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Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

Details

E·XFI

Details	
Product Status	Obsolete
Core Processor	XC800
Core Size	8-Bit
Speed	24MHz
Connectivity	CANbus, SSI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	34
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	·
RAM Size	1.75K x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-LQFP
Supplier Device Package	PG-TQFP-48
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/saf-xc886cm-8ffa-5v-ac

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Summary of Features

Features: (continued)

- Power-on reset generation
- Brownout detection for core logic supply
- On-chip OSC and PLL for clock generation
 - PLL loss-of-lock detection
- Power saving modes
 - slow-down mode
 - idle mode
 - power-down mode with wake-up capability via RXD or EXINT0
 - clock gating control to each peripheral
- Programmable 16-bit Watchdog Timer (WDT)
- Six ports
 - Up to 48 pins as digital I/O
 - 8 pins as digital/analog input
- 8-channel, 10-bit ADC
- Four 16-bit timers
 - Timer 0 and Timer 1 (T0 and T1)
 - Timer 2 and Timer 21 (T2 and T21)
- Multiplication/Division Unit for arithmetic operations (MDU)
- Software libraries to support floating point and MDU calculations
- CORDIC Coprocessor for computation of trigonometric, hyperbolic and linear functions
- MultiCAN with 2 nodes, 32 message objects
- Capture/compare unit for PWM signal generation (CCU6)
- Two full-duplex serial interfaces (UART and UART1)
- Synchronous serial channel (SSC)
- On-chip debug support
 - 1 Kbyte of monitor ROM (part of the 12-Kbyte Boot ROM)
 - 64 bytes of monitor RAM
- Packages:
 - PG-TQFP-48
 - PG-TQFP-64
- Temperature range *T*_A:
 - SAF (-40 to 85 °C)
 - SAK (-40 to 125 °C)



General Device Information

2 General Device Information

Chapter 2 contains the block diagram, pin configurations, definitions and functions of the XC886/888.

2.1 Block Diagram

The block diagram of the XC886/888 is shown in Figure 2.

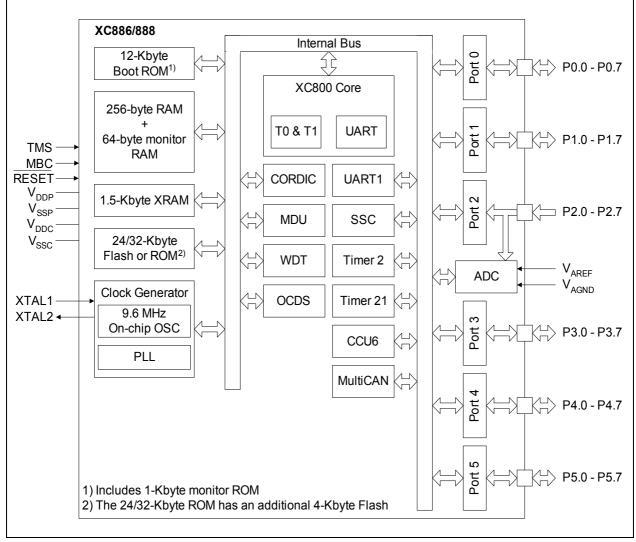


Figure 2 XC886/888 Block Diagram



General Device Information

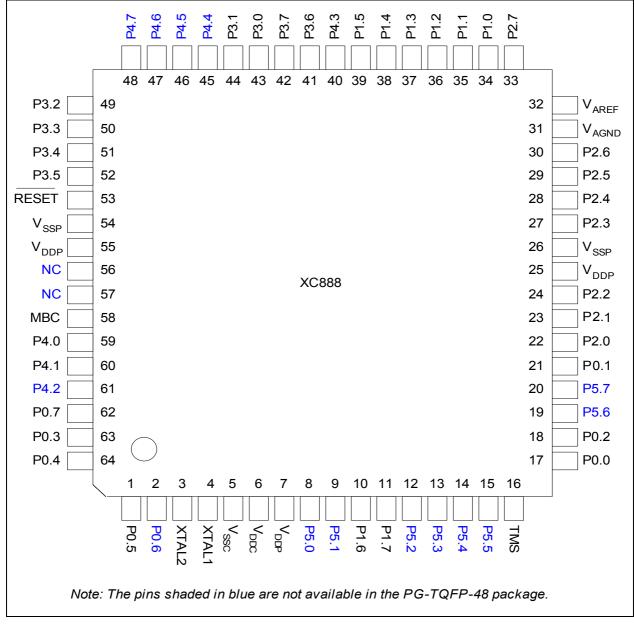


Figure 5 XC888 Pin Configuration, PG-TQFP-64 Package (top view)



General Device Information

2.4 Pin Definitions and Functions

The functions and default states of the XC886/888 external pins are provided in Table 3.

Symbol	Pin Number (TQFP-48/64)	Туре	Reset State	Function				
P0		I/O		Port 0 Port 0 is an 8-bit bidirectional general purpose I/O port. It can be used as alternate functions for the JTAG, CCU6, UART, UART1, Timer 2 Timer 21, MultiCAN and SSC.				
P0.0	11/17		Hi-Z	TCK_0 T12HR_1 CC61_1 CLKOUT_0 RXDO_1	JTAG Clock Input CCU6 Timer 12 Hardware Run Input Input/Output of Capture/Compare channel 1 Clock Output UART Transmit Data Output			
P0.1	13/21		Hi-Z	TDI_0 T13HR_1 RXD_1 RXDC1_0 COUT61_1 EXF2_1	JTAG Serial Data Input CCU6 Timer 13 Hardware Run Input UART Receive Data Input MultiCAN Node 1 Receiver Input Output of Capture/Compare channel 1 Timer 2 External Flag Output			
P0.2	12/18		PU	CTRAP_2 TDO_0 TXD_1 TXDC1_0	CCU6 Trap Input JTAG Serial Data Output UART Transmit Data Output/Clock Output MultiCAN Node 1 Transmitter Output			
P0.3	48/63		Hi-Z	SCK_1 COUT63_1 RXDO1_0	SSC Clock Input/Output Output of Capture/Compare channel 3 UART1 Transmit Data Output			

 Table 3
 Pin Definitions and Functions



Table 3

General Device Information

Pin Definitions and Functions (cont'd) Type Reset Function Pin Number Symbol (TQFP-48/64) State **P2** I Port 2 Port 2 is an 8-bit general purpose input-only port. It can be used as alternate functions for

			the digital inputs of the JTAG and CCU6. It is also used as the analog inputs for the ADC.				
P2.0	14/22	Hi-Z	CCPOS0_0 EXINT1_0 T12HR_2	CCU6 Hall Input 0 External Interrupt Input 1 CCU6 Timer 12 Hardware Run Input			
			TCK_1	JTAG Clock Input			
			CC61_3	Input of Capture/Compare channel 1			
			AN0	Analog Input 0			
P2.1	15/23	Hi-Z	CCPOS1_0 EXINT2_0 T13HR_2	CCU6 Hall Input 1 External Interrupt Input 2 CCU6 Timer 13 Hardware Run Input			
			TDI_1 CC62_3 AN1	JTAG Serial Data Input Input of Capture/Compare channel 2 Analog Input 1			
P2.2	16/24	Hi-Z	CCPOS2_0	CCU6 Hall Input 2			
1 2.2	10/24		CTRAP_1	CCU6 Trap Input			
			CC60_3	Input of Capture/Compare channel 0			
			AN2	Analog Input 2			
P2.3	19/27	Hi-Z	AN3	Analog Input 3			
P2.4	20/28	Hi-Z	AN4	Analog Input 4			
P2.5	21/29	Hi-Z	AN5	Analog Input 5			
P2.6	22/30	Hi-Z	AN6	Analog Input 6			
P2.7	25/33	Hi-Z	AN7	Analog Input 7			



code or data. Therefore, even though the ROM device contains either a 24-Kbyte or 32-Kbyte ROM, the maximum size of code that can be placed in the ROM is the given size less four bytes.

3.2.1 Memory Protection Strategy

The XC886/888 memory protection strategy includes:

- Read-out protection: The user is able to protect the contents in the Flash (for Flash devices) and ROM (for ROM devices) memory from being read
 - Flash protection is enabled by programming a valid password (8-bit non-zero value) via BSL mode 6.
 - ROM protection is fixed with the ROM mask and is always enabled.
- Flash program and erase protection: This feature is available only for Flash devices.

3.2.1.1 Flash Memory Protection

As long as a valid password is available, all external access to the device, including the Flash, will be blocked.

For additional security, the Flash hardware protection can be enabled to implement a second layer of read-out protection, as well as to enable program and erase protection.

Flash hardware protection is available only for Flash devices and comes in two modes:

- Mode 0: Only the P-Flash is protected; the D-Flash is unprotected
- Mode 1: Both the P-Flash and D-Flash are protected

The selection of each protection mode and the restrictions imposed are summarized in **Table 4**.

Flash Protection	Without hardware protection	With hardware protection					
Hardware Protection Mode	-	0	1				
Activation	Program a valid password via BSL mode 6						
Selection Bit 4 of password = 0		Bit 4 of password = 1 MSB of password = 0	Bit 4 of password = 1 MSB of password = 1				
P-Flash contents can be read by	Read instructions in any program memory	Read instructions in the P-Flash	Read instructions in the P-Flash or D- Flash				
External access to P-Flash	Not possible	Not possible	Not possible				

Table 4Flash Protection Modes



Flash Protection	Without hardware protection	With hardware protection					
P-Flash program and erase	Possible	Not possible	Not possible				
D-Flash contents can be read by	Read instructions in any program memory	Read instructions in any program memory	Read instructions in the P-Flash or D-Flash				
External access to D-Flash	Not possible	Not possible	Not possible				
D-Flash program	Possible	Possible	Not possible				
D-Flash erase	Possible	Possible, on condition that bit DFLASHEN in register MISC_CON is set to 1 prior to each erase operation	Not possible				

Table 4Flash Protection Modes (cont'd)

BSL mode 6, which is used for enabling Flash protection, can also be used for disabling Flash protection. Here, the programmed password must be provided by the user. A password match triggers an automatic erase of the protected P-Flash and D-Flash contents, including the programmed password. The Flash protection is then disabled upon the next reset.

For the ROM device, the ROM is protected at all times and BSL mode 6 is used only to block external access to the device. However, unlike the Flash device, it is not possible to disable the memory protection of the ROM device. Here, entering BSL mode 6 will result in a protection error.

Note: If ROM read-out protection is enabled, only read instructions in the ROM memory can target the ROM contents.

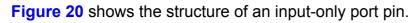
Although no protection scheme can be considered infallible, the XC886/888 memory protection strategy provides a very high level of protection for a general purpose microcontroller.



Table 11ADC Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
cc ^H	ADC_CHINSR Reset: 00 _H Channel Interrupt Set Register	Bit Field	CHINS 7	CHINS 6	CHINS 5	CHINS 4	CHINS 3	CHINS 2	CHINS 1	CHINS 0
			w	w	w	w	w	w	w	w
CDH	ADC_CHINPR Reset: 00 _H Channel Interrupt Node Pointer Register	Bit Field	CHINP 7	CHINP 6	CHINP 5	CHINP 4	CHINP 3	CHINP 2	CHINP 1	CHINP 0
	Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
Ceh	ADC_EVINFR Reset: 00 _H Event Interrupt Flag Register	Bit Field	EVINF 7	EVINF 6	EVINF 5	EVINF 4	()	EVINF 1	EVINF 0
		Туре	rh	rh	rh	rh		r	rh	rh
CF _H	ADC_EVINCR Reset: 00 _H Event Interrupt Clear Flag	Bit Field	EVINC 7	EVINC 6	EVINC 5	EVINC 4	()	EVINC 1	EVINC 0
	Register	Туре	w	w	w	w		r	w	w
D2 _H	ADC_EVINSR Reset: 00 _H Event Interrupt Set Flag Register	Bit Field	EVINS 7	EVINS 6	EVINS 5	EVINS 4	()	EVINS 1	EVINS 0
		Туре	w	w	w	w		r	w	w
D3 _H	ADC_EVINPR Reset: 00 _H Event Interrupt Node Pointer	Bit Field	EVINP 7	EVINP 6	EVINP 5	EVINP 4	0 EVINP 1		EVINP 0	
	Register	Туре	rw	rw	rw	rw		r	rw	rw
RMAP =	= 0, PAGE 6									
CA _H	ADC_CRCR1 Reset: 00 _H	Bit Field	CH7	CH6	CH5	CH4	0			
	Conversion Request Control Register 1	Туре	rwh	rwh	rwh	rwh		r		
св _Н	ADC_CRPR1 Reset: 00 _H	Bit Field	CHP7	CHP6	CHP5	CHP4	0			
	Conversion Request Pending Register 1	Туре	rwh	rwh	rwh	rwh		I	r	
сс ^н	ADC_CRMR1 Reset: 00 _H Conversion Request Mode	Bit Field	Rsv	LDEV	CLRP ND	SCAN	ENSI	ENTR	0	ENGT
	Register 1	Туре	r	w	w	rw	rw	rw	r	rw
CD _H	ADC_QMR0 Reset: 00 _H Queue Mode Register 0	Bit Field	CEV	TREV	FLUS H	CLRV	0	ENTR	0	ENGT
		Туре	w	w	w	w	r	rw	r	rw
Ceh	ADC_QSR0 Reset: 20 _H Queue Status Register 0	Bit Field	Rsv	0	EMPT Y	EV	0 FILL		LL	
		Туре	r	r	rh	rh		r	r	h
CF _H	ADC_Q0R0 Reset: 00 _H	Bit Field	EXTR	ENSI	RF	V	0	F	REQCHN	२
	Queue 0 Register 0	Туре	rh	rh	rh	rh	r		rh	
D2 _H	ADC_QBUR0 Reset: 00 _H	Bit Field	EXTR	ENSI	RF	V	0	F	REQCHN	۲
	Queue Backup Register 0	Туре	rh	rh	rh	rh	r		rh	
D2 _H	ADC_QINR0 Reset: 00 _H	Bit Field	EXTR	ENSI	RF	()	F	REQCHN	२
	Queue Input Register 0	Туре	w	w	w		r		w	





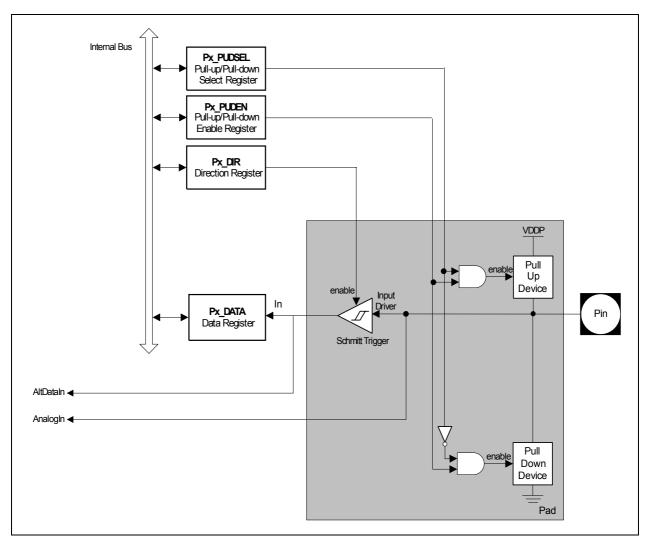


Figure 20 General Structure of Input Port



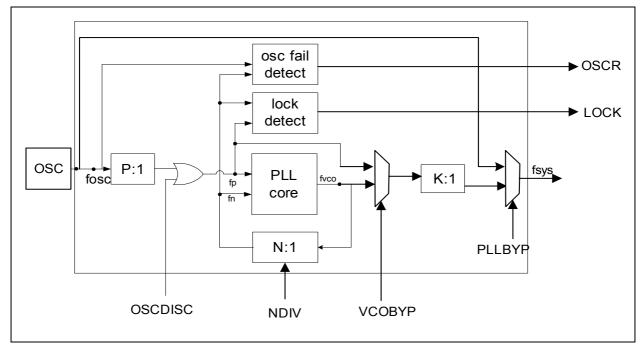


Figure 24 CGU Block Diagram

PLL Base Mode

When the oscillator is disconnected from the PLL, the system clock is derived from the VCO base (free running) frequency clock (**Table 25**) divided by the K factor.

$$f_{SYS} = f_{VCObase} \times \frac{1}{K}$$

(3.1)

Prescaler Mode (VCO Bypass Operation)

In VCO bypass operation, the system clock is derived from the oscillator clock, divided by the P and K factors.

$$f_{SYS} = f_{OSC} \times \frac{1}{P \times K}$$

(3.2)



Table 25 shows the VCO range for the XC886/888.

<i>f</i> _{vcomin}	f _{VCOmax}	$f_{\sf VCOFREEmin}$	<i>f</i> _{VCOFREEmax}	Unit
150	200	20	80	MHz
100	150	10	80	MHz

3.8.1 Recommended External Oscillator Circuits

The oscillator circuit, a Pierce oscillator, is designed to work with both, an external crystal oscillator or an external stable clock source. It basically consists of an inverting amplifier and a feedback element with XTAL1 as input, and XTAL2 as output.

When using a crystal, a proper external oscillator circuitry must be connected to both pins, XTAL1 and XTAL2. The crystal frequency can be within the range of 4 MHz to 12 MHz. Additionally, it is necessary to have two load capacitances C_{X1} and C_{X2} , and depending on the crystal type, a series resistor R_{X2} , to limit the current. A test resistor R_Q may be temporarily inserted to measure the oscillation allowance (negative resistance) of the oscillator circuitry. R_Q values are typically specified by the crystal vendor. The C_{X1} and C_{X2} values shown in **Figure 25** can be used as starting points for the negative resistance evaluation and for non-productive systems. The exact values and related operating range are dependent on the crystal frequency and have to be determined and optimized together with the final target system is strongly recommended to verify the input amplitude at XTAL1 and to determine the actual oscillation allowance (margin negative resistance) for the oscillator-crystal system.

When using an external clock signal, the signal must be connected to XTAL1. XTAL2 is left open (unconnected).

The oscillator can also be used in combination with a ceramic resonator. The final circuitry must also be verified by the resonator vendor. **Figure 25** shows the recommended external oscillator circuitries for both operating modes, external crystal mode and external input clock mode.



For power saving purposes, the clocks may be disabled or slowed down according to **Table 26**.

Table 26System frequency (f_{sys} = 96 MHz)

Power Saving Mode	Action						
Idle	Clock to the CPU is disabled.						
Slow-down	Clocks to the CPU and all the peripherals are divided by a common programmable factor defined by bit field CMCON.CLKREL.						
Power-down	Oscillator and PLL are switched off.						



fractional divider) for generating a wide range of baud rates based on its input clock f_{PCLK} , see **Figure 30**.

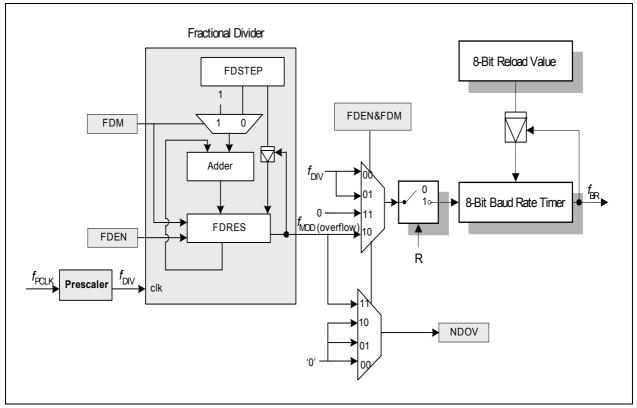


Figure 30 Baud-rate Generator Circuitry

The baud rate timer is a count-down timer and is clocked by either the output of the fractional divider (f_{MOD}) if the fractional divider is enabled (FDCON.FDEN = 1), or the output of the prescaler (f_{DIV}) if the fractional divider is disabled (FDEN = 0). For baud rate generation, the fractional divider must be configured to fractional divider mode (FDCON.FDM = 0). This allows the baud rate control run bit BCON.R to be used to start or stop the baud rate timer. At each timer underflow, the timer is reloaded with the 8-bit reload value in register BG and one clock pulse is generated for the serial channel.

Enabling the fractional divider in normal divider mode (FDEN = 1 and FDM = 1) stops the baud rate timer and nullifies the effect of bit BCON.R. See **Section 3.14**.

The baud rate (f_{BR}) value is dependent on the following parameters:

- Input clock f_{PCLK}
- Prescaling factor (2^{BRPRE}) defined by bit field BRPRE in register BCON
- Fractional divider (STEP/256) defined by register FDSTEP (to be considered only if fractional divider is enabled and operating in fractional divider mode)
- 8-bit reload value (BR_VALUE) for the baud rate timer defined by register BG



3.18 Timer 2 and Timer 21

Timer 2 and Timer 21 are 16-bit general purpose timers (THL2) that are fully compatible and have two modes of operation, a 16-bit auto-reload mode and a 16-bit one channel capture mode, see **Table 33**. As a timer, the timers count with an input clock of PCLK/12 (if prescaler is disabled). As a counter, they count 1-to-0 transitions on pin T2. In the counter mode, the maximum resolution for the count is PCLK/24 (if prescaler is disabled).

Table 33	Timer 2 Modes						
Mode	Description						
Auto-reload	 Up/Down Count Disabled Count up only Start counting from 16-bit reload value, overflow at FFFF_H Reload event configurable for trigger by overflow condition only, or by negative/positive edge at input pin T2EX as well Programmble reload value in register RC2 Interrupt is generated with reload event 						
	 Up/Down Count Enabled Count up or down, direction determined by level at input pin T2EX No interrupt is generated Count up Start counting from 16-bit reload value, overflow at FFFF_H Reload event triggered by overflow condition Programmble reload value in register RC2 Count down Start counting from FFFF_H, underflow at value defined in register RC2 Reload event triggered by underflow condition Reload event triggered by underflow condition Reload event triggered by underflow condition 						
Channel capture	 Count up only Start counting from 0000_H, overflow at FFFF_H Reload event triggered by overflow condition Reload value fixed at 0000_H Capture event triggered by falling/rising edge at pin T2EX Captured timer value stored in register RC2 Interrupt is generated with reload or capture event 						



Electrical Parameters

Parameter	Symbol		Limit	Values	Unit	Test Conditions	
			min.	max.			
Maximum current out of $V_{\rm SS}$	I _{MVSS}	SR	-	120	mA	3)	
V_{DDP} = 3.3 V Range							
Output low voltage	V_{OL}	CC	_	1.0	V	I _{OL} = 8 mA	
			-	0.4	V	I _{OL} = 2.5 mA	
Output high voltage	V _{OH}	CC	V _{DDP} - 1.0	-	V	I _{OH} = -8 mA	
			V _{DDP} - 0.4	-	V	I _{ОН} = -2.5 mA	
Input low voltage on port pins (all except P0.0 & P0.1)	V _{ILP}	SR	_	$0.3 \times V_{\text{DDP}}$	V	CMOS Mode	
Input low voltage on P0.0 & P0.1	V _{ILP0}	SR	-0.2	$0.3 \times V_{ m DDP}$	V	CMOS Mode	
Input low voltage on RESET pin	V_{ILR}	SR	-	$0.3 \times V_{\text{DDP}}$	V	CMOS Mode	
Input low voltage on TMS pin	V _{ILT}	SR	-	$0.3 \times V_{\text{DDP}}$	V	CMOS Mode	
Input high voltage on port pins (all except P0.0 & P0.1)	V _{IHP}	SR	$0.7 \times V_{\text{DDP}}$	-	V	CMOS Mode	
Input high voltage on P0.0 & P0.1	V _{IHP0}	SR	$0.7 \times V_{\text{DDP}}$	V _{DDP}	V	CMOS Mode	
Input high voltage on RESET pin	V_{IHR}	SR	$0.7 \times V_{ m DDP}$	-	V	CMOS Mode	
Input high voltage on TMS pin	V _{IHT}	SR	$0.75 \times V_{ m DDP}$	-	V	CMOS Mode	
Input Hysteresis	HYS	CC	$0.03 \times V_{ m DDP}$	-	V	CMOS Mode ¹⁾	
Input Hysteresis on XTAL1	HYSX	CC	$0.07 \times V_{ m DDC}$	-	V	1)	
Input low voltage at XTAL1	V _{ILX}	SR	V _{SS} - 0.5	$0.3 \times V_{ m DDC}$	V		



Electrical Parameters

Table 40ADC Characteristics (Operating Conditions apply; V_{DDP} = 5V Range)

				~		-	
Parameter	Symbol		Liı	Limit Values			Test Conditions/
			min.	typ .	max.		Remarks
Overload current coupling factor for	K _{OVD}	CC	-	-	5.0 x 10 ⁻³	-	$I_{\rm OV} > 0^{1)3)$
digital I/O pins			-	-	1.0 x 10 ⁻²	-	$I_{\rm OV} < 0^{1)3)}$
Switched capacitance at the reference voltage input	CAREFSW	CC	_	10	20	pF	1)4)
Switched capacitance at the analog voltage inputs	C _{AINSW}	CC	_	5	7	pF	1)5)
Input resistance of the reference input	R _{AREF}	CC	-	1	2	kΩ	1)
Input resistance of the selected analog channel	R _{AIN}	CC	_	1	1.5	kΩ	1)

1) Not subjected to production test, verified by design/characterization

2) TUE is tested at V_{AREF} = 5.0 V, V_{AGND} = 0 V, V_{DDP} = 5.0 V.

- 3) An overload current (I_{OV}) through a pin injects a certain error current (I_{INJ}) into the adjacent pins. This error current adds to the respective pin's leakage current (I_{OZ}) . The amount of error current depends on the overload current and is defined by the overload coupling factor K_{OV} . The polarity of the injected error current is inverse compared to the polarity of the overload current that produces it. The total current through a pin is $|I_{TOT}| = |I_{OZ1}| + (|I_{OV}| \times K_{OV})$. The additional error current may distort the input voltage on analog inputs.
- 4) This represents an equivalent switched capacitance. This capacitance is not switched to the reference voltage at once. Instead of this, smaller capacitances are successively switched to the reference voltage.
- 5) The sampling capacity of the conversion C-Network is pre-charged to $V_{AREF}/2$ before connecting the input to the C-Network. Because of the parasitic elements, the voltage measured at ANx is lower than $V_{AREF}/2$.



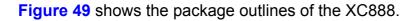
Electrical Parameters

4.2.3.1 ADC Conversion Timing

Conversion time, $t_{\rm C} = t_{\rm ADC} \times (1 + r \times (3 + n + STC))$, where r = CTC + 2 for CTC = $00_{\rm B}$, $01_{\rm B}$ or $10_{\rm B}$, r = 32 for CTC = $11_{\rm B}$, CTC = Conversion Time Control (GLOBCTR.CTC), STC = Sample Time Control (INPCR0.STC), n = 8 or 10 (for 8-bit and 10-bit conversion respectively), $t_{\rm ADC} = 1 / f_{\rm ADC}$



Package and Quality Declaration



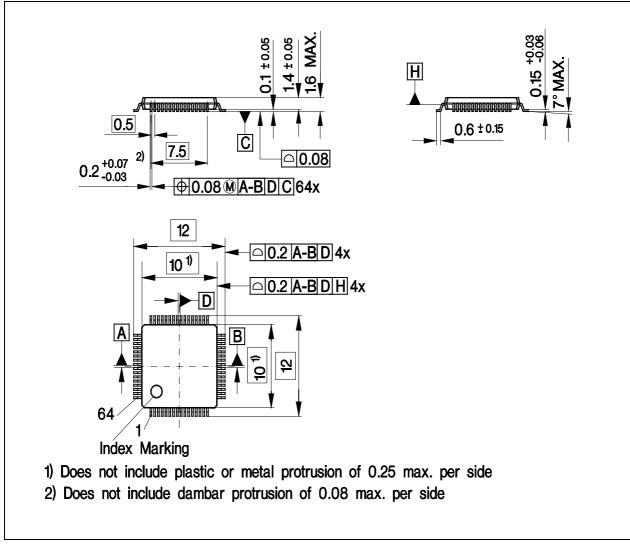


Figure 49 PG-TQFP-64 Package Outline



Package and Quality Declaration

5.3 Quality Declaration

Table 2 shows the characteristics of the quality parameters in the XC886/888.

Table 2Quality Parameters

Parameter	Symbol	Limit Values		Unit	Notes
		Min.	Max.		
ESD susceptibility according to Human Body Model (HBM)	V _{HBM}	-	2000	V	Conforming to EIA/JESD22- A114-B ¹⁾
ESD susceptibility according to Charged Device Model (CDM) pins	V _{CDM}	-	500	V	Conforming to JESD22-C101-C ¹⁾

1) Not all parameters are 100% tested, but are verified by design/characterization and test correlation.