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#### Details

Product Status	Obsolete
Core Processor	80C51
Core Size	8-Bit
Speed	40/30MHz
Connectivity	UART/USART
Peripherals	POR, PWM, WDT
Number of I/O	32
Program Memory Size	32KB (32K x 8)
Program Memory Type	OTP
EEPROM Size	<u>.</u>
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 5.5V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	44-QFP
Supplier Device Package	44-VQFP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/ts87c51rc2-vce

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# 6. TS80C51Rx2 Enhanced Features

In comparison to the original 80C52, the TS80C51Rx2 implements some new features, which are:

- The X2 option.
- The Dual Data Pointer.
- The extended RAM.
- The Programmable Counter Array (PCA).
- The Watchdog.
- The 4 level interrupt priority system.
- The power-off flag.
- The ONCE mode.
- The ALE disabling.
- Some enhanced features are also located in the UART and the timer 2.

### 6.1. X2 Feature

The TS80C51Rx2 core needs only 6 clock periods per machine cycle. This feature called "X2" provides the following advantages:

- Divide frequency crystals by 2 (cheaper crystals) while keeping same CPU power.
- Save power consumption while keeping same CPU power (oscillator power saving).
- Save power consumption by dividing dynamically operating frequency by 2 in operating and idle modes.
- Increase CPU power by 2 while keeping same crystal frequency.

In order to keep the original C51 compatibility, a divider by 2 is inserted between the XTAL1 signal and the main clock input of the core (phase generator). This divider may be disabled by software.

#### 6.1.1. Description

The clock for the whole circuit and peripheral is first divided by two before being used by the CPU core and peripherals. This allows any cyclic ratio to be accepted on XTAL1 input. In X2 mode, as this divider is bypassed, the signals on XTAL1 must have a cyclic ratio between 40 to 60%. Figure 1. shows the clock generation block diagram. X2 bit is validated on XTAL1÷2 rising edge to avoid glitches when switching from X2 to STD mode. Figure 2. shows the mode switching waveforms.

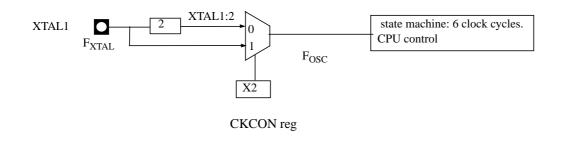


Figure 1. Clock Generation Diagram



### 6.2. Dual Data Pointer Register Ddptr

The additional data pointer can be used to speed up code execution and reduce code size in a number of ways.

The dual DPTR structure is a way by which the chip will specify the address of an external data memory location. There are two 16-bit DPTR registers that address the external memory, and a single bit called DPS = AUXR1/bit0 (See Table 4.) that allows the program code to switch between them (Refer to Figure 3).

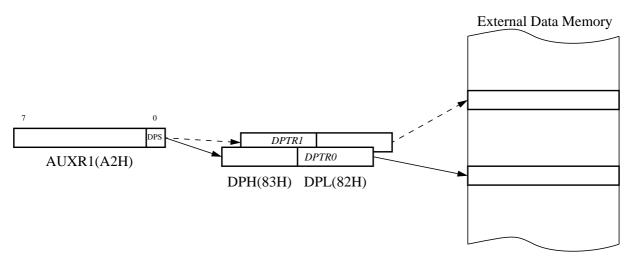


Figure 3. Use of Dual Pointer

 Table 4. AUXR1: Auxiliary Register 1

AUXR1 Address 0A2H		-	-	-	-	GF3	-	-	DPS
	Reset value	Х	Х	Х	Х	0	Х	Х	0

Symbol	Function								
-	Not implement	ted, reserved for future use. <sup>a</sup>							
DPS	Data Pointer S	ta Pointer Selection.							
	DPS	Operating Mode							
	0	DPTR0 Selected							
	1	DPTR1 Selected							
GF3	This bit is a general purpose user flag <sup>b</sup> .								

a. User software should not write 1s to reserved bits. These bits may be used in future 8051 family products to invoke new feature. In that case, the reset value of the new bit will be 0, and its active value will be 1. The value read from a reserved bit is indeterminate.

b. GF3 will not be available on first version of the RC devices.

### Application

Software can take advantage of the additional data pointers to both increase speed and reduce code size, for example, block operations (copy, compare, search ...) are well served by using one data pointer as a 'source' pointer and the other one as a "destination" pointer.



## 6.3. Expanded RAM (XRAM)

The TS80C51Rx2 provide additional Bytes of ramdom access memory (RAM) space for increased data parameter handling and high level language usage.

RA2, RB2 and RC2 devices have 256 bytes of expanded RAM, from 00H to FFH in external data space; RD2 devices have 768 bytes of expanded RAM, from 00H to 2FFH in external data space.

The TS80C51Rx2 has internal data memory that is mapped into four separate segments.

The four segments are:

- 1. The Lower 128 bytes of RAM (addresses 00H to 7FH) are directly and indirectly addressable.
- 2. The Upper 128 bytes of RAM (addresses 80H to FFH) are indirectly addressable only.
- 3. The Special Function Registers, SFRs, (addresses 80H to FFH) are directly addressable only.
- 4. The expanded RAM bytes are indirectly accessed by MOVX instructions, and with the EXTRAM bit cleared in the AUXR register. (See Table 5.)

The Lower 128 bytes can be accessed by either direct or indirect addressing. The Upper 128 bytes can be accessed by indirect addressing only. The Upper 128 bytes occupy the same address space as the SFR. That means they have the same address, but are physically separate from SFR space.

When an instruction accesses an internal location above address 7FH, the CPU knows whether the access is to the upper 128 bytes of data RAM or to SFR space by the addressing mode used in the instruction.

- Instructions that use direct addressing access SFR space. For example: MOV 0A0H, # data ,accesses the SFR at location 0A0H (which is P2).
- Instructions that use indirect addressing access the Upper 128 bytes of data RAM. For example: MOV @R0, # data where R0 contains 0A0H, accesses the data byte at address 0A0H, rather than P2 (whose address is 0A0H).
- The 256 or 768 XRAM bytes can be accessed by indirect addressing, with EXTRAM bit cleared and MOVX instructions. This part of memory which is physically located on-chip, logically occupies the first 256 or 768 bytes of external data memory.
- With <u>EXTRAM = 0</u>, the XRAM is indirectly addressed, using the MOVX instruction in combination with any of the registers R0, R1 of the selected bank or DPTR. An access to XRAM will not affect ports P0, P2, P3.6 ( $\overline{WR}$ ) and P3.7 ( $\overline{RD}$ ). For example, with EXTRAM = 0, MOVX @R0, # data where R0 contains 0A0H, accesses the XRAM at address 0A0H rather than external memory. An access to external data memory locations higher than FFH (i.e. 0100H to FFFFH) (higher than 2FFH (i.e. 0300H to FFFFH for RD devices) will be performed with the MOVX DPTR instructions in the same way as in the standard 80C51, so with P0 and P2 as data/address busses, and P3.6 and P3.7 as write and read timing signals. Refer to Figure . For RD devices, accesses to expanded RAM from 100H to 2FFH can only be done thanks to the use of DPTR.
- With <u>EXTRAM = 1</u>, MOVX @Ri and MOVX @DPTR will be similar to the standard 80C51. MOVX @ Ri will provide an eight-bit address multiplexed with data on Port0 and any output port pins can be used to output higher order address bits. This is to provide the external paging capability. MOVX @DPTR will generate a sixteen-bit address. Port2 outputs the high-order eight address bits (the contents of DPH) while Port0 multiplexes the low-order eight address bits (DPL) with data. MOVX @ Ri and MOVX @DPTR will generate either read or write signals on P3.6 (WR) and P3.7 (RD).

The stack pointer (SP) may be located anywhere in the 256 bytes RAM (lower and upper RAM) internal data memory. The stack may not be located in the XRAM.



It is possible to use timer 2 as a baud rate generator and a clock generator simultaneously. For this configuration, the baud rates and clock frequencies are not independent since both functions use the values in the RCAP2H and RCAP2L registers.

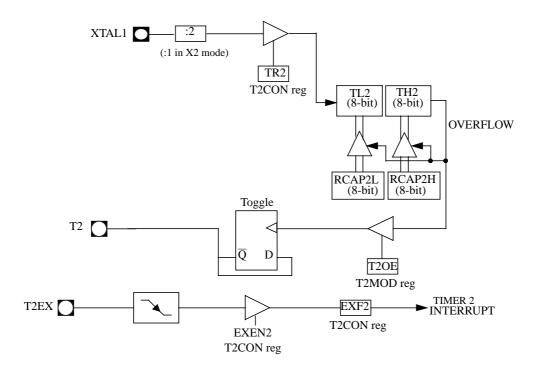


Figure 6. Clock-Out Mode  $C/\overline{T2} = 0$ 



### Table 6. T2CON Register

### T2CON - Timer 2 Control Register (C8h)

7	6	5	4	3	2	1	0			
TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2#	CP/RL2#			
Bit Number	Bit Mnemonic			Descrip	tion					
7	TF2	Timer 2 overflow Fl Must be cleared Set by hardware	by software.	w, if $\mathbf{RCLK} = 0$ and	d TCLK = 0.					
6	EXF2	When set, causes	re or a reload is ca the CPU to vecto	aused by a negative r to timer 2 interru 2 doesn't cause an i	pt routine when tin	ner 2 interrupt is e	nabled.			
5	RCLK			ceive clock for seria						
4	TCLK			nsmit clock for ser mit clock for seria						
3	EXEN2		vents on T2EX pin	n for timer 2 operat en a negative trans		is detected, if time	er 2 is not used to			
2	TR2	Timer 2 Run contro Clear to turn off Set to turn on tim	timer 2.							
1	C/T2#		peration (input fro	om internal clock sy om T2 input pin, fal	ODC/	Must be 0 for clos	ck out mode.			
0	CP/RL2#	If RCLK=1 or T Clear to auto-rel	ner 2 Capture/Reload bit If RCLK=1 or TCLK=1, CP/RL2# is ignored and timer is forced to auto-reload on timer 2 overflow. Clear to auto-reload on timer 2 overflows or negative transitions on T2EX pin if EXEN2=1. Set to capture on negative transitions on T2EX pin if EXEN2=1.							

Reset Value = 0000 0000b Bit addressable



**The CCON SFR** contains the run control bit for the PCA and the flags for the PCA timer (CF) and each module (Refer to Table 9).

- Bit CR (CCON.6) must be set by software to run the PCA. The PCA is shut off by clearing this bit.
- Bit CF: The CF bit (CCON.7) is set when the PCA counter overflows and an interrupt will be generated if the ECF bit in the CMOD register is set. The CF bit can only be cleared by software.
- Bits 0 through 4 are the flags for the modules (bit 0 for module 0, bit 1 for module 1, etc.) and are set by hardware when either a match or a capture occurs. These flags also can only be cleared by software.

	PCA Cou		CF	CR	-	CCF4	CCF3	CCF2	CCF1	CCF0		
	Rese	et value	0	0	X	0	0	0	0	0		
Sy	nbol	Function	ı									
CF		an interrup	CA Counter Overflow flag. Set by hardware when the counter rolls over. CF flags interrupt if bit ECF in CMOD is set. CF may be set by either hardware or software n only be cleared by software.									
CR		1	CA Counter Run control bit. Set by software to turn the PCA counter on. Must be cleared y software to turn the PCA counter off.									
-		Not implen	nented, res	erved for	future use	e. <sup>a</sup>						
CCF4		PCA Modu cleared by		rupt flag.	Set by ha	ardware wh	nen a matc	h or captu	are occurs	. Must be		
CCF3		PCA Modu cleared by		rupt flag.	Set by ha	ardware wł	nen a matc	h or captu	are occurs	. Must be		
CCF2		PCA Modu cleared by		rupt flag.	Set by ha	ardware wł	nen a matc	h or captu	are occurs	. Must be		
CCF1		PCA Modu cleared by		rupt flag.	Set by ha	ardware wł	nen a matc	h or captu	are occurs	. Must be		
CCF0		PCA Modu cleared by		rupt flag.	Set by ha	ardware wł	nen a matc	h or captu	are occurs	. Must be		

 Table 9. CCON: PCA Counter Control Register

a. User software should not write 1s to reserved bits. These bits may be used in future 8051 family products to invoke new features. In that case, the reset or inactive value of the new bit will be 0, and its active value will be 1. The value read from a reserved bit is indeterminate.

The watchdog timer function is implemented in module 4 (See Figure 10).

The PCA interrupt system is shown in Figure 8



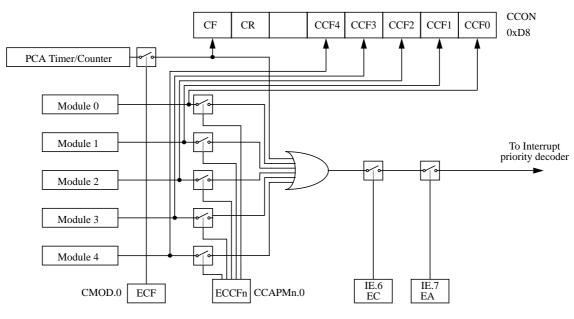


Figure 8. PCA Interrupt System

PCA Modules: each one of the five compare/capture modules has six possible functions. It can perform:

- 16-bit Capture, positive-edge triggered,
- 16-bit Capture, negative-edge triggered,
- 16-bit Capture, both positive and negative-edge triggered,
- 16-bit Software Timer,
- 16-bit High Speed Output,
- 8-bit Pulse Width Modulator.

In addition, module 4 can be used as a Watchdog Timer.

Each module in the PCA has a special function register associated with it. These registers are: CCAPM0 for module 0, CCAPM1 for module 1, etc. (See Table 10). The registers contain the bits that control the mode that each module will operate in.

- The ECCF bit (CCAPMn.0 where n=0, 1, 2, 3, or 4 depending on the module) enables the CCF flag in the CCON SFR to generate an interrupt when a match or compare occurs in the associated module.
- PWM (CCAPMn.1) enables the pulse width modulation mode.
- The TOG bit (CCAPMn.2) when set causes the CEX output associated with the module to toggle when there is a match between the PCA counter and the module's capture/compare register.
- The match bit MAT (CCAPMn.3) when set will cause the CCFn bit in the CCON register to be set when there is a match between the PCA counter and the module's capture/compare register.
- The next two bits CAPN (CCAPMn.4) and CAPP (CCAPMn.5) determine the edge that a capture input will be active on. The CAPN bit enables the negative edge, and the CAPP bit enables the positive edge. If both bits are set both edges will be enabled and a capture will occur for either transition.
- The last bit in the register ECOM (CCAPMn.6) when set enables the comparator function.

Table 11 shows the CCAPMn settings for the various PCA functions.



Table 12.	CCAPnH:	PCA	Modules	Capture/C	ompare	Registers	High

CCAPnH Address n = 0 - 4	CCAP0H=0FAH CCAP1H=0FBH CCAP2H=0FCH CCAP3H=0FDH CCAP3H=0FEH								
		7	6	5	4	3	2	1	0
	Reset value	0	0	0	0	0	0	0	0

### Table 13. CCAPnL: PCA Modules Capture/Compare Registers Low

CCAPnL Address n = 0 - 4	CCAP0L=0EAH CCAP1L=0EBH CCAP2L=0ECH CCAP3L=0EDH CCAP4L=0EEH								
		7	6	5	4	3	2	1	0
	Reset value	0	0	0	0	0	0	0	0

### Table 14. CH: PCA Counter High

CH Address 0F9H									
		7	6	5	4	3	2	1	0
	Reset value	0	0	0	0	0	0	0	0

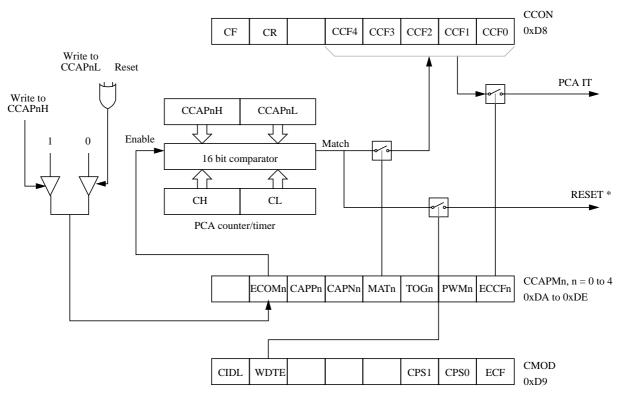
#### Table 15. CL: PCA Counter Low

CL Address 0E9H									
		7	6	5	4	3	2	1	0
	Reset value	0	0	0	0	0	0	0	0



### 6.5.2. 16-bit Software Timer / Compare Mode

The PCA modules can be used as software timers by setting both the ECOM and MAT bits in the modules CCAPMn register. The PCA timer will be compared to the module's capture registers and when a match occurs an interrupt will occur if the CCFn (CCON SFR) and the ECCFn (CCAPMn SFR) bits for the module are both set (See Figure 10).



\* Only for Module 4

#### Figure 10. PCA Compare Mode and PCA Watchdog Timer

Before enabling ECOM bit, CCAPnL and CCAPnH should be set with a non zero value, otherwise an unwanted match could happen. Writing to CCAPnH will set the ECOM bit.

Once ECOM set, writing CCAPnL will clear ECOM so that an unwanted match doesn't occur while modifying the compare value. Writing to CCAPnH will set ECOM. For this reason, user software should write CCAPnL first, and then CCAPnH. Of course, the ECOM bit can still be controlled by accessing to CCAPMn register.



## 6.5.4. Pulse Width Modulator Mode

All of the PCA modules can be used as PWM outputs. Figure 12 shows the PWM function. The frequency of the output depends on the source for the PCA timer. All of the modules will have the same frequency of output because they all share the PCA timer. The duty cycle of each module is independently variable using the module's capture register CCAPLn. When the value of the PCA CL SFR is less than the value in the module's CCAPLn SFR the output will be low, when it is equal to or greater than the output will be high. When CL overflows from FF to 00, CCAPLn is reloaded with the value in CCAPHn. This allows updating the PWM without glitches. The PWM and ECOM bits in the module's CCAPMn register must be set to enable the PWM mode.

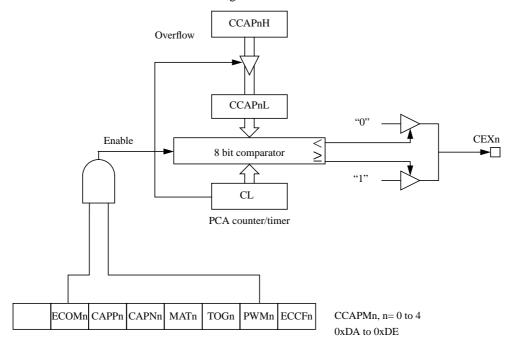


Figure 12. PCA PWM Mode

## 6.5.5. PCA Watchdog Timer

An on-board watchdog timer is available with the PCA to improve the reliability of the system without increasing chip count. Watchdog timers are useful for systems that are susceptible to noise, power glitches, or electrostatic discharge. Module 4 is the only PCA module that can be programmed as a watchdog. However, this module can still be used for other modes if the watchdog is not needed. Figure 10 shows a diagram of how the watchdog works. The user pre-loads a 16-bit value in the compare registers. Just like the other compare modes, this 16-bit value is compared to the PCA timer value. If a match is allowed to occur, an internal reset will be generated. This will not cause the RST pin to be driven high.

In order to hold off the reset, the user has three options:

- 1. periodically change the compare value so it will never match the PCA timer,
- 2. periodically change the PCA timer value so it will never match the compare values, or
- 3. disable the watchdog by clearing the WDTE bit before a match occurs and then re-enable it.

The first two options are more reliable because the watchdog timer is never disabled as in option #3. If the program counter ever goes astray, a match will eventually occur and cause an internal reset. The second option is also not recommended if other PCA modules are being used. Remember, the PCA timer is the time base for all modules; changing the time base for other modules would not be a good idea. Thus, in most applications the first solution is the best option.

This watchdog timer won't generate a reset out on the reset pin.



## 6.6. TS80C51Rx2 Serial I/O Port

The serial I/O port in the TS80C51Rx2 is compatible with the serial I/O port in the 80C52. It provides both synchronous and asynchronous communication modes. It operates as an Universal Asynchronous

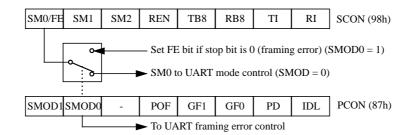
Receiver and Transmitter (UART) in three full-duplex modes (Modes 1, 2 and 3). Asynchronous transmission and reception can occur simultaneously and at different baud rates

Serial I/O port includes the following enhancements:

- Framing error detection
- Automatic address recognition

### 6.6.1. Framing Error Detection

Framing bit error detection is provided for the three asynchronous modes (modes 1, 2 and 3). To enable the framing bit error detection feature, set SMOD0 bit in PCON register (See Figure 13).



### Figure 13. Framing Error Block Diagram

When this feature is enabled, the receiver checks each incoming data frame for a valid stop bit. An invalid stop bit may result from noise on the serial lines or from simultaneous transmission by two CPUs. If a valid stop bit is not found, the Framing Error bit (FE) in SCON register (See Table 16.) bit is set.



## Table 16. SCON Register

#### SCON - Serial Control Register (98h)

7	6	5		4	3	2	1	0
FE/SM0	SM1	SN	12	REN	TB8	RB8	TI	RI
Bit Number	Bit Mnemonic				Descrip	tion	-	·
7	FE	Set by h	reset the ardware v	error state, not cle	eared by a valid sto op bit is detected. ss to the FE bit	p bit.		
	SM0		SM1 for	serial port mode	selection. access to the SM0	bit		
		Serial port I <u>SM0</u>	Mode bit <u>SM1</u>		Descripti	on Baud Rate	2	
6	SM1	0 0 1 1	0 1 0 1	0 1 2 3	Shift Reg 8-bit UAI 9-bit UAI 9-bit UAI	RT Variable RT F <sub>XTAL</sub> /6	4 or F <sub>XTAL</sub> /32 (/32	
5	SM2	Clear to Set to en	disable n	nultiprocessor cor tiprocessor comm	cessor Communic nmunication featur unication feature ir	e.	l eventually mode	1. This bit should
4	REN		disable s	t erial reception. al reception.				
3	TB8	Clear to	transmit	a logic 0 in the 9t logic 1 in the 9t		d 3.		
2	RB8	Cleared Set by h	by hardw ardware i	<b>h bit received in</b> are if 9th bit rece f 9th bit received 2 = 0, RB8 is the	ived is a logic 0.	n mode 0 RB8 is n	ot used.	
1	TI		acknowle	edge interrupt.	th bit time in mode	0 or at the beginn	ing of the stop bit	in the other
0	RI		acknowle	edge interrupt.	th bit time in mode	0, see Figure 14.	and Figure 15. in	the other modes.

Reset Value = 0000 0000b Bit addressable



#### Table 21. IPH Register

#### IPH - Interrupt Priority High Register (B7h)

7	6	5	4	3	2	1	0
-	РРСН	РТ2Н	PSH	PT1H	PX1H	РТОН	РХОН
Bit Number	Bit Mnemonic			Descrip	tion		
7	-	<b>Reserved</b> The value read f	from this bit is ind	eterminate. Do not s	et this bit.		
6	РРСН	PCA interrupt prio <u>PPCH</u> 0 1 1		<u>ority Level</u> Lowest Highest			
5	РТ2Н	<b>Timer 2 overflow in</b> <u>PT2H</u> 0 0 1 1 1	tterrupt Priority <u>PT2</u> 0 1 0 1 1 1 1 1 1 1	<b>High bit</b> <u>Priority Level</u> Lowest Highest			
4	PSH	Serial port Priority <u>PSH</u> 0 1 1 1	High bit <u>PS</u> 0 1 0 1	<u>Priority Level</u> Lowest Highest			
3	PT1H	<b>Timer 1 overflow in</b> <u>PT1H</u> 0 0 1 1 1	terrupt Priority <u>PT1</u> 0 1 0 1 1 1 1 1 1 1	High bit <u>Priority Level</u> Lowest Highest			
2	PX1H	External interrupt <u>PX1H</u> 0 0 1 1 1	1 Priority High b <u>PX1</u> 0 1 0 1 1	it <u>Priority Level</u> Lowest Highest			
1	РТОН	Timer 0 overflow in <u>PT0H</u> 0           1           1	tterrupt Priority <u>PTO</u> 0 1 0 1 1	<b>High bit</b> <u>Priority Level</u> Lowest Highest			
0	РХОН	External interrupt	0 Priority High b <u>PX0</u> 0 1 0 1	<b>it</b> <u>Priority Level</u> Lowest Highest			

Reset Value = X000 0000b Not bit addressable



#### Table 24. WDTPRG Register

7	6		5	4	3	2	1	0	
T4 T3			T2	T1	TO	<u>82</u>	<u>81</u>	<b>S0</b>	
Bit Number	Bit Mnemonic		Description						
7	T4								
6	Т3		Reserved Do not try to set or clear this bit.						
5	T2								
4	T1								
3	TO								
2	S2	WDT Ti	WDT Time-out select bit 2						
1	S1	WDT Ti	WDT Time-out select bit 1						
0	SO	WDT Ti	WDT Time-out select bit 0						
		$\frac{S2}{0}$ 0 0 0 1 1 1 1 1	<u>S1</u> 0 1 1 0 0 1 1	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	d Time-out ) machine cycles, 1( ) machine cycles, 3( ) machine cycles, 6( ) machine cycles, 1( ) machine cycles, 2( ) machine cycles, 5- ) machine cycles, 1. ) machine cycles, 2.	2.7 ms @ 12 MHz 5.5 ms @ 12 MHz 31 ms @ 12 MHz 52 ms @ 12 MHz 42 ms @ 12 MHz 05 s @ 12 MHz			

Reset value XXXX X000

### 6.10.2. WDT during Power Down and Idle

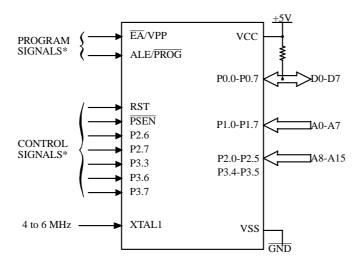
In Power Down mode the oscillator stops, which means the WDT also stops. While in Power Down mode the user does not need to service the WDT. There are 2 methods of exiting Power Down mode: by a hardware reset or via a level activated external interrupt which is enabled prior to entering Power Down mode. When Power Down is exited with hardware reset, servicing the WDT should occur as it normally should whenever the TS80C51Rx2 is reset. Exiting Power Down with an interrupt is significantly different. The interrupt is held low long enough for the oscillator to stabilize. When the interrupt is brought high, the interrupt is serviced. To prevent the WDT from resetting the device while the interrupt pin is held low, the WDT is not started until the interrupt is pulled high. It is suggested that the WDT be reset during the interrupt service routine.

To ensure that the WDT does not overflow within a few states of exiting of powerdown, it is best to reset the WDT just before entering powerdown.

In the Idle mode, the oscillator continues to run. To prevent the WDT from resetting the TS80C51Rx2 while in Idle mode, the user should always set up a timer that will periodically exit Idle, service the WDT, and re-enter Idle mode.

## WDTPRG Address (0A7h)





\* See Table 31. for proper value on these inputs

Figure 18. Set-Up Modes Configuration

### 8.3.3. Programming Algorithm

The Improved Quick Pulse algorithm is based on the Quick Pulse algorithm and decreases the number of pulses applied during byte programming from 25 to 1.

To program the TS87C51RB2/RC2/RD2 the following sequence must be exercised:

- Step 1: Activate the combination of control signals.
- Step 2: Input the valid address on the address lines.
- Step 3: Input the appropriate data on the data lines.
- Step 4: Raise  $\overline{EA}/VPP$  from VCC to VPP (typical 12.75V).
- Step 5: Pulse ALE/PROG once.
- Step 6: Lower  $\overline{EA}/VPP$  from VPP to VCC

Repeat step 2 through 6 changing the address and data for the entire array or until the end of the object file is reached (See Figure 19.).

### 8.3.4. Verify algorithm

Code array verify must be done after each byte or block of bytes is programmed. In either case, a complete verify of the programmed array will ensure reliable programming of the TS87C51RB2/RC2/RD2.

P 2.7 is used to enable data output.

To verify the TS87C51RB2/RC2/RD2 code the following sequence must be exercised:

- Step 1: Activate the combination of program and control signals.
- Step 2: Input the valid address on the address lines.
- Step 3: Read data on the data lines.

Repeat step 2 through 3 changing the address for the entire array verification (See Figure 19.)

The encryption array cannot be directly verified. Verification of the encryption array is done by observing that the code array is well encrypted.



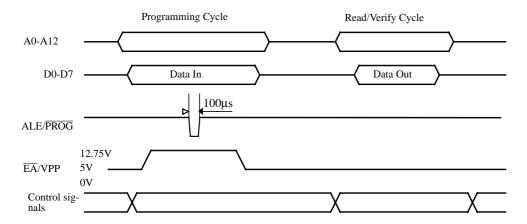


Figure 19. Programming and Verification Signal's Waveform

## 8.4. EPROM Erasure (Windowed Packages Only)

Erasing the EPROM erases the code array, the encryption array and the lock bits returning the parts to full functionality.

Erasure leaves all the EPROM cells in a 1's state (FF).

### 8.4.1. Erasure Characteristics

The recommended erasure procedure is exposure to ultraviolet light (at 2537 Å) to an integrated dose at least 15 W-sec/cm<sup>2</sup>. Exposing the EPROM to an ultraviolet lamp of 12,000  $\mu$ W/cm<sup>2</sup> rating for 30 minutes, at a distance of about 25 mm, should be sufficient. An exposure of 1 hour is recommended with most of standard erasers.

Erasure of the EPROM begins to occur when the chip is exposed to light with wavelength shorter than approximately 4,000 Å. Since sunlight and fluorescent lighting have wavelengths in this range, exposure to these light sources over an extended time (about 1 week in sunlight, or 3 years in room-level fluorescent lighting) could cause inadvertent erasure. If an application subjects the device to this type of exposure, it is suggested that an opaque label be placed over the window.



## 9. Signature Bytes

The TS83/87C51RB2/RC2/RD2 has four signature bytes in location 30h, 31h, 60h and 61h. To read these bytes follow the procedure for EPROM verify but activate the control lines provided in Table 31. for Read Signature Bytes. Table 31. shows the content of the signature byte for the TS87C51RB2/RC2/RD2.

Location	Contents Comment	
30h	58h	Manufacturer Code: Atmel Wireless & Microcontrollers
31h	57h	Family Code: C51 X2
60h	7Ch	Product name: TS83C51RD2
60h	FCh	Product name: TS87C51RD2
60h	37h	Product name: TS83C51RC2
60h	B7h	Product name: TS87C51RC2
60h	3Bh	Product name: TS83C51RB2
60h	BBh	Product name: TS87C51RB2
61h	FFh	Product revision number

### Table 31. Signature Bytes Content



## 10.3. DC Parameters for Standard Voltage

TA = 0°C to +70°C; V<sub>SS</sub> = 0 V; V<sub>CC</sub> = 5 V ± 10%; F = 0 to 40 MHz. TA = -40°C to +85°C; V<sub>SS</sub> = 0 V; V<sub>CC</sub> = 5 V ± 10%; F = 0 to 40 MHz.

#### Table 32. DC Parameters in Standard Voltage

Symbol	Parameter	Min	Тур	Max	Unit	Test Conditions
V <sub>IL</sub>	Input Low Voltage	-0.5		0.2 V <sub>CC</sub> - 0.1	v	
V <sub>IH</sub>	Input High Voltage except XTAL1, RST	0.2 V <sub>CC</sub> + 0.9		V <sub>CC</sub> + 0.5	V	
V <sub>IH1</sub>	Input High Voltage, XTAL1, RST	0.7 V <sub>CC</sub>		V <sub>CC</sub> + 0.5	V	
V <sub>OL</sub>	Output Low Voltage, ports 1, 2, 3, 4, 5 <sup>(6)</sup>			0.3 0.45 1.0	V V V	$\begin{split} I_{OL} &= 100 \; \mu A^{(4)} \\ I_{OL} &= 1.6 \; m A^{(4)} \\ I_{OL} &= 3.5 \; m A^{(4)} \end{split}$
V <sub>OL1</sub>	Output Low Voltage, port 0 <sup>(6)</sup>			0.3 0.45 1.0	V V V	$I_{OL} = 200 \ \mu A^{(4)}$ $I_{OL} = 3.2 \ m A^{(4)}$ $I_{OL} = 7.0 \ m A^{(4)}$
V <sub>OL2</sub>	Output Low Voltage, ALE, PSEN			0.3 0.45 1.0	V V V	$I_{OL} = 100 \ \mu A^{(4)}$ $I_{OL} = 1.6 \ m A^{(4)}$ $I_{OL} = 3.5 \ m A^{(4)}$
V <sub>OH</sub>	Output High Voltage, ports 1, 2, 3, 4, 5	V <sub>CC</sub> - 0.3 V <sub>CC</sub> - 0.7 V <sub>CC</sub> - 1.5			V V V	$I_{OH} = -10 \ \mu A$ $I_{OH} = -30 \ \mu A$ $I_{OH} = -60 \ \mu A$ $V_{CC} = 5 \ V \pm 10\%$
V <sub>OH1</sub>	Output High Voltage, port 0	V <sub>CC</sub> - 0.3 V <sub>CC</sub> - 0.7 V <sub>CC</sub> - 1.5			V V V	$I_{OH} = -200 \ \mu A$ $I_{OH} = -3.2 \ m A$ $I_{OH} = -7.0 \ m A$ $V_{CC} = 5 \ V \pm 10\%$
V <sub>OH2</sub>	Output High Voltage, ALE, PSEN	V <sub>CC</sub> - 0.3 V <sub>CC</sub> - 0.7 V <sub>CC</sub> - 1.5			V V V	$I_{OH} = -100 \ \mu A$ $I_{OH} = -1.6 \ m A$ $I_{OH} = -3.5 \ m A$ $V_{CC} = 5 \ V \pm 10\%$
R <sub>RST</sub>	RST Pulldown Resistor	50	90 <sup>(5)</sup>	200	kΩ	
I <sub>IL</sub>	Logical 0 Input Current ports 1, 2, 3, 4, 5			-50	μΑ	Vin = 0.45 V
I <sub>LI</sub>	Input Leakage Current			±10	μΑ	0.45 V < Vin < V <sub>CC</sub>
I <sub>TL</sub>	Logical 1 to 0 Transition Current, ports 1, 2, 3, 4, 5			-650	μΑ	Vin = 2.0 V
C <sub>IO</sub>	Capacitance of I/O Buffer			10	pF	$Fc = 1 MHz$ $TA = 25^{\circ}C$
I <sub>PD</sub>	Power Down Current		20 <sup>(5)</sup>	50	μΑ	$2.0 \text{ V} < \text{V}_{\text{CC}} < 5.5 \text{ V}^{(3)}$
I <sub>CC</sub> under RESET	Power Supply Current Maximum values, X1 mode: <sup>(7)</sup>			1 + 0.4 Freq (MHz) @12MHz 5.8 @16MHz 7.4	mA	$V_{CC} = 5.5 V^{(1)}$



## **10.5.4. External Data Memory Characteristics**

Parameter
RD Pulse Width
WR Pulse Width
RD to Valid Data In
Data Hold After RD
Data Float After RD
ALE to Valid Data In
Address to Valid Data In
ALE to $\overline{WR}$ or $\overline{RD}$
Address to $\overline{WR}$ or $\overline{RD}$
Data Valid to $\overline{WR}$ Transition
Data set-up to WR High
Data Hold After $\overline{WR}$
RD Low to Address Float
RD or WR High to ALE high

### Table 39. Symbol Description

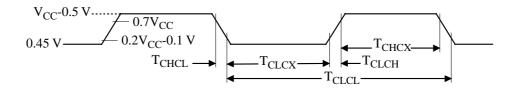


## **10.5.11. External Clock Drive Characteristics (XTAL1)**

Table	46.	AC	Parameters
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Symbol	Parameter	Min	Max	Units
T <sub>CLCL</sub>	Oscillator Period	25		ns
T <sub>CHCX</sub>	High Time	5		ns
T <sub>CLCX</sub>	Low Time	5		ns
T <sub>CLCH</sub>	Rise Time		5	ns
T <sub>CHCL</sub>	Fall Time		5	ns
T <sub>CHCX</sub> /T <sub>CLCX</sub>	Cyclic ratio in X2 mode	40	60	%

### 10.5.12. External Clock Drive Waveforms



#### Figure 30. External Clock Drive Waveforms

#### 10.5.13. AC Testing Input/Output Waveforms

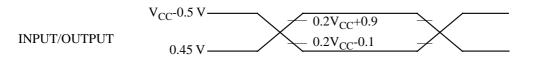


Figure 31. AC Testing Input/Output Waveforms

AC inputs during testing are driven at  $V_{CC}$  - 0.5 for a logic "1" and 0.45V for a logic "0". Timing measurement are made at  $V_{IH}$  min for a logic "1" and  $V_{IL}$  max for a logic "0".

#### **10.5.14. Float Waveforms**

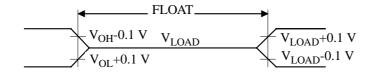


Figure 32. Float Waveforms