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Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	200
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	70
Number of Gates	1500
Voltage - Supply	4.5V ~ 5.5V
Mounting Type	Surface Mount
Operating Temperature	-55°C ~ 125°C (TC)
Package / Case	84-LCC (J-Lead)
Supplier Device Package	84-PLCC (29.31x29.31)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a1415a-plg84m

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

Plastic Device Resources

Device	Logic					User	l/Os						
Series	Modules	Gates	PL84	L84 PQ100 PQ160 PQ/RQ208		PQ/RQ208	VQ100	TQ176	BG225*	BG313			
A1415	200	1500	70	80	_	_	80	_	-	_			
A1425	310	2500	70	80	100	_	83	-	-	_			
A1440	564	4000	70	_	131	_	83	140	-	_			
A1460	848	6000	_	_	131	167	_	151	168	_			
A14100	1377	10000	-	-	_	175	1	-	_	228			

Note: *Discontinued

Hermetic Device Resources

Device	Logic			User I/Os								
Series	Modules	Gates	PG100*	PG133*	PG175*	PG175* PG207		CQ132	CQ196	CQ256		
A1415	200	1500	80	_	_	_	_	_	_	_		
A1425	310	2500	-	100	-	-	_	100	_	-		
A1440	564	4000	-	_	140	_	_	_	_	_		
A1460	848	6000	-	-	_	168	_	-	168	-		
A14100	1377	10000	-	-	-	-	228	-	-	228		

Note: *Discontinued

Contact your local Microsemi SoC Products Group (formerly Actel) representative for device availability: http://www.microsemi.com/soc/contact/default.aspx.

Revision 3 V



Logic Modules

ACT 3 logic modules are enhanced versions of the 1200XL family logic modules. As in the 1200XL family, there are two types of modules: C-modules and S-modules (Figure 2-2 and Figure 2-3). The C-module is functionally equivalent to the 1200XL C-module and implements high fanin combinatorial macros, such as 5-input AND, 5-input OR, and so on. It is available for use as the CM8 hard macro. The S-module is designed to implement high-speed sequential functions within a single module.

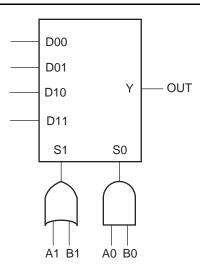


Figure 2-2 • C-Module Diagram

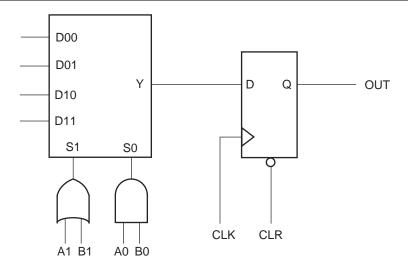


Figure 2-3 • S-Module Diagram

S-modules consist of a full C-module driving a flip-flop, which allows an additional level of logic to be implemented without additional propagation delay. It is available for use as the DFM8A/B and DLM8A/B hard macros. C-modules and S-modules are arranged in pairs called module-pairs. Module-pairs are arranged in alternating patterns and make up the bulk of the array. This arrangement allows the placement software to support two-module macros of four types (CC, CS, SC, and SS). The C-module implements the following function:

EQ 1

where: S0 = A0 * B0 and S1 = A1 + B1

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The S-module contains a full implementation of the C-module plus a clearable sequential element that can either implement a latch or flip-flop function. The S-module can therefore implement any function implemented by the C-module. This allows complex combinatorial-sequential functions to be implemented with no delay penalty. The Designer Series Development System will automatically combine any C-module macro driving an S-module macro into the S-module, thereby freeing up a logic module and eliminating a module delay.

The clear input CLR is accessible from the routing channel. In addition, the clock input may be connected to one of three clock networks: CLKA, CLKB, or HCLK. The C-module and S-module functional descriptions are shown in Figure 2-2 and Figure 2-3 on page 2-2. The clock selection is determined by a multiplexer select at the clock input to the S-module.

I/Os

I/O Modules

I/O modules provide an interface between the array and the I/O Pad Drivers. I/O modules are located in the array and access the routing channels in a similar fashion to logic modules. The I/O module schematic is shown in Figure 4. The signals DataIn and DataOut connect to the I/O pad driver.

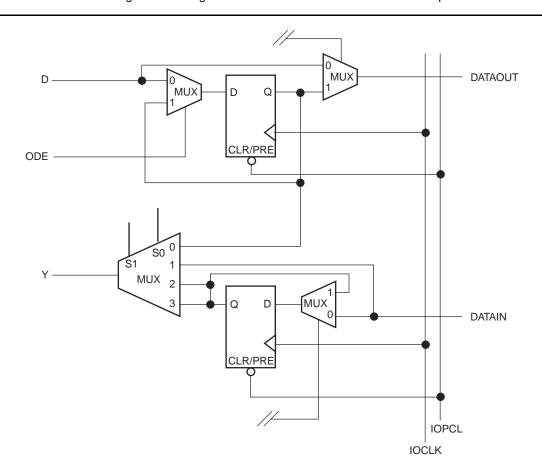


Figure 2-4 • Functional Diagram for I/O Module

Each I/O module contains two D-type flip-flops. Each flip-flop is connected to the dedicated I/O clock (IOCLK). Each flip-flop can be bypassed by nonsequential I/Os. In addition, each flip-flop contains a data enable input that can be accessed from the routing channels (ODE and IDE). The asynchronous preset/clear input is driven by the dedicated preset/clear network (IOPCL). Either preset or clear can be selected individually on an I/O module by I/O module basis.

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The I/O module output Y is used to bring Pad signals into the array or to feed the output register back into the array. This allows the output register to be used in high-speed state machine applications. Side I/O modules have a dedicated output segment for Y extending into the routing channels above and below (similar to logic modules). Top/Bottom I/O modules have no dedicated output segment. Signals coming into the chip from the top or bottom are routed using F-fuses and LVTs (F-fuses and LVTs are explained in detail in the routing section).

I/O Pad Drivers

All pad drivers are capable of being tristate. Each buffer connects to an associated I/O module with four signals: OE (Output Enable), IE (Input Enable), DataOut, and DataIn. Certain special signals used only during programming and test also connect to the pad drivers: OUTEN (global output enable), INEN (global input enable), and SLEW (individual slew selection). See Figure 2-5.

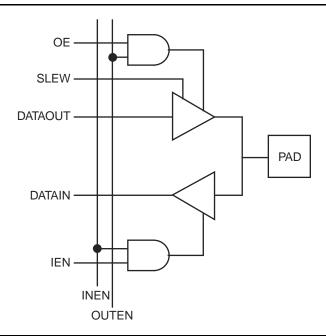


Figure 2-5 • Function Diagram for I/O Pad Driver

Special I/Os

The special I/Os are of two types: temporary and permanent. Temporary special I/Os are used during programming and testing. They function as normal I/Os when the MODE pin is inactive. Permanent special I/Os are user programmed as either normal I/Os or special I/Os. Their function does not change once the device has been programmed. The permanent special I/Os consist of the array clock input buffers (CLKA and CLKB), the hard-wired array clock input buffer (HCLK), the hard-wired I/O clock input buffer (IOCLK), and the hard-wired I/O register preset/clear input buffer (IOPCL). Their function is determined by the I/O macros selected.

Clock Networks

The ACT 3 architecture contains four clock networks: two high-performance dedicated clock networks and two general purpose routed networks. The high-performance networks function up to 200 MHz, while the general purpose routed networks function up to 150 MHz.

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3.3 V Operating Conditions

Table 2-5 • Absolute Maximum Ratings¹, Free Air Temperature Range

Symbol	Parameter	Limits	Units
VCC	DC supply voltage	−0.5 to +7.0	V
VI	Input voltage	-0.5 to VCC + 0.5	V
VO	Output voltage	-0.5 to VCC + 0.5	V
IIO	I/O source sink current ²	±20	mA
T _{STG}	Storage temperature	-65 to +150	°C

Notes:

- 1. Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum rated conditions for extended periods may affect device reliability. Device should not be operated outside the recommended operating conditions.
- 2. Device inputs are normally high impedance and draw extremely low current. However, when input voltage is greater than VCC + 0.5 V for less than GND -0.5 V, the internal protection diodes will forward bias and can draw excessive current.

Table 2-6 • Recommended Operating Conditions

Parameter	Commercial	Units
Temperature range*	0 to +70	°C
Power supply tolerance	3.0 to 3.6	V

Note: *Ambient temperature (T_A) is used for commercial.

Table 2-7 • Electrical Specifications

		С	ommercial	
Parameter		Min.	Max.	Units
VOH ¹	IOH = -4 mA	2.15	_	V
	IOH = −3.2 mA	2.4		V
VOL ¹	IOL = 6 mA		0.4	V
VIL		-0.3	0.8	V
VIH		2.0	VCC + 0.3	V
Input transition time t _R , t _F ²	VI = VCC or GND	-10	+10	μA
C _{IO} I/O Capacitance ^{2,3}			10	pF
Standby current, ICC ⁴ (typical =	0.3 mA)		0.75	mA
Leakage current ⁵		-10	10	μA

- 1. Only one output tested at a time. VCC = minimum.
- 2. Not tested; for information only.
- 3. Includes worst-case 84-pin PLCC package capacitance. VOUT = 0 V, f 1 MHz.
- 4. Typical standby current = 0.3 mA. All outputs unloaded. All inputs = VCC or GND.
- 5. VO, VIN = VCC or GND

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Power Dissipation

P = [ICC standby + lactive] * VCC * IOL * VOL * N + IOH* (VCC - VOH) * M

EQ3

where:

ICC standby is the current flowing when no inputs or outputs are changing

lactive is the current flowing due to CMOS switching.

IOL and IOH are TTL sink/source current.

VOL and VOH are TTL level output voltages.

N is the number of outputs driving TTL loads to VOL.

M equals the number of outputs driving TTL loads to VOH.

An accurate determination of N and M is problematical because their values depend on the design and on the system I/O. The power can be divided into two components: static and active.

Static Power Component

Microsemi FPGAs have small static power components that result in lower power dissipation than PALs or PLDs. By integrating multiple PALs/PLDs into one FPGA, an even greater reduction in board-level power dissipation can be achieved.

The power due to standby current is typically a small component of the overall power. Standby power is calculated in Table 2-9 for commercial, worst case conditions.

Table 2-9 • Standby Power Calculation

ICC	VCC	Power
2 mA	5.25 V	10.5 mW

The static power dissipated by TTL loads depends on the number of outputs driving high or low and the DC load current. Again, this value is typically small. For instance, a 32-bit bus sinking 4 mA at 0.33 V will generate 42 mW with all outputs driving low, and 140 mW with all outputs driving high. The actual dissipation will average somewhere between as I/Os switch states with time.

Active Power Component

Power dissipation in CMOS devices is usually dominated by the active (dynamic) power dissipation. This component is frequency dependent, a function of the logic and the external I/O. Active power dissipation results from charging internal chip capacitances of the interconnect, unprogrammed antifuses, module inputs, and module outputs, plus external capacitance due to PC board traces and load device inputs.

An additional component of the active power dissipation is the totem-pole current in CMOS transistor pairs. The net effect can be associated with an equivalent capacitance that can be combined with frequency and voltage to represent active power dissipation.

Equivalent Capacitance

The power dissipated by a CMOS circuit can be expressed by EQ 4.

Power (
$$\mu$$
W) = C_{EQ} * VCC² * F

EQ 4

Where:

C_{FO} is the equivalent capacitance expressed in pF.

VCC is the power supply in volts.

F is the switching frequency in MHz.

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Tightest Delay Distributions

Propagation delay between logic modules depends on the resistive and capacitive loading of the routing tracks, the interconnect elements, and the module inputs being driven. Propagation delay increases as the length of routing tracks, the number of interconnect elements, or the number of inputs increases.

From a design perspective, the propagation delay can be statistically correlated or modeled by the fanout (number of loads) driven by a module. Higher fanout usually requires some paths to have longer lengths of routing track. The ACT 3 family delivers the tightest fanout delay distribution of any FPGA. This tight distribution is achieved in two ways: by decreasing the delay of the interconnect elements and by decreasing the number of interconnect elements per path.

Microsemi's patented PLICE antifuse offers a very low resistive/capacitive interconnect. The ACT 3 family's antifuses, fabricated in 0.8 micron m lithography, offer nominal levels of 200Ω resistance and 6 femtofarad (fF) capacitance per antifuse. The ACT 3 fanout distribution is also tighter than alternative devices due to the low number of antifuses required per interconnect path. The ACT 3 family's proprietary architecture limits the number of antifuses per path to only four, with 90% of interconnects using only two antifuses.

The ACT 3 family's tight fanout delay distribution offers an FPGA design environment in which fanout can be traded for the increased performance of reduced logic level designs. This also simplifies performance estimates when designing with ACT 3 devices.

Speed Grade	FO = 1	FO = 2	FO = 3	FO = 4	FO = 8					
ACT 3 –3	2.9	3.2	3.4	3.7	4.8					
ACT 3 –2	3.3	3.7	3.9	4.2	5.5					
ACT 3 –1	3.7	4.2	4.4	4.8	6.2					
ACT 3 STD	4.3	4.8	5.1	5.5	7.2					

Table 2-14 • Logic Module and Routing Delay by Fanout (ns); Worst-Case Commercial Conditions

Notes:

- Obtained by added t_{RD(X=FO)} to t_{PD} from the Logic Module Timing Characteristics Tables found in this datasheet.
- 2. The –2 and –3 speed grades have been discontinued. Refer to PDN 0104, PDN 0203, PDN 0604, and PDN 1004 at http://www.microsemi.com/soc/support/notifications/default.aspx#pdn.

Timing Characteristics

Timing characteristics for ACT 3 devices fall into three categories: family dependent, device dependent, and design dependent. The input and output buffer characteristics are common to all ACT 3 family members. Internal routing delays are device dependent. Design dependency means actual delays are not determined until after placement and routing of the user's design is complete. Delay values may then be determined by using the ALS Timer utility or performing simulation with post-layout delays.

Critical Nets and Typical Nets

Propagation delays are expressed only for typical nets, which are used for initial design performance evaluation. Critical net delays can then be applied to the most time-critical paths. Critical nets are determined by net property assignment prior to placement and routing. Up to 6% of the nets in a design may be designated as critical, while 90% of the nets in a design are typical.

Long Tracks

Some nets in the design use long tracks. Long tracks are special routing resources that span multiple rows, columns, or modules. Long tracks employ three and sometimes four antifuse connections. This increases capacitance and resistance, result ng in longer net delays for macros connected to long tracks. Typically up to 6% of nets in a fully utilized device require long tracks. Long tracks contribute approximately 4 ns to 14 ns delay. This additional delay is represented statistically in higher fanout (FO=8) routing delays in the datasheet specifications section.

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A1415A, A14V15A Timing Characteristics (continued)

Table 2-20 • A1415A, A14V15A Worst-Case Commercial Conditions, VCC = 4.75 V, $T_J = 70^{\circ}$ C

I/O Mod	lule – TTL Output Timing ¹	-3 Sp	peed ²	-2 S	peed ²	-1 S	peed	Std.	Speed	3.3 V Speed ¹		Units
Parame	eter/Description	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t _{DHS}	Data to Pad, High Slew		5.0		5.6		6.4		7.5		9.8	ns
t _{DLS}	Data to Pad, Low Slew		8.0		9.0		10.2		12.0		15.6	ns
t _{ENZHS}	Enable to Pad, Z to H/L, High Slew		4.0		4.5		5.1		6.0		7.8	ns
t _{ENZLS}	Enable to Pad, Z to H/L, Low Slew		7.4		8.3		9.4		11.0		14.3	ns
t _{ENHSZ}	Enable to Pad, H/L to Z, High Slew		6.5		7.5		8.5		10.0		13.0	ns
t _{ENLSZ}	Enable to Pad, H/L to Z, Low Slew		6.5		7.5		8.5		10.0		13.0	ns
t _{CKHS}	IOCLK Pad to Pad H/L, High Slew		7.5		7.5		9.0		10.0		13.0	ns
t _{CKLS}	IOCLK Pad to Pad H/L, Low Slew		11.3		11.3		13.5		15.0		19.5	ns
d _{TLHHS}	Delta Low to High, High Slew		0.02		0.02		0.03		0.03		0.04	ns/pF
d _{TLHLS}	Delta Low to High, Low Slew		0.05		0.05		0.06		0.07		0.09	ns/pF
d _{THLHS}	Delta High to Low, High Slew		0.04		0.04		0.04		0.05		0.07	ns/pF
d _{THLLS}	Delta High to Low, Low Slew		0.05		0.05		0.06		0.07		0.09	ns/pF
I/O Mod	lule – CMOS Output Timing ¹											
t _{DHS}	Data to Pad, High Slew		6.2		7.0		7.9		9.3		12.1	ns
t _{DLS}	Data to Pad, Low Slew		11.7		13.1		14.9		17.5		22.8	ns
t _{ENZHS}	Enable to Pad, Z to H/L, High Slew		5.2		5.9		6.6		7.8		10.1	ns
t _{ENZLS}	Enable to Pad, Z to H/L, Low Slew		8.9		10.0		11.3		13.3		17.3	ns
t _{ENHSZ}	Enable to Pad, H/L to Z, High Slew		6.7		7.5		8.5		10.0		13.0	ns
t _{ENLSZ}	Enable to Pad, H/L to Z, Low Slew		6.7		7.5		9.0		10.0		13.0	ns
t _{CKHS}	IOCLK Pad to Pad H/L, High Slew		8.9		8.9		10.7		11.8		15.3	ns
t _{CKLS}	IOCLK Pad to Pad H/L, Low Slew		13.0		13.0		15.6		17.3		22.5	ns
d_TLHHS	Delta Low to High, High Slew		0.04		0.04		0.05		0.06		0.08	ns/pF
d _{TLHLS}	Delta Low to High, Low Slew		0.07		0.08		0.09		0.11		0.14	ns/pF
d _{THLHS}	Delta High to Low, High Slew		0.03		0.03		0.03		0.04		0.05	ns/pF
d _{THLLS}	Delta High to Low, Low Slew		0.04		0.04		0.04		0.05		0.07	ns/pF

Notes:

1. Delays based on 35 pF loading.

2. The -2 and -3 speed grades have been discontinued. Please refer to the Product Discontinuation Notices (PDNs) listed below:

PDN March 2001

PDN 0104

PDN 0203

PDN 0604

PDN 1004

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A1415A, A14V15A Timing Characteristics (continued)

Table 2-21 • A1415A, A14V15A Worst-Case Commercial Conditions, VCC = 4.75 V, $T_J = 70^{\circ}$ C

Dedicate	d (hardwired) I/O Clock Network	-3 S	peed	-2 S	peed	-1 Speed		Std. Speed		3.3 V Speed ¹		¹ Units
Paramete	er/Description	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t _{IOCKH}	Input Low to High (pad to I/O module input)		2.0		2.3		2.6		3.0		3.5	ns
t _{IOPWH}	Minimum Pulse Width High	1.9		2.4		3.3		3.8		4.8		ns
t _{IPOWL}	Minimum Pulse Width Low	1.9		2.4		3.3		3.8		4.8		ns
t _{IOSAPW}	Minimum Asynchronous Pulse Width	1.9		2.4		3.3		3.8		4.8		ns
t _{IOCKSW}	Maximum Skew		0.4		0.4		0.4		0.4		0.4	ns
t _{IOP}	Minimum Period	4.0		5.0		6.8		8.0		10.0		ns
f _{IOMAX}	Maximum Frequency		250		200		150		125		100	MHz
Dedicated	d (hardwired) Array Clock			•	•	•	•					
t _{HCKH}	Input Low to High (pad to S-module input)		3.0		3.4		3.9		4.5		5.5	ns
t _{HCKL}	Input High to Low (pad to S-module input)		3.0		3.4		3.9		4.5		5.5	ns
t _{HPWH}	Minimum Pulse Width High	1.9		2.4		3.3		3.8		4.8		ns
t _{HPWL}	Minimum Pulse Width Low	1.9		2.4		3.3		3.8		4.8		ns
t _{HCKSW}	Delta High to Low, Low Slew		0.3		0.3		0.3		0.3		0.3	ns
t _{HP}	Minimum Period	4.0		5.0		6.8		8.0		10.0		ns
f _{HMAX}	Maximum Frequency		250		200		150		125		100	MHz
Routed A	rray Clock Networks				•					•		
t _{RCKH}	Input Low to High (FO = 64)		3.7		4.1		4.7		5.5		9.0	ns
t _{RCKL}	Input High to Low (FO = 64)		4.0		4.5		5.1		6.0		9.0	ns
t _{RPWH}	Min. Pulse Width High (FO = 64)	3.3		3.8		4.2		4.9		6.5		ns
t _{RPWL}	Min. Pulse Width Low (FO = 64)	3.3		3.8		4.2		4.9		6.5		ns
t _{RCKSW}	Maximum Skew (FO = 128)		0.7		0.8		0.9		1.0		1.0	ns
t _{RP}	Minimum Period (FO = 64)	6.8		8.0		8.7		10.0		13.4		ns
f _{RMAX}	Maximum Frequency (FO = 64)		150		125		115		100		75	MHz
Clock-to-	Clock Skews		•	•	•		•			-		
t _{IOHCKSW}	I/O Clock to H-Clock Skew	0.0	1.7	0.0	1.8	0.0	2.0	0.0	2.2	0.0	3.0	ns
t _{IORCKSW}	I/O Clock to R-Clock Skew (FO = 64)	0.0	1.0	0.0	1.0	0.0	1.0	0.0	1.0	0.0	3.0	ns
t _{HRCKSW}	H-Clock to R-Clock Skew (FO = 64) (FO = 50% maximum)	0.0	1.0	0.0	1.0	0.0	1.0	0.0	1.0	0.0 0.0	3.0 3.0	ns

Notes:

- 1. Delays based on 35 pF loading.
- 2. The –2 and –3 speed grades have been discontinued. Refer to PDN 0104, PDN 0203, PDN 0604, and PDN 1004 at http://www.microsemi.com/soc/support/notifications/default.aspx#pdn.

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A1425A, A14V25A Timing Characteristics

Table 2-22 • A1425A, A14V25A Worst-Case Commercial Conditions, VCC = 4.75 V, T_J = 70°C¹

Logic N	Logic Module Propagation Delays ²		peed ³	-2 S _l	peed ³	–1 S	peed	Std. S	Std. Speed 3.3 V Speed		Speed ¹	Units
Parame	eter/Description	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t _{PD}	Internal Array Module		2.0		2.3		2.6		3.0		3.9	ns
t _{CO}	Sequential Clock to Q		2.0		2.3		2.6		3.0		3.9	ns
t _{CLR}	Asynchronous Clear to Q		2.0		2.3		2.6		3.0		3.9	ns
Predict	ed Routing Delays ⁴											
t _{RD1}	FO = 1 Routing Delay		0.9		1.0		1.1		1.3		1.7	ns
t _{RD2}	FO = 2 Routing Delay		1.2		1.4		1.6		1.8		2.4	ns
t _{RD3}	FO = 3 Routing Delay		1.4		1.6		1.8		2.1		2.8	ns
t _{RD4}	FO = 4 Routing Delay		1.7		1.9		2.2		2.5		3.3	ns
t _{RD8}	FO = 8 Routing Delay		2.8		3.2		3.6		4.2		5.5	ns
Logic N	Module Sequential Timing											•
t _{SUD}	Flip-Flop Data Input Setup	0.5		0.6		0.7		0.8		0.8		ns
t _{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		0.0		ns
t _{SUD}	Latch Data Input Setup	0.5		0.6		0.7		0.8		0.8		ns
t _{HD}	Latch Data Input Hold	0.0		0.0		0.0		0.0		0.0		ns
t _{WASYN}	Asynchronous Pulse Width	1.9		2.4		3.2		3.8		4.8		ns
t _{WCLKA}	Flip-Flop Clock Pulse Width	1.9		2.4		3.2		3.8		4.8		ns
t _A	Flip-Flop Clock Input Period	4.0		5.0		6.8		8.0		10.0		ns
f _{MAX}	Flip-Flop Clock Frequency		250		200		150		125		100	MHz

Notes:

- 1. VCC = 3.0 V for 3.3 V specifications.
- 2. For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn} + t_{CO} + t_{RD1} + t_{PDn}$ or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.
- 3. The -2 and -3 speed grades have been discontinued. Refer to PDN 0104, PDN 0203, PDN 0604, and PDN 1004 at http://www.microsemi.com/soc/support/notifications/default.aspx#pdn.
- 4. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

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Accelerator Series FPGAs – ACT 3 Family

SDO Serial Data Output (Output)

Serial data output for diagnostic probe. SDO is active when the MODE pin is High. This pin functions as an I/O when the MODE pin is Low.

DCLK Diagnostic Clock (Input)

Clock input for diagnostic probe and device programming. DCLK is active when the MODE pin is HIGH. This pin functions as an I/O when the MODE pin is LOW.

VCC 5 V Supply Voltage

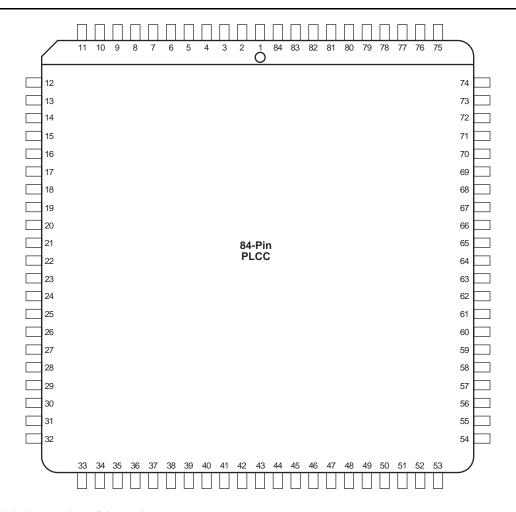
HIGH supply voltage.

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3 – Package Pin Assignments

PL84



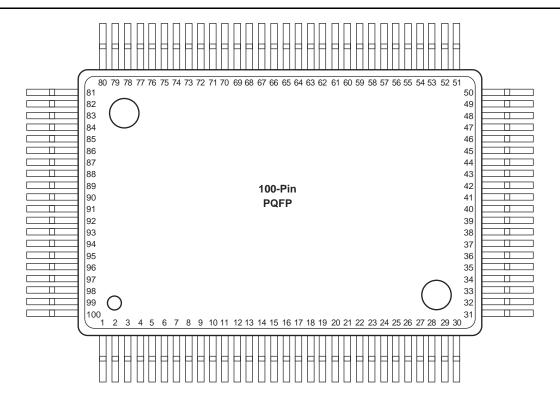
Note: This is the top view of the package.

Note

For Package Manufacturing and Environmental information, visit the Resource Center at http://www.microsemi.com/soc/products/solutions/package/docs.aspx.

Revision 3 3-1

PQ100



Note: This is the top view of the package.

Note

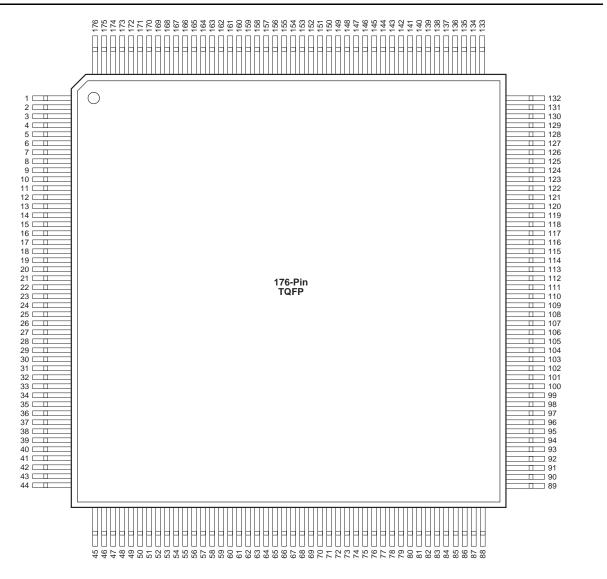
For Package Manufacturing and Environmental information, visit the Resource Center at http://www.microsemi.com/soc/products/solutions/package/docs.aspx

Revision 3 3-3



Package Pin Assignments

TQ176



Note: This is the top view.

Note

For Package Manufacturing and Environmental information, visit the Resource Center at http://www.microsemi.com/soc/products/solutions/package/docs.aspx

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	TQ176	
Pin Number	A1440, A14V40 Function	A1460, A14V60 Function
1	GND	GND
2	SDI, I/O	SDI, I/O
10	MODE	MODE
11	VCC	VCC
20	NC	I/O
21	GND	GND
22	VCC	VCC
23	GND	GND
32	VCC	VCC
33	VCC	VCC
44	GND	GND
49	NC	I/O
51	NC	I/O
63	NC	I/O
64	PRB, I/O	PRB, I/O
65	GND	GND
66	VCC	VCC
67	VCC	VCC
69	HCLK, I/O	HCLK, I/O
82	NC	I/O
83	NC	I/O
87	SDO	SDO
88	IOPCL, I/O	IOPCL, I/O

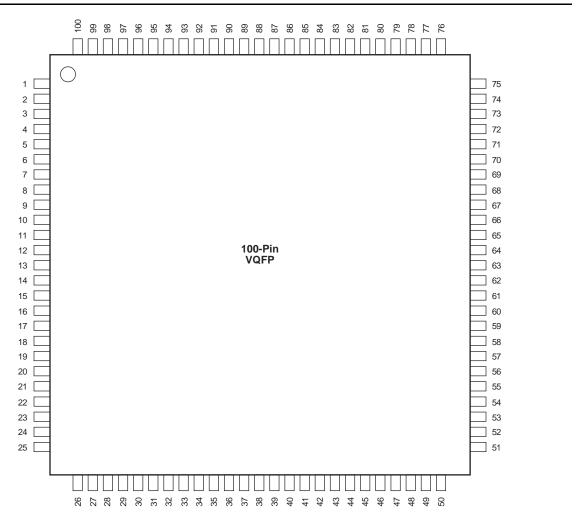
TQ176			
Pin Number	A1440, A14V40 Function	A1460, A14V60 Function	
89	GND	GND	
98	VCC	VCC	
99	VCC	VCC	
108	GND	GND	
109	VCC	VCC	
110	GND	GND	
119	NC	I/O	
121	NC	I/O	
122	VCC	VCC	
123	GND	GND	
124	VCC	VCC	
132	IOCLK, I/O	IOCLK, I/O	
133	GND	GND	
138	NC	I/O	
152	CLKA, I/O	CLKA, I/O	
153	CLKB, I/O	CLKB, I/O	
154	VCC	VCC	
155	GND	GND	
156	VCC	VCC	
157	PRA, I/O	PRA, I/O	
158	NC	I/O	
170	NC	I/O	
176	DCLK, I/O	DCLK, I/O	

Notes:

- 1. All unlisted pin numbers are user I/Os.
- 2. NC denotes no connection.
- 3. MODE should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.

Revision 3 3-11

VQ100



Note: This is the top view.

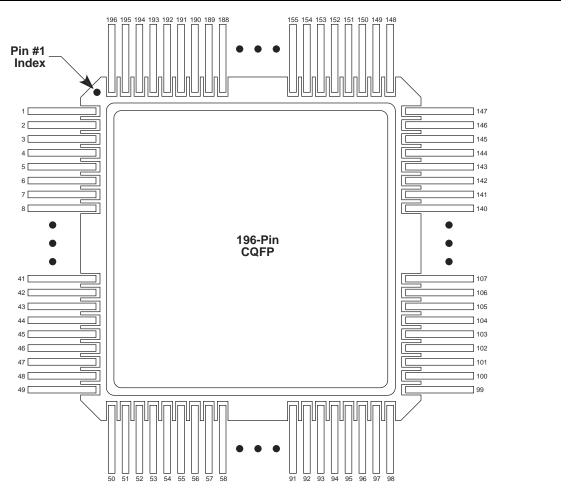
Note

For Package Manufacturing and Environmental information, visit the Resource Center at $\label{eq:http://www.microsemi.com/soc/products/solutions/package/docs.aspx$

3-12 Revision 3

Package Pin Assignments

CQ196



Note: This is the top view.

Note

For Package Manufacturing and Environmental information, visit the Resource Center at http://www.microsemi.com/soc/products/solutions/package/docs.aspx

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Accelerator Series FPGAs – ACT 3 Family

CQ196		
Pin Number	A1460 Function	
1	GND	
2	SDI, I/O	
11	MODE	
12	VCC	
13	GND	
37	GND	
38	VCC	
39	VCC	
51	GND	
52	GND	
59	VCC	
64	GND	
77	HCLK, I/O	
79	PRB, I/O	
86	GND	
94	VCC	
98	GND	
99	SDO	
100	IOPCL, I/O	

CQ196		
Pin Number	A1460 Function	
101	GND	
110	VCC	
111	VCC	
112	GND	
137	VCC	
138	GND	
139	GND	
140	VCC	
148	IOCLK, I/O	
149	GND	
155	VCC	
162	GND	
172	CLKA, I/O	
173	CLKB, I/O	
174	PRA, I/O	
183	GND	
189	VCC	
193	GND	
196	DCLK, I/O	

Notes:

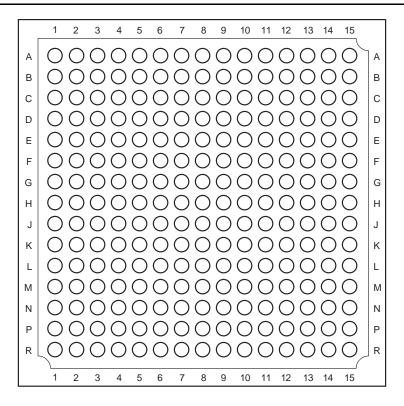
- 1. All unlisted pin numbers are user I/Os.
- 2. NC denotes no connection.
- 3. MODE should be terminated to GND through a 10K resistor to enable Actionprobe usage; otherwise it can be terminated directly to GND.

Revision 3 3-17



Package Pin Assignments

BG225



Note: This is the top view.

Note

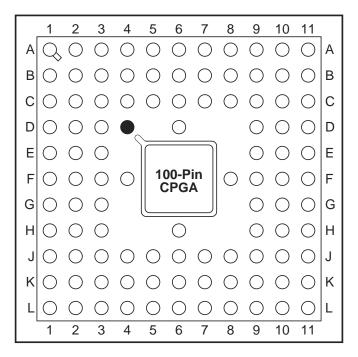
For Package Manufacturing and Environmental information, visit the Resource Center at http://www.microsemi.com/soc/products/solutions/package/docs.aspx

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Package Pin Assignments

PG100



Orientation Pin

Note: This is the top view.

Note

For Package Manufacturing and Environmental information, visit the Resource Center at http://www.microsemi.com/soc/products/solutions/package/docs.aspx

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