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Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	684
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	104
Number of Gates	4000
Voltage - Supply	4.5V ~ 5.5V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	144-BQFP
Supplier Device Package	144-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a1240a-1pqg144i

1 – ACT 2 Family Overview

General Description

The ACT 2 family represents Actel's second generation of field programmable gate arrays (FPGAs). The ACT 2 family presents a two-module architecture, consisting of C-modules and S-modules. These modules are optimized for both combinatorial and sequential designs. Based on Actel's patented channeled array architecture, the ACT 2 family provides significant enhancements to gate density and performance while maintaining downward compatibility with the ACT 1 design environment and upward compatibility with the ACT 3 design environment. The devices are implemented in silicon gate, 1.0- μm , two-level metal CMOS, and employ Actel's PLICE® antifuse technology. This revolutionary architecture offers gate array design flexibility, high performance, and fast time-to-production with user programming. The ACT 2 family is supported by the Designer and Designer Advantage Systems, which offers automatic pin assignment, validation of electrical and design rules, automatic placement and routing, timing analysis, user programming, and diagnostic probe capabilities. The systems are supported on the following platforms: 386/486™ PC, Sun™, and HP™ workstations. The systems provide CAE interfaces to the following design environments: Cadence, Viewlogic®, Mentor Graphics®, and OrCAD™.

2 – Detailed Specifications

Operating Conditions

Table 2-1 • Absolute Maximum Ratings¹

Symbol	Parameter	Limits	Units
VCC	DC supply voltage	–0.5 to +7.0	V
VI	Input voltage	–0.5 to VCC + 0.5	V
VO	Output voltage	–0.5 to VCC + 0.5	V
IIO	I/O source sink current ²	±20	mA
T _{STG}	Storage temperature	–65 to +150	°C

Notes:

1. Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum rated conditions for extended periods may affect device reliability. Device should not be operated outside the recommended operating conditions.
2. Device inputs are normally high impedance and draw extremely low current. However, when input voltage is greater than VCC + 0.5 V for less than GND –0.5 V, the internal protection diodes will be forward biased and can draw excessive current.

Table 2-2 • Recommended Operating Conditions

Parameter	Commercial	Industrial	Military	Units
Temperature range*	0 to +70	–40 to +85	–55 to +125	°C
Power supply tolerance	±5	±10	±10	%VCC

Note: *Ambient temperature (T_A) is used for commercial and industrial; case temperature (T_C) is used for military.

Package Thermal Characteristics

The device junction to case thermal characteristic is θ_{jc} , and the junction to ambient air characteristic is θ_{ja} . The thermal characteristics for θ_{ja} are shown with two different air flow rates.

Maximum junction temperature is 150°C.

A sample calculation of the absolute maximum power dissipation allowed for a PQ160 package at commercial temperature and still air is as follows:

$$\frac{\text{Max. junction temp. (°C)} - \text{Max. ambient temp. (°C)}}{\theta_{ja} \text{ °C/W}} = \frac{150^{\circ}\text{C} - 70^{\circ}\text{C}}{33^{\circ}\text{C/W}} = 2.4 \text{ W}$$

EQ 1

Table 2-4 • Package Thermal Characteristics

Package Type*	Pin Count	θ_{jc}	θ_{ja} Still Air	θ_{ja} 300 ft./min.	Units
Ceramic Pin Grid Array	100	5	35	17	°C/W
	132	5	30	15	°C/W
	176	8	23	12	°C/W
Ceramic Quad Flatpack	172	8	25	15	°C/W
Plastic Quad Flatpack ¹	100	13	48	40	°C/W
	144	15	40	32	°C/W
	160	15	38	30	°C/W
Plastic Leaded Chip Carrier	84	12	37	28	°C/W
Very Thin Quad Flatpack	100	12	43	35	°C/W
Thin Quad Flatpack	176	15	32	25	°C/W

Notes: (Maximum Power in Still Air)

1. Maximum power dissipation values for PQFP packages are 1.9 W (PQ100), 2.3 W (PQ144), and 2.4 W (PQ160).
2. Maximum power dissipation for PLCC packages is 2.7 W.
3. Maximum power dissipation for VQFP packages is 2.3 W.
4. Maximum power dissipation for TQFP packages is 3.1 W.

Power Dissipation

$$P = [\text{ICC standby} + \text{ICC active}] * V_{CC} + \text{IOL} * \text{VOL} * N + \text{IOH} * (V_{CC} - \text{VOH}) * M$$

EQ 2

where:

ICC standby is the current flowing when no inputs or outputs are changing

ICC active is the current flowing due to CMOS switching.

IOL and IOH are TTL sink/source currents.

VOL and VOH are TTL level output voltages.

N is the number of outputs driving TTL loads to VOL.

M is the number of outputs driving TTL loads to VOH.

An accurate determination of N and M is problematical because their values depend on the family type, design details, and on the system I/O. The power can be divided into two components: static and active.

Static Power Component

Microsemi FPGAs have small static power components that result in lower power dissipation than PALs or PLDs. By integrating multiple PALs/PLDs into one FPGA, an even greater reduction in board-level power dissipation can be achieved.

The power due to standby current is typically a small component of the overall power. Standby power is calculated in Table 2-5 for commercial, worst case conditions.

Table 2-5 • Standby Power Calculation

ICC	VCC	Power
2 mA	5.25 V	10.5 mW

The static power dissipated by TTL loads depends on the number of outputs driving high or low and the DC load current. Again, this value is typically small. For instance, a 32-bit bus sinking 4 mA at 0.33 V will generate 42 mW with all outputs driving low, and 140 mW with all outputs driving high. The actual dissipation will average somewhere between as I/Os switch states with time.

Active Power Component

Power dissipation in CMOS devices is usually dominated by the active (dynamic) power dissipation. This component is frequency dependent, a function of the logic and the external I/O. Active power dissipation results from charging internal chip capacitances of the interconnect, unprogrammed antifuses, module inputs, and module outputs, plus external capacitance due to PC board traces and load device inputs.

An additional component of the active power dissipation is the totem-pole current in CMOS transistor pairs. The net effect can be associated with an equivalent capacitance that can be combined with frequency and voltage to represent active power dissipation.

Equivalent Capacitance

The power dissipated by a CMOS circuit can be expressed by EQ 3.

$$\text{Power } (\mu\text{W}) = C_{\text{EQ}} * V_{\text{CC}}^2 * F$$

EQ 3

Where:

C_{EQ} is the equivalent capacitance expressed in pF.

VCC is the power supply in volts.

F is the switching frequency in MHz.

Equivalent capacitance is calculated by measuring ICC active at a specified frequency and voltage for each circuit component of interest. Measurements have been made over a range of frequencies at a fixed value of VCC. Equivalent capacitance is frequency independent so that the results may be used over a wide range of operating conditions. Equivalent capacitance values are shown in Table 2-6.

Table 2-6 • CEQ Values for Microsemi FPGAs

Item	CEQ Value
Modules (C_{EQM})	5.8
Input Buffers (C_{EQI})	12.9
Output Buffers (C_{EQO})	23.8
Routed Array Clock Buffer Loads (C_{EQCR})	3.9

Determining Average Switching Frequency

To determine the switching frequency for a design, you must have a detailed understanding of the data input values to the circuit. The following guidelines are meant to represent worst-case scenarios so that they can be generally used to predict the upper limits of power dissipation. These guidelines are given in Table 2-8.

Table 2-8 • Guidelines for Predicting Power Dissipation

Data	Value
Logic Modules (m)	80% of modules
Inputs switching (n)	# inputs/4
Outputs switching (p)	# output/4
First routed array clock loads (q1)	40% of sequential modules
Second routed array clock loads (q2)	40% of sequential modules
Load capacitance (C_L)	35 pF
Average logic module switching rate (f_m)	F/10
Average input switching rate (f_n)	F/5
Average output switching rate (f_p)	F/10
Average first routed array clock rate (f_{q1})	F
Average second routed array clock rate (f_{q2})	F/2

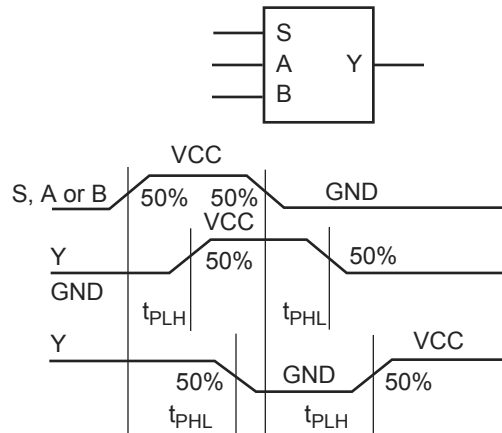
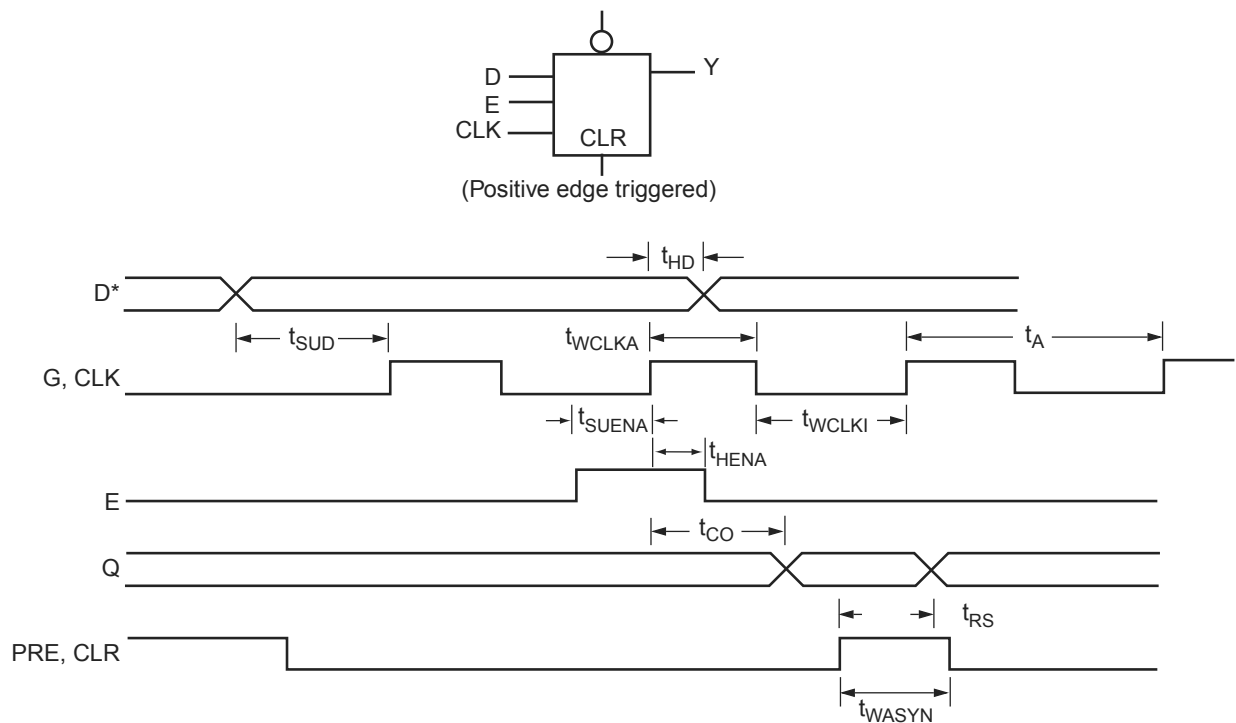


Figure 2-5 • Module Delays

Sequential Module Timing Characteristics



Note: D represents all data functions involving A, B, and S for multiplexed flip-flops.

Figure 2-6 • Flip-Flops and Latches

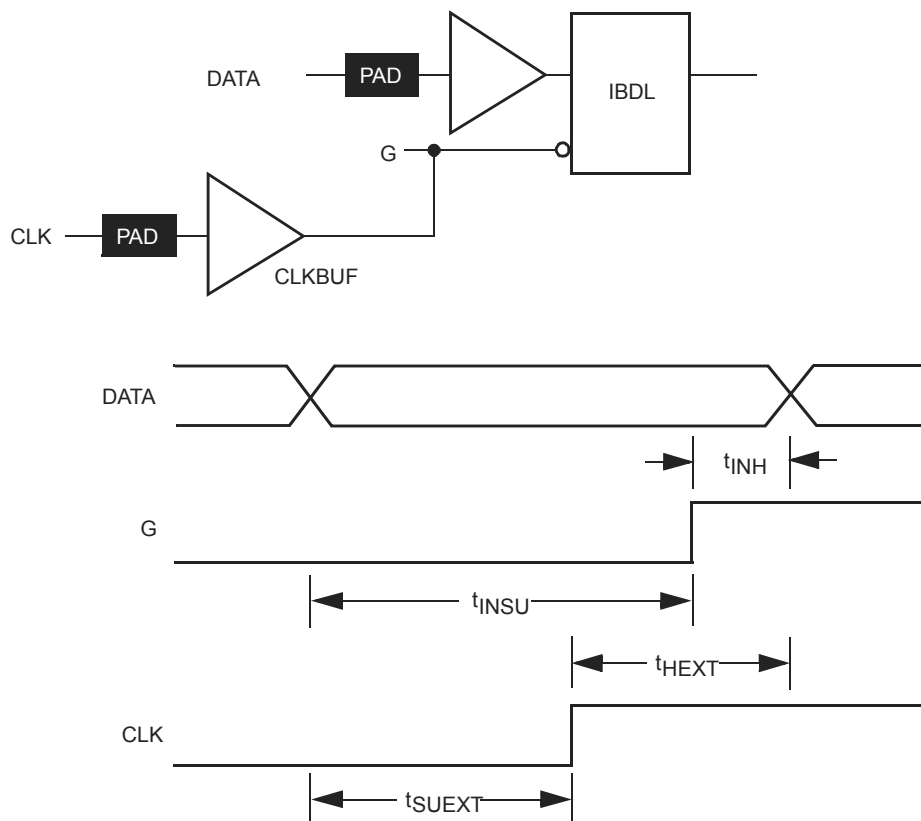


Figure 2-7 • Input Buffer Latches

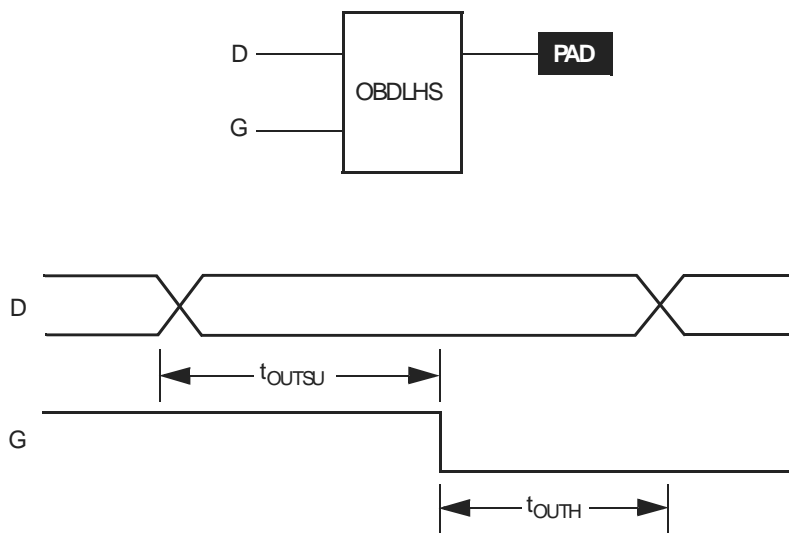


Figure 2-8 • Output Buffer Latches

Timing Derating Factor (Temperature and Voltage)

Table 2-9 • Timing Derating Factor (Temperature and Voltage)

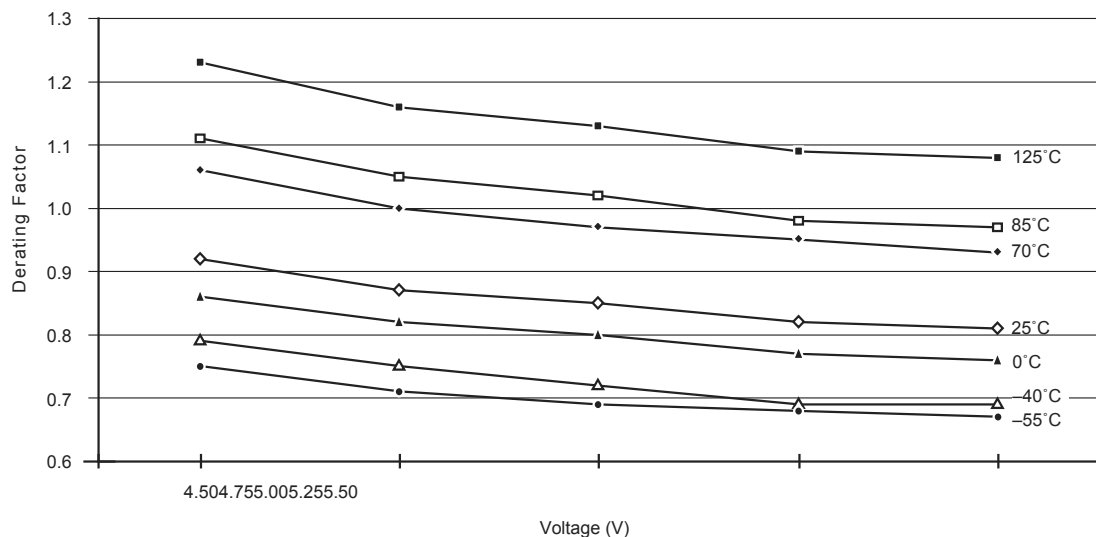
(Commercial Minimum/Maximum Specification) x	Industrial		Military	
	Min.	Max.	Min.	Max.
	0.69	1.11	0.67	1.23

Table 2-10 • Timing Derating Factor for Designs at Typical Temperature ($T_J = 25^\circ\text{C}$) and Voltage (5.0 V)

(Commercial Maximum Specification) x	0.85
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**Table 2-11 • Temperature and Voltage Derating Factors
(normalized to Worst-Case Commercial, $T_J = 4.75\text{ V}$, 70°C)**

	-55	-40	0	25	70	85	125
4.50	0.75	0.79	0.86	0.92	1.06	1.11	1.23
4.75	0.71	0.75	0.82	0.87	1.00	1.05	1.13
5.00	0.69	0.72	0.80	0.85	0.97	1.02	1.13
5.25	0.68	0.69	0.77	0.82	0.95	0.98	1.09
5.50	0.67	0.69	0.76	0.81	0.93	0.97	1.08



Note: This derating factor applies to all routing and propagation delays.

**Figure 2-9 • Junction Temperature and Voltage Derating Curves
(normalized to Worst-Case Commercial, $T_J = 4.75\text{ V}$, 70°C)**

A1240A Timing Characteristics

Table 2-15 • A1240A Worst-Case Commercial Conditions, VCC = 4.75 V, T_J = 70°C

Logic Module Propagation Delays ¹		–2 Speed ³		–1 Speed		Std. Speed		Units
Parameter/Description		Min.	Max.	Min.	Max.	Min.	Max.	
t _{PD1}	Single Module		3.8		4.3		5.0	ns
t _{CO}	Sequential Clock to Q		3.8		4.3		5.0	ns
t _{GO}	Latch G to Q		3.8		4.3		5.0	ns
t _{RS}	Flip-Flop (Latch) Reset to Q		3.8		4.3		5.0	ns
Predicted Routing Delays²								
t _{RD1}	FO = 1 Routing Delay		1.4		1.5		1.8	ns
t _{RD2}	FO = 2 Routing Delay		1.7		2.0		2.3	ns
t _{RD3}	FO = 3 Routing Delay		2.3		2.6		3.0	ns
t _{RD4}	FO = 4 Routing Delay		3.1		3.5		4.1	ns
t _{RD8}	FO = 8 Routing Delay		4.7		5.4		6.3	ns
Sequential Timing Characteristics^{3,4}								
t _{SUD}	Flip-Flop (Latch) Data Input Setup	0.4		0.4		0.5		ns
t _{HD}	Flip-Flop (Latch) Data Input Hold	0.0		0.0		0.0		ns
t _{SUENA}	Flip-Flop (Latch) Enable Setup	0.8		0.9		1.0		ns
t _{HENA}	Flip-Flop (Latch) Enable Hold	0.0		0.0		0.0		ns
t _{WCLKA}	Flip-Flop (Latch) Clock Active Pulse Width	4.5		6.0		6.5		ns
t _{WASYN}	Flip-Flop (Latch) Clock Asynchronous Pulse Width	4.5		6.0		6.5		ns
t _A	Flip-Flop Clock Input Period	9.8		12.0		15.0		ns
t _{INH}	Input Buffer Latch Hold	0.0		0.0		0.0		ns
t _{INSU}	Input Buffer Latch Setup	0.4		0.4		0.5		ns
t _{OUTH}	Output Buffer Latch Hold	0.0		0.0		0.0		ns
t _{OUTSU}	Output Buffer Latch Setup	0.4		0.4		0.5		ns
f _{MAX}	Flip-Flop (Latch) Clock Frequency		100.0		80.0		66.0	MHz

Notes:

1. For dual-module macros, use t_{PD1} + t_{RD1} + t_{PDn}, t_{CO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}—whichever is appropriate.
2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.
3. Data applies to macros based on the S-module. Timing parameters for sequential macros constructed from C-modules can be obtained from the DirectTime Analyzer utility.
4. Setup and hold timing parameters for the Input Buffer Latch are defined with respect to the PAD and the D input. External setup/hold timing parameters must account for delay from an external PAD signal to the G inputs. Delay from an external PAD signal to the G input subtracts (adds) to the internal setup (hold) time.

Table 2-20 • A1280A Worst-Case Commercial Conditions, VCC = 4.75 V, T_J = 70°C

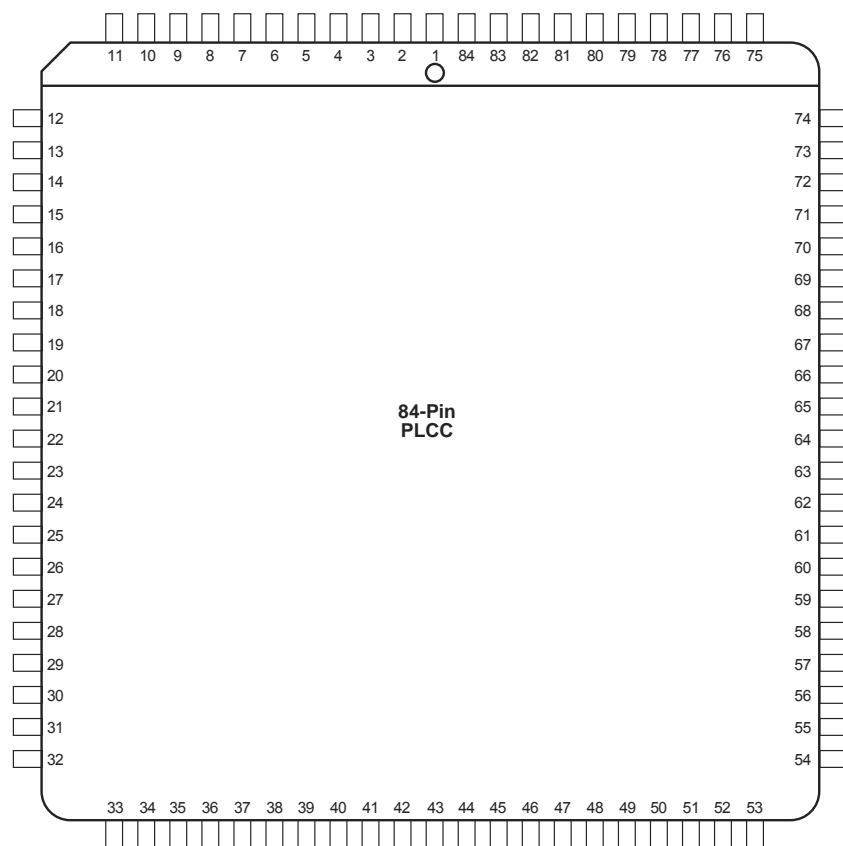
TTL Output Module Timing ¹		–2 Speed		–1 Speed		Std. Speed		Units
Parameter/Description		Min.	Max.	Min.	Max.	Min.	Max.	
t _{DLH}	Data to Pad High		8.1		9.0		10.6	ns
t _{DHL}	Data to Pad Low		10.2		11.4		13.4	ns
t _{ENZH}	Enable Pad Z to High		9.0		10.0		11.8	ns
t _{ENZL}	Enable Pad Z to Low		11.8		13.2		15.5	ns
t _{ENHZ}	Enable Pad High to Z		7.1		8.0		9.4	ns
t _{ENLZ}	Enable Pad Low to Z		8.4		9.5		11.1	ns
t _{GLH}	G to Pad High		9.0		10.2		11.9	ns
t _{GHL}	G to Pad Low		11.3		12.7		14.9	ns
d _{TLH}	Delta Low to High		0.07		0.08		0.09	ns/pF
d _{THL}	Delta High to Low		0.12		0.13		0.16	ns/pF
CMOS Output Module Timing ¹								
t _{DLH}	Data to Pad High		10.3		11.5		13.5	ns
t _{DHL}	Data to Pad Low		8.5		9.6		11.2	ns
t _{ENZH}	Enable Pad Z to High		9.0		10.0		11.8	ns
t _{ENZL}	Enable Pad Z to Low		11.8		13.2		15.5	ns
t _{ENHZ}	Enable Pad High to Z		7.1		8.0		9.4	ns
t _{ENLZ}	Enable Pad Low to Z		8.4		9.5		11.1	ns
t _{GLH}	G to Pad High		9.0		10.2		11.9	ns
t _{GHL}	G to Pad Low		11.3		12.7		14.9	ns
d _{TLH}	Delta Low to High		0.12		0.13		0.16	ns/pF
d _{THL}	Delta High to Low		0.09		0.10		0.12	ns/pF

Notes:

1. Delays based on 50 pF loading.
2. SSO information can be found at www.microsemi.com/soc/techdocs/appnotes/board_consideration.aspx.

3 – Package Pin Assignments

PL84



Note

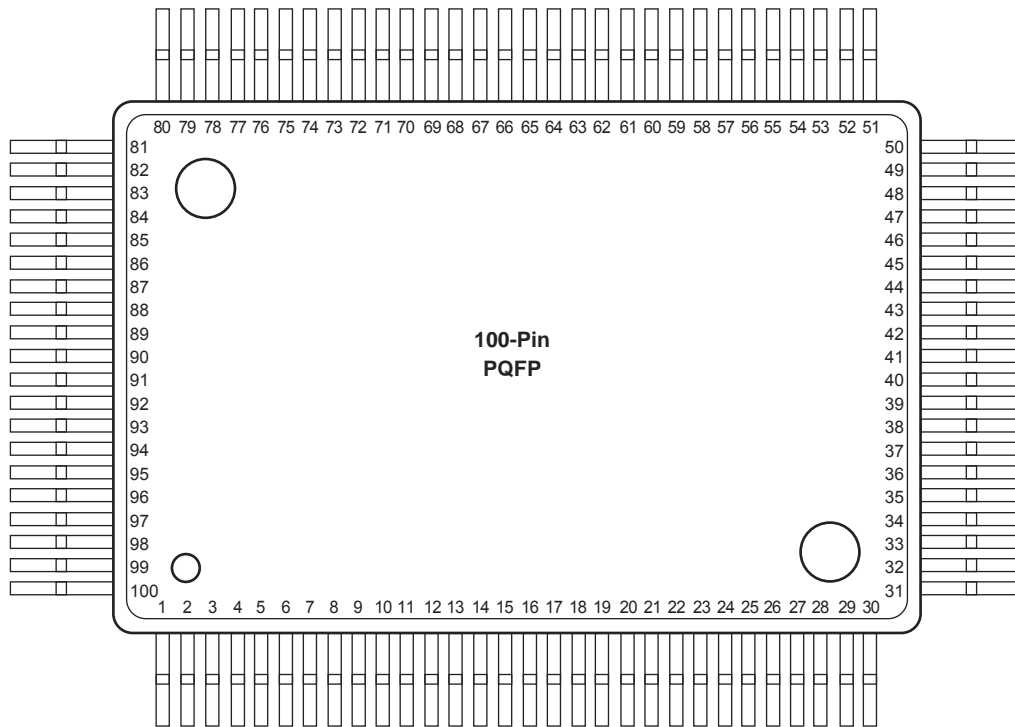
For Package Manufacturing and Environmental information, visit the Resource Center at <http://www.microsemi.com/soc/products/solutions/package/docs.aspx>.

PL84			
Pin Number	A1225A Function	A1240A Function	A1280A Function
2	CLKB, I/O	CLKB, I/O	CLKB, I/O
4	PRB, I/O	PRB, I/O	PRB, I/O
6	GND	GND	GND
10	DCLK, I/O	DCLK, I/O	DCLK, I/O
12	MODE	MODE	MODE
22	VCC	VCC	VCC
23	VCC	VCC	VCC
28	GND	GND	GND
43	VCC	VCC	VCC
49	GND	GND	GND
52	SDO	SDO	SDO
63	GND	GND	GND
64	VCC	VCC	VCC
65	VCC	VCC	VCC
70	GND	GND	GND
76	SDI, I/O	SDI, I/O	SDI, I/O
81	PRA, I/O	PRA, I/O	PRA, I/O
83	CLKA, I/O	CLKA, I/O	CLKA, I/O
84	VCC	VCC	VCC

Notes:

1. All unlisted pin numbers are user I/Os.
2. MODE pin should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.

PQ100



Note

For Package Manufacturing and Environmental information, visit the Resource Center at <http://www.microsemi.com/soc/products/solutions/package/docs.aspx>

PQ160	
Pin Number	A1280A Function
2	DCLK, I/O
6	VCC
11	GND
16	PRB, I/O
18	CLKB, I/O
20	VCC
21	CLKA, I/O
23	PRA, I/O
30	GND
35	VCC
38	SDI, I/O
40	GND
44	GND
49	GND
54	VCC
57	VCC
58	VCC
59	GND
60	VCC
61	GND
64	GND

PQ160	
Pin Number	A1280A Function
69	GND
80	GND
82	SDO
86	VCC
89	GN
98	GND
99	GND
109	GND
114	VCC
120	GND
125	GND
130	GND
135	VCC
138	VCC
139	VCC
140	GND
145	GND
150	VCC
155	GND
159	MODE
160	GND

Notes:

1. All unlisted pin numbers are user I/Os.
2. MODE pin should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.

VQ100	
Pin Number	A1225A Function
2	MODE
7	GND
14	VCC
15	VCC
20	GND
32	GND
38	VCC
44	GND
50	SDO
55	GND
62	GND
63	VCC

VQ100	
Pin Number	A1225A Function
64	VCC
65	VCC
70	GND
77	SDI, I/O
82	GND
85	PRA, I/O
87	CLKA, I/O
88	VCC
90	CLKB, I/O
92	PRB, I/O
94	GND
100	DCLK, I/O

Notes:

1. All unlisted pin numbers are user I/Os.
2. MODE pin should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.

TQ176		
Pin Number	A1240A Function	A1280A Function
155	VCC	VCC
156	GND	GND
158	CLKB, I/O	CLKB, I/O
160	PRB, I/O	PRB, I/O
161	NC	I/O
165	NC	NC
166	NC	I/O
168	NC	I/O
170	NC	VCC
173	NC	I/O
175	DCLK, I/O	DCLK, I/O

Notes:

1. NC denotes no connection.
2. All unlisted pin numbers are user I/Os.
3. MODE pin should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.

PG100		PG100	
Pin Number	A1225A Function	Pin Number	A1225A Function
A4	PRB, I/O	E11	VCC
A7	PRA, I/O	F3	VCC
B6	VCC	F9	VCC
C2	MODE	F10	VCC
C3	DCLK, I/O	F11	GND
C5	GND	G1	VCC
C6	CLKA, I/O	G3	GND
C7	GND	G9	GND
C8	SDI, I/O	J5	GND
D6	CLKB, I/O	J7	GND
D10	GND	J9	SDO
E3	GND	K6	VCC

Notes:

1. All unlisted pin numbers are user I/Os.
2. MODE pin should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.

4 – Datasheet Information

List of Changes

The following table lists critical changes that were made in each version of the datasheet.

Revision	Changes	Page
Revision 8 (January 2012)	The ACT 2 datasheet was formatted newly in the style used for current datasheets. The same information is present (other than noted in the list of changes for this revision) but divided into chapters.	N/A
	Package names used in Table 1 • ACT 2 Product Family Profile and throughout the document were revised to match standards given in <i>Package Mechanical Drawings</i> (SAR 27395).	I
	The description for SDO pins had earlier been removed from the datasheet and has now been included again, in the "Pin Descriptions" section (SAR 35819).	2-21
	SDO pin numbers had earlier been removed from package pin assignment tables in the datasheet, and have now been restored to the pin tables (SAR 35819).	3-2
Revision 7 (June 2006)	The "Ordering Information" section was revised to include RoHS information.	II
Revision 6 (December 2000)	In the "PG176" package, pin A3 was incorrectly assigned as CLKA, I/O. A3 is a user I/O. Pin A9 is CLKA, I/O.	3-21



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