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#### Details

Product Status	Active
Core Processor	AVR
Core Size	8-Bit
Speed	16MHz
Connectivity	I <sup>2</sup> C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	32
Program Memory Size	32KB (16K x 16)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/atmega32a-aur

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

# 2. Configuration Summary

Features	ATmega32A
Pin count	44
Flash (KB)	32
SRAM (KB)	2
EEPROM (KB)	1
General Purpose I/O pins	32
SPI	1
TWI (I <sup>2</sup> C)	1
USART	1
ADC	10-bit, up to 76.9ksps (15ksps at max resolution)
ADC channels	8
AC propagation delay	Typ 400ns
8-bit Timer/Counters	2
16-bit Timer/Counters	1
PWM channels	4
RC Oscillator	+/-3%
VREF Bandgap	
Operating voltage	2.7 - 5.5V
Max operating frequency	16MHz
Temperature range	-55°C to +125°C
JTAG	Yes



as an address pointer for look up tables in Flash Program memory. These added function registers are the 16-bit X-, Y-, and Z-register, described later in this section.

The ALU supports arithmetic and logic operations between registers or between a constant and a register. Single register operations can also be executed in the ALU. After an arithmetic operation, the Status Register is updated to reflect information about the result of the operation.

The Program flow is provided by conditional and unconditional jump and call instructions, able to directly address the whole address space. Most AVR instructions have a single 16-bit word format. Every Program memory address contains a 16- or 32-bit instruction.

Program Flash memory space is divided in two sections, the Boot program section and the Application program section. Both sections have dedicated Lock Bits for write and read/write protection. The SPM instruction that writes into the Application Flash memory section must reside in the Boot program section.

During interrupts and subroutine calls, the return address Program Counter (PC) is stored on the Stack. The Stack is effectively allocated in the general data SRAM, and consequently the Stack size is only limited by the total SRAM size and the usage of the SRAM. All user programs must initialize the SP in the reset routine (before subroutines or interrupts are executed). The Stack Pointer SP is read/write accessible in the I/O space. The data SRAM can easily be accessed through the five different addressing modes supported in the AVR architecture.

The memory spaces in the AVR architecture are all linear and regular memory maps.

A flexible interrupt module has its control registers in the I/O space with an additional global interrupt enable bit in the Status Register. All interrupts have a separate Interrupt Vector in the Interrupt Vector table. The interrupts have priority in accordance with their Interrupt Vector position. The lower the Interrupt Vector address, the higher the priority.

The I/O memory space contains 64 addresses for CPU peripheral functions as Control Registers, SPI, and other I/O functions. The I/O Memory can be accessed directly, or as the Data Space locations following those of the Register File, 0x20 - 0x5F.

## 10.2. ALU – Arithmetic Logic Unit

The high-performance Atmel AVR ALU operates in direct connection with all the 32 general purpose working registers. Within a single clock cycle, arithmetic operations between general purpose registers or between a register and an immediate are executed. The ALU operations are divided into three main categories – arithmetic, logical, and bit-functions. Some implementations of the architecture also provide a powerful multiplier supporting both signed/unsigned multiplication and fractional format. See the "Instruction Set" section for a detailed description.

## 10.3. Status Register

The Status Register contains information about the result of the most recently executed arithmetic instruction. This information can be used for altering program flow in order to perform conditional operations. Note that the Status Register is updated after all ALU operations, as specified in the Instruction Set Reference. This will in many cases remove the need for using the dedicated compare instructions, resulting in faster and more compact code.

The Status Register is not automatically stored when entering an interrupt routine and restored when returning from an interrupt. This must be handled by software.



# 12.6. External RC Oscillator

For timing insensitive applications, the external RC configuration shown in the figure below can be used. The frequency is roughly estimated by the equation f = 1/(3RC). C should be at least 22pF. By programming the CKOPT Fuse, the user can enable an internal 36pF capacitor between XTAL1 and GND, thereby removing the need for an external capacitor.

#### Figure 12-3. External RC Configuration



The Oscillator can operate in four different modes, each optimized for a specific frequency range. The operating mode is selected by the fuses CKSEL3:0 as shown in the following table.

Table 12-6.	External	RC	Oscillator	Operating	Modes
-------------	----------	----	------------	-----------	-------

CKSEL3:0	Frequency Range (MHz)
0101	0.1 - 0.9
0110	0.9 - 3.0
0111	3.0 - 8.0
1000	8.0 - 12.0

When this Oscillator is selected, start-up times are determined by the SUT Fuses as shown in the table below.

Table 12-7. Start-up Times for the External RC Oscillator Clock Selection

SUT1:0	Start-up Time from Power-down and Power-save	Additional Delay from Reset (V <sub>CC</sub> = 5.0V)	Recommended Usage
00	18 CK	-	BOD enabled
01	18 CK	4.1ms	Fast rising power
10	18 CK	65ms	Slowly rising power
11	6 CK <sup>(1)</sup>	4.1ms	Fast rising power or BOD enabled

**Note:** 1. This option should not be used when operating close to the maximum frequency of the device.

# 12.7. Calibrated Internal RC Oscillator

The calibrated internal RC Oscillator provides a fixed 1.0, 2.0, 4.0, or 8.0MHz clock. All frequencies are nominal values at 5V and 25°C. This clock may be selected as the system clock by programming the CKSEL Fuses as shown in the following table. If selected, it will operate with no external components.



#### 14.5.1. MCUCSR – MCU Control and Status Register

When using the I/O specific commands IN and OUT, the I/O addresses 0x00 - 0x3F must be used. When addressing I/O Registers as data space using LD and ST instructions, 0x20 must be added to these offset addresses.

The MCU Control and Status Register provides information on which reset source caused an MCU Reset.

Name:MCUCSROffset:0x34Reset:0x00Property:When addressing I/O Registers as data space the offset address is 0x54

Bit	7	6	5	4	3	2	1	0
				JTRF	WDRF	BORF	EXTRF	PORF
Access				R/W	R/W	R/W	R/W	R/W
Reset				-	-	-	-	-

#### Bit 4 – JTRF: JTAG Reset Flag

This bit is set if a reset is being caused by a logic one in the JTAG Reset Register selected by the JTAG instruction AVR\_RESET. This bit is reset by a Power-on Reset, or by writing a logic zero to the flag.

#### Bit 3 – WDRF: Watchdog Reset Flag

This bit is set if a Watchdog Reset occurs. The bit is reset by a Power-on Reset, or by writing a logic zero to the flag.

#### Bit 2 – BORF: Brown-out Reset Flag

This bit is set if a Brown-out Reset occurs. The bit is reset by a Power-on Reset, or by writing a logic zero to the flag.

#### Bit 1 – EXTRF: External Reset Flag

This bit is set if an External Reset occurs. The bit is reset by a Power-on Reset, or by writing a logic zero to the flag.

#### Bit 0 – PORF: Power-on Reset Flag

This bit is set if a Power-on Reset occurs. The bit is reset only by writing a logic zero to the flag. To make use of the Reset Flags to identify a reset condition, the user should read and then reset the MCUCSR as early as possible in the program. If the register is cleared before another reset occurs, the source of the reset can be found by examining the Reset Flags.



 $t_{pd,max}$  and  $t_{pd,min}$ , a single signal transition on the pin will be delayed between  $\frac{1}{2}$  and  $1-\frac{1}{2}$  system clock period depending upon the time of assertion.

When reading back a software assigned pin value, a *nop* instruction must be inserted as indicated in the figure below. The *out* instruction sets the "SYNC LATCH" signal at the positive edge of the clock. In this case, the delay t<sub>pd</sub> through the synchronizer is 1 system clock period.



Figure 17-4. Synchronization when Reading a Software Assigned Pin Value

The following code example shows how to set port B pins 0 and 1 high, 2 and 3 low, and define the port pins from 4 to 7 as input with pull-ups assigned to port pins 6 and 7. The resulting pin values are read back again, but as previously discussed, a *nop* instruction is included to be able to read back the value recently assigned to some of the pins.

```
Assembly Code Example<sup>(1)</sup>
```

```
:.
; Define pull-ups and set outputs high
; Define directions for port pins
ldi r16,(1<<PB7)|(1<<PB6)|(1<<PB1)|(1<<PB0)
ldi r17,(1<<DDB3)|(1<<DDB2)|(1<<DDB1)|(1<<DDB0)
out PORTB,r16
out DDRB,r17
; Insert nop for synchronization
nop
; Read port pins
in r16,PINB
:.</pre>
```

#### C Code Example<sup>(1)</sup>

```
unsigned char i;
:.
/* Define pull-ups and set outputs high */
/* Define directions for port pins */
PORTB = (1<<PB7) | (1<<PB6) | (1<<PB1) | (1<<PB0);
DDRB = (1<<DDB3) | (1<<DDB2) | (1<<DDB1) | (1<<DDB0);
/* Insert nop for synchronization*/
NOP();
/* Read port pins */
i = PINB;
:.
```



#### 19.11.11. TIMSK – Timer/Counter Interrupt Mask Register

When using the I/O specific commands IN and OUT, the I/O addresses 0x00 - 0x3F must be used. When addressing I/O Registers as data space using LD and ST instructions, 0x20 must be added to these offset addresses.

**Note:** 1. This register contains interrupt control bits for several Timer/Counters, but only Timer1 bits are described in this section. The remaining bits are described in their respective timer sections.

Name:TIMSKOffset:0x39Reset:0x00Property:When addressing I/O Registers as data space the offset address is 0x59

Bit	7	6	5	4	3	2	1	0
			TICIE1	OCIE1A	OCIE1B	TOIE1		
Access			R/W	R/W	R/W	R/W		
Reset			0	0	0	0		

#### Bit 5 – TICIE1: Timer/Counter1, Input Capture Interrupt Enable

When this bit is written to one, and the I-flag in the Status Register is set (interrupts globally enabled), the Timer/Counter1 Input Capture interrupt is enabled. The corresponding Interrupt Vector (see Interrupts) is executed when the ICF1 Flag, located in TIFR, is set.

#### Bit 4 – OCIE1A: Timer/Counter1, Output Compare A Match Interrupt Enable

When this bit is written to one, and the I-flag in the Status Register is set (interrupts globally enabled), the Timer/Counter1 Output Compare A match interrupt is enabled. The corresponding Interrupt Vector (see Interrupts) is executed when the OCF1A Flag, located in TIFR, is set.

#### Bit 3 – OCIE1B: Timer/Counter1, Output Compare B Match Interrupt Enable

When this bit is written to one, and the I-flag in the Status Register is set (interrupts globally enabled), the Timer/Counter1 Output Compare B match interrupt is enabled. The corresponding Interrupt Vector (see Interrupts) is executed when the OCF1B Flag, located in TIFR, is set.

#### Bit 2 – TOIE1: Timer/Counter1, Overflow Interrupt Enable

When this bit is written to one, and the I-flag in the Status Register is set (interrupts globally enabled), the Timer/Counter1 Overflow Interrupt is enabled. The corresponding Interrupt Vector (see Interrupts) is executed when the TOV1 Flag, located in TIFR, is set.



#### 20.2.1. Registers

The Timer/Counter (TCNT2) and Output Compare Register (OCR2) are 8-bit registers. Interrupt request (shorten as Int.Req.) signals are all visible in the Timer Interrupt Flag Register (TIFR). All interrupts are individually masked with the Timer Interrupt Mask Register (TIMSK). TIFR and TIMSK are not shown in the figure since these registers are shared by other timer units.

The Timer/Counter can be clocked internally, via the prescaler, or asynchronously clocked from the TOSC1/2 pins, as detailed later in this section. The asynchronous operation is controlled by the Asynchronous Status Register (ASSR). The Clock Select logic block controls which clock source the Timer/Counter uses to increment (or decrement) its value. The Timer/Counter is inactive when no clock source is selected. The output from the clock select logic is referred to as the timer clock ( $clk_{T2}$ ).

The double buffered Output Compare Register (OCR2) is compared with the Timer/Counter value at all times. The result of the compare can be used by the waveform generator to generate a PWM or variable frequency output on the Output Compare Pin (OC2). For details, see <u>Output Compare Unit</u>. The Compare Match event will also set the Compare Flag (OCF2) which can be used to generate an Output Compare interrupt request.

#### 20.2.2. Definitions

Many register and bit references in this document are written in general form. A lower case "n" replaces the Timer/Counter number, in this case 2. However, when using the register or bit defines in a program, the precise form must be used (i.e., TCNT2 for accessing Timer/Counter2 counter value and so on).

The definitions in the following table are also used extensively throughout the document.

BOTTOM	The counter reaches the BOTTOM when it becomes zero (0x00).
MAX	The counter reaches its MAXimum when it becomes 0xFF (decimal 255).
ТОР	The counter reaches the TOP when it becomes equal to the highest value in the count sequence. The TOP value can be assigned to be the fixed value 0xFF (MAX) or the value stored in the OCR2 Register. The assignment is dependent on the mode of operation.

#### Table 20-1. Definitions

## 20.3. Timer/Counter Clock Sources

The Timer/Counter can be clocked by an internal synchronous or an external asynchronous clock source. The clock source  $clk_{T2}$  is by default equal to the MCU clock,  $clk_{I/O}$ . When the AS2 bit in the ASSR Register is written to logic one, the clock source is taken from the Timer/Counter Oscillator connected to TOSC1 and TOSC2. For details on asynchronous operation, refer to Asynchronous Operation of the Timer/Counter. For details on clock sources and prescaler, refer to Timer/Counter Prescaler.

#### **Related Links**

Timer/Counter0 and Timer/Counter1 Prescalers on page 102

# 20.4. Counter Unit

The main part of the 8-bit Timer/Counter is the programmable bi-directional counter unit. The following figure shows a block diagram of the counter and its surrounding environment.



#### 23.11.2. UCSRA – USART Control and Status Register A

When using the I/O specific commands IN and OUT, the I/O addresses 0x00 - 0x3F must be used. When addressing I/O Registers as data space using LD and ST instructions, 0x20 must be added to these offset addresses.

Name:UCSRAOffset:0x0BReset:0x20Property:When addressing I/O Registers as data space the offset address is 0x2B

Bit	7	6	5	4 3 2 1	5 4 3 2 1	1	0	
	RXC	TXC	UDRE	FE	DOR	PE	U2X	MPCM
Access	R	R/W	R	R	R	R	R/W	R/W
Reset	0	0	1	0	0	0	0	0

#### Bit 7 – RXC: USART Receive Complete

This flag bit is set when there are unread data in the receive buffer and cleared when the receive buffer is empty (i.e. does not contain any unread data). If the Receiver is disabled, the receive buffer will be flushed and consequently the RXC bit will become zero. The RXC Flag can be used to generate a Receive Complete interrupt (see description of the RXCIE bit).

#### Bit 6 – TXC: USART Transmit Complete

This flag bit is set when the entire frame in the Transmit Shift Register has been shifted out and there are no new data currently present in the transmit buffer (UDR). The TXC Flag bit is automatically cleared when a transmit complete interrupt is executed, or it can be cleared by writing a one to its bit location. The TXC Flag can generate a Transmit Complete interrupt (see description of the TXCIE bit).

#### Bit 5 – UDRE: USART Data Register Empty

The UDRE Flag indicates if the transmit buffer (UDR) is ready to receive new data. If UDRE is one, the buffer is empty, and therefore ready to be written. The UDRE Flag can generate a Data Register Empty interrupt (see description of the UDRIE bit).

UDRE is set after a reset to indicate that the Transmitter is ready.

#### Bit 4 – FE: Frame Error

This bit is set if the next character in the receive buffer had a Frame Error when received (i.e., when the first stop bit of the next character in the receive buffer is zero). This bit is valid until the receive buffer (UDR) is read. The FE bit is zero when the stop bit of received data is one. Always set this bit to zero when writing to UCSRA.

#### Bit 3 – DOR: Data OverRun

This bit is set if a Data OverRun condition is detected. A Data OverRun occurs when the receive buffer is full (two characters), it is a new character waiting in the Receive Shift Register, and a new start bit is detected. This bit is valid until the receive buffer (UDR) is read. Always set this bit to zero when writing to UCSRA.

#### Bit 2 – PE: Parity Error

This bit is set if the next character in the receive buffer had a Parity Error when received and the parity checking was enabled at that point (UPM1 = 1). This bit is valid until the receive buffer (UDR) is read. Always set this bit to zero when writing to UCSRA.



#### 23.11.3. UCSRB – USART Control and Status Register B

When using the I/O specific commands IN and OUT, the I/O addresses 0x00 - 0x3F must be used. When addressing I/O Registers as data space using LD and ST instructions, 0x20 must be added to these offset addresses.

Name:UCSRBOffset:0x0AReset:0x00Property:When addressing I/O Registers as data space the offset address is 0x2A

Bit	7	6	5	4	3	3 2		0
[	RXCIE	TXCIE	UDRIE	RXEN	TXEN	UCSZ2	RXB8	TXB8
Access	R/W	R/W	R/W	R/W	R/W	R/W	R	R/W
Reset	0	0	0	0	0	0	0	0

#### Bit 7 – RXCIE: RX Complete Interrupt Enable

Writing this bit to one enables interrupt on the RXC Flag. A USART Receive Complete interrupt will be generated only if the RXCIE bit is written to one, the Global Interrupt Flag in SREG is written to one and the RXC bit in UCSRA is set.

#### Bit 6 – TXCIE: TX Complete Interrupt Enable

Writing this bit to one enables interrupt on the TXC Flag. A USART Transmit Complete interrupt will be generated only if the TXCIE bit is written to one, the Global Interrupt Flag in SREG is written to one and the TXC bit in UCSRA is set.

#### Bit 5 – UDRIE: USART Data Register Empty Interrupt Enable

Writing this bit to one enables interrupt on the UDRE Flag. A Data Register Empty interrupt will be generated only if the UDRIE bit is written to one, the Global Interrupt Flag in SREG is written to one and the UDRE bit in UCSRA is set.

#### Bit 4 – RXEN: Receiver Enable

Writing this bit to one enables the USART Receiver. The Receiver will override normal port operation for the RxD pin when enabled. Disabling the Receiver will flush the receive buffer invalidating the FE, DOR and PE Flags.

#### Bit 3 – TXEN: Transmitter Enable

Writing this bit to one enables the USART Transmitter. The Transmitter will override normal port operation for the TxD pin when enabled. The disabling of the Transmitter (writing TXEN to zero) will not become effective until ongoing and pending transmissions are completed (i.e., when the Transmit Shift Register and Transmit Buffer Register do not contain data to be transmitted). When disabled, the Transmitter will no longer override the TxD port.

#### Bit 2 – UCSZ2: Character Size

The UCSZ2 bits combined with the UCSZ1:0 bit in UCSRC sets the number of data bits (Character Size) in a frame the Receiver and Transmitter use.

#### Bit 1 – RXB8: Receive Data Bit 8

RXB8 is the ninth data bit of the received character when operating with serial frames with nine data bits. Must be read before reading the low bits from UDR.



#### 23.11.5. UBRRL – USART Baud Rate Register Low

When using the I/O specific commands IN and OUT, the I/O addresses 0x00 - 0x3F must be used. When addressing I/O Registers as data space using LD and ST instructions, 0x20 must be added to these offset addresses.

Name:UBRRLOffset:0x09Reset:0x00Property:When addressing I/O Registers as data space the offset address is 0x29

Bit	7	6	5	4	3	2	1	0		
[	UBBR[7:0]									
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		
Reset	0	0	0	0	0	0	0	0		

#### Bits 7:0 – UBBR[7:0]: USART Baud Rate Register

This is a 12-bit register which contains the USART baud rate. The UBRRH contains the four most significant bits, and the UBRRL contains the eight least significant bits of the USART baud rate. Ongoing transmissions by the Transmitter and Receiver will be corrupted if the baud rate is changed. Writing UBRRL will trigger an immediate update of the baud rate prescaler.



#### Figure 24-9. Arbitration Between Two Masters



Note that arbitration is not allowed between:

- A REPEATED START condition and a data bit.
- A STOP condition and a data bit.
- A REPEATED START and a STOP condition.

It is the user software's responsibility to ensure that these illegal arbitration conditions never occur. This implies that in multi-master systems, all data transfers must use the same composition of SLA+R/W and data packets. In other words: All transmissions must contain the same number of data packets, otherwise the result of the arbitration is undefined.

#### 24.6. Using the TWI

The AVR TWI is byte-oriented and interrupt based. Interrupts are issued after all bus events, like reception of a byte or transmission of a START condition. Because the TWI is interrupt-based, the application software is free to carry on other operations during a TWI byte transfer. Note that the TWI Interrupt Enable (TWIE) bit in TWCR together with the Global Interrupt Enable bit in SREG allow the application to decide whether or not assertion of the TWINT Flag should generate an interrupt request. If the TWIE bit is cleared, the application must poll the TWINT Flag in order to detect actions on the TWI bus.

When the TWINT Flag is asserted, the TWI has finished an operation and awaits application response. In this case, the TWI Status Register (TWSR) contains a value indicating the current state of the TWI bus. The application software can then decide how the TWI should behave in the next TWI bus cycle by manipulating the TWCR and TWDR Registers.

The following figure is a simple example of how the application can interface to the TWI hardware. In this example, a Master wishes to transmit a single data byte to a Slave. This description is quite abstract, a more detailed explanation follows later in this section. A simple code example implementing the desired behavior is also presented.



to switch between Slaves, Master Transmitter mode and Master Receiver mode without losing control of the bus.

Status Code	Status of the 2-wire Serial Bus	Application Software Resp	oonse			Next Action Taken by TWI Hardware	
(IWSK) Prescaler Bits	Hardware	To/from TWDR	To TW	CR			
are 0			STA	ѕто	TWIN T	TWE A	
0x08	A START condition has been transmitted	Load SLA+W	0	0	1	х	SLA+W will be transmitted; ACK or NOT ACK will be received
0x10	A repeated START condition has been transmitted	Load SLA+W or Load SLA+R	0 0	0 0	1 1	X X	SLA+W will be transmitted; ACK or NOT ACK will be received SLA+R will be transmitted; Logic will switch to Master Receiver mode
0x18	SLA+W has been transmitted; ACK has been received	Load data byte or No TWDR action or No TWDR action or No TWDR action	0 1 0 1	0 0 1 1	1 1 1	x x x x	Data byte will be transmitted and ACK or NOT ACK will be received Repeated START will be transmitted STOP condition will be transmitted and TWSTO Flag will be reset STOP condition followed by a START condition will be transmitted and TWSTO Flag will be reset
0x20	SLA+W has been transmitted; NOT ACK has been received	Load data byte or No TWDR action or No TWDR action or No TWDR action	0 1 0 1	0 0 1 1	1 1 1	x x x x	Data byte will be transmitted and ACK or NOT ACK will be received Repeated START will be transmitted STOP condition will be transmitted and TWSTO Flag will be reset STOP condition followed by a START condition will be transmitted and TWSTO Flag will be reset
0x28	Data byte has been transmitted; ACK has been received	Load data byte or No TWDR action or No TWDR action or No TWDR action	0 1 0 1	0 0 1 1	1 1 1	x x x x	Data byte will be transmitted and ACK or NOT ACK will be received Repeated START will be transmitted STOP condition will be transmitted and TWSTO Flag will be reset STOP condition followed by a START condition will be transmitted and TWSTO Flag will be reset
0x30	Data byte has been transmitted; NOT ACK has been received	Load data byte or No TWDR action or No TWDR action or No TWDR action	0 1 0 1	0 0 1 1	1 1 1	x x x x	Data byte will be transmitted and ACK or NOT ACK will be received Repeated START will be transmitted STOP condition will be transmitted and TWSTO Flag will be reset STOP condition followed by a START condition will be transmitted and TWSTO Flag will be reset
0x38	Arbitration lost in SLA+W or data bytes	No TWDR action or No TWDR action	0 1	0 0	1 1	X X	2-wire Serial Bus will be released and not addressed Slave mode entered A START condition will be transmitted when the bus becomes free

Table 24-3. Status Codes for Master Transmitter Mode



Figure 26-1. Analog to Digital Converter Block Schematic Operation



The ADC converts an analog input voltage to a 10-bit digital value through successive approximation. The minimum value represents GND and the maximum value represents the voltage on the AREF pin minus 1 LSB. Optionally, AVCC or an internal 2.56V reference voltage may be connected to the AREF pin by writing to the REFSn bits in the ADMUX Register. The internal voltage reference may thus be decoupled by an external capacitor at the AREF pin to improve noise immunity.

The analog input channel and differential gain are selected by writing to the MUX bits in ADMUX. Any of the ADC input pins, as well as GND and a fixed bandgap voltage reference, can be selected as single ended inputs to the ADC. A selection of ADC input pins can be selected as positive and negative inputs to the differential gain amplifier.

If differential channels are selected, the differential gain stage amplifies the voltage difference between the selected input channel pair by the selected gain factor. This amplified value then becomes the analog input to the ADC. If single ended channels are used, the gain amplifier is bypassed altogether.

# Atmel

(writing ADEN in ADCSRA to "0" then to "1"), only extended conversions are performed. The result from the extended conversions will be valid. Refer to Prescaling and Conversion Timing for timing details.

# 26.5. Changing Channel or Reference Selection

The MUXn and REFS1:0 bits in the ADMUX Register are single buffered through a temporary register to which the CPU has random access. This ensures that the channels and reference selection only takes place at a safe point during the conversion. The channel and reference selection is continuously updated until a conversion is started. Once the conversion starts, the channel and reference selection is locked to ensure a sufficient sampling time for the ADC. Continuous updating resumes in the last ADC clock cycle before the conversion completes (ADIF in ADCSRA is set). Note that the conversion starts on the following rising ADC clock edge after ADSC is written. The user is thus advised not to write new channel or reference selection values to ADMUX until one ADC clock cycle after ADSC is written.

If Auto Triggering is used, the exact time of the triggering event can be indeterministic. Special care must be taken when updating the ADMUX Register, in order to control which conversion will be affected by the new settings.

If both ADATE and ADEN is written to one, an interrupt event can occur at any time. If the ADMUX Register is changed in this period, the user cannot tell if the next conversion is based on the old or the new settings. ADMUX can be safely updated in the following ways:

- 1. When ADATE or ADEN is cleared.
- 2. During conversion, minimum one ADC clock cycle after the trigger event.
- 3. After a conversion, before the interrupt flag used as trigger source is cleared.

When updating ADMUX in one of these conditions, the new settings will affect the next ADC conversion.

Special care should be taken when changing differential channels. Once a differential channel has been selected, the gain stage may take as much as 125µs to stabilize to the new value. Thus conversions should not be started within the first 125µs after selecting a new differential channel. Alternatively, conversion results obtained within this period should be discarded.

The same settling time should be observed for the first differential conversion after changing ADC reference (by changing the REFS1:0 bits in ADMUX).

## 26.5.1. ADC Input Channels

When changing channel selections, the user should observe the following guidelines to ensure that the correct channel is selected:

- In Single Conversion mode, always select the channel before starting the conversion. The channel selection may be changed one ADC clock cycle after writing one to ADSC. However, the simplest method is to wait for the conversion to complete before changing the channel selection.
- In Free Running mode, always select the channel before starting the first conversion. The channel selection may be changed one ADC clock cycle after writing one to ADSC. However, the simplest method is to wait for the first conversion to complete, and then change the channel selection. Since the next conversion has already started automatically, the next result will reflect the previous channel selection. Subsequent conversions will reflect the new channel selection.

When switching to a differential gain channel, the first conversion result may have a poor accuracy due to the required settling time for the automatic offset cancellation circuitry. The user should preferably disregard the first conversion result.



Figure 27-2. TAP Controller State Diagram



## 27.4. TAP Controller

The TAP controller is a 16-state finite state machine that controls the operation of the Boundary-scan circuitry, JTAG programming circuitry, or On-chip Debug system. The state transitions depicted in Figure 27-2 depend on the signal present on TMS (shown adjacent to each state transition) at the time of the rising edge at TCK. The initial state after a Power-on Reset is Test-Logic-Reset.

As a definition in this document, the LSB is shifted in and out first for all Shift Registers.

Assuming Run-Test/Idle is the present state, a typical scenario for using the JTAG interface is:

• At the TMS input, apply the sequence 1, 1, 0, 0 at the rising edges of TCK to enter the Shift Instruction Register – Shift-IR state. While in this state, shift the 4 bits of the JTAG instructions into the JTAG instruction register from the TDI input at the rising edge of TCK. The TMS input must be



Table 27-5.	Boundar	y-scan 🗄	Signals	for the	ADC
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Signal Name	Direction as Seen from the ADC	Description	Recommended Input when not in Use	Output Values when Recommended Inputs are Used, and CPU is not Using the ADC
COMP	Output	Comparator Output	0	0
ACLK	Input	Clock signal to gain stages implemented as Switch-cap filters	0	0
ACTEN	Input	Enable path from gain stages to the comparator	0	0
ADCBGEN	Input	Enable Band-gap reference as negative input to comparator	0	0
ADCEN	Input	Power-on signal to the ADC	0	0
AMPEN	Input	Power-on signal to the gain stages	0	0
DAC_9	Input	Bit 9 of digital value to DAC	1	1
DAC_8	Input	Bit 8 of digital value to DAC	0	0
DAC_7	Input	Bit 7 of digital value to DAC	0	0
DAC_6	Input	Bit 6 of digital value to DAC	0	0
DAC_5	Input	Bit 5 of digital value to DAC	0	0
DAC_4	Input	Bit 4 of digital value to DAC	0	0
DAC_3	Input	Bit 3 of digital value to DAC	0	0
DAC_2	Input	Bit 2 of digital value to DAC	0	0
DAC_1	Input	Bit 1 of digital value to DAC	0	0
DAC_0	Input	Bit 0 of digital value to DAC	0	0
EXTCH	Input	Connect ADC channels 0 - 3 to by-pass path around gain stages	1	1
G10	Input	Enable 10x gain	0	0
G20	Input	Enable 20x gain	0	0
GNDEN	Input	Ground the negative input to comparator when true	0	0



- To protect only the Application Flash section from a software update by the MCU
- Allow software update in the entire Flash

The Boot Lock bits can be set in software and in Serial or Parallel Programming mode, but they can be cleared by a Chip Erase command only. The general Write Lock (Lock Bit mode 2) does not control the programming of the Flash memory by SPM instruction. Similarly, the general Read/Write Lock (Lock Bit mode 1) does not control reading nor writing by LPM/SPM, if it is attempted.

BLB0 Mode	BLB02	BLB01	Protection
1	1	1	No restrictions for SPM or LPM accessing the Application section.
2	1	0	SPM is not allowed to write to the Application section.
3	0	0	SPM is not allowed to write to the Application section, and LPM executing from the Boot Loader section is not allowed to read from the Application section. If Interrupt Vectors are placed in the Boot Loader section, interrupts are disabled while executing from the Application section.
4	0	1	LPM executing from the Boot Loader section is not allowed to read from the Application section. If Interrupt Vectors are placed in the Boot Loader section, interrupts are disabled while executing from the Application section.

Table 28-2. Boot Lock Bit0 Protection Modes (Application Section)

**Note:** "1" means unprogrammed, "0" means programmed.

Table 28-3.	Boot Lock Bit1	<b>Protection Modes</b>	(Boot Loader Section	on)
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BLB1 Mode	BLB12	BLB11	Protection
1	1	1	No restrictions for SPM or LPM accessing the Boot Loader section.
2	1	0	SPM is not allowed to write to the Boot Loader section.
3	0	0	SPM is not allowed to write to the Boot Loader section, and LPM executing from the Application section is not allowed to read from the Boot Loader section. If Interrupt Vectors are placed in the Application section, interrupts are disabled while executing from the Boot Loader section.
4	0	1	LPM executing from the Application section is not allowed to read from the Boot Loader section. If Interrupt Vectors are placed in the Application section, interrupts are disabled while executing from the Boot Loader section.

Note: "1" means unprogrammed, "0" means programmed.

# 28.6. Entering the Boot Loader Program

Entering the Boot Loader takes place by a jump or call from the application program. This may be initiated by a trigger such as a command received via USART, or SPI interface. Alternatively, the Boot Reset Fuse can be programmed so that the Reset Vector is pointing to the Boot Flash start address after a reset. In this case, the Boot Loader is started after a reset. After the application code is loaded, the program can start executing the application code. The fuses cannot be changed by the MCU itself. This means that



Figure 29-10. Serial Programming and Verify<sup>(1)</sup>



#### Note:

- 1. If the device is clocked by the Internal Oscillator, it is no need to connect a clock source to the XTAL1 pin.
- 2.  $V_{CC}$  0.3 < AV<sub>CC</sub> < V<sub>CC</sub> + 0.3, however, AV<sub>CC</sub> should always be within 2.7 5.5V.

When programming the EEPROM, an auto-erase cycle is built into the self-timed programming operation (in the Serial mode ONLY) and there is no need to first execute the Chip Erase instruction. The Chip Erase operation turns the content of every memory location in both the Program and EEPROM arrays into 0xFF.

Depending on CKSEL Fuses, a valid clock must be present. The minimum low and high periods for the Serial Clock (SCK) input are defined as follows:

Low: > 2 CPU clock cycles for  $f_{ck}$  < 12MHz, 3 CPU clock cycles for  $f_{ck}$  ≥ 12MHz

High: > 2 CPU clock cycles for  $f_{ck}$  < 12MHz, 3 CPU clock cycles for  $f_{ck} \ge$  12MHz

#### 29.9.1. Serial Programming Algorithm

When writing serial data to the ATmega32A, data is clocked on the rising edge of SCK.

When reading data from the ATmega32A, data is clocked on the falling edge of SCK. Please refer to the figure, Figure 29-11 in SPI Serial Programming Characteristics section for timing details.

To program and verify the ATmega32A in the serial programming mode, the following sequence is recommended (See Serial Programming Instruction set in Table 29-16 Serial Programming Waveforms:

1. Power-up sequence:

Apply power between  $V_{CC}$  and GND while RESET and SCK are set to "0". In some systems, the programmer can not guarantee that SCK is held low during power-up. In this case, RESET must be given a positive pulse of at least two CPU clock cycles duration after SCK has been set to "0".



Instruction	TDI sequence	TDO sequence	Notes
2g. Write Flash Page	0110111_00000000 0110101_00000000	xxxxxxx_xxxxxxx xxxxxxx_xxxxxxx	(1)
	0110111_00000000	xxxxxxx_xxxxxxx	
	0110111_00000000	xxxxxxx_xxxxxxx	
2h. Poll for Page Write complete	0110111_00000000	xxxxxox_xxxxxxx	(2)
3a. Enter Flash Read	0100011_00000010	xxxxxxx_xxxxxxx	
3b. Load Address High Byte	0000111_aaaaaaaa	xxxxxxx_xxxxxxx	(9)
3c. Load Address Low Byte	0000011_bbbbbbbb	xxxxxxx_xxxxxxx	
3d. Read Data Low and High Byte	0110010_0000000 0110110_00000000	xxxxxxx_xxxxxxx xxxxxxx_00000000	low byte high byte
	0110111_00000000	xxxxxxx_00000000	<b>3 • 9 • •</b>
4a. Enter EEPROM Write	0100011_00010001	xxxxxxx_xxxxxxx	
4b. Load Address High Byte	0000111_aaaaaaaa	xxxxxxx_xxxxxxx	(9)
4c. Load Address Low Byte	0000011_bbbbbbbb	xxxxxxx_xxxxxxx	
4d. Load Data Byte	0010011_iiiiiiiii	xxxxxxx_xxxxxxx	
4e. Latch Data	0110111_00000000	xxxxxxx_xxxxxxx	(1)
	1110111_00000000	xxxxxxx_xxxxxxx	
	0110111_00000000	XXXXXXXX_XXXXXXXX	
4f. Write EEPROM Page	0110011_00000000 0110001_00000000	xxxxxxx_xxxxxxxxxxxxxxxxxxxxxxxxxxxxxx	(1)
	0110011 00000000	xxxxxxx xxxxxxx	
	0110011 00000000	-	
4g. Poll for Page Write complete	0110011 00000000		(2)
5a. Enter EEPROM Read	0100011_00000011		
5b. Load Address High Byte	0000111_aaaaaaaa		(9)
5c. Load Address Low Byte	0000011 bbbbbbbb		
5d Read Data Byte	0110011 bbbbbbbb		
	0110010_00000000	xxxxxxx_xxxxxxx	
	0110011_00000000	xxxxxxx_00000000	
6a. Enter Fuse Write	0100011_01000000	xxxxxxx_xxxxxxx	
6b. Load Data Low Byte <sup>(6)</sup>	0010011_iiiiiiiii	xxxxxxx_xxxxxxx	(3)



Symbol	Parameter	Condition	Min	Тур	Max	Units
	Clock Frequency		50		1000	kHz
AV <sub>CC</sub>	Analog Analog Supply Voltage		V <sub>CC</sub> - 0.3 <sup>(1)</sup>		V <sub>CC</sub> + 0.3 <sup>(2)</sup>	V
$V_{REF}$	Reference Voltage		2.0		AV <sub>CC</sub>	V
V <sub>IN</sub>	Input voltage		GND		V <sub>REF</sub>	V
	ADC conversion output		0		1023	LSB
	Input bandwidth			38.5		kHz
V <sub>INT</sub>	Internal Voltage Reference		2.3	2.56	2.7	V
R <sub>REF</sub>	Reference Input Resistance			32		kΩ
R <sub>AIN</sub>	Analog Input Resistance			100		MΩ

#### Note:

- 1. Minimum for  $AV_{CC}$  is 2.7V.
- 2. Maximum for AV<sub>CC</sub> is 5.5V.

## Table 30-9. ADC Characteristics

Symbol	Parameter	Condition	Min	Тур	Max	Units
	Resolution	Gain = 1x			10	Bits
		Gain = 10x			10	Bits
		Gain = 200x			10	Bits
	Absolute Accuracy	Gain = 1x		17		LSB
		$V_{REF}$ = 4V, $V_{CC}$ = 5V				
		ADC clock = 50 - 200kHz				
		Gain = 10x		16		LSB
		$V_{REF}$ = 4V, $V_{CC}$ = 5V				
		ADC clock = 50 - 200kHz				
		Gain = 200x		7		LSB
		$V_{REF}$ = 4V, $V_{CC}$ = 5V				
		ADC clock = 50 - 200kHz				

