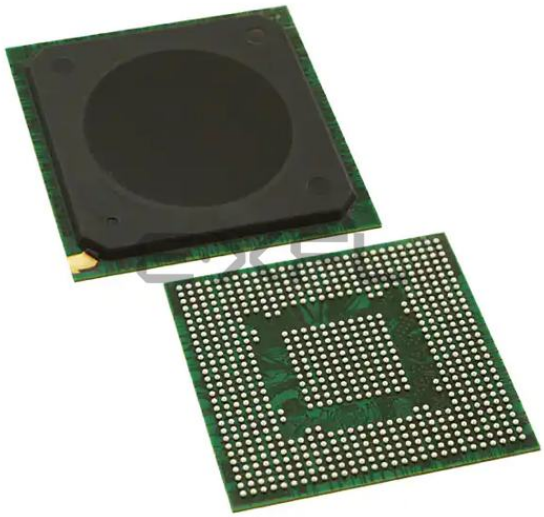

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300c4s
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	400MHz
Co-Processors/DSP	Security; SEC 3.0
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	SATA 3Gbps (2)
USB	USB 2.0 + PHY (1)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	-40°C ~ 125°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	689-BBGA Exposed Pad
Supplier Device Package	689-TEPBGA II (31x31)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8377ecvragd

Table 2. Absolute Maximum Ratings[1] (continued)

Characteristic		Symbol	Max Value	Unit	Note
Input voltage	DDR DRAM signals	MV_{IN}	–0.3 to (GV_{DD} + 0.3)	V	2, 4
	DDR DRAM reference	MV_{REF}	–0.3 to (GV_{DD} + 0.3)	V	2, 4
	Three-speed Ethernet signals	LV_{IN}	–0.3 to (LV_{DD} + 0.3)	V	—
	PCI, DUART, CLKIN, system control and power management, I^2C, and JTAG signals	OV_{IN}	–0.3 to (OV_{DD} + 0.3)	V	3, 4, 5
	Local Bus	LB_{IN}	–0.3 to (LBV_{DD} + 0.3)	V	—
Storage temperature range		T_{STG}	–55 to 150	°C	—

Notes:
1. Functional and tested operating conditions are given in Table 3. Absolute maximum ratings are stress ratings only, and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect device reliability or cause permanent damage to the device.
2. **Caution:** MV_{IN} must not exceed GV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
3. **Caution:** OV_{IN} must not exceed OV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
4. (M,O)V_{IN} and MV_{REF} may overshoot/undershoot to a voltage and for a maximum duration as shown in Figure 2.
5. Overshoot/undershoot by OV_{IN} on the PCI interface does not comply to the *PCI Electrical Specification* for 3.3-V operation, as shown in Figure 2.
6. L[1,2]_$n$$V_{DD}$ includes $SDAV_{DD\_0}$, $XCOREV_{DD}$, and $XPADV_{DD}$ power inputs.

2.1.2 Power Supply Voltage Specification

This table provides recommended operating conditions for the device. Note that the values in this table are the recommended and tested operating conditions. Proper device operation outside of these conditions is not guaranteed.

Table 3. Recommended Operating Conditions

Characteristic		Symbol	Recommended Value	Unit	Note
Core supply voltage	up to 667 MHz	V_{DD}	1.0 ± 50 mV	V	1
	800 MHz		1.05 ± 50 mV	V	1
PLL supply voltage (e300 core, eLBC and system)	up to 667 MHz	AV_{DD}	1.0 ± 50 mV	V	1, 2
	800 MHz		1.05 ± 50 mV	V	1, 2
DDR1 and DDR2 DRAM I/O voltage		GV_{DD}	2.5 V ± 125 mV 1.8 V ± 90 mV	V	1
Three-speed Ethernet I/O, MII management voltage		LV_{DD}[1,2]	3.3 V ± 165 mV 2.5 V ± 125 mV	V	—
PCI, local bus, DUART, system control and power management, I^2C, and JTAG I/O voltage		OV_{DD}	3.3 V ± 165 mV	V	1
Local Bus		LBV_{DD}	1.8 V ± 90 mV 2.5 V ± 125 mV 3.3 V ± 165 mV	V	—

6.2 DDR1 and DDR2 SDRAM AC Electrical Characteristics

This section provides the AC electrical characteristics for the DDR SDRAM interface.

6.2.1 DDR1 and DDR2 SDRAM Input AC Timing Specifications

This table provides the input AC timing specifications for the DDR2 SDRAM when GVDD(typ) = 1.8 V.

Table 18. DDR2 SDRAM Input AC Timing Specifications for 1.8-V Interface

Parameter	Symbol	Min	Max	Unit
AC input low voltage	V_{IL}	—	MV_{REF} – 0.25	V
AC input high voltage	V_{IH}	MV_{REF} + 0.25	—	V

This table provides the input AC timing specifications for the DDR1 SDRAM when GV_{DD}(typ) = 2.5 V.

Table 19. DDR1 SDRAM Input AC Timing Specifications for 2.5-V Interface

Parameter	Symbol	Min	Max	Unit
AC input low voltage	V_{IL}	—	MV_{REF} – 0.31	V
AC input high voltage	V_{IH}	MV_{REF} + 0.31	—	V

This table provides the input AC timing specifications for the DDR1 and DDR2 SDRAM interface.

Table 20. DDR1 and DDR2 SDRAM Input AC Timing Specifications

Parameter	Symbol	Min	Max	Unit	Note
Controller skew for MDQS-MDQ/MECC/MDM	t_{CISKEW}			ps	1, 2
400 MHz data rate		–500	500		3
333 MHz data rate		–750	750		—
266 MHz data rate		–750	750		—

Note:

1. t_{CISKEW} represents the total amount of skew consumed by the controller between MDQS*n* and any corresponding bit that will be captured with MDQS*n*. This should be subtracted from the total timing budget.
2. The amount of skew that can be tolerated from MDQS to a corresponding MDQ signal is called t_{DISKEW}. This can be determined by the following equation: $t_{DISKEW} = \pm[T/4 - |t_{CISKEW}|]$ where T is the MCK clock period and $|t_{CISKEW}|$ is the absolute value of t_{CISKEW}.
3. This specification applies only to DDR2 interface.

Due to the special implementation of the eSDHC, there are constraints regarding the clock and data signals propagation delay on the user board. The constraints are for minimum and maximum delays, as well as skew between the CLK and DAT/CMD signals.

In full speed mode, there is no need to add special delay on the data or clock signals. The user should make sure to meet the timing requirements as described further within this document.

If the system is designed to support both high-speed and full-speed cards, the high-speed constraints should be fulfilled. If the systems is designed to operate up to 25 MHz only, full-speed mode is recommended.

11.1 eSDHC DC Electrical Characteristics

This table provides the DC electrical characteristics for the eSDHC (SD/MMC) interface of the device.

Table 41. eSDHC interface DC Electrical Characteristics

Parameter	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}	—	$0.625 \times OV_{DD}$	$OV_{DD} + 0.3$	V
Input low voltage	V_{IL}	—	–0.3	$0.25 \times OV_{DD}$	V
Input current	I_{IN}	—	—	±30	μA
Output high voltage	V_{OH}	I_{OH} = –100 uA, at OV_{DD}(min)	$0.75 \times OV_{DD}$	—	V
Output low voltage	V_{OL}	I_{OL} = +100 uA, at OV_{DD}(min)	—	$0.125 \times OV_{DD}$	V

11.2 eSDHC AC Timing Specifications (Full-Speed Mode)

This section describes the AC electrical specifications for the eSDHC (SD/MMC) interface of the device. This table provides the eSDHC AC timing specifications for full-speed mode as defined in Figure 27 and Figure 28.

Table 42. eSDHC AC Timing Specifications for Full-Speed Mode

At recommended operating conditions OV_{DD} = 3.3 V ± 165 mV.

Parameter	Symbol[1]	Min	Max	Unit	Note
SD_CLK clock frequency—full speed mode	f_{SFSCK}	0	25	MHz	—
SD_CLK clock cycle	t_{SFSCK}	40	—	ns	—
SD_CLK clock frequency—identification mode	f_{SIDCK}	0	400	KHz	—
SD_CLK clock low time	t_{SFSCKL}	15	—	ns	2
SD_CLK clock high time	t_{SFSCKH}	15	—	ns	2
SD_CLK clock rise and fall times	t_{SFSCKR}/ t_{SFSCKF}	—	5	ns	2
Input setup times: SD_CMD, SD_DATx, SD_CD to SD_CLK	$t_{SFSIVKH}$	5	—	ns	2

11.2.1 Full-Speed Output Path (Write)

This figure provides the data and command output timing diagram.

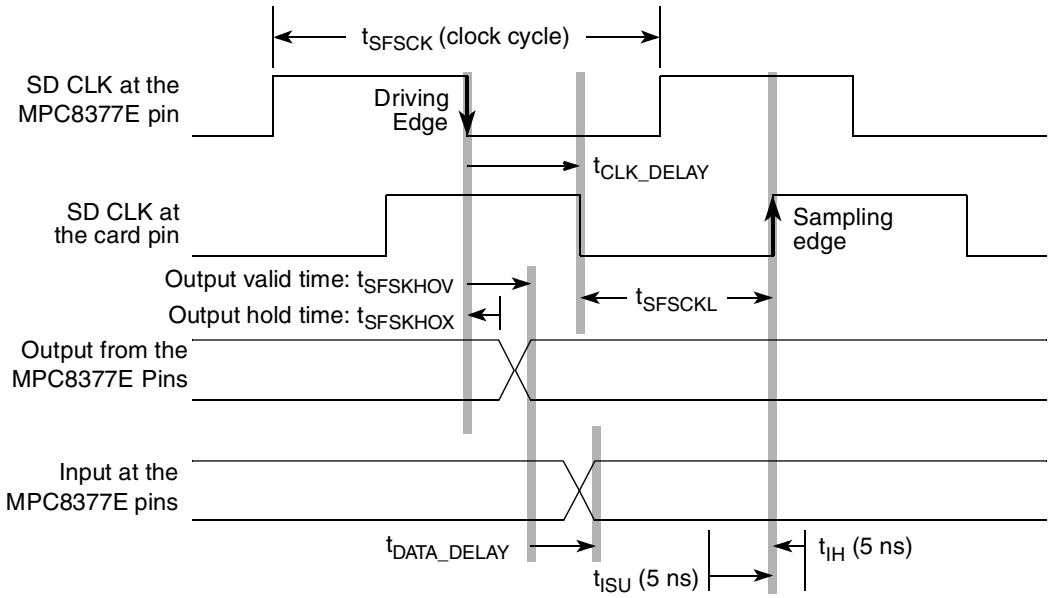


Figure 27. Full Speed Output Path

11.2.1.1 Full-Speed Write Meeting Setup (Maximum Delay)

The following equations show how to calculate the allowed skew range between the SD_CLK and SD_DAT/CMD signals on the PCB.

No clock delay:

$$t_{SFSKHOV} + t_{DATA\_DELAY} + t_{ISU} < t_{SFSCKL} \quad \textit{Eqn. 1}$$

With clock delay:

$$t_{SFSKHOV} + t_{DATA\_DELAY} + t_{ISU} < t_{SFSCKL} + t_{CLK\_DELAY} \quad \textit{Eqn. 2}$$

$$t_{DATA\_DELAY} + t_{SFSCKL} < t_{SFSCK} + t_{CLK\_DELAY} - t_{ISU} - t_{SFSKHOV} \quad \textit{Eqn. 3}$$

This means that data can be delayed versus clock up to 11 ns in ideal case of t_{SFSCKL} = 20 ns:

$t_{DATA\_DELAY} + 20 < 40 + t_{CLK\_DELAY} - 5 - 4$

$t_{DATA\_DELAY} < 11 + t_{CLK\_DELAY}$

11.2.1.2 Full-Speed Write Meeting Hold (Minimum Delay)

The following equations show how to calculate the allowed skew range between the SD_CLK and SD_DAT/CMD signals on the PCB.

$$t_{CLK\_DELAY} < t_{SFSCKL} + t_{SFSKHOX} + t_{DATA\_DELAY} - t_{IH} \quad \textit{Eqn. 4}$$

12.1 JTAG DC Electrical Characteristics

This table provides the DC electrical characteristics for the IEEE 1149.1 (JTAG) interface of the chip.

Table 44. JTAG interface DC Electrical Characteristics

Parameter	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}	—	2.5	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	—	—	±30	μA
Output high voltage	V_{OH}	I_{OH} = –8.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 8.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

12.2 JTAG AC Timing Specifications

This section describes the AC electrical specifications for the IEEE 1149.1 (JTAG) interface of the device. This table provides the JTAG AC timing specifications as defined in Figure 33 through Figure 36.

Table 45. JTAG AC Timing Specifications (Independent of CLKIN) [1]

Parameter	Symbol[2]	Min	Max	Unit	Note
JTAG external clock frequency of operation	f_{JTG}	0	33.3	MHz	—
JTAG external clock cycle time	t_{JTG}	30	—	ns	—
JTAG external clock pulse width measured at 1.4 V	t_{JTKHKL}	15	—	ns	—
JTAG external clock rise and fall times	t_{JTGR} & t_{JTGF}	0	2	ns	—
$\overline{TRST}$ assert time	t_{TRST}	25	—	ns	3
Input setup times: Boundary-scan data TMS, TDI	 t_{JTDVKH} t_{JTIVKH}	 4 4	 — —	ns	4
Input hold times: Boundary-scan data TMS, TDI	 t_{JTDXKH} t_{JTIXKH}	 10 10	 — —	ns	4
Valid times: Boundary-scan data TDO	 t_{JTKLDV} t_{JTKLOV}	 2 2	 11 11	ns	—
Output hold times: Boundary-scan data TDO	 t_{JTKLDX} t_{JTKLOX}	 2 2	 — —	ns	—

Table 49. PCI AC Timing Specifications at 66 MHz (continued)

PCI_SYNC_IN clock input levels are with next levels: VIL = $0.1 \times OV_{DD}$, VIH = $0.7 \times OV_{DD}$.

Parameter	Symbol[1]	Min	Max	Unit	Note
Input hold from cock	t_{PCIXKH}	0.25	—	ns	2, 4, 6
Output clock skew	t_{PCKOSK}	—	0.5	ns	5

Notes:

1. Note that the symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{PCIVKH} symbolizes PCI timing (PC) with respect to the time the input signals (I) reach the valid state (V) relative to the PCI_SYNC_IN clock, t_{SYS}, reference (K) going to the high (H) state or setup time. Also, t_{PCRHFV} symbolizes PCI timing (PC) with respect to the time hard reset (R) went high (H) relative to the frame signal (F) going to the valid (V) state.
2. See the timing measurement conditions in the *PCI 2.3 Local Bus Specifications*.
3. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
4. Input timings are measured at the pin.
5. PCI specifications allows 1 ns skew for 66 MHz but includes the total allowed skew, board, connectors, etc.
6. Value does not comply with the *PCI 2.3 Local Bus Specifications*.

This table shows the PCI AC timing specifications at 33 MHz.

Table 50. PCI AC Timing Specifications at 33 MHz

PCI_SYNC_IN clock input levels are with next levels: VIL = $0.1 \times OV_{DD}$, $V_{IH} = 0.7 \times OV_{DD}$.

Parameter	Symbol[1]	Min	Max	Unit	Note
Clock to output valid	t_{PCKHOV}	—	11	ns	2
Output hold from clock	t_{PCKHOX}	2	—	ns	2
Clock to output high impedance	t_{PCKHOZ}	—	14	ns	2, 3
Input setup to clock	t_{PCIVKH}	3.0	—	ns	2, 4
Input hold from clock	t_{PCIXKH}	0.25	—	ns	2, 4, 6
Output clock skew	t_{PCKOSK}	—	0.5	ns	5

Notes:

1. Note that the symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{PCIVKH} symbolizes PCI timing (PC) with respect to the time the input signals (I) reach the valid state (V) relative to the PCI_SYNC_IN clock, t_{SYS}, reference (K) going to the high (H) state or setup time. Also, t_{PCRHFV} symbolizes PCI timing (PC) with respect to the time hard reset (R) went high (H) relative to the frame signal (F) going to the valid (V) state.
2. See the timing measurement conditions in the *PCI 2.3 Local Bus Specifications*.
3. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
4. Input timings are measured at the pin.
5. PCI specifications allows 2 ns skew for 33 MHz but includes the total allowed skew, board, connectors, etc.
6. Value does not comply with the *PCI 2.3 Local Bus Specifications*.

Table 52. Differential Transmitter (Tx) Output Specifications (continued)

Parameter	Conditions	Symbol	Min	Typical	Max	Units	Note
Common mode return loss	Measured over 50 MHz to 1.25 GHz.	RL_{TX-CM}	6	—	—	dB	4
DC differential Tx impedance	Tx DC differential mode low impedance	$Z_{TX-DIFF-DC}$	80	100	120	Ω	—
Transmitter DC impedance	Required Tx D+ as well as D– DC impedance during all states	Z_{TX-DC}	40	—	—	Ω	—
Lane-to-Lane output skew	Static skew between any two transmitter lanes within a single link	$L_{TX-SKEW}$	—	—	500 + 2 UI	ps	—
AC coupling capacitor	All transmitters should be AC coupled. The AC coupling is required either within the media or within the transmitting component itself.	C_{TX}	75	—	200	nF	—
Crosslink random timeout	This random timeout helps resolve conflicts in crosslink configuration by eventually resulting in only one downstream and one upstream port.	$T_{crosslink}$	0	—	1	ms	7

Notes:
1. No test load is necessarily associated with this value.
2. Specified at the measurement point into a timing and voltage compliance test load as shown in Figure 44 and measured over any 250 consecutive Tx UIs. (Also refer to the transmitter compliance eye diagram shown in Figure 42.)
3. A T_{TX-EYE} = 0.70 UI provides for a total sum of deterministic and random jitter budget of $T_{TX-JITTER-MAX}$ = 0.30 UI for the transmitter collected over any 250 consecutive Tx UIs. The $T_{TX-EYE-MEDIAN-to-MAX-JITTER}$ median is less than half of the total Tx jitter budget collected over any 250 consecutive Tx UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value.
4. The transmitter input impedance will result in a differential return loss greater than or equal to 12 dB and a common mode return loss greater than or equal to 6 dB over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements is 50 Ω to ground for both the D+ and D– line (that is, as measured by a vector network analyzer with 50-Ω probes, see Figure 44). Note that the series capacitors, C_{TX}, is optional for the return loss measurement.
5. Measured between 20%–80% at transmitter package pins into a test load as shown in Figure 44 for both V_{TX-D+} and V_{TX-D-}.
6. See Section 4.3.1.8 of the *PCI Express Base Specifications*, Rev 1.0a.
7. See Section 4.2.6.3 of the *PCI Express Base Specifications*, Rev 1.0a.

15.4.2 Transmitter Compliance Eye Diagrams

The Tx eye diagram in Figure 42 is specified using the passive compliance/test measurement load (see Figure 44) in place of any real PCI Express interconnect + Rx component. There are two eye diagrams that must be met for the transmitter. Both diagrams must be aligned in time using the jitter median to locate the center of the eye diagram. The different eye diagrams differ in voltage depending on whether it is a transition bit or a de-emphasized bit. The exact reduced voltage level of the de-emphasized bit is always relative to the transition bit.

The eye diagram must be valid for any 250 consecutive UIs.

A recovered Tx UI is calculated over 3500 consecutive unit intervals of sample data. The eye diagram is created using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the Tx UI.

NOTE

It is recommended that the recovered Tx UI be calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function (that is, least squares and median deviation fits).

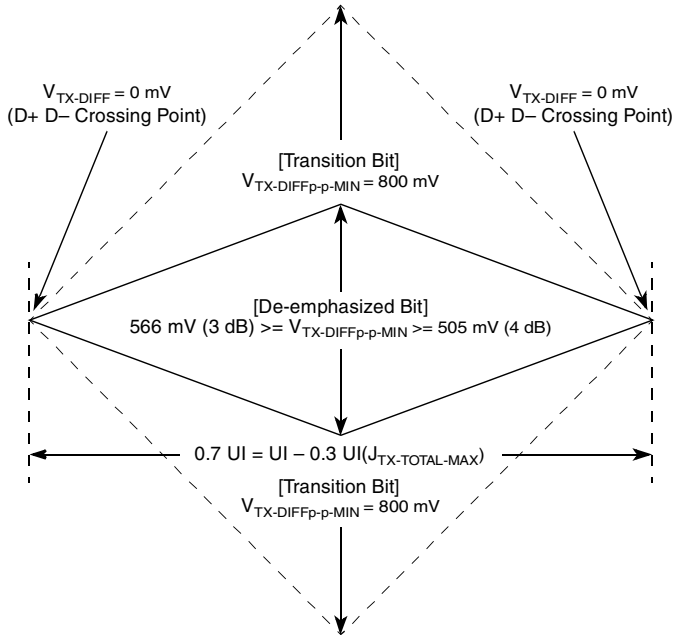


Figure 42. Minimum Transmitter Timing and Voltage Output Compliance Specifications

15.4.3 Differential Receiver (Rx) Input Specifications

This table defines the specifications for the differential input at all receivers. The parameters are specified at the component pins.

Table 53. Differential Receiver (Rx) Input Specifications

Parameter	Comments	Symbol	Min	Typical	Max	Units	Note
Unit interval	Each U_{PERX} is 400 ps ± 300 ppm. U_{PERX} does not account for Spread Spectrum Clock dictated variations.	UI	399.88	400	400.12	ps	1
Differential peak-to-peak output voltage	$V_{PEDPPRX} = 2 \times \lvert V_{RX-D+} - V_{RX-D-} \rvert$	$V_{RX-DIFFp-p}$	0.175	—	1.200	V	2

Table 54. SATA Reference Clock Input Requirements (continued)

Parameter	Condition	Symbol	Min	Typical	Max	Unit	Note
SD_REF_CLK/ $\overline{\text{SD\_REF\_CLK}}$ cycle to cycle Clock jitter (period jitter)	Cycle-to-cycle at ref clock input	t_{CLK_CJ}	—	—	100	ps	—
SD_REF_CLK/ $\overline{\text{SD\_REF\_CLK}}$ total reference clock jitter, phase jitter (peak-peak)	Peak-to-peak jitter at ref clock input	t_{CLK_PJ}	–50	—	+50	ps	2, 3

Notes:
1. Only 100/125/150 MHz have been tested, other in between values will not work correctly with the rest of the system.
2. In a frequency band from 150 kHz to 15 MHz at BER of 10^{-12}.
3. Total peak to peak Deterministic Jitter "D_J" should be less than or equal to 50 ps.

This figure shows the SATA reference clock timing waveform.

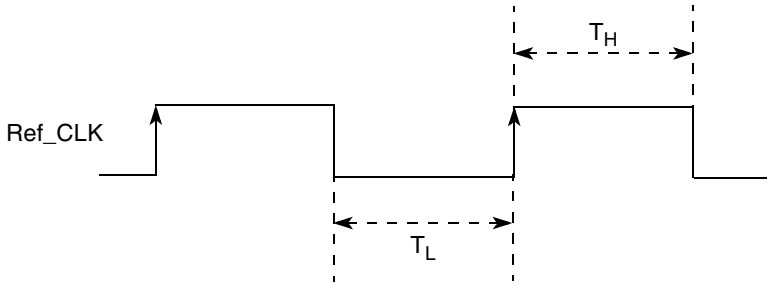


Figure 45. SATA Reference Clock Timing Waveform

16.2 Transmitter (Tx) Output Characteristics

This section discusses the Gen1i/1.5G and Gen2i/3G transmitter output characteristics for the SATA interface.

16.2.1 Gen1i/1.5G Transmitter Specifications

This table provides the DC differential transmitter output DC characteristics for the SATA interface at Gen1i or 1.5 Gbits/s transmission.

Table 55. Gen1i/1.5G Transmitter (Tx) DC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Note
Tx differential output voltage	V_{SATA_TXDIFF}	400	500	600	mV_{p-p}	1
Tx differential pair impedance	$Z_{SATA\_TXDIFFIM}$	85	100	115	Ω	—

Note:
1. Terminated by 50 Ω load.

This table provides the differential transmitter output AC characteristics for the SATA interface at Gen1i or 1.5 Gbits/s transmission.

Table 56. Gen1i/1.5G Transmitter AC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Note
Channel speed	t_{CH_SPEED}	—	1.5	—	Gbps	—
Unit interval	T_{UI}	666.4333	666.667	670.2333	ps	—
Total jitter, data-data 5 UI	$U_{SATA\_TXTJ5UI}$	—	—	0.355	UI_{p-p}	1
Total jitter, data-data 250 UI	$U_{SATA\_TXTJ250UI}$	—	—	0.47	UI_{p-p}	1
Deterministic jitter, data-data 5 UI	$U_{SATA\_TXDJ5UI}$	—	—	0.175	UI_{p-p}	1
Deterministic jitter, data-data 250 UI	$U_{SATA\_TXDJ250UI}$	—	—	0.22	UI_{p-p}	1

Note:
1. Measured at Tx output pins peak to peak phase variation, random data pattern.

16.2.2 Gen2i/3G Transmitter Specifications

This table provides the differential transmitter output DC characteristics for the SATA interface at Gen2i or 3.0 Gbits/s transmission.

Table 57. Gen 2i/3G Transmitter DC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Note
Tx differential output voltage	V_{SATA_TXDIFF}	400	550	700	mV_{p-p}	1
Tx differential pair impedance	$Z_{SATA\_TXDIFFIM}$	85	100	115	Ω	—

Note:
1. Terminated by 50 Ω load.

This table provides the differential transmitter output AC characteristics for the SATA interface at Gen2i or 3.0 Gbits/s transmission.

Table 58. Gen 2i/3G Transmitter AC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Note
Channel speed	t_{CH_SPEED}	—	3.0	—	Gbps	—
Unit interval	T_{UI}	333.2	333.33	335.11	ps	—
Total jitter $f_{C3dB}=f_{BAUD}/10$	$U_{SATA\_TXTJfB/10}$	—	—	0.3	UI_{p-p}	1
Total jitter $f_{C3dB} = f_{BAUD}/500$	$U_{SATA\_TXTJfB/500}$	—	—	0.37	UI_{p-p}	1

Table 69. SPI DC Electrical Characteristics (continued)

Parameter	Condition	Symbol	Min	Max	Unit
Output low voltage	I_{OL} = 8.0 mA	V_{OL}	—	0.5	V
Output low voltage	I_{OL} = 3.2 mA	V_{OL}	—	0.4	V

20.2 SPI AC Timing Specifications

This table provides the SPI input and output AC timing specifications.

Table 70. SPI AC Timing Specifications

Parameter	Symbol[1]	Min	Max	Unit
SPI outputs—Master mode (internal clock) delay	t_{NIKHOV}	0.5	6	ns
SPI outputs—Slave mode (external clock) delay	t_{NEKHOV}	2	8	ns
SPI inputs—Master mode (internal clock) input setup time	t_{NIIVKH}	4	—	ns
SPI inputs—Master mode (internal clock) input hold time	t_{NIIXKH}	0	—	ns
SPI inputs—Slave mode (external clock) input setup time	t_{NEIVKH}	4	—	ns
SPI inputs—Slave mode (external clock) input hold time	t_{NEIXKH}	2	—	ns

Notes:
1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{NIKHOV} symbolizes the internal timing (NI) for the time SPICLK clock reference (K) goes to the high state (H) until outputs (O) are invalid (X).
2. Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin. The maximum SPICLK input frequency is 66.666 MHz.

This figure provides the AC test load for the SPI.

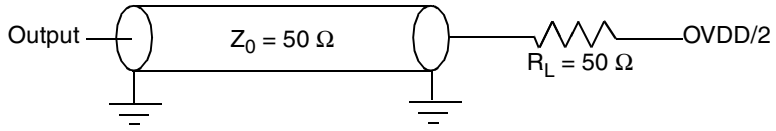


Figure 48. SPI AC Test Load

These figures represent the AC timing from Table 70. Note that although the specifications generally reference the rising edge of the clock, these AC timing diagrams also apply when the falling edge is the active edge.

Using this waveform, the definitions are as follows. To simplify illustration, the following definitions assume that the SerDes transmitter and receiver operate in a fully symmetrical differential signaling environment.

- **Single-Ended Swing**

  The transmitter output signals and the receiver input signals SD*n*_TX, $\overline{\text{SD}n\text{\_TX}}$, SD*n*_RX and $\overline{\text{SD}n\text{\_RX}}$ each have a peak-to-peak swing of A – B volts. This is also referred as each signal wire's single-ended swing.

- **Differential Output Voltage, V_{OD}** (or **Differential Output Swing**):

  The differential output voltage (or swing) of the transmitter, V_{OD}, is defined as the difference of the two complimentary output voltages: $V_{SDn\_TX} - V_{\overline{SDn\_TX}}$. The V_{OD} value can be either positive or negative.

- **Differential Input Voltage, V_{ID}** (or **Differential Input Swing**):

  The differential input voltage (or swing) of the receiver, V_{ID}, is defined as the difference of the two complimentary input voltages: $V_{SDn\_RX} - V_{\overline{SDn\_RX}}$. The V_{ID} value can be either positive or negative.

- **Differential Peak Voltage**, V_{DIFFp}

  The peak value of the differential transmitter output signal or the differential receiver input signal is defined as differential peak voltage, $V_{DIFFp} = |A - B|$ volts.

- **Differential Peak-to-Peak**, $V_{DIFFp-p}$

  Since the differential output signal of the transmitter and the differential input signal of the receiver each range from A – B to –(A – B) volts, the peak-to-peak value of the differential transmitter output signal or the differential receiver input signal is defined as differential peak-to-peak voltage, $V_{DIFFp-p} = 2 \times V_{DIFFp} = 2 \times |(A - B)|$ volts, which is twice of differential swing in amplitude, or twice of the differential peak. For example, the output differential peak-peak voltage can also be calculated as $V_{TX-DIFFp-p} = 2 \times |V_{OD}|$.

- **Differential Waveform**

  The differential waveform is constructed by subtracting the inverting signal ($\overline{\text{SD}n\text{\_TX}}$, for example) from the non-inverting signal (SD*n*_TX, for example) within a differential pair. There is only one signal trace curve in a differential waveform. The voltage represented in the differential waveform is not referenced to ground. Refer to Figure 60 as an example for differential waveform.

- **Common Mode Voltage, V_{cm}**

  The common mode voltage is equal to one half of the sum of the voltages between each conductor of a balanced interchange circuit and ground. In this example, for SerDes output, $V_{cm\_out} = (V_{SDn\_TX} + V_{\overline{SDn\_TX}}) \div 2 = (A + B) \div 2$, which is the arithmetic mean of the two complimentary output voltages within a differential pair. In a system, the common mode voltage may often differ from one component's output to the other's input. Sometimes it may be even different between the receiver input and driver output circuits within the same component. It is also referred as the DC offset in some occasion.

— The SerDes reference clock input can be either differential or single-ended. Refer to the Differential Mode and Single-ended Mode description below for further detailed requirements.

- The maximum average current requirement that also determines the common mode voltage range
  — When the SerDes reference clock differential inputs are DC coupled externally with the clock driver chip, the maximum average current allowed for each input pin is 8 mA. In this case, the exact common mode input voltage is not critical as long as it is within the range allowed by the maximum average current of 8 mA (refer to the following bullet for more detail), since the input is AC-coupled on-chip.
  — This current limitation sets the maximum common mode input voltage to be less than 0.4 V (0.4 V ÷ 50 = 8 mA) while the minimum common mode input level is 0.1 V above SGND_SRDS*n* (xcorevss). For example, a clock with a 50/50 duty cycle can be produced by a clock driver with output driven by its current source from 0 mA to 16 mA (0–0.8 V), such that each phase of the differential input has a single-ended swing from 0 V to 800 mV with the common mode voltage at 400 mV.
  — If the device driving the SD*n*_REF_CLK and $\overline{\text{SD}n\text{\_REF\_CLK}}$ inputs cannot drive 50 Ω to SGND_SRDS*n* (xcorevss) DC, or it exceeds the maximum input current limitations, then it must be AC-coupled off-chip.
- The input amplitude requirement
  — This requirement is described in detail in the following sections.

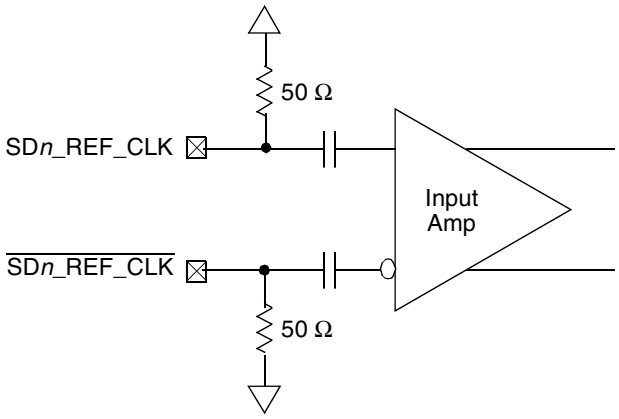


Figure 52. Receiver of SerDes Reference Clocks

21.2.2 DC Level Requirement for SerDes Reference Clocks

The DC level requirement for the device SerDes reference clock inputs is different depending on the signaling mode used to connect the clock driver chip and SerDes reference clock inputs as described below.

- **Differential Mode**
  — The input amplitude of the differential clock must be between 400 mV and 1600 mV differential peak-peak (or between 200 mV and 800 mV differential peak). In other words, each signal wire of the differential pair must have a single-ended swing less than 800 mV and

Table 72. TePBGA II Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Note
MDQ2	C7	I/O	GVDD	11
MDQ3	D8	I/O	GVDD	11
MDQ4	A7	I/O	GVDD	11
MDQ5	A5	I/O	GVDD	11
MDQ6	A3	I/O	GVDD	11
MDQ7	C6	I/O	GVDD	11
MDQ8	D7	I/O	GVDD	11
MDQ9	E8	I/O	GVDD	11
MDQ10	B1	I/O	GVDD	11
MDQ11	D5	I/O	GVDD	11
MDQ12	B3	I/O	GVDD	11
MDQ13	D6	I/O	GVDD	11
MDQ14	C3	I/O	GVDD	11
MDQ15	C2	I/O	GVDD	11
MDQ16	D4	I/O	GVDD	11
MDQ17	E6	I/O	GVDD	11
MDQ18	F6	I/O	GVDD	11
MDQ19	G4	I/O	GVDD	11
MDQ20	F8	I/O	GVDD	11
MDQ21	E4	I/O	GVDD	11
MDQ22	C1	I/O	GVDD	11
MDQ23	G6	I/O	GVDD	11
MDQ24	F2	I/O	GVDD	11
MDQ25	G5	I/O	GVDD	11
MDQ26	H6	I/O	GVDD	11
MDQ27	H4	I/O	GVDD	11
MDQ28	D1	I/O	GVDD	11
MDQ29	G3	I/O	GVDD	11
MDQ30	H5	I/O	GVDD	11
MDQ31	F1	I/O	GVDD	11
MDQ32	W6	I/O	GVDD	11
MDQ33	AC1	I/O	GVDD	11
MDQ34	AC3	I/O	GVDD	11

Table 72. TePBGA II Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Note
TSEC2_RXD0/GPIO1[16]	AE28	I/O	LVDD2	16
TSEC2_RXD1/GPIO1[15]	AE29	I/O	LVDD2	16
TSEC2_RXD2/GPIO1[14]	AH26	I/O	LVDD2	16
TSEC2_RXD3/GPIO1[13]	AH25	I/O	LVDD2	16
TSEC2_TX_CLK/GPIO2[24]/ TSEC1_TMR_GCLK	AG28	I/O	LVDD2	16
TSEC2_TX_EN/GPIO1[12]/ TSEC1_TMR_ALARM2	AJ26	I/O	LVDD2	16
TSEC2_TX_ER/GPIO1[24]/ TSEC1_TMR_ALARM1	AG26	I/O	LVDD2	16
TSEC2_TXD0/GPIO1[20]	AH28	I/O	LVDD2	16
TSEC2_TXD1/GPIO1[19]/ TSEC1_TMR_PP1	AF27	I/O	LVDD2	16
TSEC2_TXD2/GPIO1[18]/ TSEC1_TMR_PP2	AJ28	I/O	LVDD2	16
TSEC2_TXD3/GPIO1[17]/ TSEC1_TMR_PP3	AF29	I/O	LVDD2	16
GPIO1 Interface				
GPIO1[0]/GTM1_TIN1/ GTM2_TIN2/DREQ0_B	P25	I/O	OVDD	—
GPIO1[1]/GTM1_TGATE1_B/ GTM2_TGATE2_B/DACK0_B	N25	I/O	OVDD	—
GPIO1[2]/GTM1_TOUT1_B/ DDONE0_B	N26	I/O	OVDD	—
GPIO1[3]/GTM1_TIN2/ GTM2_TIN1/DREQ1_B	B9	I/O	OVDD	—
GPIO1[4]/GTM1_TGATE2_B/ GTM2_TGATE1_B/DACK1_B	N29	I/O	OVDD	—
GPIO1[5]/GTM1_TOUT2_B/ GTM2_TOUT1_B/DDONE1_B	M29	I/O	OVDD	—
GPIO1[6]/GTM1_TIN3/ GTM2_TIN4/DREQ2_B	A9	I/O	OVDD	—
GPIO1[7]/GTM1_TGATE3_B/ GTM2_TGATE4_B/DACK2_B	B10	I/O	OVDD	—
GPIO1[8]/GTM1_TOUT3_B/ DDONE2_B	J26	I/O	OVDD	—
GPIO1[9]/GTM1_TIN4/ GTM2_TIN3/DREQ3_B	J24	I/O	OVDD	—

Table 72. TePBGA II Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Note
Pull Down	B16, AH18	—	—	7

Notes:
1. This pin is an open drain signal. A weak pull-up resistor (1 kΩ) should be placed on this pin to OVDD.
2. This pin is an open drain signal. A weak pull-up resistor (2–10 kΩ) should be placed on this pin to OVDD.
3. This output is actively driven during reset rather than being released to high impedance during reset.
4. These JTAG pins have weak internal pull-up P-FETs that are always enabled.
5. This pin should have a weak pull up if the chip is in PCI host mode. Follow PCI Specification recommendation and see AN3665, “MPC837xE Design Checklist,” for more details.
6. These are On Die Termination pins, used to control DDR2 memories internal termination resistance.
7. This pin must always be tied to GND using a 0 Ω resistor.
8. This pin must always be left not connected.
9. For DDR2 operation, it is recommended that MDIC0 be tied to GND using an 18.2 Ω resistor and MDIC1 be tied to DDR power using an 18.2 Ω resistor.
10. This pin must always be tied low. If it is left floating it may cause the device to malfunction.
11. See AN3665, “MPC837xE Design Checklist,” for proper DDR termination.
12. This pin must not be pulled down during PORESET.
13. This pin must always be tied to OVDD.
14. Open or tie to GND.
15. Voltage settings are dependent on the frequency used; see Table 3.
16. See AN3665, “MPC837xE Design Checklist,” for proper termination.

Table 81. Package Thermal Characteristics for TePBGA II (continued)

Parameter	Symbol	Value	Unit	Note
Junction-to-package natural convection on top	ψ_{JT}	6	°C/W	6

Notes:
1. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.
2. Per JEDEC JESD51-2 with the single layer board horizontal. Board meets JESD51-9 specification.
3. Per JEDEC JESD51-6 with the board horizontal.
4. Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
5. Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1).
6. Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2. When Greek letters are not available, the thermal characterization parameter is written as Psi-JT.

24.2 Thermal Management Information

For the following sections, $P_D = (V_{DD} \times I_{DD}) + P_{I/O}$ where $P_{I/O}$ is the power dissipation of the I/O drivers.

24.2.1 Estimation of Junction Temperature with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature, T_J, can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_J = junction temperature (°C)

T_A = ambient temperature for the package (°C)

$R_{\theta JA}$ = junction to ambient thermal resistance (°C/W)

P_D = power dissipation in the package (W)

The junction to ambient thermal resistance is an industry-standard value that provides a quick and easy estimation of thermal performance. Generally, the value obtained on a single layer board is appropriate for a tightly packed printed circuit board. The value obtained on the board with the internal planes is usually appropriate if the board has low power dissipation and the components are well separated. Test cases have demonstrated that errors of a factor of two (in the quantity $T_J - T_A$) are possible.

24.2.2 Estimation of Junction Temperature with Junction-to-Board Thermal Resistance

NOTE

The heat sink cannot be mounted on the package.

This table shows the heat sink thermal resistance for TePBGA II package with heat sinks, simulated in a standard JEDEC environment, per JESD 51-6.

Table 82. Thermal Resistance with Heat Sink in Open Flow (TePBGA II)

Heat Sink Assuming Thermal Grease	Air Flow	Thermal Resistance (°/W)
AAVID 30 × 30 × 9.4 mm Pin Fin	Natural Convection	13.1
	0.5 m/s	10.6
	1 m/s	9.3
	2 m/s	8.2
	4 m/s	7.5
AAVID 31 × 35 × 23 mm Pin Fin	Natural Convection	11.1
	0.5 m/s	8.5
	1 m/s	7.7
	2 m/s	7.2
	4 m/s	6.8
AAVID 43× 41× 16.5mm Pin Fin	Natural Convection	11.3
	0.5 m/s	9.0
	1 m/s	7.8
	2 m/s	7.0
	4 m/s	6.5
Wakefield, 53 × 53 × 25 mm Pin Fin	Natural Convection	9.7
	0.5 m/s	7.7
	1 m/s	6.8
	2 m/s	6.4
	4 m/s	6.1

Heat sink vendors include the following:

Aavid Thermalloy
www.aavidthermalloy.com

Alpha Novatech
www.alphanovatech.com

International Electronic Research Corporation (IERC)
www.ctscorp.com

Millennium Electronics (MEI)
www.mei-thermal.com

Tyco Electronics
Chip Coolers™
www.chipcoolers.com

Wakefield Engineering
www.wakefield.com

Interface material vendors include the following:

Chomerics, Inc.
www.chomerics.com

Dow-Corning Corporation
Dow-Corning Electronic Materials
www.dowcorning.com

Shin-Etsu MicroSi, Inc.
www.microsi.com

The Bergquist Company
www.bergquistcompany.com

24.3 Heat Sink Attachment

The device requires the use of heat sinks. When heat sinks are attached, an interface material is required, preferably thermal grease and a spring clip. The spring clip should connect to the printed circuit board, either to the board itself, to hooks soldered to the board, or to a plastic stiffener. Avoid attachment forces that can lift the edge of the package or peel the package from the board. Such peeling forces reduce the solder joint lifetime of the package. The recommended maximum compressive force on the top of the package is 10 lb force (4.5 kg force). Any adhesive attachment should attach to painted or plastic surfaces, and its performance should be verified under the application requirements.

24.3.1 Experimental Determination of the Junction Temperature with a Heat Sink

When a heat sink is used, the junction temperature is determined from a thermocouple inserted at the interface between the case of the package and the interface material. A clearance slot or hole is normally required in the heat sink. Minimize the size of the clearance to minimize the change in thermal performance caused by removing part of the thermal interface to the heat sink. Because of the experimental difficulties with this technique, many engineers measure the heat sink temperature and then back calculate the case temperature using a separate measurement of the thermal resistance of the interface. From this case temperature, the junction temperature is determined from the junction to case thermal resistance.

$$T_J = T_C + (R_{\theta JC} \times P_D)$$

where:

T_J = junction temperature (°C)

T_C = case temperature of the package (°C)

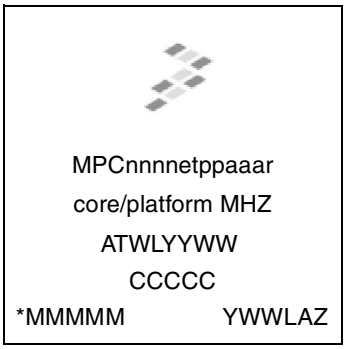
This table shows the SVR and PVR settings by device.

Table 86. SVR and PVR Settings by Product Revision

Device	Package	SVR		PVR	
		Rev 1.0	Rev. 2.1	Rev. 1.0	Rev. 2.1
MPC8377	TePBGA II	0x80C7_0010	0x80C7_0021	0x8086_1010	0x8086_1011
MPC8377E		0x80C6_0010	0x80C6_0021		
MPC8378		0x80C5_0010	0x80C5_0021		
MPC8378E		0x80C4_0010	0x80C4_0021		
MPC8379		0x80C3_0010	0x80C3_0021		
MPC8379E		0x80C2_0010	0x80C2_0021		

26.2 Part Marking

Parts are marked as in the example as shown in this figure.



TePBGA II

Notes:
ATWLYYWW is the traceability code.
CCCCC is the country code.
MMMMM is the mask number.
YWWLAZ is the assembly traceability code.

Figure 67. Freescale Part Marking for TePBGA II Devices

27 Document Revision History

This table provides a revision history for this document.

Table 87. Document Revision History

Revision	Date	Substantive Change(s)
8	05/2012	In Table 15, “DDR SDRAM DC Electrical Characteristics for GV_{DD} (typ) = 2.5 V,” updated Output leakage current (I_{OZ}) min and max values.
7	10/2011	• In Table 84, “Part Numbering Nomenclature,” updated “Revision Level description”and added footnote 4.In Section 21.2.4, “AC Requirements for SerDes Reference Clocks,” modified the introductory sentence for Table 71, “SerDes Reference Clock Common AC Parameters.”