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Understanding <u>Embedded - FPGAs (Field</u> <u>Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Details	
Product Status	Obsolete
Number of LABs/CLBs	684
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	104
Number of Gates	4000
Voltage - Supply	4.5V ~ 5.5V
Mounting Type	Through Hole
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	132-BCPGA
Supplier Device Package	132-CPGA (34.54x34.54)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a1240a-pg132c

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

1 – ACT 2 Family Overview

General Description

The ACT 2 family represents Actel's second generation of field programmable gate arrays (FPGAs). The ACT 2 family presents a two-module architecture, consisting of C-modules and S-modules. These modules are optimized for both combinatorial and sequential designs. Based on Actel's patented channeled array architecture, the ACT 2 family provides significant enhancements to gate density and performance while maintaining downward compatibility with the ACT 1 design environment and upward compatibility with the ACT 3 design environment. The devices are implemented in silicon gate, 1.0- μ m, two-level metal CMOS, and employ Actel's PLICE® antifuse technology. This revolutionary architecture offers gate array design flexibility, high performance, and fast time-to-production with user programming. The ACT 2 family is supported by the Designer and Designer Advantage Systems, which offers automatic pin assignment, validation of electrical and design rules, automatic placement and routing, timing analysis, user programming, and diagnostic probe capabilities. The systems are supported on the following platforms: 386/486TM PC, SunTM, and HPTM workstations. The systems provide CAE interfaces to the following design environments: Cadence, Viewlogic[®], Mentor Graphics[®], and OrCADTM.

2 – Detailed Specifications

Operating Conditions

Table 2-1 • Absolute Maximum Ratings¹

Symbol	Parameter	Limits	Units
VCC	DC supply voltage	–0.5 to +7.0	V
VI	Input voltage	-0.5 to VCC + 0.5	V
VO	Output voltage	-0.5 to VCC + 0.5	V
IIO	I/O source sink current ²	±20	mA
T _{STG}	Storage temperature	-65 to +150	°C

Notes:

1. Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum rated conditions for extended periods may affect device reliability. Device should not be operated outside the recommended operating conditions.

2. Device inputs are normally high impedance and draw extremely low current. However, when input voltage is greater than VCC + 0.5 V for less than GND –0.5 V, the internal protection diodes will be forward biased and can draw excessive current.

Table 2-2 • Recommended Operating Conditions

Parameter	Commercial	Industrial	Military	Units
Temperature range*	0 to +70	-40 to +85	–55 to +125	°C
Power supply tolerance	±5	±10	±10	%VCC

Note: *Ambient temperature (T_A) is used for commercial and industrial; case temperature (T_C) is used for military.



Detailed Specifications

Table 2-3 • Electrical Specifications

	Parameter	Con	nmercial	In	dustrial	N		
Symbol		Min.	Max.	Min.	Max.	Min.	Max.	Units
VOH ¹	$(IOH = -10 \text{ mA})^2$	2.4	-	_	_	_	-	V
	(IOH = –6 mA)	3.84	-	_	_	_	-	V
	(IOH = -4 mA)	-	-	3.7	_	3.7	-	V
VOL ¹	(IOL = 10 mA) ²	-	0.5	_	-	_	-	V
	(IOL = 6 mA)	-	0.33	_	0.40	_	0.40	V
VIL		-0.3	0.8	-0.3	0.8	-0.3	0.8	V
VIH		2.0	VCC + 0.3	2.0	VCC + 0.3	2.0	VCC + 0.3	V
Input Tran	sition Time t _R , t _F ²	-	500	_	500	-	500	ns
C _{IO} I/O caj	pacitance ^{2,3}	-	10	_	10	-	10	pF
Standby Current, ICC ⁴ (typical = 1 mA)		-	2	_	10	_	20	mA
Leakage Current ⁵		-10	+10	-10	+10	-10	+10	μA
ICC(D)	Dynamic VCC supply curren	t. See the	Power Dissip	ation see	ction.		1	1

Notes:

1. Only one output tested at a time. VCC = minimum.

2. Not tested, for information only.

3. Includes worst-case PG176 package capacitance. VOUT = 0 V, f = 1 MHz

4. All outputs unloaded. All inputs = VCC or GND, typical ICC = 1 mA. ICC limit includes IPP and ISV during normal operations.

5. VOUT, VIN = VCC or GND.



Static Power Component

Microsemi FPGAs have small static power components that result in lower power dissipation than PALs or PLDs. By integrating multiple PALs/PLDs into one FPGA, an even greater reduction in board-level power dissipation can be achieved.

The power due to standby current is typically a small component of the overall power. Standby power is calculated in Table 2-5 for commercial, worst case conditions.

Table 2-5 • Standby Power Calculation

ICC	VCC	Power
2 mA	5.25 V	10.5 mW

The static power dissipated by TTL loads depends on the number of outputs driving high or low and the DC load current. Again, this value is typically small. For instance, a 32-bit bus sinking 4 mA at 0.33 V will generate 42 mW with all outputs driving low, and 140 mW with all outputs driving high. The actual dissipation will average somewhere between as I/Os switch states with time.

Active Power Component

Power dissipation in CMOS devices is usually dominated by the active (dynamic) power dissipation. This component is frequency dependent, a function of the logic and the external I/O. Active power dissipation results from charging internal chip capacitances of the interconnect, unprogrammed antifuses, module inputs, and module outputs, plus external capacitance due to PC board traces and load device inputs.

An additional component of the active power dissipation is the totem-pole current in CMOS transistor pairs. The net effect can be associated with an equivalent capacitance that can be combined with frequency and voltage to represent active power dissipation.

EQ 3

Equivalent Capacitance

The power dissipated by a CMOS circuit can be expressed by EQ 3.

Power (μ W) = C_{EQ} * VCC² * F

Where:

C_{EQ} is the equivalent capacitance expressed in pF.

VCC is the power supply in volts.

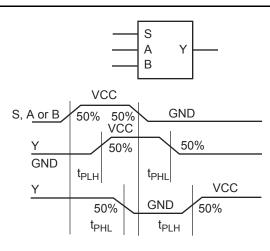
F is the switching frequency in MHz.

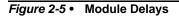
Equivalent capacitance is calculated by measuring ICC active at a specified frequency and voltage for each circuit component of interest. Measurements have been made over a range of frequencies at a fixed value of VCC. Equivalent capacitance is frequency independent so that the results may be used over a wide range of operating conditions. Equivalent capacitance values are shown in Table 2-6.

Table 2-6 • CEQ Values for Microsemi FPGAs

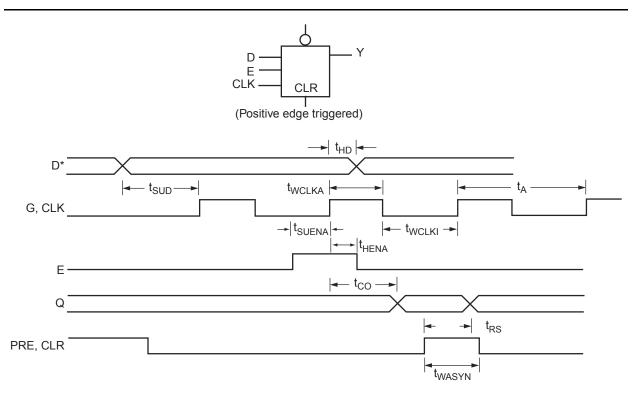
Item	CEQ Value
Modules (C _{EQM})	5.8
Input Buffers (C _{EQI})	12.9
Output Buffers (C _{EQO})	23.8
Routed Array Clock Buffer Loads (C _{EQCR})	3.9







Sequential Module Timing Characteristics



Note: D represents all data functions involving A, B, and S for multiplexed flip-flops.

Figure 2-6 • Flip-Flops and Latches

A1225A Timing Characteristics (continued)

Table 2-13 • A1225A Worst-Case Commercial Conditions, VCC = 4.75 V, T_J = 70°C

I/O Module Input Propagation Delays			-2 S	peed	–1 S	peed	Std. Speed		Units
Parameter/Description		Min.	Max.	Min.	Max.	Min.	Max.		
t _{INYH}	Pad to Y High			2.9		3.3		3.8	ns
t _{INYL}	Pad to Y Low			2.6		3.0		3.5	ns
t _{INGH}	G to Y High			5.0		5.7		6.6	ns
t _{INGL}	G to Y Low			4.7		5.4		6.3	ns
Input M	odule Predicted Input Routing Del	ays [*]							
t _{IRD1}	FO = 1 Routing Delay			4.1		4.6		5.4	ns
t _{IRD2}	FO = 2 Routing Delay			4.6		5.2		6.1	ns
t _{IRD3}	FO = 3 Routing Delay			5.3		6.0		7.1	ns
t _{IRD4}	FO = 4 Routing Delay			5.7		6.4		7.6	ns
t _{IRD8}	FO = 8 Routing Delay			7.4		8.3		9.8	ns
Global (Clock Network						-	-	
t _{скн}	Input Low to High	FO = 32		10.2		11.0		12.8	ns
		FO = 256		11.8		13.0		15.7	
t _{CKL}	Input High to Low	FO = 32		10.2		11.0		12.8	ns
		FO = 256		12.0		13.2		15.9	
t _{PWH}	Minimum Pulse Width High	FO = 32	3.4		4.1		4.5		ns
		FO = 256	3.8		4.5		5.0		
t _{PWL}	Minimum Pulse Width Low	FO = 32	3.4		4.1		4.5		ns
		FO = 256	3.8		4.5		5.0		
t _{CKSW}	Maximum Skew	FO = 32		0.7		0.7		0.7	ns
		FO = 256		3.5		3.5		3.5	
t _{SUEXT}	Input Latch External Setup	FO = 32	0.0		0.0		0.0		ns
		FO = 256	0.0		0.0		0.0		
t _{HEXT}	Input Latch External Hold	FO = 32	7.0		7.0		7.0		ns
		FO = 256	11.2		11.2		11.2		
t _P	Minimum Period	FO = 32	7.7		8.3		9.1		ns
		FO = 256	8.1		8.8		10.0		
f _{MAX}	Maximum Frequency	FO = 32		130.0		120.0		110.0	ns
		FO = 256		125.0		115.0		100.0]

Note: *These parameters should be used for estimating device performance. Optimization techniques may further reduce delays by 0 to 4 ns. Routing delays are for typical designs across worst-case operating conditions. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.



Detailed Specifications

A1225A Timing Characteristics (continued)

Table 2-14 • A1225A Worst-Case Commercial Conditions, VCC = 4.75 V, T_J = 70°C

TTL Ou	tput Module Timing ¹	–2 S	peed	–1 S	peed	Std. Speed		Units
Parame	ter/Description	Min.	Max.	Min.	Max.	Min.	Max.	
t _{DLH}	Data to Pad High		8.0		9.0		10.6	ns
t _{DHL}	Data to Pad Low		10.1		11.4		13.4	ns
t _{ENZH}	Enable Pad Z to High		8.9		10.0		11.8	ns
t _{ENZL}	Enable Pad Z to Low		11.6		13.2		15.5	ns
t _{ENHZ}	Enable Pad High to Z		7.1		8.0		9.4	ns
t _{ENLZ}	Enable Pad Low to Z		8.3		9.5		11.1	ns
t _{GLH}	G to Pad High		8.9		10.2		11.9	ns
t _{GHL}	G to Pad Low		11.2		12.7		14.9	ns
d _{TLH}	Delta Low to High		0.07		0.08		0.09	ns/pF
d _{THL}	Delta High to Low		0.12		0.13		0.16	ns/pF
CMOS	Output Module Timing ¹	•						
t _{DLH}	Data to Pad High		10.1		11.5		13.5	ns
t _{DHL}	Data to Pad Low		8.4		9.6		11.2	ns
t _{ENZH}	Enable Pad Z to High		8.9		10.0		11.8	ns
t _{ENZL}	Enable Pad Z to Low		11.6		13.2		15.5	ns
t _{ENHZ}	Enable Pad High to Z		7.1		8.0		9.4	ns
t _{ENLZ}	Enable Pad Low to Z		8.3		9.5		11.1	ns
t _{GLH}	G to Pad High		8.9		10.2		11.9	ns
t _{GHL}	G to Pad Low		11.2		12.7		14.9	ns
d _{TLH}	Delta Low to High		0.12		0.13		0.16	ns/pF
d _{THL}	Delta High to Low		0.09		0.10		0.12	ns/pF

Notes:

1. Delays based on 50 pF loading.

2. SSO information can be found at www.microsemi.com/soc/techdocs/appnotes/board_consideration.aspx.

A1240A Timing Characteristics

Logic M	odule Propagation Delays ¹	–2 Sj	beed ³	-1 Speed		Std. Speed		Units
Parameter/Description		Min.	Max.	Min.	Max.	Min.	Max.	
t _{PD1}	Single Module		3.8		4.3		5.0	ns
t _{CO}	Sequential Clock to Q		3.8		4.3		5.0	ns
t _{GO}	Latch G to Q		3.8		4.3		5.0	ns
t _{RS}	Flip-Flop (Latch) Reset to Q		3.8		4.3		5.0	ns
Predicte	d Routing Delays ²							
t _{RD1}	FO = 1 Routing Delay		1.4		1.5		1.8	ns
t _{RD2}	FO = 2 Routing Delay		1.7		2.0		2.3	ns
t _{RD3}	FO = 3 Routing Delay		2.3		2.6		3.0	ns
t _{RD4}	FO = 4 Routing Delay		3.1		3.5		4.1	ns
t _{RD8}	FO = 8 Routing Delay		4.7		5.4		6.3	ns
Sequent	ial Timing Characteristics ^{3,4}							
t _{SUD}	Flip-Flop (Latch) Data Input Setup	0.4		0.4		0.5		ns
t _{HD}	Flip-Flop (Latch) Data Input Hold	0.0		0.0		0.0		ns
t _{SUENA}	Flip-Flop (Latch) Enable Setup	0.8		0.9		1.0		ns
t _{HENA}	Flip-Flop (Latch) Enable Hold	0.0		0.0		0.0		ns
t _{WCLKA}	Flip-Flop (Latch) Clock Active Pulse Width	4.5		6.0		6.5		ns
t _{WASYN}	Flip-Flop (Latch) Clock Asynchronous Pulse Width	4.5		6.0		6.5		ns
t _A	Flip-Flop Clock Input Period	9.8		12.0		15.0		ns
t _{INH}	Input Buffer Latch Hold	0.0		0.0		0.0		ns
t _{INSU}	Input Buffer Latch Setup	0.4		0.4		0.5		ns
t _{OUTH}	Output Buffer Latch Hold	0.0		0.0		0.0		ns
t _{outsu}	Output Buffer Latch Setup	0.4		0.4		0.5		ns
f _{MAX}	Flip-Flop (Latch) Clock Frequency		100.0		80.0		66.0	MHz

Notes:

1. For dual-module macros, use $t_{PD1} + t_{RD1} + t_{PDn}$, $t_{CO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$ —whichever is appropriate.

 Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

3. Data applies to macros based on the S-module. Timing parameters for sequential macros constructed from C-modules can be obtained from the DirectTime Analyzer utility.

4. Setup and hold timing parameters for the Input Buffer Latch are defined with respect to the PAD and the D input. External setup/hold timing parameters must account for delay from an external PAD signal to the G inputs. Delay from an external PAD signal to the G input subtracts (adds) to the internal setup (hold) time.

A1280A Timing Characteristics (continued)

Table 2-19 • A1280A Worst-Case Commercial Conditions, VCC = 4.75 V, T_J = 70°C

I/O Module Input Propagation Delays			-2 S	peed	–1 S	peed	Std. Speed		Units
Parameter/Description		Min.	Max.	Min.	Max.	Min.	Max.	1	
t _{INYH}	Pad to Y High			2.9		3.3		3.8	ns
t _{INYL}	Pad to Y Low			2.7		3.0		3.5	ns
t _{INGH}	G to Y High			5.0		5.7		6.6	ns
t _{INGL}	G to Y Low			4.8		5.4		6.3	ns
Input M	odule Predicted Input Routing Del	ays [*]	-				-	-	
t _{IRD1}	FO = 1 Routing Delay			4.6		5.1		6.0	ns
t _{IRD2}	FO = 2 Routing Delay			5.2		5.9		6.9	ns
t _{IRD3}	FO = 3 Routing Delay			5.6		6.3		7.4	ns
t _{IRD4}	FO = 4 Routing Delay			6.5		7.3		8.6	ns
t _{IRD8}	FO = 8 Routing Delay			9.4		10.5		12.4	ns
Global (Clock Network		-				-	-	
t _{скн}	Input Low to High	FO = 32		10.2		11.0		12.8	ns
		FO = 256		13.1		14.6		17.2	1
t _{CKL}	Input High to Low	FO = 32		10.2		11.0		12.8	ns
		FO = 256		13.3		14.9		17.5	
t _{PWH}	Minimum Pulse Width High	FO = 32	5.0		5.5		6.6		ns
		FO = 256	5.8		6.4		7.6		
t _{PWL}	Minimum Pulse Width Low	FO = 32	5.0		5.5		6.6		ns
		FO = 256	5.8		6.4		7.6		
t _{CKSW}	Maximum Skew	FO = 32		0.5		0.5		0.5	ns
		FO = 256		2.5		2.5		2.5	
t _{SUEXT}	Input Latch External Setup	FO = 32	0.0		0.0		0.0		ns
		FO = 256	0.0		0.0		0.0		
t _{HEXT}	Input Latch External Hold	FO = 32	7.0		7.0		7.0		ns
		FO = 256	11.2		11.2		11.2		
t _P	Minimum Period	FO = 32	9.6		11.2		13.3		ns
		FO = 256	10.6		12.6		15.3]
f _{MAX}	Maximum Frequency	FO = 32		105.0		90.0		75.0	ns
		FO = 256		95.0		80.0		65.0	1

Note: *These parameters should be used for estimating device performance. Optimization techniques may further reduce delays by 0 to 4 ns. Routing delays are for typical designs across worst-case operating conditions. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

A1280A Timing Characteristics (continued)



Detailed Specifications

TTL Ou	tput Module Timing ¹	–2 S	peed	-1 S	peed	Std.	Speed	Units
Parame	ter/Description	Min.	Max.	Min.	Max.	Min.	Max.	
t _{DLH}	Data to Pad High		8.1		9.0		10.6	ns
t _{DHL}	Data to Pad Low		10.2		11.4		13.4	ns
t _{ENZH}	Enable Pad Z to High		9.0		10.0		11.8	ns
t _{ENZL}	Enable Pad Z to Low		11.8		13.2		15.5	ns
t _{ENHZ}	Enable Pad High to Z		7.1		8.0		9.4	ns
t _{ENLZ}	Enable Pad Low to Z		8.4		9.5		11.1	ns
t _{GLH}	G to Pad High		9.0		10.2		11.9	ns
t _{GHL}	G to Pad Low		11.3		12.7		14.9	ns
d _{TLH}	Delta Low to High		0.07		0.08		0.09	ns/pF
d _{THL}	Delta High to Low		0.12		0.13		0.16	ns/pF
CMOS	Output Module Timing ¹							
t _{DLH}	Data to Pad High		10.3		11.5		13.5	ns
t _{DHL}	Data to Pad Low		8.5		9.6		11.2	ns
t _{ENZH}	Enable Pad Z to High		9.0		10.0		11.8	ns
t _{ENZL}	Enable Pad Z to Low		11.8		13.2		15.5	ns
t _{ENHZ}	Enable Pad High to Z		7.1		8.0		9.4	ns
t _{ENLZ}	Enable Pad Low to Z		8.4		9.5		11.1	ns
t _{GLH}	G to Pad High		9.0		10.2		11.9	ns
t _{GHL}	G to Pad Low		11.3		12.7		14.9	ns
d _{TLH}	Delta Low to High		0.12		0.13		0.16	ns/pF
d _{THL}	Delta High to Low		0.09		0.10		0.12	ns/pF

Table 2-20 • A1280A Worst-Case Commercial Conditions, VCC = 4.75 V, $T_J = 70^{\circ}C$

Notes:

1. Delays based on 50 pF loading.

2. SSO information can be found at www.microsemi.com/soc/techdocs/appnotes/board_consideration.aspx.



PL84			
Pin Number	A1225A Function	A1240A Function	A1280A Function
2	CLKB, I/O	CLKB, I/O	CLKB, I/O
4	PRB, I/O	PRB, I/O	PRB, I/O
6	GND	GND	GND
10	DCLK, I/O	DCLK, I/O	DCLK, I/O
12	MODE	MODE	MODE
22	VCC	VCC	VCC
23	VCC	VCC	VCC
28	GND	GND	GND
43	VCC	VCC	VCC
49	GND	GND	GND
52	SDO	SDO	SDO
63	GND	GND	GND
64	VCC	VCC	VCC
65	VCC	VCC	VCC
70	GND	GND	GND
76	SDI, I/O	SDI, I/O	SDI, I/O
81	PRA, I/O	PRA, I/O	PRA, I/O
83	CLKA, I/O	CLKA, I/O	CLKA, I/O
84	VCC	VCC	VCC

Notes:

1. All unlisted pin numbers are user I/Os.

2. MODE pin should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.



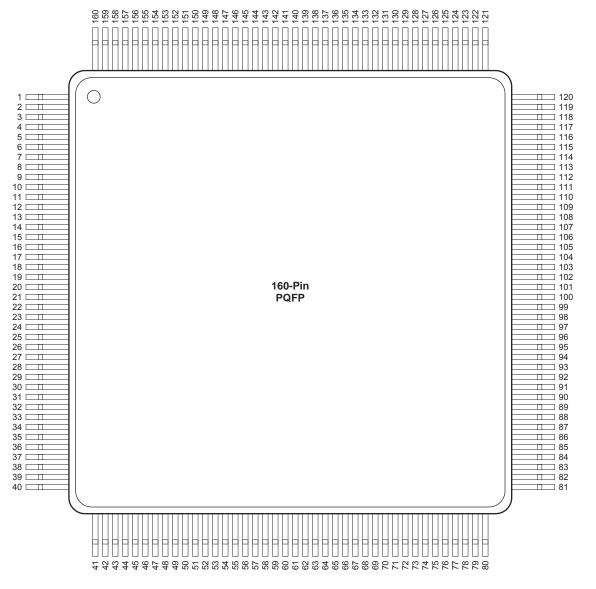
PQ100		PQ100	
Pin Number	A1225A Function	Pin Number	A1225A Function
2	DCLK, I/O	65	VCC
4	MODE	66	VCC
9	GND	67	VCC
16	VCC	72	GND
17	VCC	79	SDI, I/O
22	GND	84	GND
34	GND	87	PRA, I/O
40	VCC	89	CLKA, I/O
46	GND	90	VCC
52	SDO	92	CLKB, I/O
57	GND	94	PRB, I/O
64	GND	96	GND

- 1. All unlisted pin numbers are user I/Os.
- 2. MODE pin should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.



PQ144		PQ144	
Pin Number	A1240A Function	Pin Number	A1240A Function
2	MODE	89	VCC
9	GND	90	VCC
10	GND	91	VCC
11	GND	92	VCC
18	VCC	93	VCC
19	VCC	100	GND
20	VCC	101	GND
21	VCC	102	GND
28	GND	110	SDI, I/O
29	GND	116	GND
30	GND	117	GND
44	GND	118	GND
45	GND	123	PRA, I/O
46	GND	125	CLKA, I/O
54	VCC	126	VCC
55	VCC	127	VCC
56	VCC	128	VCC
64	GND	130	CLKB, I/O
65	GND	132	PRB, I/O
71	SDO	136	GND
79	GND	137	GND
80	GND	138	GND
81	GND	144	DCLK, I/O
88	GND		

- 1. All unlisted pin numbers are user I/Os.
- 2. MODE pin should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.



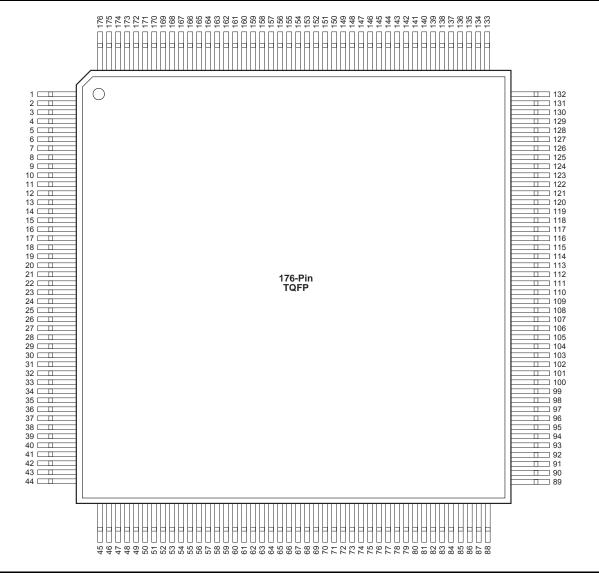
Note: This is the top view of the package

Note

For Package Manufacturing and Environmental information, visit the Resource Center at http://www.microsemi.com/soc/products/solutions/package/docs.aspx

Microsemi. ACT 2 Family FPGAs





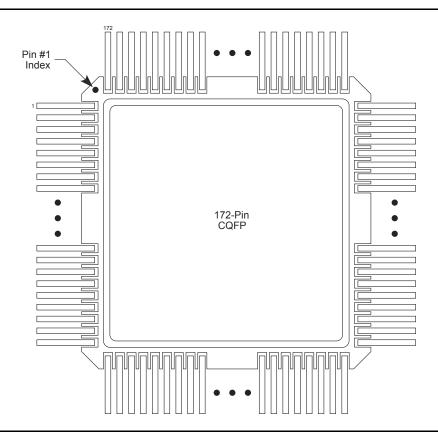
Note

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Microsemi. ACT 2 Family FPGAs



CQ172



Note

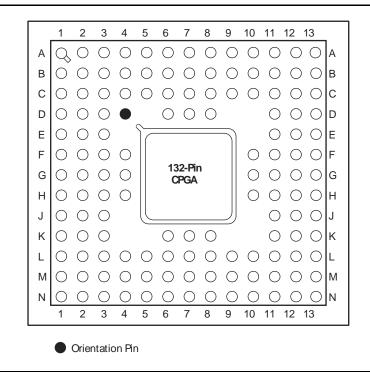
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PG100		PG100	
Pin Number	A1225A Function	Pin Number	A1225A Function
A4	PRB, I/O	E11	VCC
A7	PRA, I/O	F3	VCC
B6	VCC	F9	VCC
C2	MODE	F10	VCC
C3	DCLK, I/O	F11	GND
C5	GND	G1	VCC
C6	CLKA, I/O	G3	GND
C7	GND	G9	GND
C8	SDI, I/O	J5	GND
D6	CLKB, I/O	J7	GND
D10	GND	J9	SDO
E3	GND	K6	VCC

- 1. All unlisted pin numbers are user I/Os.
- 2. MODE pin should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.



PG132



Note

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PG132		
Pin Number	A1240A Function	
A1	MODE	
B5	GND	
B6	CLKB, I/O	
B7	CLKA, I/O	
B8	PRA, I/O	
B9	GND	
B12	SDI, I/O	
C3	DCLK, I/O	
C5	GND	
C6	PRB, I/O	
C7	VCC	
C9	GND	
D7	VCC	
E3	GND	
E11	GND	
E12	GND	
F4	GND	
G2	VCC	

PG132		
Pin Number	A1240A Function	
G3	VCC	
G4	VCC	
G10	VCC	
G11	VCC	
G12	VCC	
G13	VCC	
H13	GND	
J2	GND	
J3	GND	
J11	GND	
K7	VCC	
K12	GND	
L5	GND	
L7	VCC	
L9	GND	
M9	GND	
N12	SDO	

- 1. All unlisted pin numbers are user I/Os.
- 2. MODE pin should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.