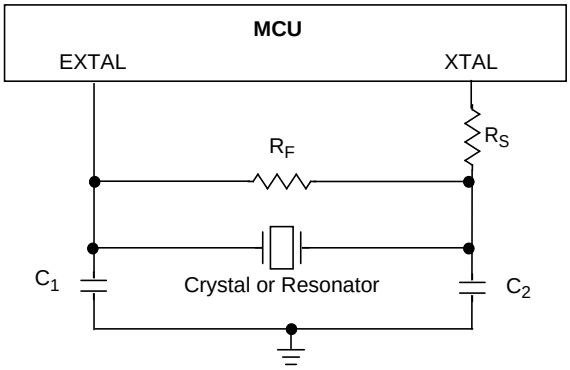
1 Typical data was characterized at 3.0 V, 25°C or is recommended value.

2 When MCG is configured for FEE or FBE mode, the input clock source must be divisible using RDIV to within the range of 31.25 kHz to 39.0625 kHz.

3 When MCG is configured for PEE or PBE mode, input clock source must be divisible using RDIV to within the range of 1 MHz to 2 MHz.

4 This parameter is characterized and not tested on each device. Proper PC board layout procedures must be followed to achieve specifications. This data will vary based upon the crystal manufacturer and board design. The crystal should be characterized by the crystal manufacturer.

5 4 MHz crystal.



A.11 MCG Specifications

Table A-12. MCG Frequency Specifications (Temperature Range = –40 to 125°C Ambient)

Num	C	Rating	Symbol	Min	Typical	Max	Unit
1	P	Internal reference frequency - factory trimmed at V_{DD} = 5 V and temperature = 25 °C	f_{int_ft}	—	31.25	—	kHz
2	P	Average internal reference frequency - untrimmed [1]	f_{int_ut}	25	32.7	41.66	kHz
3	P	Average internal reference frequency - user trimmed	f_{int_t}	31.25	—	39.0625	kHz
4	D	Internal reference startup time	t_{irefst}	—	60	100	us
5	—	DCO output frequency range - untrimmed [1] value provided for reference: f_{dco_ut} = 1024 X f_{int_ut}	f_{dco_ut}	25.6	33.48	42.66	MHz
6	P	DCO output frequency range - trimmed	f_{dco_t}	32	—	40	MHz
7	C	Resolution of trimmed DCO output frequency at fixed voltage and temperature (using FTRIM)	$\Delta f_{dco\_res\_t}$	—	± 0.1	± 0.2	$\%f_{dco}$
8	C	Resolution of trimmed DCO output frequency at fixed voltage and temperature (not using FTRIM)	$\Delta f_{dco\_res\_t}$	—	± 0.2	± 0.4	$\%f_{dco}$
9	P	Total deviation of trimmed DCO output frequency over voltage and temperature	Δf_{dco_t}	—	+ 0.5 -1.0	± 2	$\%f_{dco}$
10	C	Total deviation of trimmed DCO output frequency over fixed voltage and temperature range of 0 - 70 °C	Δf_{dco_t}	—	± 0.5	± 1	$\%f_{dco}$
11	C	FLL acquisition time [2]	$t_{fll\_acquire}$	—	—	1	ms
12	D	PLL acquisition time [3]	$t_{pll\_acquire}$	—	—	1	ms
13	C	Long term Jitter of DCO output clock (averaged over 2ms interval) [4]	C_{Jitter}	—	0.02	0.2	$\%f_{dco}$
14	D	VCO operating frequency	f_{vco}	7.0	—	55.0	MHz
15	D	PLL reference frequency range	f_{pll_ref}	1.0	—	2.0	MHz
16	T	RMS frequency variation of a single clock cycle measured 2 ms after reference edge.[5]	$f_{pll\_cycjit\_2ms}$	—	0.590[4]	—	$\%f_{pll}$
17	T	Maximum frequency variation averaged over 2 ms window.	$f_{pll\_maxjit\_2ms}$	—	0.001	—	$\%f_{pll}$

at the transition from the value set in the modulus register and the next lower count value. This corresponds to the end of a PWM period. (The 0x0000 count value corresponds to the center of a period.)

B.4.3 Channel Event Interrupt Description

The meaning of channel interrupts depends on the current mode of the channel (input capture, output compare, edge-aligned PWM, or center-aligned PWM).

When a channel is configured as an input capture channel, the ELSnB:ELSnA control bits select rising edges, falling edges, any edge, or no edge (off) as the edge that triggers an input capture event. When the selected edge is detected, the interrupt flag is set. The flag is cleared by the 2-step sequence described in Section B.4.1, “Clearing Timer Interrupt Flags.”

When a channel is configured as an output compare channel, the interrupt flag is set each time the main timer counter matches the 16-bit value in the channel value register. The flag is cleared by the 2-step sequence described in Section B.4.1, “Clearing Timer Interrupt Flags.”

B.4.4 PWM End-of-Duty-Cycle Events

For channels that are configured for PWM operation, there are two possibilities:

- When the channel is configured for edge-aligned PWM, the channel flag is set when the timer counter matches the channel value register that marks the end of the active duty cycle period.
- When the channel is configured for center-aligned PWM, the timer count matches the channel value register twice during each PWM cycle. In this CPWM case, the channel flag is set at the start and at the end of the active duty cycle, which are the times when the timer counter matches the channel value register.

The flag is cleared by the 2-step sequence described in Section B.4.1, “Clearing Timer Interrupt Flags.”