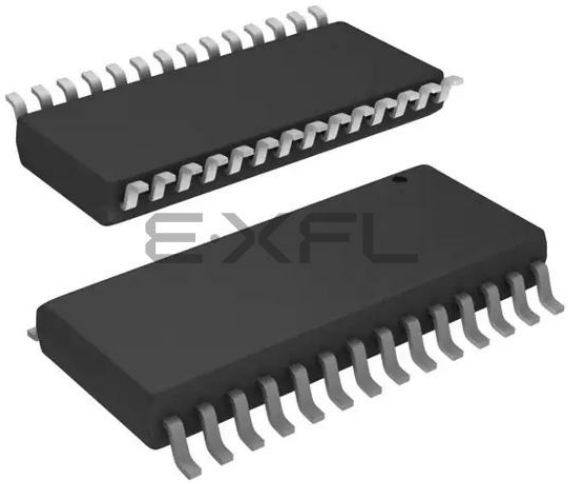

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	48MHz
Connectivity	I²C, LINbus, SPI, UART/USART, USB
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	22
Program Memory Size	64KB (32K x 16)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	3.8K x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 2.75V
Data Converters	A/D 10x10b/12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-SOIC (0.295", 7.50mm Width)
Supplier Device Package	28-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf26j53t-i-so

6.3.5 SPECIAL FUNCTION REGISTERS

The SFRs are registers used by the CPU and peripheral modules for controlling the desired operation of the device. These registers are implemented as static RAM. SFRs start at the top of data memory (FFFh) and extend downward to occupy more than the top half of Bank 15 (F40h to FFFh). Table 6-2, Table 6-3 and Table 6-4 provide a list of these registers.

The SFRs can be classified into two sets: those associated with the "core" device functionality (ALU, Resets and interrupts) and those related to the peripheral functions. The Reset and Interrupt registers are described in their corresponding chapters, while the ALU's STATUS register is described later in this section. Registers related to the operation of the peripheral features are described in the chapter for that peripheral.

The SFRs are typically distributed among the peripherals whose functions they control. Unused SFR locations are unimplemented and read as '0's

Note:	The SFRs located between EB0h and F5Fh are not part of the Access Bank. Either BANKED instructions (using BSR) or the MOVFF instruction should be used to access these locations. When programming in MPLAB® C18, the compiler will automatically use the appropriate addressing mode.

TABLE 6-2: ACCESS BANK SPECIAL FUNCTION REGISTER MAP

Address	Name	Address	Name	Address	Name	Address	Name	Address	Name
FFFh	TOSU	FDFh	INDF2(1)	FBFh	PSTR1CON	F9Fh	IPR1	F7Fh	SPBRGH1
FFEh	TOSH	FDEh	POSTINC2(1)	FBEh	ECCP1AS	F9Eh	PIR1	F7Eh	BAUDCON1
FFDh	TOSL	FDDh	POSTDEC2(1)	FBDh	ECCP1DEL	F9Dh	PIE1	F7Dh	SPBRGH2
FFCh	STKPTR	FDCh	PREINC2(1)	FBCh	CCPR1H	F9Ch	RCSTA2	F7Ch	BAUDCON2
FFBh	PCLATU	FDBh	PLUSW2(1)	FBBh	CCPR1L	F9Bh	OSCTUNE	F7Bh	TMR3H
FFAh	PCLATH	FDAh	FSR2H	FBAh	CCP1CON	F9Ah	T1GCON	F7Ah	TMR3L
FF9h	PCL	FD9h	FSR2L	FB9h	PSTR2CON	F99h	IPR5	F79h	T3CON
FF8h	TBLPTRU	FD8h	STATUS	FB8h	ECCP2AS	F98h	PIR5	F78h	TMR4
FF7h	TBLPTRH	FD7h	TMR0H	FB7h	ECCP2DEL	F97h	T3GCON	F77h	PR4
FF6h	TBLPTRL	FD6h	TMR0L	FB6h	CCPR2H	F96h	TRISE	F76h	T4CON
FF5h	TABLAT	FD5h	T0CON	FB5h	CCPR2L	F95h	TRISD	F75h	SSP2BUF
FF4h	PRODH	FD4h	—(5)	FB4h	CCP2CON	F94h	TRISC	F74h	SSP2ADD(3)
FF3h	PRODL	FD3h	OSCCON	FB3h	CTMUCONH	F93h	TRISB	F73h	SSP2STAT
FF2h	INTCON	FD2h	CM1CON	FB2h	CTMUCONL	F92h	TRISA	F72h	SSP2CON1
FF1h	INTCON2	FD1h	CM2CON	FB1h	CTMUICON	F91h	PIE5	F71h	SSP2CON2
FF0h	INTCON3	FD0h	RCON	FB0h	SPBRG1	F90h	IPR4	F70h	CMSTAT
FEFh	INDF0(1)	FCFh	TMR1H	FAFh	RCREG1	F8Fh	PIR4	F6Fh	PMADDRH(2,4)
FEEh	POSTINC0(1)	FCEh	TMR1L	FAEh	TXREG1	F8Eh	PIE4	F6Eh	PMADDRL(2,4)
FEDh	POSTDEC0(1)	FCDh	T1CON	FADh	TXSTA1	F8Dh	LATE(2)	F6Dh	PMDIN1H(2)
FECh	PREINC0(1)	FCCh	TMR2	FACh	RCSTA1	F8Ch	LATD(2)	F6Ch	PMDIN1L(2)
FEBh	PLUSW0(1)	FCBh	PR2	FABh	SPBRG2	F8Bh	LATC	F6Bh	TXADDRL
FEAh	FSR0H	FCAh	T2CON	FAAh	RCREG2	F8Ah	LATB	F6Ah	TXADDRH
FE9h	FSR0L	FC9h	SSP1BUF	FA9h	TXREG2	F89h	LATA	F69h	RXADDRL
FE8h	WREG	FC8h	SSP1ADD(3)	FA8h	TXSTA2	F88h	DMACON1	F68h	RXADDRH
FE7h	INDF1(1)	FC7h	SSP1STAT	FA7h	EECON2	F87h	OSCCON2(5)	F67h	DMABCL
FE6h	POSTINC1(1)	FC6h	SSP1CON1	FA6h	EECON1	F86h	DMACON2	F66h	DMABCH
FE5h	POSTDEC1(1)	FC5h	SSP1CON2	FA5h	IPR3	F85h	HLVDCON	F65h	UCON
FE4h	PREINC1(1)	FC4h	ADRESH	FA4h	PIR3	F84h	PORTE(2)	F64h	USTAT
FE3h	PLUSW1(1)	FC3h	ADRESL	FA3h	PIE3	F83h	PORTD(2)	F63h	UEIR
FE2h	FSR1H	FC2h	ADCON0	FA2h	IPR2	F82h	PORTC	F62h	UIR
FE1h	FSR1L	FC1h	ADCON1	FA1h	PIR2	F81h	PORTB	F61h	UFRMH
FE0h	BSR	FC0h	WDTCON	FA0h	PIE2	F80h	PORTA	F60h	UFRML

Note 1: This is not a physical register.

2: This register is not available on 28-pin devices.

3: SSPxADD and SSPxMSK share the same address.

4: PMADDRH and PMDOUTH share the same address and PMADDRL and PMDOUTL share the same address. PMADDRx is used in Master modes and PMDOUTx is used in Slave modes.

5: Reserved: Do not write to this location.

REGISTER 9-17: IPR4: PERIPHERAL INTERRUPT PRIORITY REGISTER 4 (ACCESS F90h)

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
CCP10IP	CCP9IP	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-1 **CCP10IP:CCP4IP:** CCP<10:4> Interrupt Priority bits
1 = High priority
0 = Low priority

bit 0 **CCP3IP:** ECCP3 Interrupt Priority bit
1 = High priority
0 = Low priority

REGISTER 9-18: IPR5: PERIPHERAL INTERRUPT PRIORITY REGISTER 5 (ACCESS F99h)

U-0	U-0	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
—	—	CM3IP	TMR8IP	TMR6IP	TMR5IP	TMR5GIP	TMR1GIP
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-6 **Unimplemented:** Read as ‘0’

bit 5 **CM3IP:** Comparator3 Interrupt Priority bit
1 = High priority
0 = Low priority

bit 4 **TMR8IP:** TMR8 to PR8 Match Interrupt Priority bit
1 = High priority
0 = Low priority

bit 3 **TMR6IP:** TMR6 to PR6 Match Interrupt Priority bit
1 = High priority
0 = Low priority

bit 2 **TMR5IP:** TMR5 Overflow Interrupt Priority bit
1 = High priority
0 = Low priority

bit 1 **TMR5GIP:** TMR5 Gate Interrupt Priority bit
1 = High priority
0 = Low priority

bit 0 **TMR1GIP:** TMR1 Gate Interrupt Priority bit
1 = High priority
0 = Low priority

REGISTER 10-8: RPINR3: PERIPHERAL PIN SELECT INPUT REGISTER 3 (BANKED EE3h)

U-0	U-0	U-0	R/$\overline{W}$-1	R/$\overline{W}$-1	R/$\overline{W}$-1	R/$\overline{W}$-1	R/$\overline{W}$-1
—	—	—	INTR3R4	INTR3R3	INTR3R2	INTR3R1	INTR3R0
bit 7							bit 0

Legend:	R/$\overline{W}$ = Readable bit, Writable bit if IOLOCK = 0		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5 **Unimplemented:** Read as '0'

bit 4-0 **INTR3R<4:0>:** Assign External Interrupt 3 (INT3) to the Corresponding RPn Pin bits

REGISTER 10-9: RPINR4: PERIPHERAL PIN SELECT INPUT REGISTER 4 (BANKED EE4h)

U-0	U-0	U-0	R/$\overline{W}$-1	R/$\overline{W}$-1	R/$\overline{W}$-1	R/$\overline{W}$-1	R/$\overline{W}$-1
—	—	—	T0CKR4	T0CKR3	T0CKR2	T0CKR1	T0CKR0
bit 7							bit 0

Legend:	R/$\overline{W}$ = Readable bit, Writable bit if IOLOCK = 0		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5 **Unimplemented:** Read as '0'

bit 4-0 **T0CKR<4:0>:** Timer0 External Clock Input (T0CKI) to the Corresponding RPn Pin bits

REGISTER 10-10: RPINR6: PERIPHERAL PIN SELECT INPUT REGISTER 6 (BANKED EE6h)

U-0	U-0	U-0	R/$\overline{W}$-1	R/$\overline{W}$-1	R/$\overline{W}$-1	R/$\overline{W}$-1	R/$\overline{W}$-1
—	—	—	T3CKR4	T3CKR3	T3CKR2	T3CKR1	T3CKR0
bit 7							bit 0

Legend:	R/$\overline{W}$ = Readable bit, Writable bit if IOLOCK = 0		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5 **Unimplemented:** Read as '0'

bit 4-0 **T3CKR<4:0>:** Timer3 External Clock Input (T3CKI) to the Corresponding RPn Pin bits

REGISTER 10-33: RPOR9: PERIPHERAL PIN SELECT OUTPUT REGISTER 9 (BANKED EC9h)

U-0	U-0	U-0	R/$\overline{W}$-0	R/$\overline{W}$-0	R/$\overline{W}$-0	R/$\overline{W}$-0	R/$\overline{W}$-0
—	—	—	RP9R4	RP9R3	RP9R2	RP9R1	RP9R0
bit 7							bit 0

Legend:	R/$\overline{W}$ = Readable bit, Writable bit if IOLOCK = 0		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5 **Unimplemented:** Read as '0'

bit 4-0 **RP9R<4:0>:** Peripheral Output Function is Assigned to RP9 Output Pin bits
(see Table 10-14 for peripheral function numbers)

REGISTER 10-34: RPOR10: PERIPHERAL PIN SELECT OUTPUT REGISTER 10 (BANKED ECAh)

U-0	U-0	U-0	R/$\overline{W}$-0	R/$\overline{W}$-0	R/$\overline{W}$-0	R/$\overline{W}$-0	R/$\overline{W}$-0
—	—	—	RP10R4	RP10R3	RP10R2	RP10R1	RP10R0
bit 7							bit 0

Legend:	R/$\overline{W}$ = Readable bit, Writable bit if IOLOCK = 0		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5 **Unimplemented:** Read as '0'

bit 4-0 **RP10R<4:0>:** Peripheral Output Function is Assigned to RP10 Output Pin bits
(see Table 10-14 for peripheral function numbers)

REGISTER 10-35: RPOR11: PERIPHERAL PIN SELECT OUTPUT REGISTER 11 (BANKED ECBh)

U-0	U-0	U-0	R/$\overline{W}$-0	R/$\overline{W}$-0	R/$\overline{W}$-0	R/$\overline{W}$-0	R/$\overline{W}$-0
—	—	—	RP11R4	RP11R3	RP11R2	RP11R1	RP11R0
bit 7							bit 0

Legend:	R/$\overline{W}$ = Readable bit, Writable bit if IOLOCK = 0		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5 **Unimplemented:** Read as '0'

bit 4-0 **RP11R<4:0>:** Peripheral Output Function is Assigned to RP11 Output Pin bits
(see Table 10-14 for peripheral function numbers)

11.3 MASTER PORT MODES

In its Master modes, the PMP module provides an 8-bit data bus, up to 16 bits of address, and all the necessary control signals to operate a variety of external parallel devices, such as memory devices, peripherals and slave microcontrollers. To use the PMP as a master, the module must be enabled (PMPEN = 1) and the mode must be set to one of the two possible Master modes (PMMODEH<1:0> = 10 or 11).

Because there are a number of parallel devices with a variety of control methods, the PMP module is designed to be extremely flexible to accommodate a range of configurations. Some of these features include:

- 8-Bit and 16-Bit Data modes on an 8-bit data bus
- Configurable address/data multiplexing
- Up to two chip select lines
- Up to 16 selectable address lines
- Address auto-increment and auto-decrement
- Selectable polarity on all control lines
- Configurable Wait states at different stages of the read/write cycle

11.3.1 PMP AND I/O PIN CONTROL

Multiple control bits are used to configure the presence or absence of control and address signals in the module. These bits are PTBEEN, PTWREN, PTRDEN and PTEN<15:0>. They give the user the ability to conserve pins for other functions and allow flexibility to control the external address. When any one of these bits is set, the associated function is present on its associated pin; when clear, the associated pin reverts to its defined I/O port function.

Setting a PTENx bit will enable the associated pin as an address pin and drive the corresponding data contained in the PMADDR register. Clearing a PTENx bit will force the pin to revert to its original I/O function.

For the pins configured as chip select (PMCS1 or PMCS2) with the corresponding PTENx bit set, the PTEN0 and PTEN1 bits will also control the PMALL and PMALH signals. When multiplexing is used, the associated address latch signals should be enabled.

11.3.2 READ/WRITE-CONTROL

The PMP module supports two distinct read/write signaling methods. In Master Mode 1, read and write strobes are combined into a single control line, PMRD/PMWR. A second control line, PMENB, determines when a read or write action is to be taken. In Master Mode 2, separate read and write strobes (PMRD and PMWR) are supplied on separate pins.

All control signals (PMRD, PMWR, PMBE, PMENB, PMAL and PMCSx) can be individually configured as either positive or negative polarity. Configuration is controlled by separate bits in the PMCONL register. Note that the polarity of control signals that share the same output pin (for example, PMWR and PMENB) are controlled by the same bit; the configuration depends on which Master Port mode is being used.

11.3.3 DATA WIDTH

The PMP supports data widths of both 8 bits and 16 bits. The data width is selected by the MODE16 bit (PMMODEH<2>). Because the data path into and out of the module is only 8 bits wide, 16-bit operations are always handled in a multiplexed fashion, with the Least Significant Byte (LSB) of data being presented first. To differentiate data bytes, the byte enable control strobe, PMBE, is used to signal when the Most Significant Byte (MSB) of data is being presented on the data lines.

11.3.4 ADDRESS MULTIPLEXING

In either of the Master modes (PMMODEH<1:0> = 1x), the user can configure the address bus to be multiplexed together with the data bus. This is accomplished by using the ADRMUX<1:0> bits (PMCONH<4:3>). There are three address multiplexing modes available. Typical pinout configurations for these modes are displayed in Figure 11-9, Figure 11-10 and Figure 11-11.

In Demultiplexed mode (PMCONH<4:3> = 00), data and address information are completely separated. Data bits are presented on PMD<7:0> and address bits are presented on PMADDRH<6:0> and PMADDRL<7:0>.

In Partially Multiplexed mode (PMCONH<4:3> = 01), the lower eight bits of the address are multiplexed with the data pins on PMD<7:0>. The upper eight bits of the address are unaffected and are presented on PMADDRH<6:0>. The PMA0 pin is used as an address latch and presents the Address Latch Low (PMALL) enable strobe. The read and write sequences are extended by a complete CPU cycle during which the address is presented on the PMD<7:0> pins.

In Fully Multiplexed mode (PMCONH<4:3> = 10), the entire 16 bits of the address are multiplexed with the data pins on PMD<7:0>. The PMA0 and PMA1 pins are used to present Address Latch Low (PMALL) enable and Address Latch High (PMALH) enable strobes, respectively. The read and write sequences are extended by two complete CPU cycles. During the first cycle, the lower eight bits of the address are presented on the PMD<7:0> pins with the PMALL strobe active. During the second cycle, the upper eight bits of the address are presented on the PMD<7:0> pins with the PMALH strobe active. In the event the upper address bits are configured as chip select pins, the corresponding address bits are automatically forced to '0'.

TABLE 15-5: REGISTERS ASSOCIATED WITH TIMER3/5 AS A TIMER/COUNTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR5	—	—	CM3IF	TMR8IF	TMR6IF	TMR5IF	TMR5GIF	TMR1GIF
PIE5	—	—	CM3IE	TMR8IE	TMR6IE	TMR5IE	TMR5GIE	TMR1GIE
PIR2	OSCFIF	CM2IF	CM1IF	USBIF	BCL1IF	HLVDIF	TMR3IF	CCP2IF
PIE2	OSCFIE	CM2IE	CM1IE	USBIE	BCL1IE	HLVDIE	TMR3IE	CCP2IE
TMR3H	Timer3 Register High Byte							
TMR3L	Timer3 Register Low Byte							
T3GCON	TMR3GE	T3GPOL	T3GTM	T3GSPM	T3GGO/$\overline{\text{T3DONE}}$	T3GVAL	T3GSS1	T3GSS0
T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	T3OSCEN	$\overline{\text{T3SYNC}}$	RD16	TMR3ON
TMR5H	Timer5 Register High Byte							
TMR5L	Timer5 Register Low Byte							
T5GCON	TMR5GE	T5GPOL	T5GTM	T5GSPM	T5GGO/$\overline{\text{T5DONE}}$	T5GVAL	T5GSS1	T5GSS0
T5CON	TMR5CS1	TMR5CS0	T5CKPS1	T5CKPS0	T5OSCEN	$\overline{\text{T5SYNC}}$	RD16	TMR5ON
OSCCON2	—	SOSCRUN	—	SOSCDRV	SOSCGO	PRISD	—	—
CCPTMRS0	C3TSEL1	C3TSEL0	C2TSEL2	C2TSEL1	C2TSEL0	C1TSEL2	C1TSEL1	C1TSEL0
CCPTMRS1	C7TSEL1	C7TSEL0	—	C6TSEL0	—	C5TSEL0	C4TSEL1	C4TSEL0
CCPTMRS1	—	—	—	C10TSEL0	—	C9TSEL0	C8TSEL1	C8TSEL0
CCPTMRS2	—	—	—	C10TSEL0	—	C9TSEL0	C8TSEL1	C8TSEL0

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Timer3 module.

REGISTER 20-3: DMACON1: DMA CONTROL REGISTER 1 (ACCESS F88h)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
SSCON1	SSCON0	TXINC	RXINC	DUPLEX1	DUPLEX0	DLYINTEN	DMAEN
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 **SSCON<1:0>:** $\overline{\text{SSDMA}}$ Output Control bits (Master modes only)
11 = $\overline{\text{SSDMA}}$ is asserted for the duration of 4 bytes; DLYINTEN is always reset low
01 = $\overline{\text{SSDMA}}$ is asserted for the duration of 2 bytes; DLYINTEN is always reset low
10 = $\overline{\text{SSDMA}}$ is asserted for the duration of 1 byte; DLYINTEN is always reset low
00 = $\overline{\text{SSDMA}}$ is not controlled by the DMA module; DLYINTEN bit is software programmable

bit 5 **TXINC:** Transmit Address Increment Enable bit
Allows the transmit address to increment as the transfer progresses.
1 = The transmit address is to be incremented from the initial value of TXADDR<11:0>
0 = The transmit address is always set to the initial value of TXADDR<11:0>

bit 4 **RXINC:** Receive Address Increment Enable bit
Allows the receive address to increment as the transfer progresses.
1 = The received address is to be incremented from the initial value of RXADDR<11:0>
0 = The received address is always set to the initial value of RXADDR<11:0>

bit 3-2 **DUPLEX<1:0>:** Transmit/Receive Operating Mode Select bits
10 = SPI DMA operates in Full-Duplex mode, data is simultaneously transmitted and received
01 = DMA operates in Half-Duplex mode, data is transmitted only
00 = DMA operates in Half-Duplex mode, data is received only

bit 1 **DLYINTEN:** Delay Interrupt Enable bit
Enables the interrupt to be invoked after the number of T_{CY} cycles specified in DLYCYC<2:0> has elapsed from the latest completed transfer.
1 = The interrupt is enabled, SSCON<1:0> must be set to '00'
0 = The interrupt is disabled

bit 0 **DMAEN:** DMA Operation Start/Stop bit
This bit is set by the users' software to start the DMA operation. It is reset back to zero by the DMA engine when the DMA operation is completed or aborted.
1 = DMA is in session
0 = DMA is not in session

REGISTER 20-7: SSPxCON2: MSSPx CONTROL REGISTER 2 (I^2C MASTER MODE) (1, ACCESS FC5h; 2, F71h)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
GCEN(3)	ACKSTAT	ACKDT(1)	ACKEN(2)	RCEN(2)	PEN(2)	RSEN(2)	SEN(2)
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **GCEN:** General Call Enable bit (Slave mode only)(3)
1 = Enable interrupt when a general call address (0000h) is received in the SSPxSR
0 = General call address disabled

bit 6 **ACKSTAT:** Acknowledge Status bit (Master Transmit mode only)
1 = Acknowledge was not received from slave
0 = Acknowledge was received from slave

bit 5 **ACKDT:** Acknowledge Data bit (Master Receive mode only)(1)
1 = Not Acknowledge
0 = Acknowledge

bit 4 **ACKEN:** Acknowledge Sequence Enable bit(2)
1 = Initiates Acknowledge sequence on SDAx and SCLx pins, and transmits ACKDT data bit; automatically cleared by hardware
0 = Acknowledge sequence Idle

bit 3 **RCEN:** Receive Enable bit (Master Receive mode only)(2)
1 = Enables Receive mode for I^2C
0 = Receive Idle

bit 2 **PEN:** Stop Condition Enable bit(2)
1 = Initiates Stop condition on SDAx and SCLx pins; automatically cleared by hardware
0 = Stop condition Idle

bit 1 **RSEN:** Repeated Start Condition Enable bit(2)
1 = Initiates Repeated Start condition on SDAx and SCLx pins; automatically cleared by hardware
0 = Repeated Start condition Idle

bit 0 **SEN:** Start Condition Enable bit(2)
1 = Initiates Start condition on SDAx and SCLx pins; automatically cleared by hardware
0 = Start condition Idle

Note 1: Value that will be transmitted when the user initiates an Acknowledge sequence at the end of a receive.
2: If the I^2C module is active, these bits may not be set (no spooling) and the SSPxBUF may not be written (or writes to the SSPxBUF are disabled).
3: This bit is not implemented in I^2C Master mode.

20.5.4 CLOCK STRETCHING

Both 7-Bit and 10-Bit Slave modes implement automatic clock stretching during a transmit sequence.

The SEN bit (SSPxCON2<0>) allows clock stretching to be enabled during receives. Setting SEN will cause the SCLx pin to be held low at the end of each data receive sequence.

20.5.4.1 Clock Stretching for 7-Bit Slave Receive Mode (SEN = 1)

In 7-Bit Slave Receive mode, on the falling edge of the ninth clock at the end of the ACK sequence, if the BF bit is set, the CKP bit in the SSPxCON1 register is automatically cleared, forcing the SCLx output to be held low. The CKP bit being cleared to '0' will assert the SCLx line low. The CKP bit must be set in the user's ISR before reception is allowed to continue. By holding the SCLx line low, the user has time to service the ISR and read the contents of the SSPxBUF before the master device can initiate another receive sequence. This will prevent buffer overruns from occurring (see Figure 20-15).

Note 1: If the user reads the contents of the SSPxBUF before the falling edge of the ninth clock, thus clearing the BF bit, the CKP bit will not be cleared and clock stretching will not occur.

2: The CKP bit can be set in software regardless of the state of the BF bit. The user should be careful to clear the BF bit in the ISR before the next receive sequence in order to prevent an overflow condition.

20.5.4.2 Clock Stretching for 10-Bit Slave Receive Mode (SEN = 1)

In 10-Bit Slave Receive mode, during the address sequence, clock stretching automatically takes place but CKP is not cleared. During this time, if the UA bit is set after the ninth clock, clock stretching is initiated. The UA bit is set after receiving the upper byte of the 10-bit address and following the receive of the second byte of the 10-bit address with the R/$\overline{W}$ bit cleared to '0'. The release of the clock line occurs upon updating SSPxADD. Clock stretching will occur on each data receive sequence, as described in 7-bit mode.

Note: If the user polls the UA bit and clears it by updating the SSPxADD register before the falling edge of the ninth clock occurs, and if the user has not cleared the BF bit by reading the SSPxBUF register before that time, then the CKP bit will still NOT be asserted low. Clock stretching on the basis of the state of the BF bit only occurs during a data sequence, not an address sequence.

20.5.4.3 Clock Stretching for 7-Bit Slave Transmit Mode

The 7-Bit Slave Transmit mode implements clock stretching by clearing the CKP bit, after the falling edge of the ninth clock, if the BF bit is clear. This occurs regardless of the state of the SEN bit.

The user's Interrupt Service Routine (ISR) must set the CKP bit before transmission is allowed to continue. By holding the SCLx line low, the user has time to service the ISR and load the contents of the SSPxBUF before the master device can initiate another transmit sequence (see Figure 20-10).

Note 1: If the user loads the contents of SSPxBUF, setting the BF bit before the falling edge of the ninth clock, the CKP bit will not be cleared and clock stretching will not occur.

2: The CKP bit can be set in software regardless of the state of the BF bit.

20.5.4.4 Clock Stretching for 10-Bit Slave Transmit Mode

In 10-Bit Slave Transmit mode, clock stretching is controlled during the first two address sequences by the state of the UA bit, just as it is in 10-Bit Slave Receive mode. The first two addresses are followed by a third address sequence, which contains the high-order bits of the 10-bit address and the R/$\overline{W}$ bit set to '1'. After the third address sequence is performed, the UA bit is not set, the module is now configured in Transmit mode and clock stretching is controlled by the BF flag as in 7-Bit Slave Transmit mode (see Figure 20-13).

21.2.2 EUSART ASYNCHRONOUS RECEIVER

The receiver block diagram is displayed in Figure 21-6. The data is received on the RXx pin and drives the data recovery block. The data recovery block is actually a high-speed shifter, operating at x16 times the baud rate, whereas the main receive serial shifter operates at the bit rate or at FOSC. This mode would typically be used in RS-232 systems.

To set up an Asynchronous Reception:

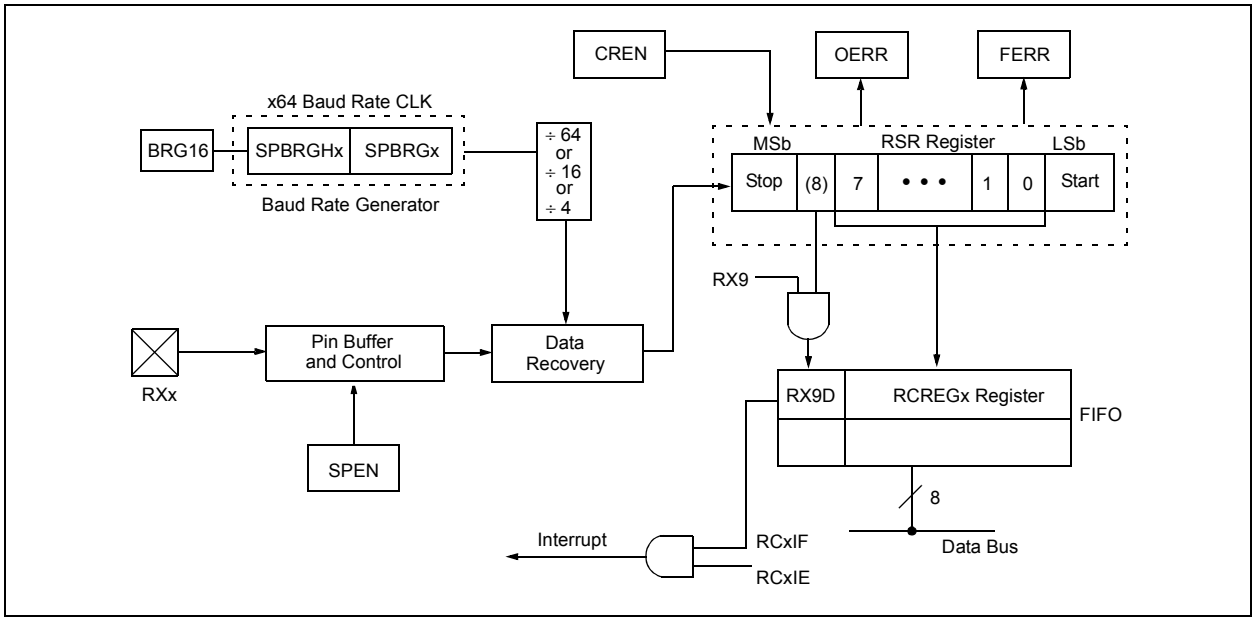
1. Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
3. If interrupts are desired, set enable bit, RCxIE.
4. If 9-bit reception is desired, set bit, RX9.
5. Enable the reception by setting bit, CREN.
6. Flag bit, RCxIF, will be set when reception is complete and an interrupt will be generated if enable bit, RCxIE, was set.
7. Read the RCSTAx register to get the ninth bit (if enabled) and determine if any error occurred during reception.
8. Read the 8-bit received data by reading the RCREGx register.
9. If any error occurred, clear the error by clearing enable bit, CREN.
10. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

21.2.3 SETTING UP 9-BIT MODE WITH ADDRESS DETECT

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

1. Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
3. If interrupts are required, set the RCEN bit and select the desired priority level with the RCxIP bit.
4. Set the RX9 bit to enable 9-bit reception.
5. Set the ADDEN bit to enable address detect.
6. Enable reception by setting the CREN bit.
7. The RCxIF bit will be set when reception is complete. The interrupt will be Acknowledged if the RCxIE and GIE bits are set.
8. Read the RCSTAx register to determine if any error occurred during reception, as well as read bit 9 of data (if applicable).
9. Read RCREGx to determine if the device is being addressed.
10. If any error occurred, clear the CREN bit.
11. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and interrupt the CPU.

FIGURE 21-6: EUSARTx RECEIVE BLOCK DIAGRAM



23.5.3 USB ERROR INTERRUPT STATUS REGISTER (UEIR)

The USB Error Interrupt Status register (Register 23-9) contains the flag bits for each of the error sources within the USB peripheral. Each of these sources is controlled by a corresponding interrupt enable bit in the UEIE register. All of the USB error flags are ORed together to generate the USB Error Interrupt Flag (UERRIF) at the top level of the interrupt logic.

Each error bit is set as soon as the error condition is detected. Thus, the interrupt will typically not correspond with the end of a token being processed.

Once an interrupt bit has been set by the SIE, it must be cleared in software by writing a '0'.

REGISTER 23-9: UEIR: USB ERROR INTERRUPT STATUS REGISTER (ACCESS F63h)

R/C-0	U-0	U-0	R/C-0	R/C-0	R/C-0	R/C-0	R/C-0
BTSEF	—	—	BTOEF	DFN8EF	CRC16EF	CRC5EF	PIDEF
bit 7							bit 0

Legend:			
R = Readable bit	C = Clearable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **BTSEF:** Bit Stuff Error Flag bit
1 = A bit stuff error has been detected
0 = No bit stuff error has been detected

bit 6-5 **Unimplemented:** Read as '0'

bit 4 **BTOEF:** Bus Turnaround Time-out Error Flag bit
1 = Bus turnaround time-out has occurred (more than 16 bit times of Idle from previous EOP elapsed)
0 = No bus turnaround time-out has occurred

bit 3 **DFN8EF:** Data Field Size Error Flag bit
1 = The data field was not an integral number of bytes
0 = The data field was an integral number of bytes

bit 2 **CRC16EF:** CRC16 Failure Flag bit
1 = The CRC16 failed
0 = The CRC16 passed

bit 1 **CRC5EF:** CRC5 Host Error Flag bit
1 = The token packet was rejected due to a CRC5 error
0 = The token packet was accepted

bit 0 **PIDEF:** PID Check Failure Flag bit
1 = PID check failed
0 = PID check passed

EXAMPLE 27-1: SETUP FOR CTMU CALIBRATION ROUTINES

```
#include <p18cxxx.h>
/*****************************************************************************/
/*Setup CTMU ****************************************************************/
/*****************************************************************************/
void setup(void)

{ //CTMUCON - CTMU Control register

    CTMUCONH = 0x00;              //make sure CTMU is disabled
    CTMUCONL = 0x90;
    //CTMU continues to run when emulator is stopped,CTMU continues
    //to run in idle mode,Time Generation mode disabled, Edges are blocked
    //No edge sequence order, Analog current source not grounded, trigger
    //output disabled, Edge2 polarity = positive level, Edge2 source =
    //source 0, Edge1 polarity = positive level, Edge1 source = source 0,


    //CTMUICON - CTMU Current Control Register
    CTMUICON = 0x01;                  //0.55uA, Nominal - No Adjustment

/*****************************************************************************/
//Setup AD converter;
/*****************************************************************************/

    TRISA=0x04;                       //set channel 2 as an input

    // Configured AN2 as an analog channel
    // ANCON0
    ANCON0 = 0XFB;
    // ANCON1
    ANCON1 = 0X1F;

    // ADCON1
    ADCON1bits.ADFM=1;            // Resulst format 1= Right justified
    ADCON1bits.ADCAL=0;           // Normal A/D conversion operation
    ADCON1bits.ACQT=1;            // Acquition time 7 = 20TAD 2 = 4TAD 1=2TAD
    ADCON1bits.ADCS=2;            // Clock conversion bits 6= FOSC/64 2=FOSC/32


       // ADCON0
    ADCON0bits.VCFG0 =0;          // Vref+ = AVdd
    ADCON0bits.VCFG1 =0;          // Vref- = AVss
    ADCON0bits.CHS=2;             // Select ADC channel

    ADCON0bits.ADON=1;            // Turn on ADC

}
```

27.3.2 CAPACITANCE CALIBRATION

There is a small amount of capacitance from the internal A/D Converter sample capacitor as well as stray capacitance from the circuit board traces and pads that affect the precision of capacitance measurements. A measurement of the stray capacitance can be taken by making sure the desired capacitance to be measured has been removed. The measurement is then performed using the following steps:

1. Initialize the A/D Converter and the CTMU.
2. Set EDG1STAT (= 1).
3. Wait for a fixed delay of time, *t*.
4. Clear EDG1STAT.
5. Perform an A/D conversion.
6. Calculate the stray and A/D sample capacitances:

$$C_{OFFSET} = C_{STRAY} + C_{AD} = (I \cdot t)/V$$

where *I* is known from the current source measurement step, *t* is a fixed delay and *V* is measured by performing an A/D conversion.

This measured value is then stored and used for calculations of time measurement or subtracted for capacitance measurement. For calibration, it is expected that the capacitance of $C_{STRAY} + C_{AD}$ is approximately known. C_{AD} is approximately 4 pF.

An iterative process may need to be used to adjust the time, *t*, that the circuit is charged to obtain a reasonable voltage reading from the A/D Converter. The value of *t* may be determined by setting C_{OFFSET} to a theoretical value, then solving for *t*. For example, if C_{STRAY} is theoretically calculated to be 11 pF, and *V* is expected to be 70% of V_{DD} or 2.31V, then *t* would be:

$$(4\ pF + 11\ pF) \bullet 2.31V/0.55\ \mu A$$

or 63 μs.

See Example 27-3 for a typical routine for CTMU capacitance calibration.

28.2 Watchdog Timer (WDT)

PIC18F47J53 family devices have both a conventional WDT circuit and a dedicated, Deep Sleep capable Watchdog Timer. When enabled, the conventional WDT operates in normal Run, Idle and Sleep modes. This data sheet section describes the conventional WDT circuit.

The dedicated, Deep Sleep capable WDT can only be enabled in Deep Sleep mode. This timer is described in **Section 4.6.4 "Deep Sleep Watchdog Timer (DSWDT)"**.

The conventional WDT is driven by the INTRC oscillator. When the WDT is enabled, the clock source is also enabled. The nominal WDT period is 4 ms and has the same stability as the INTRC oscillator.

The 4 ms period of the WDT is multiplied by a 16-bit postscaler. Any output of the WDT postscaler is selected by a multiplexer, controlled by the WDTPS bits in Configuration Register 2H. Available periods range from about 4 ms to 135 seconds (2.25 minutes depending on voltage, temperature and WDT postscaler). The WDT and postscaler are cleared whenever a SLEEP or CLRWDT instruction is executed, or a clock failure (primary or Timer1 oscillator) has occurred.

Note 1:	The CLRWDT and SLEEP instructions clear the WDT and postscaler counts when executed.
2:	When a CLRWDT instruction is executed, the postscaler count will be cleared.

28.2.1 CONTROL REGISTER

The WDTCON register (Register 28-11) is a readable and writable register. The SWDTEN bit enables or disables WDT operation. This allows software to override the WDTEN Configuration bit and enable the WDT only if it has been disabled by the Configuration bit.

LVDSTAT is a read-only status bit that is continuously updated and provides information about the current level of V_{DDCORE}. This bit is only valid when the on-chip voltage regulator is enabled.

FIGURE 28-1: WDT BLOCK DIAGRAM

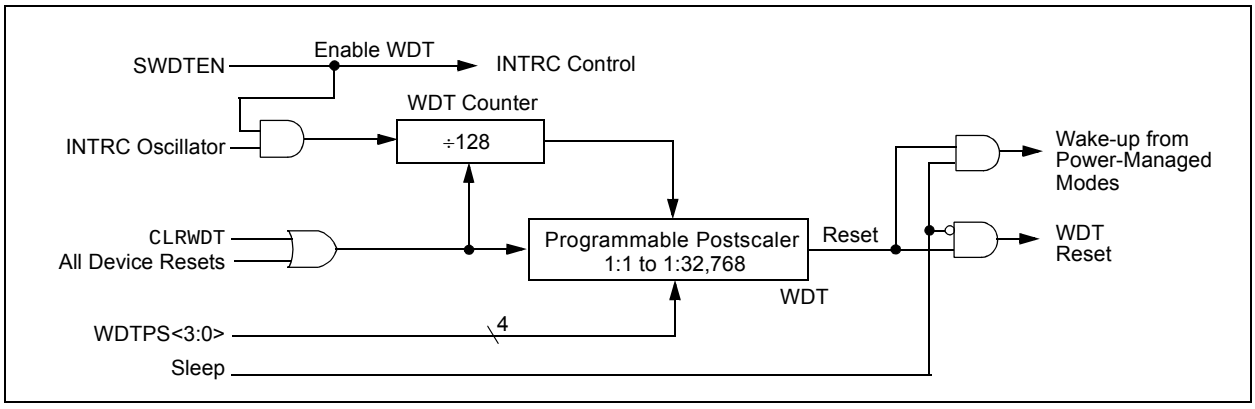
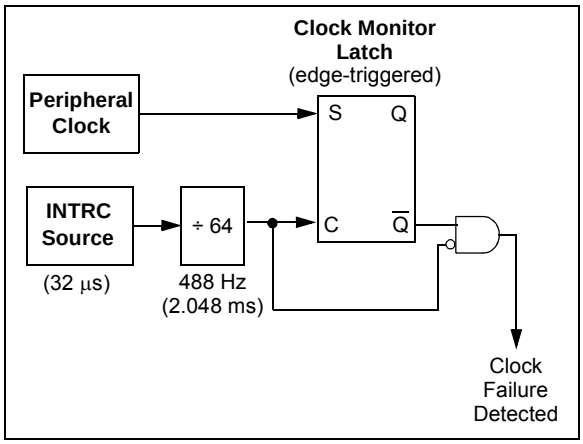


FIGURE 28-4: FSCM BLOCK DIAGRAM



Clock failure is tested for on the falling edge of the sample clock. If a sample clock falling edge occurs while the clock monitor is still set, and a clock failure has been detected (Figure 28-5), the following results:

- The FSCM generates an oscillator fail interrupt by setting bit, OSCFIF (PIR2<7>);
- The device clock source is switched to the internal oscillator block (OSCCON is not updated to show the current clock source – this is the Fail-safe condition); and
- The WDT is reset.

During switchover, the postscaler frequency from the internal oscillator block may not be sufficiently stable for timing-sensitive applications. In these cases, it may be desirable to select another clock configuration and enter an alternate power-managed mode. This can be done to attempt a partial recovery or execute a controlled shutdown. See **Section 4.1.4 "Multiple Sleep Commands"** and **Section 28.4.1 "Special Considerations For Using Two-speed Start-up"** for more details.

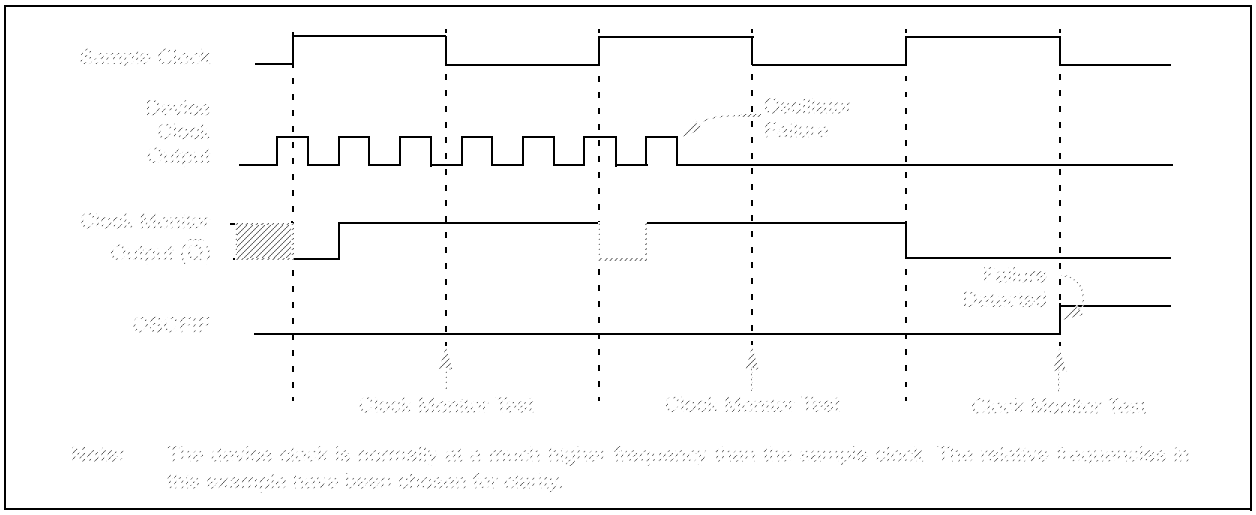
The FSCM will detect failures of the primary or secondary clock sources only. If the internal oscillator block fails, no failure would be detected, nor would any action be possible.

28.5.1 FSCM AND THE WATCHDOG TIMER

Both the FSCM and the WDT are clocked by the INTRC oscillator. Since the WDT operates with a separate divider and counter, disabling the WDT has no effect on the operation of the INTRC oscillator when the FSCM is enabled.

As already noted, the clock source is switched to the INTRC clock when a clock failure is detected; this may mean a substantial change in the speed of code execution. If the WDT is enabled with a small prescale value, a decrease in clock speed allows a WDT time-out to occur and a subsequent device Reset. For this reason, Fail-Safe Clock Monitor events also reset the WDT and postscaler, allowing it to start timing from when execution speed was changed and decreasing the likelihood of an erroneous time-out.

FIGURE 28-5: FSCM TIMING DIAGRAM



CLRF **Clear f**

Syntax: CLRF f {,a}

Operands: $0 \le f \le 255$
$a \in [0,1]$

Operation: 000h → f,
1 → Z

Status Affected: Z

Encoding:

0110	101a	ffff	ffff

Description: Clears the contents of the specified register.

If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See **Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write register 'f'

Example: CLRF FLAG_REG,1

Before Instruction
FLAG_REG = 5Ah
After Instruction
FLAG_REG = 00h

CLRWDT **Clear Watchdog Timer**

Syntax: CLRWDT

Operands: None

Operation: 000h → WDT,
000h → WDT postscaler,
1 → $\overline{TO}$,
1 → $\overline{PD}$

Status Affected: $\overline{TO}$, $\overline{PD}$

Encoding:

0000	0000	0000	0100

Description: CLRWDT instruction resets the Watchdog Timer. It also resets the postscaler of the WDT. Status bits, $\overline{TO}$ and $\overline{PD}$, are set.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	No operation	Process Data	No operation

Example: CLRWDT

Before Instruction
WDT Counter = ?
After Instruction
WDT Counter = 00h
WDT Postscaler = 0
$\overline{TO}$ = 1
$\overline{PD}$ = 1

31.3 DC Characteristics:PIC18F47J53 Family (Industrial)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial			
Param. No.	**Symbol**	**Characteristic**	**Min.**	**Max.**	**Units**	**Conditions**
	VIL	**Input Low Voltage**				
		All I/O ports:				
D030		with TTL Buffer(4)	VSS	0.15 VDD	V	VDD < 3.3V
D030A		with TTL Buffer(4)	VSS	0.8	V	3.3V < VDD < 3.6V
D031		with Schmitt Trigger Buffer	VSS	0.2 VDD	V	
D031A		SCLx/SDAx	—	0.3 VDD	V	I2C enabled
D031B		SCLx/SDAx	—	0.8	V	SMBus enabled
D032		$\overline{MCLR}$	VSS	0.2 VDD	V	
D033		OSC1	VSS	0.3 VDD	V	HS, HSPLL modes
D033A		OSC1	VSS	0.2 VDD	V	EC, ECPLL modes
D034		T1OSI	VSS	0.3	V	T1OSCEN = 1
	VIH	**Input High Voltage**				
		I/O Ports without 5.5V Tolerance:				
D040		with TTL Buffer(4)	0.25 VDD + 0.8V	VDD	V	VDD < 3.3V
D040A		with TTL Buffer(4)	2.0	VDD	V	3.3V < VDD < 3.6V
D041		with Schmitt Trigger Buffer	0.8 VDD	VDD	V	
		I/O Ports with 5.5V Tolerance:(5)				
Dxxx		with TTL Buffer	0.25 VDD + 0.8V	5.5	V	VDD < 3.3V
DxxxA			2.0	5.5	V	3.3V ≤ VDD ≤ 3.6V
Dxxx		with Schmitt Trigger Buffer	0.8 VDD	5.5	V	
D041A		SCLx/SDAx	0.7 VDD	—	V	I2C enabled
D041B		SCLx/SDAx	2.1	—	V	SMBus enabled; VDD ≥ 3V
D042		$\overline{MCLR}$	0.8 VDD	5.5	V	
D043		OSC1	0.7 VDD	VDD	V	HS, HSPLL modes
D043A		OSC1	0.8 VDD	VDD	V	EC, ECPLL modes
D044		T1OSI	1.6	VDD	V	T1OSCEN = 1
	IIL	**Input Leakage Current(1,2)**				
D060		I/O Ports	±5	±200	nA	VSS ≤ VPIN ≤ VDD, Pin at high-impedance
D061		$\overline{MCLR}$	±5	±200	nA	Vss ≤ VPIN ≤ VDD
D062		D+/D-	±75 typical	±500	nA	Vss ≤ VPIN ≤ VDD
D063		OSC1	±5	±200	nA	Vss ≤ VPIN ≤ VDD
	IPU	**Weak Pull-up Current**				
D070	IPURB	PORTB, PORTD(3) and PORTE(3) Weak Pull-up Current	80	400	µA	VDD = 3.3V, VPIN = VSS

Note 1: The leakage current on the $\overline{MCLR}$ pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

2: Negative current is defined as current sourced by the pin.

3: Only available in 44-pin devices.

4: When used as general purpose inputs, the RC4 and RC5 thresholds are referenced to VUSB instead of VDD.

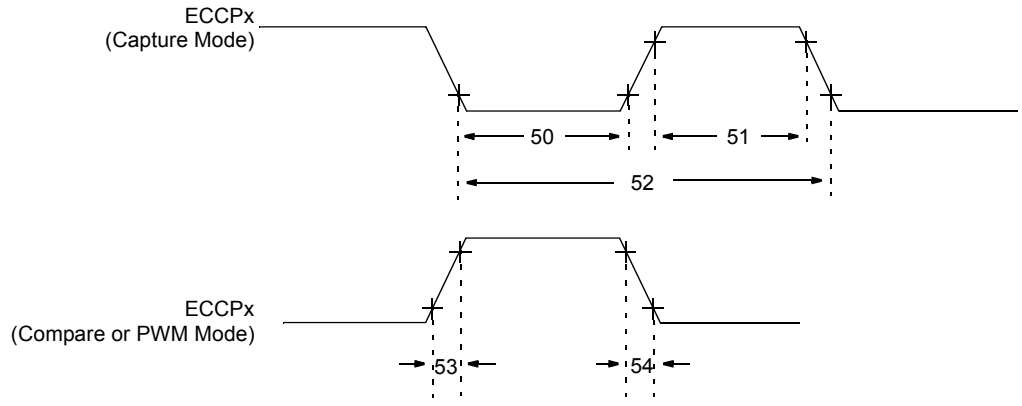
5: Refer to Table 10-2 for pin tolerance levels.

TABLE 31-16: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

Param. No.	Symbol	Characteristic			Min.	Max.	Units	Conditions
40	TT0H	T0CKI High Pulse Width		No prescaler	0.5 TCY + 20	—	ns	
				With prescaler	10	—	ns	
41	TT0L	T0CKI Low Pulse Width		No prescaler	0.5 TCY + 20	—	ns	
				With prescaler	10	—	ns	
42	TT0P	T0CKI Period		No prescaler	TCY + 10	—	ns	
				With prescaler	Greater of: 20 ns or (TCY + 40)/N	—	ns	N = prescale value (1, 2, 4,..., 256)
45	TT1H	T1CKI/T3CKI High Time	Synchronous, no prescaler		0.5 TCY + 20	—	ns	
			Synchronous, with prescaler		10	—	ns	
			Asynchronous		30	—	ns	
46	TT1L	T1CKI/T3CKI Low Time	Synchronous, no prescaler		0.5 TCY + 5	—	ns	
			Synchronous, with prescaler		10	—	ns	
			Asynchronous		30	—	ns	
47	TT1P	T1CKI/T3CKI Input Period	Synchronous		Greater of: 20 ns or (TCY + 40)/N	—	ns	N = prescale value (1, 2, 4, 8)
			Asynchronous		83	—	ns	
	FT1	T1CKI Input Frequency Range(1)			DC	12	MHz	
48	TCKE2TMRI	Delay from External T1CKI Clock Edge to Timer Increment			2 TOSC	7 TOSC	—	

Note 1: The Timer1 oscillator is designed to drive 32.768 kHz crystals. When T1CKI is used as a digital input, frequencies up to 12 MHz are supported.

FIGURE 31-9: ENHANCED CAPTURE/COMPARE/PWM TIMINGS



Note: Refer to Figure 31-4 for load conditions.

FIGURE 31-15: EXAMPLE SPI SLAVE MODE TIMING (CKE = 1)

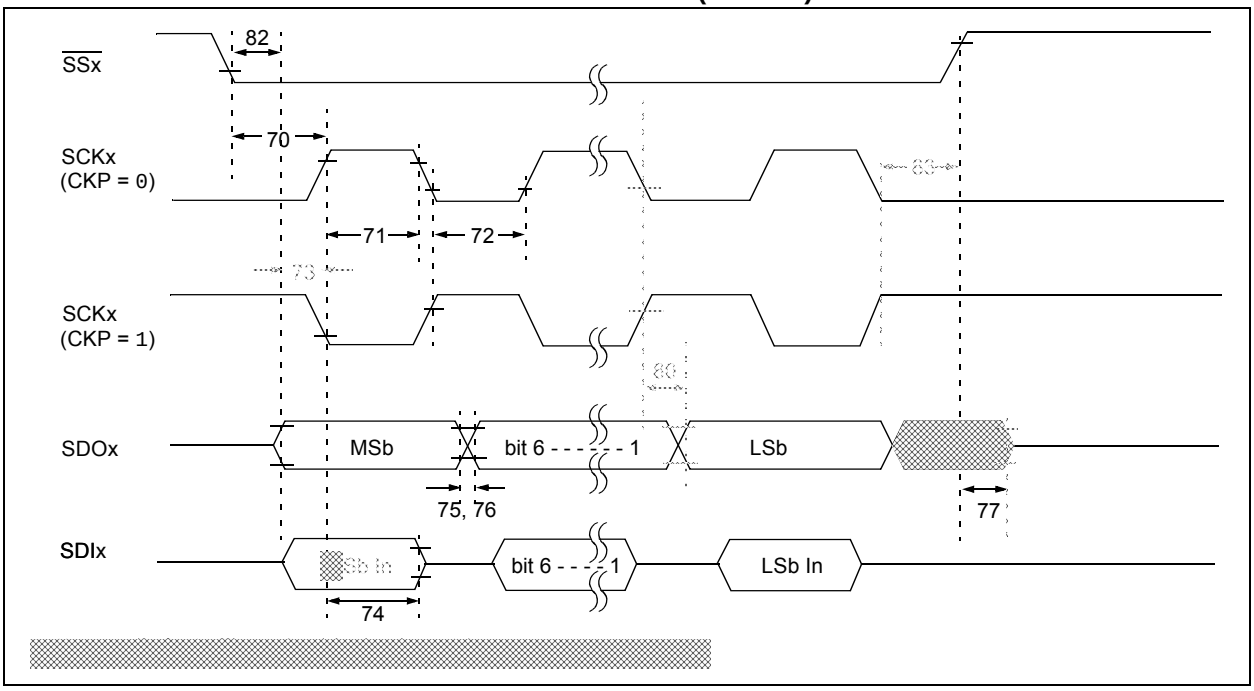


TABLE 31-23: EXAMPLE SPI SLAVE MODE REQUIREMENTS (CKE = 1)

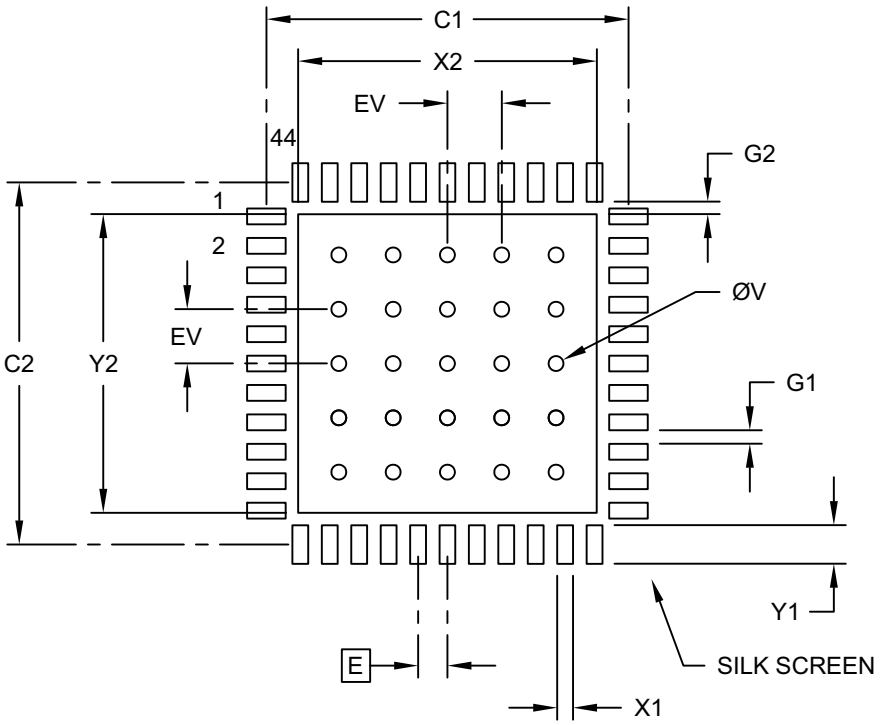
Param. No.	Symbol	Characteristic		Min.	Max.	Units	Conditions
70	TssL2scH, TssL2scL	$\overline{SSx}$ ↓ to SCKx ↓ or SCKx ↑ Input		3 TCY	—	ns	
70A	TssL2WB	$\overline{SSx}$ ↓ to Write to SSPxBUF		3 TCY	—	ns	
71	TscH	SCKx Input High Time (Slave mode)	Continuous	1.25 TCY + 30	—	ns	
71A			Single byte	40	—	ns	**(Note 1)**
72	TscL	SCKx Input Low Time (Slave mode)	Continuous	1.25 TCY + 30	—	ns	
72A			Single byte	40	—	ns	**(Note 1)**
73	TdiV2scH, TdiV2scL	Setup Time of SDIx Data Input to SCKx Edge		25	—	ns	
73A	Tb2b	Last Clock Edge of Byte 1 to the First Clock Edge of Byte 2		1.5 TCY + 40	—	ns	**(Note 2)**
74	TscH2diL, TscL2diL	Hold Time of SDIx Data Input to SCKx Edge		35	—	ns	VDD = 3.3V, VDDCORE = 2.5V
				100	—	ns	VDD = 2.15V
75	TdoR	SDOx Data Output Rise Time		—	25	ns	
76	TdoF	SDOx Data Output Fall Time		—	25	ns	
77	TssH2doZ	$\overline{SSx}$ ↑ to SDOx Output High-Impedance		10	70	ns	
80	TscH2doV, TscL2doV	SDOx Data Output Valid after SCKx Edge		—	50	ns	VDD = 3.3V, VDDCORE = 2.5V
				—	100	ns	VDD = 2.15V
81	TdoV2scH, TdoV2scL	SDOx Data Output Setup to SCKx Edge		TCY	—	ns	
82	TssL2doV	SDOx Data Output Valid after $\overline{SSx}$ ↓ Edge		—	50	ns	
83	TscH2ssH, TscL2ssH	$\overline{SSx}$ ↑ after SCKx Edge		1.5 TCY + 40	—	ns	

Note 1: Requires the use of Parameter #73A.

2: Only if Parameter #71A and #72A are used.

44-Lead Plastic Quad Flat, No Lead Package (ML) - 8x8 mm Body [QFN or VQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E	0.65 BSC		
Optional Center Pad Width	X2			6.60
Optional Center Pad Length	Y2			6.60
Contact Pad Spacing	C1		8.00	
Contact Pad Spacing	C2		8.00	
Contact Pad Width (X44)	X1			0.35
Contact Pad Length (X44)	Y1			0.85
Contact Pad to Contact Pad (X40)	G1	0.30		
Contact Pad to Center Pad (X44)	G2	0.28		
Thermal Via Diameter	V		0.33	
Thermal Via Pitch	EV		1.20	

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

2. For best soldering results, thermal vias, if used, should be filled or tented to avoid solder loss during reflow process

Microchip Technology Drawing No. C04-2103C