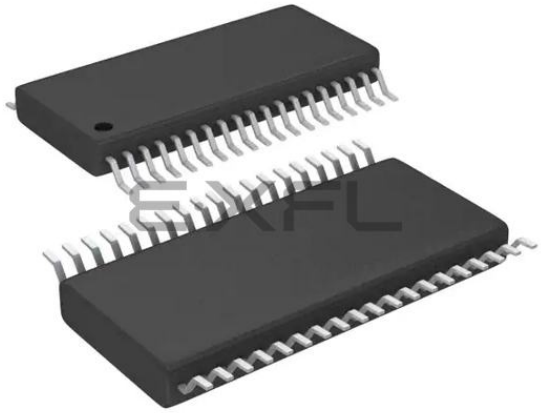

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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	C166SV2
Core Size	16/32-Bit
Speed	40MHz
Connectivity	EBI/EMI, I²C, LINbus, SPI, UART/USART
Peripherals	DMA, I²S, POR, PWM, WDT
Number of I/O	28
Program Memory Size	64KB (64K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	8K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 8x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	38-TFSOP (0.173", 4.40mm Width)
Supplier Device Package	PG-TSSOP-38-8
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/xc2310s8f40raakxuma1

2 General Device Information

The XC2310S series (16/32-Bit Single-Chip Microcontroller with 32-Bit Performance) is a part of the Infineon XC2000 Family of full-feature single-chip CMOS microcontrollers. These devices extend the functionality and performance of the C166 Family in terms of instructions (MAC unit), peripherals, and speed. They combine high CPU performance (up to 66 million instructions per second) with extended peripheral functionality and enhanced IO capabilities. Optimized peripherals can be adapted flexibly to meet the application requirements. These derivatives utilize clock generation via PLL and internal or external clock sources. On-chip memory modules include program Flash, program RAM, and data RAM.

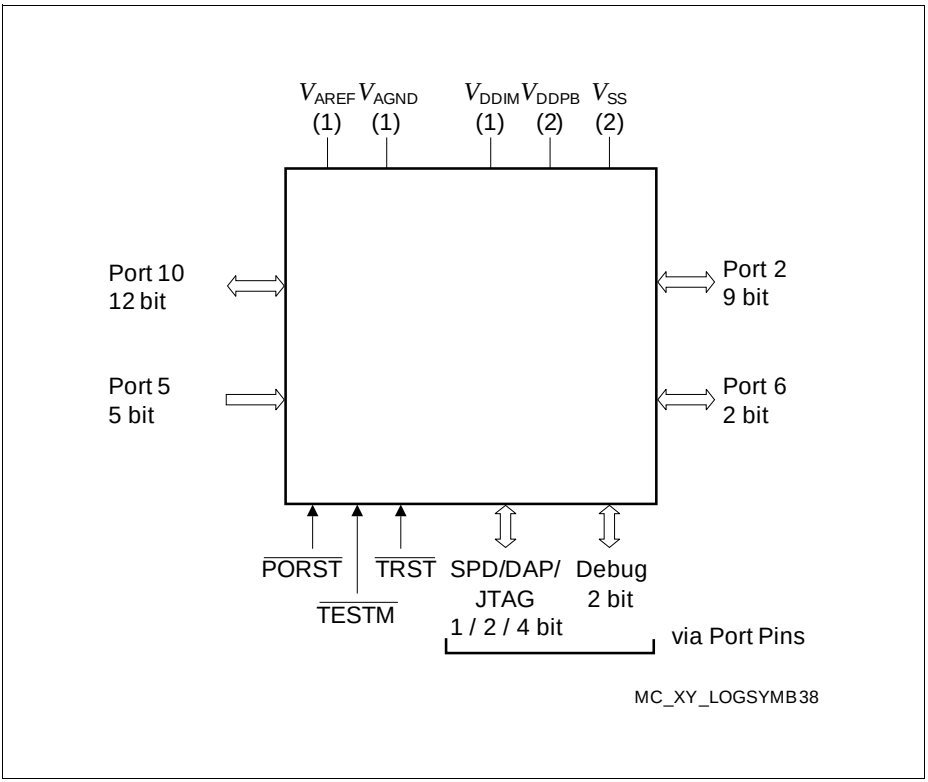

Figure 1 XC2310S Logic Symbol

3.5 Interrupt System

The architecture of the XC2310S supports several mechanisms for fast and flexible response to service requests; these can be generated from various sources internal or external to the microcontroller. Any of these interrupt requests can be programmed to be serviced by the Interrupt Controller or by the Peripheral Event Controller (PEC).

Using a standard interrupt service the current program execution is suspended and a branch to the interrupt vector table is performed. With the PEC just one cycle is 'stolen' from the current CPU activity to perform the PEC service. A PEC service implies a single byte or word data transfer between any two memory locations with an additional increment of either the PEC source pointer, the destination pointer, or both. An individual PEC transfer counter is implicitly decremented for each PEC service except when performing in the continuous transfer mode. When this counter reaches zero, a standard interrupt is performed to the corresponding source-related vector location. PEC services are particularly well suited to supporting the transmission or reception of blocks of data. The XC2310S has eight PEC channels, each with fast interrupt-driven data transfer capabilities.

With a minimum interrupt response time of 7/11[1)] CPU clocks, the XC2310S can react quickly to the occurrence of non-deterministic events.

Interrupt Nodes and Source Selection

The interrupt system provides 46 physical nodes with separate control register containing an interrupt request flag, an interrupt enable flag and an interrupt priority bit field. Most interrupt sources are assigned to a dedicated node. A particular subset of interrupt sources shares a set of nodes. The source selection can be programmed using the interrupt source selection (ISSR) registers.

External Request Unit (ERU)

A dedicated External Request Unit (ERU) is provided to route and preprocess selected on-chip peripheral and external interrupt requests. The ERU features 4 programmable input channels with event trigger logic (ETL) a routing matrix and 4 output gating units (OGU). The ETL features rising edge, falling edge, or both edges event detection. The OGU combines the detected interrupt events and provides filtering capabilities depending on a programmable pattern match or miss.

Trap Processing

The XC2310S provides efficient mechanisms to identify and process exceptions or error conditions that arise during run-time, the so-called 'Hardware Traps'. A hardware trap causes an immediate system reaction similar to a standard interrupt service (branching

1) Depending if the jump cache is used or not.

3.7 Capture/Compare Unit (CC2)

The CAPCOM unit supports generation and control of timing sequences on up to 16 channels with a maximum resolution of one system clock cycle (eight cycles in staggered mode). The CAPCOM unit is typically used to handle high-speed I/O tasks such as pulse and waveform generation, pulse width modulation (PWM), digital to analog (D/A) conversion, software timing, or time recording with respect to external events.

Two 16-bit timers with reload registers provide two independent time bases for the capture/compare register array.

The input clock for the timers is programmable to several prescaled values of the internal system clock, or may be derived from an overflow/underflow of timer T6 in module GPT2. This provides a wide range of variation for the timer period and resolution and allows precise adjustments to the application specific requirements. In addition, external count inputs allow event scheduling for the capture/compare registers relative to external events.

The capture/compare register array contains 16 dual purpose capture/compare registers, each of which may be individually allocated to either CAPCOM timer and programmed for capture or compare function.
All registers have each one port pin associated with it which serves as an input pin for triggering the capture function, or as an output pin to indicate the occurrence of a compare event.

When a capture/compare register has been selected for capture mode, the current contents of the allocated timer will be latched (‘captured’) into the capture/compare register in response to an external event at the port pin which is associated with this register. In addition, a specific interrupt request for this capture/compare register is generated. Either a positive, a negative, or both a positive and a negative external signal transition at the pin can be selected as the triggering event.

The contents of all registers which have been selected for one of the five compare modes are continuously compared with the contents of the allocated timers.

When a match occurs between the timer value and the value in a capture/compare register, specific actions will be taken based on the selected compare mode.

Table 9 Compare Modes

Compare Modes	Function
Mode 0	Interrupt-only compare mode; Several compare interrupts per timer period are possible
Mode 1	Pin toggles on each compare match; Several compare events per timer period are possible

The RTC module can be used for different purposes:

- System clock to determine the current time and date
- Cyclic time-based interrupt, to provide a system time tick independent of CPU frequency and other resources
- 48-bit timer for long-term measurements
- Alarm interrupt at a defined time

3.11 A/D Converters

For analog signal measurement, a 12-bit A/D converters (ADC0) with 8 multiplexed input channels and a sample and hold circuit have been integrated on-chip. Conversions use the successive approximation method. The sample time (to charge the capacitors) and the conversion time are programmable so that they can be adjusted to the external circuit. The A/D converters can also operate in 8-bit and 10-bit conversion mode, further reducing the conversion time.

Several independent conversion result registers, selectable interrupt requests, and highly flexible conversion sequences provide a high degree of programmability to meet the application requirements.

For applications that require more analog input channels, external analog multiplexers can be controlled automatically. For applications that require fewer analog input channels, the remaining channel inputs can be used as digital input port pins.

The A/D converters of the XC2310S support two types of request sources which can be triggered by several internal and external events.

- Parallel requests are activated at the same time and then executed in a predefined sequence.
- Queued requests are executed in a user-defined sequence.

In addition, the conversion of a specific channel can be inserted into a running sequence without disturbing that sequence. All requests are arbitrated according to the priority level assigned to them.

Data reduction features reduce the number of required CPU access operations allowing the precise evaluation of analog inputs (high conversion rate) even at a low CPU speed. Result data can be reduced by limit checking or accumulation of results. Two cascadable filters build the hardware to generate a configurable moving average.

The Peripheral Event Controller (PEC) can be used to control the A/D converters or to automatically store conversion results to a table in memory for later evaluation, without requiring the overhead of entering and exiting interrupt routines for each data transfer. Each A/D converter contains eight result registers which can be concatenated to build a result FIFO. Wait-for-read mode can be enabled for each result register to prevent the loss of conversion data.

In order to decouple analog inputs from digital noise and to avoid input trigger noise, those pins used for analog input can be disconnected from the digital input stages. This can be selected for each pin separately with the Port x Digital Input Disable registers.

The Auto-Power-Down feature of the A/D converters minimizes the power consumption when no conversion is in progress.

Broken wire detection for each channel and a multiplexer test mode provide information to verify the proper operation of the analog signal sources (e.g. a sensor system).

Table 11 Instruction Set Summary (cont'd)

Mnemonic	Description	Bytes
NOP	Null operation	2
CoMUL/CoMAC	Multiply (and accumulate)	4
CoADD/CoSUB	Add/Subtract	4
Co(A)SHR	(Arithmetic) Shift right	4
CoSHL	Shift left	4
CoLOAD/STORE	Load accumulator/Store MAC register	4
CoCMP	Compare	4
CoMAX/MIN	Maximum/Minimum	4
CoABS/CoRND	Absolute value/Round accumulator	4
CoMOV	Data move	4
CoNEG/NOP	Negate accumulator/Null operation	4

1) The Enter Power Down Mode instruction is not used in the XC2310S, due to the enhanced power control scheme. PWRDN will be correctly decoded, but will trigger no action.

4.3.3 Power Consumption

The power consumed by the XC2310S depends on several factors such as supply voltage, operating frequency, active circuits, and operating temperature. The power consumption specified here consists of two components:

- The switching current I_S depends on the device activity
- The leakage current I_{LK} depends on the device temperature

To determine the actual power consumption, always both components, switching current I_S and leakage current I_{LK} must be added:

$I_{DDP} = I_S + I_{LK}$.

Note: The power consumption values are not subject to production test. They are verified by design/characterization.
To determine the total power consumption for dimensioning the external power supply, also the pad driver currents must be considered.

The given power consumption parameters and their values refer to specific operating conditions:

- **Active mode**:
  Regular operation, i.e. peripherals are active, code execution out of Flash.
- **Stopover mode**:
  Crystal oscillator and PLL stopped, Flash switched off, clock in most parts of domain DMP_M stopped.

Note: The maximum values cover the complete specified operating range of all manufactured devices.
The typical values refer to average devices under typical conditions, such as nominal supply voltage, room temperature, application-oriented activity.
After a power reset, the decoupling capacitors for V_{DDIM} are charged with the maximum possible current.

For additional information, please refer to **Section 5.2**, **Thermal Considerations**.

Note: Operating Conditions apply.

$I_{LK1} = 440{,}000 + e^{-\alpha}$ with $\alpha = 5000 / (273 + B \times T_J)$

Parameter B must be replaced by

- 1.0 for typical values
- 1.3 for maximum values

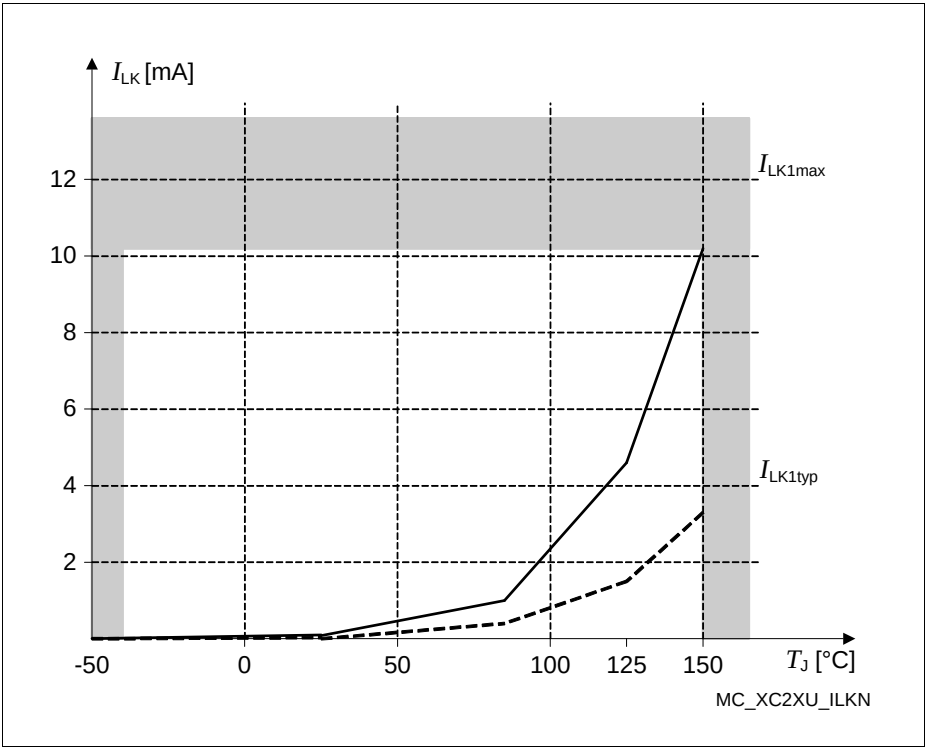

Figure 13 Leakage Supply Current as a Function of Temperature

4.4 Analog/Digital Converter Parameters

These parameters describe the conditions for optimum ADC performance.

Note: Operating Conditions apply.

Table 20 ADC Parameters for All Voltage Ranges

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Switched capacitance at an analog input	C_{AINSW} CC	–	9	20	pF	not subject to production test [1]
Total capacitance at an analog input	C_{AINT} CC	–	20	30	pF	not subject to production test [1]
Switched capacitance at the reference input	C_{AREFSW} CC	–	15	30	pF	not subject to production test [1]
Total capacitance at the reference input	C_{AREFT} CC	–	20	40	pF	not subject to production test [1]
Broken wire detection delay against VAGND[2]	t_{BWG} CC	–	–	50[3]		
Broken wire detection delay against VAREF[2]	t_{BWR} CC	–	–	50[4]		
Conversion time for 8-bit result[2]	t_{c8} CC	(10 + STC x t_{ADCI} + 2 x t_{SYS}				
Conversion time for 10-bit result[2]	t_{c10} CC	(12 + STC x t_{ADCI} + 2 x t_{SYS}				
Conversion time for 12-bit result[2]	t_{c12} CC	(16 + STC x t_{ADCI} + 2 x t_{SYS}				
Analog reference ground	V_{AGND} SR	V_{SS} - 0.05	–	1.5	V	
Analog input voltage range	V_{AIN} SR	V_{AGND}	–	V_{AREF}	V	[5]
Analog reference voltage	V_{AREF} SR	V_{AGND} + 1.0	–	V_{DDPB} + 0.05	V	

1) These parameter values cover the complete operating range. Under relaxed operating conditions (temperature, supply voltage) typical values can be used for calculation.

Table 22 ADC Parameters for Lower Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Wakeup time from analog powerdown, fast mode	t_{WAF} CC	–	–	8.5	µs	
Wakeup time from analog powerdown, slow mode	t_{WAS} CC	–	–	15.0	µs	

1) These parameter values cover the complete operating range. Under relaxed operating conditions (temperature, supply voltage) typical values can be used for calculation.
2) The sum of DNL/INL/GAIN/OFF errors does not exceed the related TUE total unadjusted error.
3) If a reduced analog reference voltage between 1V and V_{DDPB} / 2 is used, then there are additional decrease in the ADC speed and accuracy.
4) If the analog reference voltage range is below V_{DDPB} but still in the defined range of V_{DDPB} / 2 and V_{DDPB} is used, then the ADC converter errors increase. If the reference voltage is reduced by the factor k (k<1), TUE, DNL, INL, Gain and Offset errors increase also by the factor 1/k.
5) If the analog reference voltage is > V_{DDPB}, then the ADC converter errors increase.
6) TUE is based on 12-bit conversion.
7) TUE is tested at V_{AREF} = V_{DDPB} = 3.3 V, V_{AGND} = 0 V. It is verified by design for all other voltages within the defined voltage range. The specified TUE is valid only if the absolute sum of input overload currents on analog port pins (see I_{OV} specification) does not exceed 10 mA, and if V_{AREF} and V_{AGND} remain stable during the measurement time.

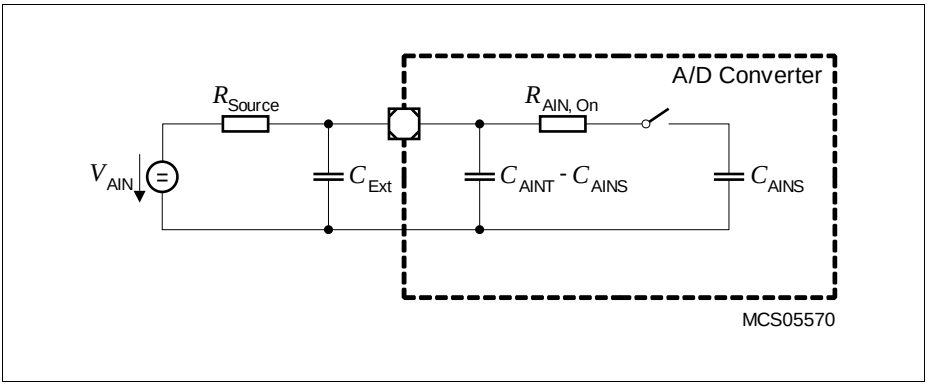

Figure 14 Equivalent Circuitry for Analog Inputs

Sample time and conversion time of the XC2310S's A/D converters are programmable. The timing above can be calculated using **Table 23**.
The limit values for f_{ADCI} must not be exceeded when selecting the prescaler value.

Table 23 A/D Converter Computation Table

GLOBCTR.5-0 (DIVA)	A/D Converter Analog Clock f_{ADCI}	INPCRx.7-0 (STC)	Sample Time[1] t_S
000000_B	f_{SYS}	00_H	$t_{ADCI} \times 2$
000001_B	f_{SYS} / 2	01_H	$t_{ADCI} \times 3$
000010_B	f_{SYS} / 3	02_H	$t_{ADCI} \times 4$
:	f_{SYS} / (DIVA+1)	:	$t_{ADCI} \times$ (STC+2)
111110_B	f_{SYS} / 63	FE_H	$t_{ADCI} \times 256$
111111_B	f_{SYS} / 64	FF_H	$t_{ADCI} \times 257$

1) The selected sample time is doubled if broken wire detection is active (due to the presampling phase).

Converter Timing Example A:

Assumptions: f_{SYS} = 66 MHz (i.e. t_{SYS} = 15.2 ns), DIVA = 03_H, STC = 00_H
Analog clock f_{ADCI} = f_{SYS} / 4 = 16.5 MHz, i.e. t_{ADCI} = 60.6 ns
Sample time t_S = $t_{ADCI} \times 2$ = 121.2 ns

Conversion 12-bit:

$t_{C12} = 16 \times t_{ADCI} + 2 \times t_{SYS} = 16 \times 60.6\text{ ns} + 2 \times 15.2\text{ ns} = 1.0\ \mu s$

Conversion 10-bit:

$t_{C10} = 12 \times t_{ADCI} + 2 \times t_{SYS} = 12 \times 60.6\text{ ns} + 2 \times 15.2\text{ ns} = 0.758\ \mu s$

Conversion 8-bit:

$t_{C8} = 10 \times t_{ADCI} + 2 \times t_{SYS} = 10 \times 60.6\text{ ns} + 2 \times 15.2\text{ ns} = 0.636\ \mu s$

Converter Timing Example B:

Assumptions: f_{SYS} = 40 MHz (i.e. t_{SYS} = 25 ns), DIVA = 01_H, STC = 00_H
Analog clock f_{ADCI} = f_{SYS} / 2 = 20 MHz, i.e. t_{ADCI} = 50 ns
Sample time t_S = $t_{ADCI} \times 2$ = 100 ns

Conversion 12-bit:

$t_{C12} = 16 \times t_{ADCI} + 2 \times t_{SYS} = 16 \times 50\text{ ns} + 2 \times 25\text{ ns} = 0.85\ \mu s$

Conversion 10-bit:

$t_{C10} = 12 \times t_{ADCI} + 2 \times t_{SYS} = 12 \times 50\text{ ns} + 2 \times 25\text{ ns} = 0.65\ \mu s$

4.7.2 Definition of Internal Timing

The internal operation of the XC2310S is controlled by the internal system clock f_{SYS}.

Because the system clock signal f_{SYS} can be generated from a number of internal and external sources using different mechanisms, the duration of the system clock periods (TCSs) and their variation (as well as the derived external timing) depend on the mechanism used to generate f_{SYS}. This must be considered when calculating the timing for the XC2310S.

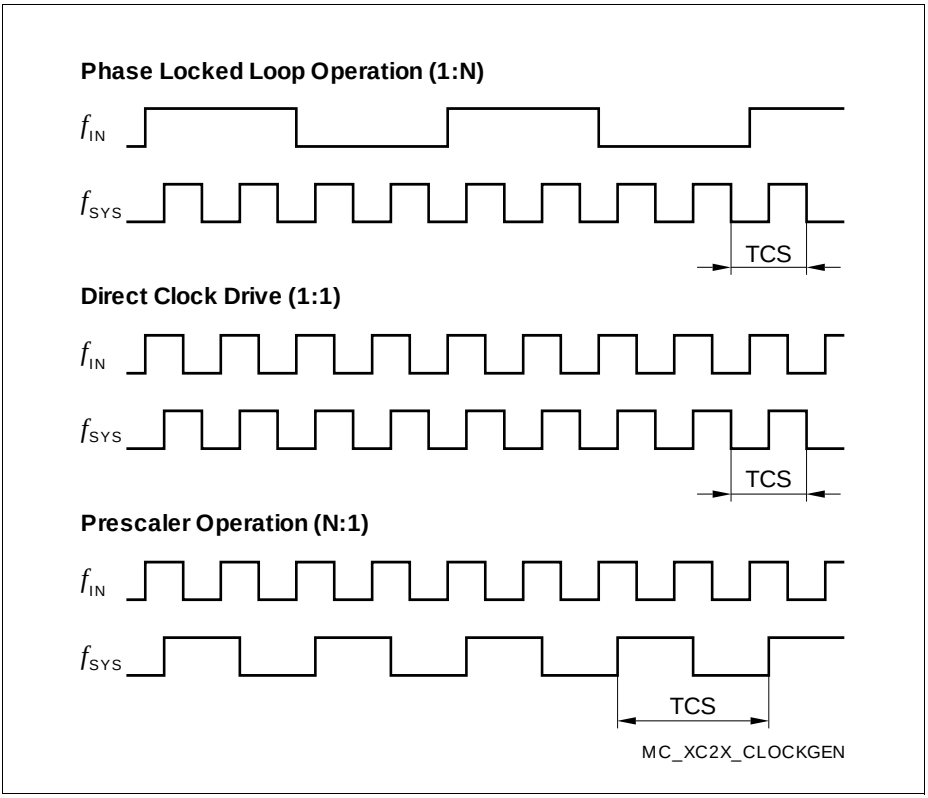

Figure 17 Generation Mechanisms for the System Clock

*Note: The example of PLL operation shown in **Figure 17** uses a PLL factor of 1:4; the example of prescaler operation uses a divider factor of 2:1.*

The specification of the external timing (AC Characteristics) depends on the period of the system clock (TCS).

Direct Drive

When direct drive operation is selected (SYSCON0.CLKSEL = 11_B), the system clock is derived directly from the input clock signal CLKIN1:

$f_{SYS} = f_{IN}$.

The frequency of f_{SYS} is the same as the frequency of f_{IN}. In this case the high and low times of f_{SYS} are determined by the duty cycle of the input clock f_{IN}.

Prescaler Operation

When prescaler operation is selected (SYSCON0.CLKSEL = 10_B, PLLCON0.VCOBY = 1_B), the system clock is derived from the internal clock source through the output prescaler K1 (= K1DIV+1):

$f_{SYS} = f_{OSC} / K1$.

If a divider factor of 1 is selected, the frequency of f_{SYS} equals the frequency of f_{OSC}. In this case the high and low times of f_{SYS} are determined by the duty cycle of the input clock f_{OSC} (external or internal).

The lowest system clock frequency results from selecting the maximum value for the divider factor K1:

$f_{SYS} = f_{OSC} / 1024$.

4.7.2.1 Phase Locked Loop (PLL)

When PLL operation is selected (SYSCON0.CLKSEL = 10_B, PLLCON0.VCOBY = 0_B), the on-chip phase locked loop is enabled and provides the system clock. The PLL multiplies the input frequency by the factor **F** ($f_{SYS} = f_{IN} \times$ **F**).
F is calculated from the input divider P (= PDIV+1), the multiplication factor N (= NDIV+1), and the output divider K2 (= K2DIV+1):
(**F** = N / (P $\times$ K2)).

The input clock can be derived from the on-chip clock source.

The PLL circuit synchronizes the system clock to the input clock. This synchronization is performed smoothly so that the system clock frequency does not change abruptly.

Adjustment to the input clock continuously changes the frequency of f_{SYS} so that it is locked to f_{IN}. The slight variation causes a jitter of f_{SYS} which in turn affects the duration of individual TCSs.

The timing in the AC Characteristics refers to TCSs. Timing must be calculated using the minimum TCS possible under the given circumstances.
The actual minimum value for TCS depends on the jitter of the PLL. Because the PLL is constantly adjusting its output frequency to correspond to the input frequency (from crystal or oscillator), the accumulated jitter is limited. This means that the relative deviation for periods of more than one TCS is lower than for a single TCS (see formulas and **Figure 18**).

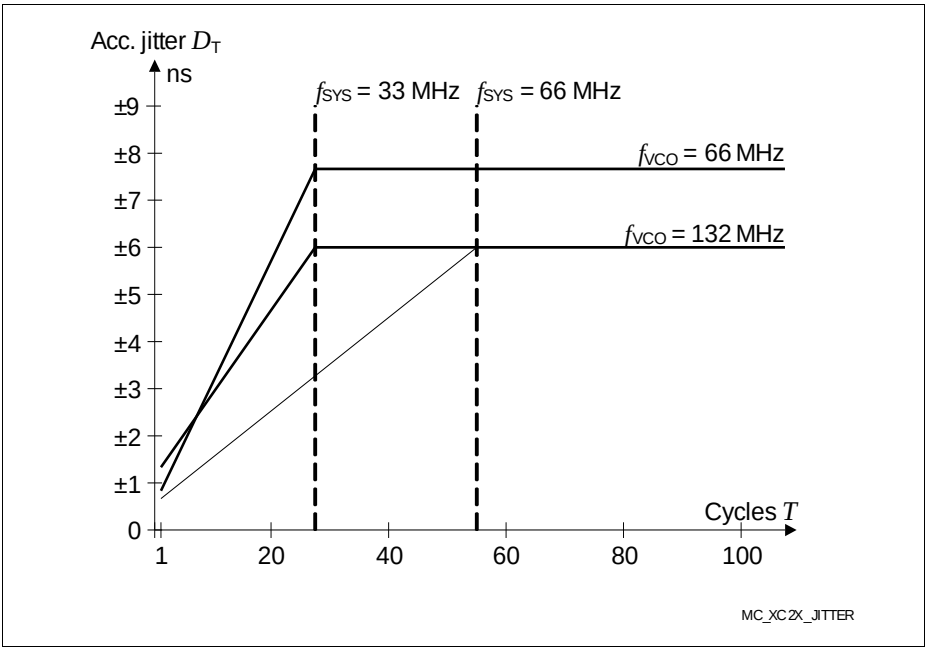

Figure 18 Approximated Accumulated PLL Jitter

Note: The specified PLL jitter values are valid if the capacitive load per pin does not exceed C_L = 20 pF.
The maximum peak-to-peak noise on the pad supply voltage (measured between V_{DDPB} pin and V_{SS} pin) is limited to a peak-to-peak voltage of V_{PP} = 50 mV. This can be achieved by appropriate blocking of the supply voltage as close as possible to the supply pins and using PCB supply and ground planes.

PLL frequency band selection

Different frequency bands can be selected for the VCO so that the operation of the PLL can be adjusted to a wide range of input and output frequencies:

4.7.4 Pad Properties

The output pad drivers of the XC2310S can operate in several user-selectable modes. Strong driver mode allows controlling external components requiring higher currents such as power bridges or LEDs. Reducing the driving power of an output pad reduces electromagnetic emissions (EME). In strong driver mode, selecting a slower edge reduces EME.

The dynamic behavior, i.e. the rise time and fall time, depends on the applied external capacitance that must be charged and discharged. Timing values are given for a capacitance of 20 pF, unless otherwise noted.

In general, the performance of a pad driver depends on the available supply voltage V_{DDP}. Therefore the following tables list the pad parameters for the upper voltage range and the lower voltage range, respectively.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 30 is valid under the following conditions: $V_{DDP} \leq$ 5.5 V; V_{DDP}typ. 5 V; $V_{DDP} \geq$ 4.5 V

Table 30 Standard Pad Parameters for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Maximum output driver current (absolute value)[1)]	I_{Omax} CC	–	–	3.0	mA	Driver_Strength = Medium
		–	–	5.0	mA	Driver_Strength = Strong
		–	–	0.5	mA	Driver_Strength = Weak
Nominal output driver current (absolute value)	I_{Onom} CC	–	–	1.0	mA	Driver_Strength = Medium
		–	–	1.6	mA	Driver_Strength = Strong
		–	–	0.25	mA	Driver_Strength = Weak

Table 30 Standard Pad Parameters for Upper Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Rise and Fall times (10% - 90%)	t_{RF} CC	–	–	38 + 0.6 x C_L	ns	$C_L \geq$ 20 pF; $C_L \leq$ 100 pF; Driver_Strength = Medium
		–	–	1 + 0.45 x C_L	ns	$C_L \geq$ 20 pF; $C_L \leq$ 100 pF; Driver_Strength = Strong ; Driver_Edge= Soft
		–	–	16 + 0.45 x C_L	ns	$C_L \geq$ 20 pF; $C_L \leq$ 100 pF; Driver_Strength = Strong ; Driver_Edge= Slow
		–	–	200 + 2.5 x C_L	ns	$C_L \geq$ 20 pF; $C_L \leq$ 100 pF; Driver_Strength = Weak

1) The total output current that may be drawn at a given time must be limited to protect the supply rails from damage. For any group of 16 neighboring output pins, the total output current in each direction (ΣI_{OL} and ΣI_{OH}) must remain below 25 mA.

Table 31 Standard Pad Parameters for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Maximum output driver current (absolute value)[1)]	I_{Omax} CC	–	–	1.8	mA	Driver_Strength = Medium
		–	–	3.0	mA	Driver_Strength = Strong
		–	–	0.3	mA	Driver_Strength = Weak

Debug via JTAG

The following parameters are applicable for communication through the JTAG debug interface. The JTAG module is fully compliant with IEEE1149.1-2000.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 38 is valid under the following conditions: C_L= 20 pF; voltage_range= upper

Table 38 JTAG Interface Timing for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
TCK clock period	t_1 SR	100[1]	–	–	ns	[2]
TCK high time	t_2 SR	16	–	–	ns	
TCK low time	t_3 SR	16	–	–	ns	
TCK clock rise time	t_4 SR	–	–	8	ns	
TCK clock fall time	t_5 SR	–	–	8	ns	
TDI/TMS setup to TCK rising edge	t_6 SR	6	–	–	ns	
TDI/TMS hold after TCK rising edge	t_7 SR	6	–	–	ns	
TDO valid from TCK falling edge (propagation delay)[3]	t_8 CC	–	29	32	ns	
TDO high impedance to valid output from TCK falling edge[4][3]	t_9 CC	–	29	32	ns	
TDO valid output to high impedance from TCK falling edge[3]	t_{10} CC	–	29	32	ns	
TDO hold after TCK falling edge[3]	t_{18} CC	5	–	–	ns	

1) The debug interface cannot operate faster than the overall system, therefore $t_1 \geq t_{SYS}$.

2) Under typical conditions, the JTAG interface can operate at transfer rates up to 10 MHz.

3) The falling edge on TCK is used to generate the TDO timing.

4) The setup time for TDO is given implicitly by the TCK cycle time.

5.2 Thermal Considerations

When operating the XC2310S in a system, the total heat generated in the chip must be dissipated to the ambient environment to prevent overheating and the resulting thermal damage.

The maximum heat that can be dissipated depends on the package and its integration into the target board. The “Thermal resistance $R_{\Theta JA}$” quantifies these parameters. The power dissipation must be limited so that the average junction temperature does not exceed 150 °C.

The difference between junction temperature and ambient temperature is determined by
$\Delta T = (P_{INT} + P_{IOSTAT} + P_{IODYN}) \times R_{\Theta JA}$

The internal power consumption is defined as
$P_{INT} = V_{DDP} \times I_{DDP}$ (switching current and leakage current).

The static external power consumption caused by the output drivers is defined as
$P_{IOSTAT} = \Sigma((V_{DDP}-V_{OH}) \times I_{OH}) + \Sigma(V_{OL} \times I_{OL})$

The dynamic external power consumption caused by the output drivers (P_{IODYN}) depends on the capacitive load connected to the respective pins and their switching frequencies.

If the total power dissipation for a given system configuration exceeds the defined limit, countermeasures must be taken to ensure proper system operation:

- Reduce V_{DDP}, if possible in the system
- Reduce the system frequency
- Reduce the number of output pins
- Reduce the load on active output drivers

5.3 Quality Declarations

The operation lifetime of the XC2310S depends on the applied temperature profile in application. For a typical example, please refer to **Table 42**; for other profiles, please contact your Infineon counterpart to calculate the specific lifetime within your application.

Table 41 Quality Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Operation lifetime	t_{OP} CC	–	–	20	a	See **Table 42** and **Table 43**
ESD susceptibility according to Human Body Model (HBM)	V_{HBM} SR	–	–	2000	V	EIA/JESD22-A114-B
Moisture sensitivity level	MSL CC	–	–	3	–	JEDEC J-STD-020C

Table 42 Typical Usage Temperature Profile

Operating Time (Sum = 20 years)	Operating Temperature	Notes
1200 h	T_J = 150°C	Normal operation
3600 h	T_J = 125°C	Normal operation
7200 h	T_J = 110°C	Normal operation
12000 h	T_J = 100°C	Normal operation
7 × 21600 h	T_J = 0…10°C, …, 60…70°C	Power reduction

Table 43 Long Time Maximum Storage Temperature Profile

Operating Time (Sum = 20 years)	Operating Temperature	Notes
2000 h	T_J = 150°C	Normal operation
16000 h	T_J = 125°C	Normal operation
6000 h	T_J = 110°C	Normal operation
151200 h	$T_J \leq$ 150°C	No operation

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