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Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	200
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	80
Number of Gates	1500
Voltage - Supply	4.5V ~ 5.5V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	100-BQFP
Supplier Device Package	100-PQFP (20x14)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a1415a-1pqg100i

Table 1-1 • Chip-to-Chip Performance (worst-case commercial)

Device and Speed Grade	t _{CKHS} (ns)	t _{TRACE} (ns)	t _{INSU} (ns)	Total (ns)	MHz
A1425A -3	7.5	1.0	1.8	10.3	97
A1460A -3	9.0	1.0	1.3	11.3	88
A1425A -2	7.5	1.0	2.0	10.5	95
A1460A -2	9.0	1.0	1.5	11.5	87
A1425A -1	9.0	1.0	2.3	12.3	81
A1460A -1	10.0	1.0	1.8	12.8	78
A1425A STD	10.0	1.0	2.7	13.7	73
A1460A STD	11.5	1.0	2.0	14.5	69

Note: The –2 and –3 speed grades have been discontinued. Refer to PDN 0104, PDN 0203, PDN 0604, and PDN 1004 at <http://www.microsemi.com/soc/support/notifications/default.aspx#pdn>.

Logic Modules

ACT 3 logic modules are enhanced versions of the 1200XL family logic modules. As in the 1200XL family, there are two types of modules: C-modules and S-modules (Figure 2-2 and Figure 2-3). The C-module is functionally equivalent to the 1200XL C-module and implements high fanin combinatorial macros, such as 5-input AND, 5-input OR, and so on. It is available for use as the CM8 hard macro. The S-module is designed to implement high-speed sequential functions within a single module.

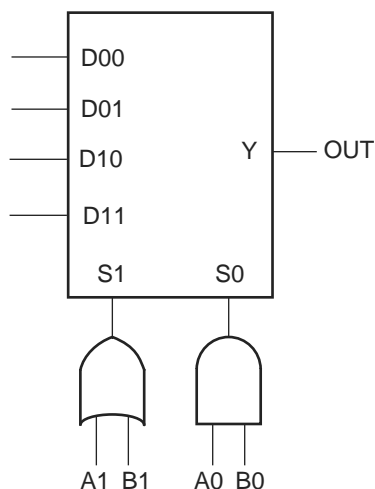


Figure 2-2 • C-Module Diagram

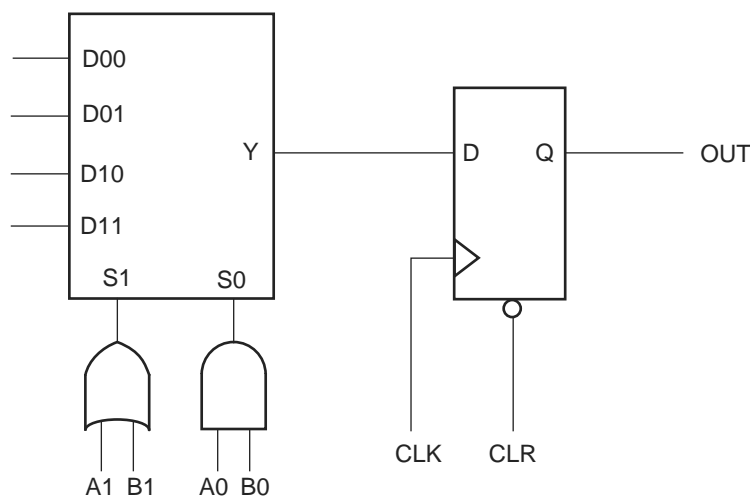


Figure 2-3 • S-Module Diagram

S-modules consist of a full C-module driving a flip-flop, which allows an additional level of logic to be implemented without additional propagation delay. It is available for use as the DFM8A/B and DLM8A/B hard macros. C-modules and S-modules are arranged in pairs called module-pairs. Module-pairs are arranged in alternating patterns and make up the bulk of the array. This arrangement allows the placement software to support two-module macros of four types (CC, CS, SC, and SS). The C-module implements the following function:

$$Y = !S1 * !S0 * D00 + !S1 * S0 * D01 + S1 * !S0 * D10 + S1 * S0 * D11$$

EQ 1

where: $S0 = A0 * B0$ and $S1 = A1 + B1$

The S-module contains a full implementation of the C-module plus a clearable sequential element that can either implement a latch or flip-flop function. The S-module can therefore implement any function implemented by the C-module. This allows complex combinatorial-sequential functions to be implemented with no delay penalty. The Designer Series Development System will automatically combine any C-module macro driving an S-module macro into the S-module, thereby freeing up a logic module and eliminating a module delay.

The clear input CLR is accessible from the routing channel. In addition, the clock input may be connected to one of three clock networks: CLKA, CLKB, or HCLK. The C-module and S-module functional descriptions are shown in Figure 2-2 and Figure 2-3 on page 2-2. The clock selection is determined by a multiplexer select at the clock input to the S-module.

I/Os

I/O Modules

I/O modules provide an interface between the array and the I/O Pad Drivers. I/O modules are located in the array and access the routing channels in a similar fashion to logic modules. The I/O module schematic is shown in Figure 4. The signals DataIn and DataOut connect to the I/O pad driver.

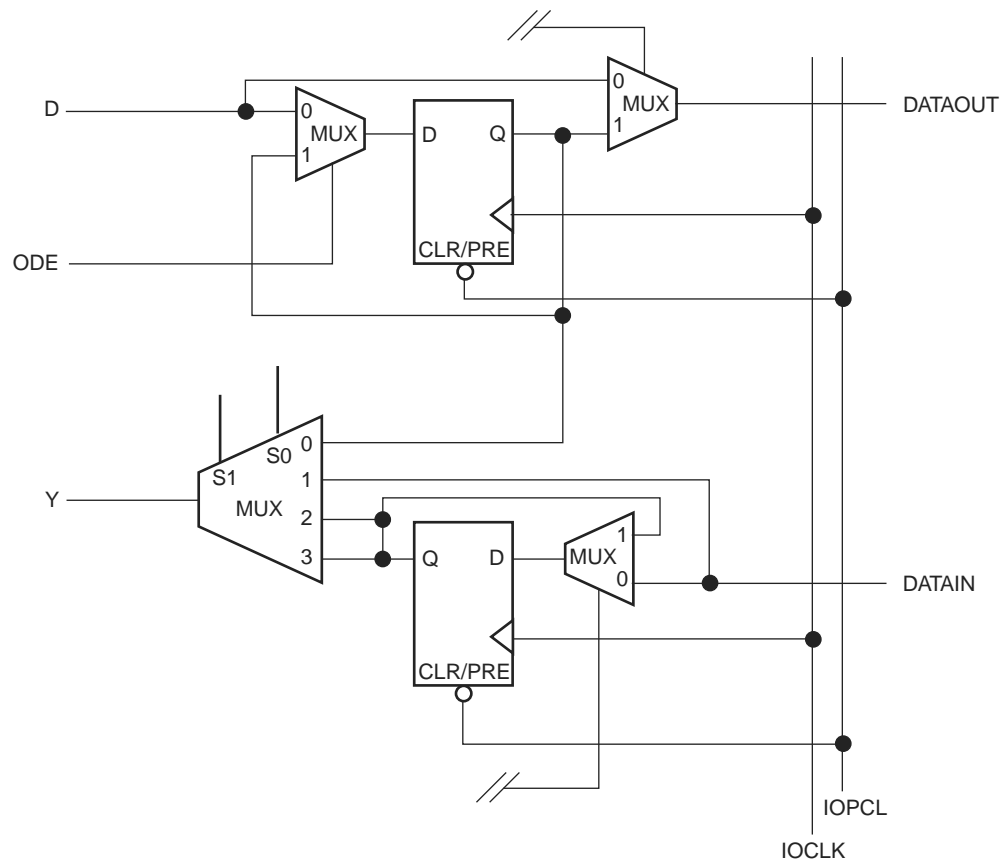


Figure 2-4 • Functional Diagram for I/O Module

Each I/O module contains two D-type flip-flops. Each flip-flop is connected to the dedicated I/O clock (IOCLK). Each flip-flop can be bypassed by nonsequential I/Os. In addition, each flip-flop contains a data enable input that can be accessed from the routing channels (ODE and IDE). The asynchronous preset/clear input is driven by the dedicated preset/clear network (IOPCL). Either preset or clear can be selected individually on an I/O module by I/O module basis.

The I/O module output Y is used to bring Pad signals into the array or to feed the output register back into the array. This allows the output register to be used in high-speed state machine applications. Side I/O modules have a dedicated output segment for Y extending into the routing channels above and below (similar to logic modules). Top/Bottom I/O modules have no dedicated output segment. Signals coming into the chip from the top or bottom are routed using F-fuses and LVTs (F-fuses and LVTs are explained in detail in the routing section).

I/O Pad Drivers

All pad drivers are capable of being tristate. Each buffer connects to an associated I/O module with four signals: OE (Output Enable), IE (Input Enable), DataOut, and DataIn. Certain special signals used only during programming and test also connect to the pad drivers: OUTEN (global output enable), INEN (global input enable), and SLEW (individual slew selection). See Figure 2-5.

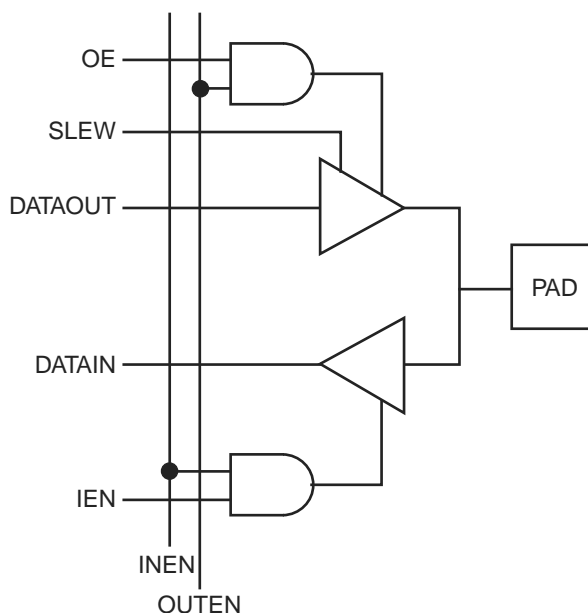


Figure 2-5 • Function Diagram for I/O Pad Driver

Special I/Os

The special I/Os are of two types: temporary and permanent. Temporary special I/Os are used during programming and testing. They function as normal I/Os when the MODE pin is inactive. Permanent special I/Os are user programmed as either normal I/Os or special I/Os. Their function does not change once the device has been programmed. The permanent special I/Os consist of the array clock input buffers (CLKA and CLKB), the hard-wired array clock input buffer (HCLK), the hard-wired I/O clock input buffer (IOCLK), and the hard-wired I/O register preset/clear input buffer (IOPCL). Their function is determined by the I/O macros selected.

Clock Networks

The ACT 3 architecture contains four clock networks: two high-performance dedicated clock networks and two general purpose routed networks. The high-performance networks function up to 200 MHz, while the general purpose routed networks function up to 150 MHz.

Module Output Connections

Module outputs have dedicated output segments. Output segments extend vertically two channels above and two channels below, except at the top or bottom of the array. Output segments twist, as shown in Figure 10, so that only four vertical tracks are required.

LVT Connections

Outputs may also connect to nondedicated segments called Long Vertical Tracks (LVTs). Each module pair in the array shares four LVTs that span the length of the column. Any module in the column pair can connect to one of the LVTs in the column using an FF connection. The FF connection uses antifuses connected directly to the driver stage of the module output, bypassing the isolation transistor. FF antifuses are programmed at a higher current level than HF, VF, or XF antifuses to produce a lower resistance value.

Antifuse Connections

In general every intersection of a vertical segment and a horizontal segment contains an unprogrammed antifuse (XF-type). One exception is in the case of the clock networks.

Clock Connections

To minimize loading on the clock networks, a subset of inputs has antifuses on the clock tracks. Only a few of the C-module and S-module inputs can be connected to the clock networks. To further reduce loading on the clock network, only a subset of the horizontal routing tracks can connect to the clock inputs of the S-module.

Programming and Test Circuits

The array of logic and I/O modules is surrounded by test and programming circuits controlled by the temporary special I/O pins MODE, SDI, and DCLK. The function of these pins is similar to all ACT family devices. The ACT 3 family also includes support for two Actionprobe[®] circuits, allowing complete observability of any logic or I/O module in the array using the temporary special I/O pins, PRA and PRB.

Equivalent capacitance is calculated by measuring ICC active at a specified frequency and voltage for each circuit component of interest. Measurements have been made over a range of frequencies at a fixed value of VCC. Equivalent capacitance is frequency independent so that the results may be used over a wide range of operating conditions. Equivalent capacitance values are shown in Figure 2-10.

Table 2-10 • CEQ Values for Microsemi FPGAs

Item	CEQ Value
Modules (C _{EQM})	6.7
Input Buffers (C _{EQI})	7.2
Output Buffers (C _{EQO})	10.4
Routed Array Clock Buffer Loads (C _{EQCR})	1.6
Dedicated Clock Buffer Loads (C _{EQCD})	0.7
I/O Clock Buffer Loads (C _{EQCI})	0.9

To calculate the active power dissipated from the complete design, the switching frequency of each part of the logic must be known. EQ 5 shows a piece-wise linear summation over all components.

$$\begin{aligned}
 \text{Power} = & VCC^2 * [(m * C_{EQM} * f_m)_{\text{modules}} + (n * C_{EQI} * f_n)_{\text{inputs}} \\
 & + (p * (C_{EQO} + C_L) * f_p)_{\text{outputs}} \\
 & + 0.5 * (q_1 * C_{EQCR} * f_{q1})_{\text{routed_Clk1}} + (r_1 * f_{q1})_{\text{routed_Clk1}} \\
 & + 0.5 * (q_2 * C_{EQCR} * f_{q2})_{\text{routed_Clk2}} \\
 & + (r_2 * f_{q2})_{\text{routed_Clk2}} + 0.5 * (s_1 * C_{EQCD} * f_{s1})_{\text{dedicated_Clk}} \\
 & + (s_2 * C_{EQCI} * f_{s2})_{\text{IO_Clk}}]
 \end{aligned}$$

EQ 5

Where:

m = Number of logic modules switching at f_m

n = Number of input buffers switching at f_n

p = Number of output buffers switching at f_p

q₁ = Number of clock loads on the first routed array clock

q₂ = Number of clock loads on the second routed array clock

r₁ = Fixed capacitance due to first routed array clock

r₂ = Fixed capacitance due to second routed array clock

s₁ = Fixed number of clock loads on the dedicated array clock

s₂ = Fixed number of clock loads on the dedicated I/O clock

C_{EQM} = Equivalent capacitance of logic modules in pF

C_{EQI} = Equivalent capacitance of input buffers in pF

C_{EQO} = Equivalent capacitance of output buffers in pF

C_{EQCR} = Equivalent capacitance of routed array clock in pF

C_{EQCD} = Equivalent capacitance of dedicated array clock in pF

C_{EQCI} = Equivalent capacitance of dedicated I/O clock in pF

C_L = Output lead capacitance in pF

f_m = Average logic module switching rate in MHz

f_n = Average input buffer switching rate in MHz

f_p = Average output buffer switching rate in MHz

f_{q1} = Average first routed array clock rate in MHz

f_{q2} = Average second routed array clock rate in MHz

f_{s1} = Average dedicated array clock rate in MHz

f_{s2} = Average dedicated I/O clock rate in MHz

Tightest Delay Distributions

Propagation delay between logic modules depends on the resistive and capacitive loading of the routing tracks, the interconnect elements, and the module inputs being driven. Propagation delay increases as the length of routing tracks, the number of interconnect elements, or the number of inputs increases.

From a design perspective, the propagation delay can be statistically correlated or modeled by the fanout (number of loads) driven by a module. Higher fanout usually requires some paths to have longer lengths of routing track. The ACT 3 family delivers the tightest fanout delay distribution of any FPGA. This tight distribution is achieved in two ways: by decreasing the delay of the interconnect elements and by decreasing the number of interconnect elements per path.

Microsemi's patented PLICE antifuse offers a very low resistive/capacitive interconnect. The ACT 3 family's antifuses, fabricated in 0.8 micron m lithography, offer nominal levels of 200Ω resistance and 6 femtofarad (fF) capacitance per antifuse. The ACT 3 fanout distribution is also tighter than alternative devices due to the low number of antifuses required per interconnect path. The ACT 3 family's proprietary architecture limits the number of antifuses per path to only four, with 90% of interconnects using only two antifuses.

The ACT 3 family's tight fanout delay distribution offers an FPGA design environment in which fanout can be traded for the increased performance of reduced logic level designs. This also simplifies performance estimates when designing with ACT 3 devices.

Table 2-14 • Logic Module and Routing Delay by Fanout (ns); Worst-Case Commercial Conditions

Speed Grade	FO = 1	FO = 2	FO = 3	FO = 4	FO = 8
ACT 3 –3	2.9	3.2	3.4	3.7	4.8
ACT 3 –2	3.3	3.7	3.9	4.2	5.5
ACT 3 –1	3.7	4.2	4.4	4.8	6.2
ACT 3 STD	4.3	4.8	5.1	5.5	7.2

Notes:

1. Obtained by added $t_{RD(x=FO)}$ to t_{PD} from the Logic Module Timing Characteristics Tables found in this datasheet.
2. The –2 and –3 speed grades have been discontinued. Refer to PDN 0104, PDN 0203, PDN 0604, and PDN 1004 at <http://www.microsemi.com/soc/support/notifications/default.aspx#pdn>.

Timing Characteristics

Timing characteristics for ACT 3 devices fall into three categories: family dependent, device dependent, and design dependent. The input and output buffer characteristics are common to all ACT 3 family members. Internal routing delays are device dependent. Design dependency means actual delays are not determined until after placement and routing of the user's design is complete. Delay values may then be determined by using the ALS Timer utility or performing simulation with post-layout delays.

Critical Nets and Typical Nets

Propagation delays are expressed only for typical nets, which are used for initial design performance evaluation. Critical net delays can then be applied to the most time-critical paths. Critical nets are determined by net property assignment prior to placement and routing. Up to 6% of the nets in a design may be designated as critical, while 90% of the nets in a design are typical.

Long Tracks

Some nets in the design use long tracks. Long tracks are special routing resources that span multiple rows, columns, or modules. Long tracks employ three and sometimes four antifuse connections. This increases capacitance and resistance, resulting in longer net delays for macros connected to long tracks. Typically up to 6% of nets in a fully utilized device require long tracks. Long tracks contribute approximately 4 ns to 14 ns delay. This additional delay is represented statistically in higher fanout (FO = 8) routing delays in the datasheet specifications section.

A1440A, A14V40A Timing Characteristics

Table 2-26 • A1440A, A14V40A Worst-Case Commercial Conditions, VCC = 4.75 V, T_J = 70°C¹

Logic Module Propagation Delays ²		–3 Speed ³		–2 Speed ³		–1 Speed		Std. Speed		3.3 V Speed ¹		Units
Parameter/Description		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t _{PD}	Internal Array Module		2.0		2.3		2.6		3.0		3.9	ns
t _{CO}	Sequential Clock to Q		2.0		2.3		2.6		3.0		3.9	ns
t _{CLR}	Asynchronous Clear to Q		2.0		2.3		2.6		3.0		3.9	ns
Predicted Routing Delays⁴												
t _{RD1}	FO = 1 Routing Delay		0.9		1.0		1.1		1.3		1.7	ns
t _{RD2}	FO = 2 Routing Delay		1.2		1.4		1.6		1.8		2.4	ns
t _{RD3}	FO = 3 Routing Delay		1.4		1.6		1.8		2.1		2.8	ns
t _{RD4}	FO = 4 Routing Delay		1.7		1.9		2.2		2.5		3.3	ns
t _{RD8}	FO = 8 Routing Delay		2.8		3.2		3.6		4.2		5.5	ns
Logic Module Sequential Timing												
t _{SUD}	Flip-Flop Data Input Setup	0.5		0.6		0.7		0.8		0.8		ns
t _{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		0.0		ns
t _{SUD}	Latch Data Input Setup	0.5		0.6		0.7		0.8		0.8		ns
t _{HD}	Latch Data Input Hold	0.0		0.0		0.0		0.0		0.0		ns
t _{WASYN}	Asynchronous Pulse Width	1.9		2.4		3.2		3.8		4.8		ns
t _{WCLKA}	Flip-Flop Clock Pulse Width	1.9		2.4		3.2		3.8		4.8		ns
t _A	Flip-Flop Clock Input Period	4.0		5.0		6.8		8.0		10.0		ns
f _{MAX}	Flip-Flop Clock Frequency		250		200		150		125		100	MHz

Notes:

1. VCC = 3.0 V for 3.3 V specifications.
2. For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn} + t_{CO} + t_{RD1} + t_{PDn} or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.
3. The –2 and –3 speed grades have been discontinued. Refer to PDN 0104, PDN 0203, PDN 0604, and PDN 1004 at <http://www.microsemi.com/soc/support/notifications/default.aspx#pdn>.
4. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

A1460A, A14V60A Timing Characteristics (continued)

Table 2-32 • A1460A, A14V60A Worst-Case Commercial Conditions, VCC = 4.75 V, T_J = 70°C

I/O Module – TTL Output Timing ¹		–3 Speed ²		–2 Speed ²		–1 Speed		Std. Speed		3.3 V Speed ¹		Units
Parameter/Description		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t _{DHS}	Data to Pad, High Slew		5.0		5.6		6.4		7.5		9.8	ns
t _{DLS}	Data to Pad, Low Slew		8.0		9.0		10.2		12.0		15.6	ns
t _{ENZHS}	Enable to Pad, Z to H/L, High Slew		4.0		4.5		5.1		6.0		7.8	ns
t _{ENZLS}	Enable to Pad, Z to H/L, Low Slew		7.4		8.3		9.4		11.0		14.3	ns
t _{ENHSZ}	Enable to Pad, H/L to Z, High Slew		7.8		8.7		9.9		11.6		15.1	ns
t _{ENLSZ}	Enable to Pad, H/L to Z, Low Slew		7.4		8.3		9.4		11.0		14.3	ns
t _{CKHS}	IOCLK Pad to Pad H/L, High Slew		9.0		9.0		10.0		11.5		15.0	ns
t _{CKLS}	IOCLK Pad to Pad H/L, Low Slew		12.8		12.8		15.3		17.0		22.1	ns
d _{TLHHS}	Delta Low to High, High Slew		0.02		0.02		0.03		0.03		0.04	ns/pF
d _{TLHLS}	Delta Low to High, Low Slew		0.05		0.05		0.06		0.07		0.09	ns/pF
d _{THLHS}	Delta High to Low, High Slew		0.04		0.04		0.04		0.05		0.07	ns/pF
d _{THLLS}	Delta High to Low, Low Slew		0.05		0.05		0.06		0.07		0.09	ns/pF
I/O Module – CMOS Output Timing ¹												
t _{DHS}	Data to Pad, High Slew		6.2		7.0		7.9		9.3		12.1	ns
t _{DLS}	Data to Pad, Low Slew		11.7		13.1		14.9		17.5		22.8	ns
t _{ENZHS}	Enable to Pad, Z to H/L, High Slew		5.2		5.9		6.6		7.8		10.1	ns
t _{ENZLS}	Enable to Pad, Z to H/L, Low Slew		8.9		10.0		11.3		13.3		17.3	ns
t _{ENHSZ}	Enable to Pad, H/L to Z, High Slew		7.4		8.3		9.4		11.0		14.3	ns
t _{ENLSZ}	Enable to Pad, H/L to Z, Low Slew		7.4		8.3		9.4		11.0		14.3	ns
t _{CKHS}	IOCLK Pad to Pad H/L, High Slew		10.4		10.4		12.1		13.8		17.9	ns
t _{CKLS}	IOCLK Pad to Pad H/L, Low Slew		14.5		14.5		17.4		19.3		25.1	ns
d _{TLHHS}	Delta Low to High, High Slew		0.04		0.04		0.05		0.06		0.08	ns/pF
d _{TLHLS}	Delta Low to High, Low Slew		0.07		0.08		0.09		0.11		0.14	ns/pF
d _{THLHS}	Delta High to Low, High Slew		0.03		0.03		0.03		0.04		0.05	ns/pF
d _{THLLS}	Delta High to Low, Low Slew		0.04		0.04		0.04		0.05		0.07	ns/pF

Notes:

1. Delays based on 35 pF loading.
2. The –2 and –3 speed grades have been discontinued. Refer to PDN 0104, PDN 0203, PDN 0604, and PDN 1004 at <http://www.microsemi.com/soc/support/notifications/default.aspx#pdn>.

A14100A, A14V100A Timing Characteristics (continued)

Table 2-35 • A14100A, A14V100A Worst-Case Commercial Conditions, VCC = 4.75 V, T_J = 70°C

I/O Module Input Propagation Delays		–3 Speed ¹		–2 Speed ¹		–1 Speed		Std. Speed		3.3 V Speed ¹		Units
Parameter/Description		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t _{INY}	Input Data Pad to Y		2.8		3.2		3.6		4.2		5.5	ns
t _{ICKY}	Input Reg IOCLK Pad to Y		4.7		5.3		6.0		7.0		9.2	ns
t _{OCKY}	Output Reg IOCLK Pad to Y		4.7		5.3		6.0		7.0		9.2	ns
t _{ICLRY}	Input Asynchronous Clear to Y		4.7		5.3		6.0		7.0		9.2	ns
t _{OCLRY}	Output Asynchronous Clear to Y		4.7		5.3		6.0		7.0		9.2	ns
Predicted Input Routing Delays²												
t _{RD1}	FO = 1 Routing Delay		0.9		1.0		1.1		1.3		1.7	ns
t _{RD2}	FO = 2 Routing Delay		1.2		1.4		1.6		1.8		2.4	ns
t _{RD3}	FO = 3 Routing Delay		1.4		1.6		1.8		2.1		2.8	ns
t _{RD4}	FO = 4 Routing Delay		1.7		1.9		2.2		2.5		3.3	ns
t _{RD8}	FO = 8 Routing Delay		2.8		3.2		3.6		4.2		5.5	ns
I/O Module Sequential Timing (wrt IOCLK pad)												
t _{INH}	Input F-F Data Hold	0.0		0.0		0.0		0.0		0.0		ns
t _{INSU}	Input F-F Data Setup	1.2		1.4		1.5		1.8		1.8		ns
t _{IDEH}	Input Data Enable Hold	0.0		0.0		0.0		0.0		0.0		ns
t _{IDESU}	Input Data Enable Setup	5.8		6.5		7.5		8.6		8.6		ns
t _{OUTH}	Output F-F Data hold	0.7		0.8		1.0		1.0		1.0		ns
t _{OUTSU}	Output F-F Data Setup	0.7		0.8		1.0		1.0		1.0		ns
t _{ODEH}	Output Data Enable Hold	0.3		0.4		0.5		0.5		0.5		ns
f _{ODESU}	Output Data Enable Setup	1.3		1.5		2.0		2.0		2.0		ns

Notes: *

1. The –2 and –3 speed grades have been discontinued. Refer to PDN 0104, PDN 0203, PDN 0604, and PDN 1004 at <http://www.microsemi.com/soc/support/notifications/default.aspx#pdn>.
2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

SDO Serial Data Output (Output)

Serial data output for diagnostic probe. SDO is active when the MODE pin is High. This pin functions as an I/O when the MODE pin is Low.

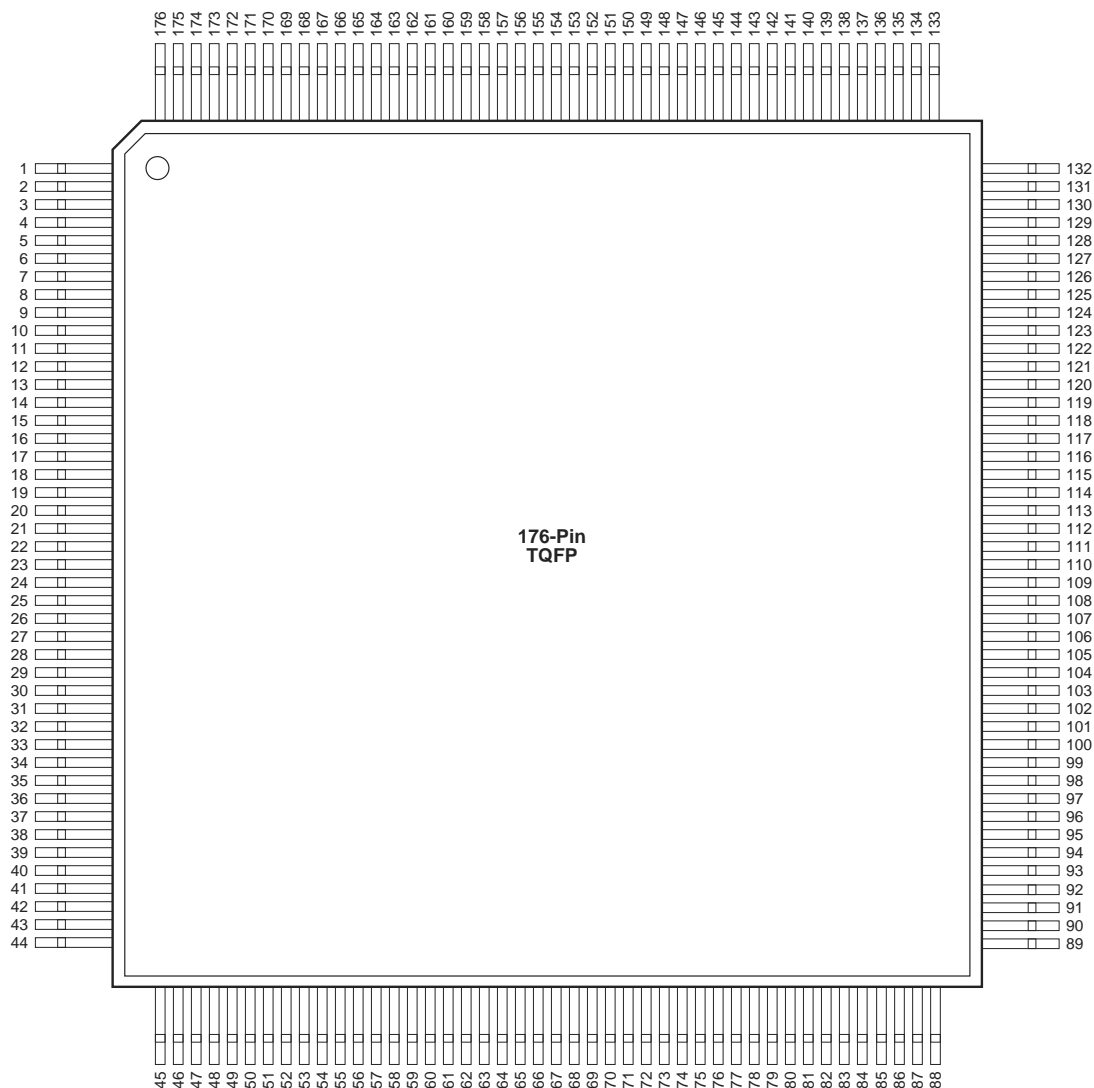
DCLK Diagnostic Clock (Input)

Clock input for diagnostic probe and device programming. DCLK is active when the MODE pin is HIGH. This pin functions as an I/O when the MODE pin is LOW.

VCC 5 V Supply Voltage

HIGH supply voltage.

TQ176

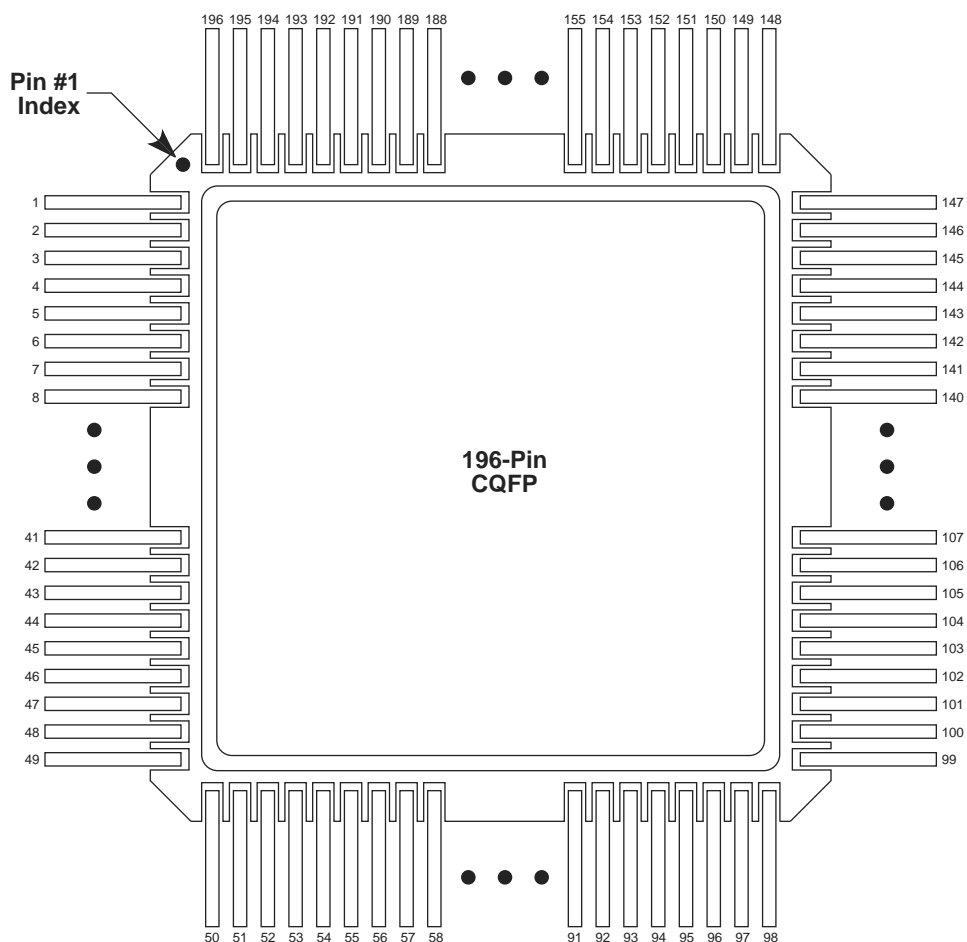


Note: This is the top view.

Note

For Package Manufacturing and Environmental information, visit the Resource Center at <http://www.microsemi.com/soc/products/solutions/package/docs.aspx>

CQ196

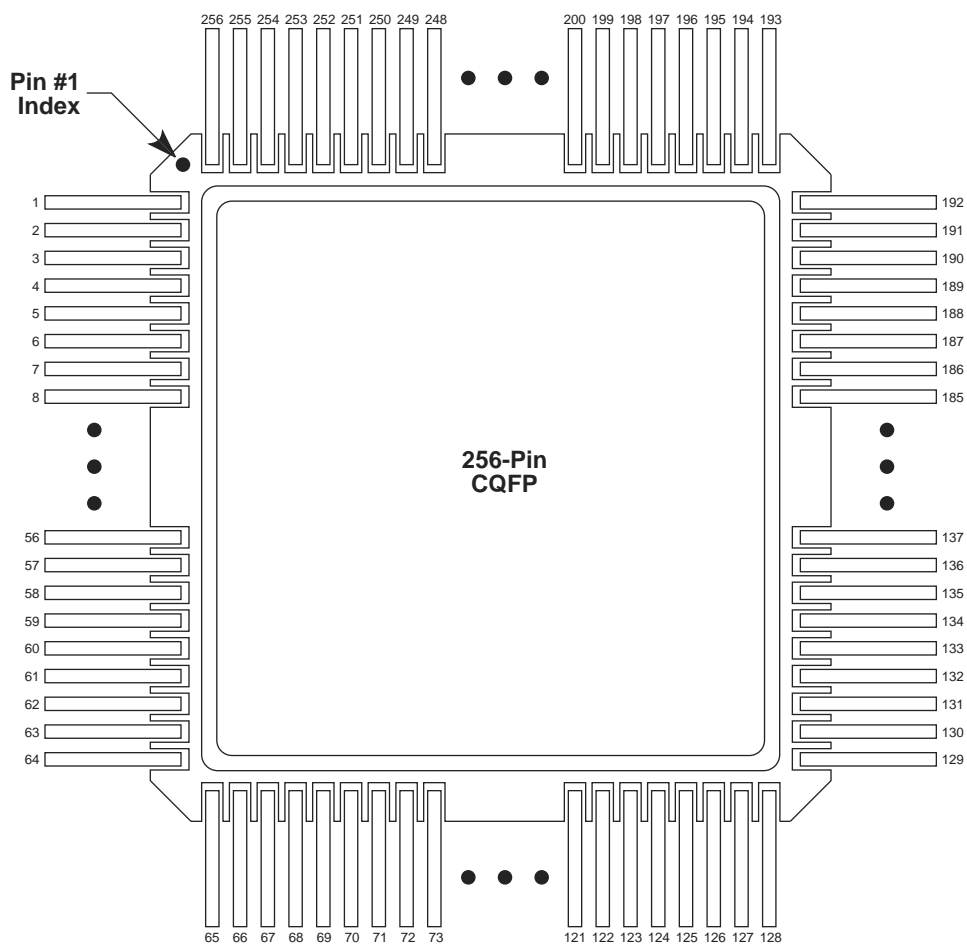


Note: This is the top view.

Note

For Package Manufacturing and Environmental information, visit the Resource Center at <http://www.microsemi.com/soc/products/solutions/package/docs.aspx>

CQ256



Note: This is the top view.

Note

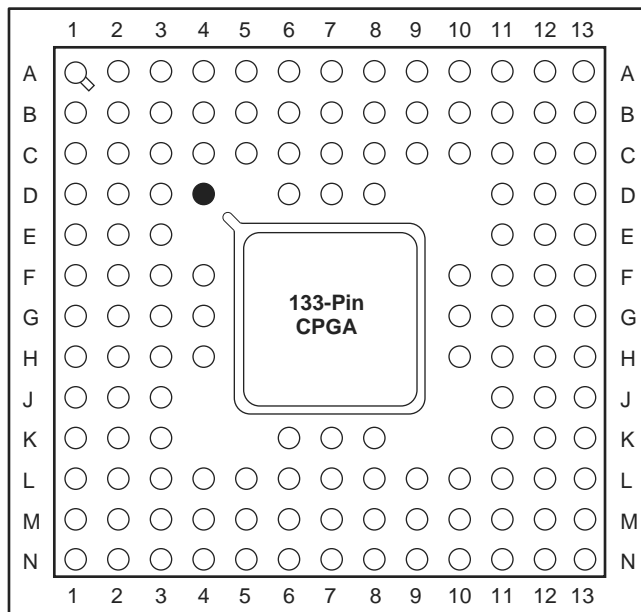
For Package Manufacturing and Environmental information, visit the Resource Center at <http://www.microsemi.com/soc/products/solutions/package/docs.aspx>

PG100	
A1415 Function	Location
CLKA or I/O	C7
CLKB or I/O	D6
DCLK or I/O	C4
GND	C3, C6, C9, E9, F3, F9, J3, J6, J8, J9
HCLK or I/O	H6
IOCLK or I/O	C10
IOPCL or I/O	K9
MODE	C2
PRA or I/O	A6
PRB or I/O	L3
SDI or I/O	B3
SDO	L9
VCC	B6, B10, E11, F2, F10, G2, K2, K6, K10

Notes:

1. All unlisted pin numbers are user I/Os.
2. NC denotes no connection.
3. MODE should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.
4. The PG100 package has been discontinued.

PG133

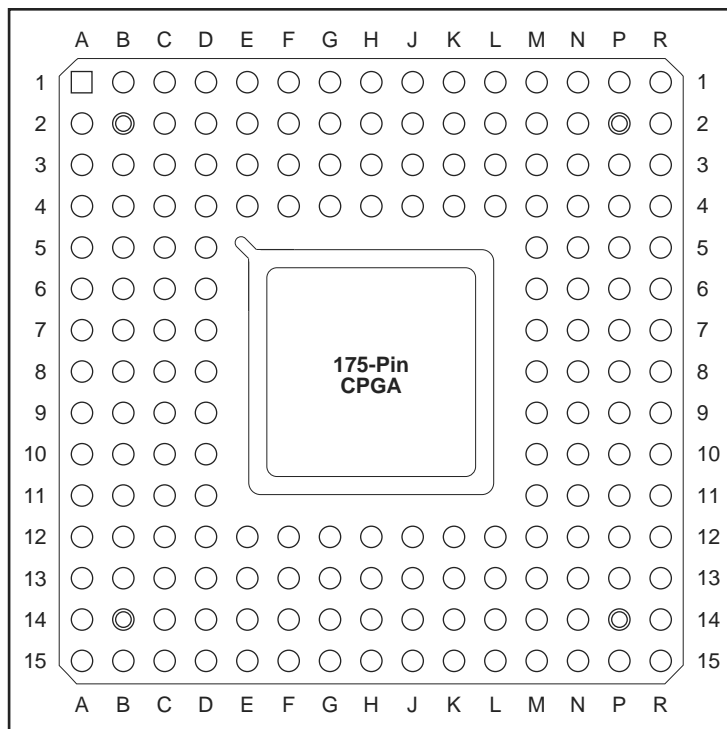


Note: This is the top view.

Note

For Package Manufacturing and Environmental information, visit the Resource Center at <http://www.microsemi.com/soc/products/solutions/package/docs.aspx>

PG175



Note: This is the top view.

Note

For Package Manufacturing and Environmental information, visit the Resource Center at <http://www.microsemi.com/soc/products/solutions/package/docs.aspx>

PG207	
A1460 Function	Location
CLKA or I/O	K1
CLKB or I/O	J3
DCLK or I/O	E4
GND	C14, D4, D5, D9, D14, J4, J14, P3, P4, P7, P9, P14, R15
HCLK or I/O	J15
IOCLK or I/O	P5
IOPCL or I/O	N14
MODE	D7
NC	A1, A2, A16, A17, B1, B17, C1, C2, S1, S3, S17, T1, T2, T16, T17
PRA or I/O	H1
PRB or I/O	K16
SDI or I/O	C3
SDO	P15
VCC	B2, B9, B16, D11, J2, J16, P12, S2, S9, S16, T5

Notes:

1. All unlisted pin numbers are user I/Os.
2. NC denotes no connection.
3. MODE should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.



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