# E·XFL



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#### What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

#### Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

#### Details

Product Status	Obsolete
Core Processor	AVR
Core Size	8-Bit
Speed	4MHz
Connectivity	SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	20
Program Memory Size	2KB (1K x 16)
Program Memory Type	FLASH
EEPROM Size	128 x 8
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 6V
Data Converters	A/D 6x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C
Mounting Type	Surface Mount
Package / Case	32-TQFP
Supplier Device Package	32-TQFP (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/at90ls2333-4ai

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

# AT90S/LS2333 and AT90S/LS4433

The I/O memory space contains 64 addresses for CPU peripheral functions as Control Registers, Timer/Counters, A/D-converters, and other I/O functions. The I/O Memory can be accessed directly, or as the Data Space locations following those of the register file, \$20 - \$5F.

The AVR uses a Harvard architecture concept - with separate memories and buses for program and data. The program memory is executed with a two stage pipeline. While one instruction is being executed, the next instruction is pre-fetched from the program memory. This concept enables instructions to be executed in every clock cycle.

The program memory is In-System Programmable Flash memory.

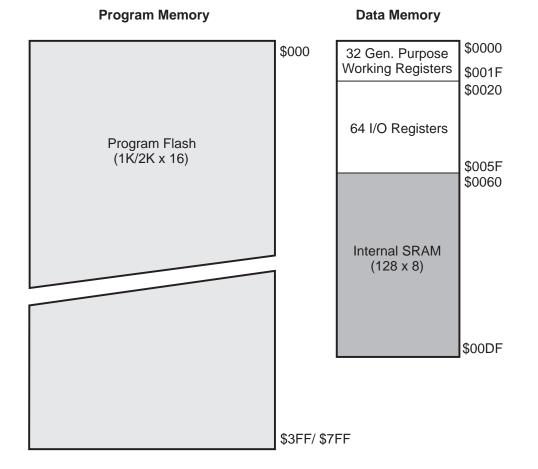
With the relative jump and call instructions, the whole 1K/2K word address space is directly accessed. Most AVR instructions have a single 16-bit word format. Every program memory address contains a 16- or 32-bit instruction.

During interrupts and subroutine calls, the return address program counter (PC) is stored on the stack. The stack is effectively allocated in the general data SRAM, and consequently the stack size is only limited by the total SRAM size and the usage of the SRAM. All user programs must initialize the SP in the reset routine (before subroutines or interrupts are executed). The 8-bit stack pointer SP is read/write accessible in the I/O space.

The 128 bytes data SRAM can be easily accessed through the five different addressing modes supported in the AVR architecture.

The memory spaces in the AVR architecture are all linear and regular memory maps.

Figure 6. AT90S2333/4433 Memory Maps

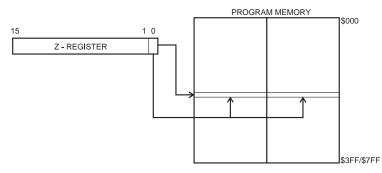


A flexible interrupt module has its control registers in the I/O space with an additional global interrupt enable bit in the status register. All the different interrupts have a separate interrupt vector in the interrupt vector table at the beginning of the program memory. The different interrupts have priority in accordance with their interrupt vector position. The lower the interrupt vector address, the higher the priority.



### **Constant Addressing Using the LPM Instruction**

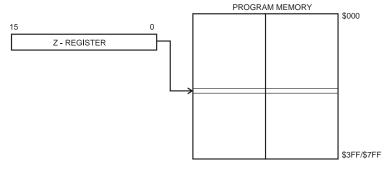
Figure 18. Code Memory Constant Addressing



Constant byte address is specified by the Z-register contents. The 15 MSBs select word address (0 - 1K/2K), the LSB selects low byte if cleared (LSB = 0) or high byte if set (LSB = 1).

### Indirect Program Addressing, IJMP and ICALL

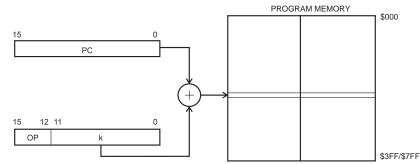
Figure 19. Indirect Program Memory Addressing



Program execution continues at address contained by the Z-register (i.e. the PC is loaded with the contents of the Z-register).

#### Relative Program Addressing, RJMP and RCALL

Figure 20. Relative Program Memory Addressing



Program execution continues at address PC + k + 1. The relative address k is from -2048 to 2047.





# **EEPROM Data Memory**

The AT90S2333/4433 contains 128/256 bytes of data EEPROM memory. It is organized as a separate data space, in which single bytes can be read and written. The EEPROM has an endurance of at least 100,000 write/erase cycles per location. The access between the EEPROM and the CPU is described on page 44 specifying the EEPROM address registers, the EEPROM data register, and the EEPROM control register.

For the SPI data downloading, see page 78 for a detailed description. The EEPROM data memory is In-System Programmable through the SPI port. Please refer to the "EEPROM Read/Write Access" section on page 38 for a thorough description on EEPROM access.

## **Memory Access Times and Instruction Execution Timing**

This section describes the general access timing concepts for instruction execution and internal memory access.

The AVR CPU is driven by the System Clock  $\emptyset$ , directly generated from the external clock crystal for the chip. No internal clock division is used.

Figure 21 shows the parallel instruction fetches and instruction executions enabled by the Harvard architecture and the fast-access register file concept. This is the basic pipelining concept to obtain up to 1 MIPS per MHz with the corresponding unique results for functions per cost, functions per clocks, and functions per power-unit.



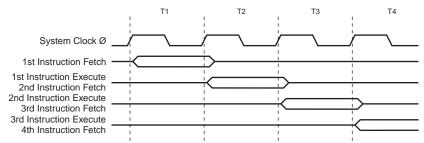
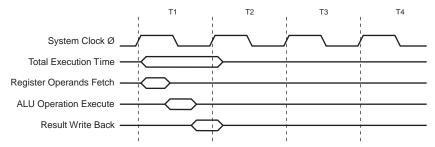


Figure 22 shows the internal timing concept for the register file. In a single clock cycle an ALU operation using two register operands is executed, and the result is stored back to the destination register.

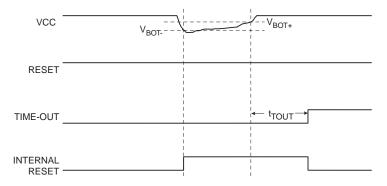
#### Figure 22. Single Cycle ALU Operation



The internal data SRAM access is performed in two System Clock cycles as described in Figure 23.



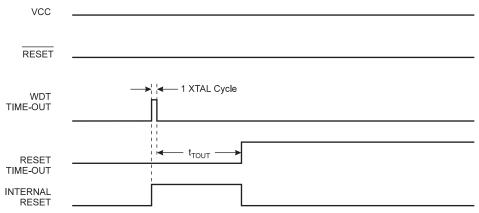
#### Figure 28. Brown-Out Reset During Operation



#### Watchdog Reset

When the Watchdog times out, it will generate a short reset pulse of 1 XTAL cycle duration. On the falling edge of this pulse, the delay timer starts counting the Time-out period  $t_{TOUT}$ . Refer to Page page 36 for details on operation of the Watchdog.





#### **MCU Status Register - MCUSR**

The MCU Status Register provides information on which reset source caused an MCU reset.



• Bits 7..4 - Res: Reserved Bits

These bits are reserved bits in the AT90S2333 and always read as zero.

• Bit 3 - WDRF: Watchdog Reset Flag

This bit is set if a watchdog reset occurs. The bit is cleared by a power-on reset, or by writing a logic zero to the flag. • **Bit 2 - BORF: Brown-Out Reset Flag** 

This bit is set if a brown-out reset occurs. The bit is cleared by a power-on reset, or by writing a logic zero to the flag.

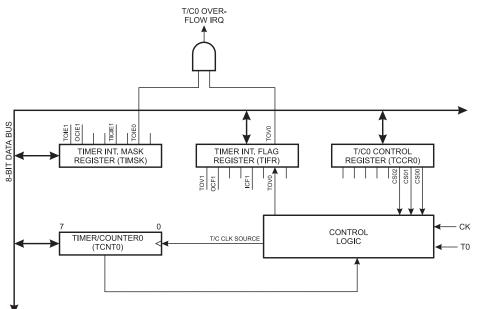
#### • Bit 1 - EXTRF: External Reset Flag

This bit is set if an external reset occurs. The bit is cleared by a power-on reset, or by writing a logic zero to the flag.

Bit 0 - PORF: Power-on Reset Flag

This bit is set if a power-on reset occurs. The bit is cleared only by writing a logic zero to the flag.

#### Figure 31. Timer/Counter0 Block Diagram



#### Timer/Counter0 Control Register - TCCR0

Bit	7	6	5	4	3	2	1	0	
\$33 (\$53)	-	-	-	-	-	CS02	CS01	CS00	TCCR0
Read/Write	R	R	R	R	R	R/W	R/W	R/W	-
Initial value	0	0	0	0	0	0	0	0	

• Bits 7-3 - Res: Reserved bits

These bits are reserved bits in the AT90S2333/4433 and always read as zero.

#### - Bits 2,1,0 - CS02, CS01, CS00: Clock Select0, bit 2,1 and 0

The Clock Select0 bits 2,1, and 0 define the prescaling source of Timer0.

CS02	CS01	CS00	Description	
0	0	0	Stop, Timer/Counter0 is stopped.	
0	0	1	СК	
0	1	0	CK / 8	
0	1	1	CK / 64	
1	0	0	СК / 256	
1	0	1	СК / 1024	
1	1	0	External Pin T0, falling edge	
1	1	1	External Pin T0, rising edge	

#### Table 9. Clock 0 Prescale Select

The Stop condition provides a Timer Enable/Disable function. The prescaled CK modes are scaled directly from the CK oscillator clock. If the external pin modes are used, for Timer/Counter0, transitions on PD4/(T0) will clock the counter even if the pin is configured as an output. This feature can give the user SW control of the counting.



When Timer/Counter1 is externally clocked, the external signal is synchronized with the oscillator frequency of the CPU. To assure proper sampling of the external clock, the minimum time between two external clock transitions must be at least one internal CPU clock period. The external clock signal is sampled on the rising edge of the internal CPU clock.

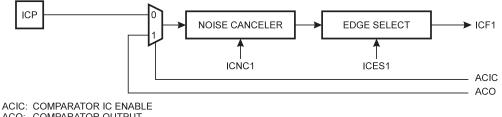
The 16-bit Timer/Counter1 features both a high resolution and a high accuracy usage with the lower prescaling opportunities. Similarly, the high prescaling opportunities makes the Timer/Counter1 useful for lower speed functions or exact timing functions with infrequent actions.

The Timer/Counter1 supports an Output Compare function using the Output Compare Register 1 - OCR1 as the data source to be compared to the Timer/Counter1 contents. The Output Compare functions include optional clearing of the counter on compare matches, and actions on the Output Compare pin 1 on compare matches.

Timer/Counter1 can also be used as a 8, 9 or 10-bit Pulse With Modulator. In this mode the counter and the OCR1 register serve as a glitch-free stand-alone PWM with centered pulses. Refer to page 34 for a detailed description on this function.

The Input Capture function of Timer/Counter1 provides a capture of the Timer/Counter1 contents to the Input Capture Register - ICR1, triggered by an external event on the Input Capture Pin - ICP. The actual capture event settings are defined by the Timer/Counter1 Control Register - TCCR1. In addition, the Analog Comparator can be set to trigger the Input Capture. Refer to the section, "The Analog Comparator", for details on this. The ICP pin logic is shown in Figure 33.

Figure 33. ICP Pin Schematic Diagram



ACO: COMPARATOR OUTPUT

If the noise canceler function is enabled, the actual trigger condition for the capture event is monitored over 4 samples, and all 4 must be equal to activate the capture flag. The input pin signal is sampled at XTAL clock frequency.

#### **Timer/Counter1 Control Register A - TCCR1A**

Bit	7	6	5	4	3	2	1	0	_
\$2F (\$4F)	COM11	COM10	-	-	-	-	PWM11	PWM10	TCCR1A
Read/Write	R/W	R/W	R	R	R	R	R/W	R/W	-
Initial value	0	0	0	0	0	0	0	0	

#### • Bits 7,6 - COM11, COM10: Compare Output Mode1, bits 1 and 0

The COM11 and COM10 control bits determine any output pin action following a compare match in Timer/Counter1. Any output pin actions affect pin OC1 - Output Compare pin 1. This is an alternative function to an I/O port, and the corresponding direction control bit must be set (one) to control an output pin. The control configuration is shown in Table 10.

#### Table 10. Compare 1 Mode Select

COM11	COM10	Description			
0	0	Timer/Counter1 disconnected from output pin OC1			
0	1	oggle the OC1 output line.			
1	0	Clear the OC1 output line (to zero).			
1	1	Set the OC1 output line (to one).			

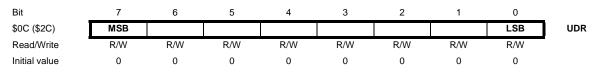
In PWM mode, these bits have a different function. Refer to Table 11 for a detailed description.



- 3. Each slave MCU reads the UDR register and determines if it has been selected. If so, it clears the MPCM bit in UCSRA, otherwise it waits for the next address byte.
- 4. For each received data byte, the receiving MCU will set the receive complete flag (RXC in UCSRA). In 8-bit mode, the receiving MCU will also generate a framing error (FE in UCSRA set), since the stop bit is zero. The other slave MCUs, which still have the MPCM bit set, will ignore the data byte. In this case, the UDR register and the RXC or FE flags will not be affected.
- 5. After the last byte has been transferred, the process repeats from step 2.

# **UART Control**

#### UART I/O Data Register - UDR



The UDR register is actually two physically separate registers sharing the same I/O address. When writing to the register, the UART Transmit Data register is written. When reading from UDR, the UART Receive Data register is read.

#### **UART Control and Status Registers - UCSRA**

Bit	7	6	5	4	3	2	1	0	_
\$0B (\$2B)	RXC	TXC	UDRE	FE	OR	-	-	MPCM	UCSRA
Read/Write	R	R/W	R	R	R	R	R	R/W	-
Initial value	0	0	1	0	0	0	0	0	

#### • Bit 7 - RXC: UART Receive Complete

This bit is set (one) when a received character is transferred from the Receiver Shift register to UDR. The bit is set regardless of any detected framing errors. When the RXCIE bit in UCR is set, the UART Receive Complete interrupt will be executed when RXC is set(one). RXC is cleared by reading UDR. When interrupt-driven data reception is used, the UART Receive Complete Interrupt routine must read UDR in order to clear RXC, otherwise a new interrupt will occur once the interrupt routine terminates.

#### • Bit 6 - TXC: UART Transmit Complete

This bit is set (one) when the entire character (including the stop bit) in the Transmit Shift register has been shifted out and no new data has been written to UDR. This flag is especially useful in half-duplex communications interfaces, where a transmitting application must enter receive mode and free the communications bus immediately after completing the transmission.

When the TXCIE bit in UCR is set, setting of TXC causes the UART Transmit Complete interrupt to be executed. TXC is cleared by hardware when executing the corresponding interrupt handling vector. Alternatively, the TXC bit is cleared (zero) by writing a logical one to the bit.

#### • Bit 5 - UDRE: UART Data Register Empty

This bit is set (one) when a character written to UDR is transferred to the Transmit shift register. Setting of this bit indicates that the transmitter is ready to receive a new character for transmission.

When the UDRIE bit in UCR is set, the UART Transmit Complete interrupt to be executed as long as UDRE is set. UDRE is cleared by writing UDR. When interrupt-driven data transmittal is used, the UART Data Register Empty Interrupt routine must write UDR in order to clear UDRE, otherwise a new interrupt will occur once the interrupt routine terminates.

UDRE is set (one) during reset to indicate that the transmitter is ready.

#### • Bit 4 - FE: Framing Error

This bit is set if a Framing Error condition is detected, i.e. when the stop bit of an incoming character is zero.

The FE bit is cleared when the stop bit of received data is one.





#### **UART Baud Rate Register - UBRR**

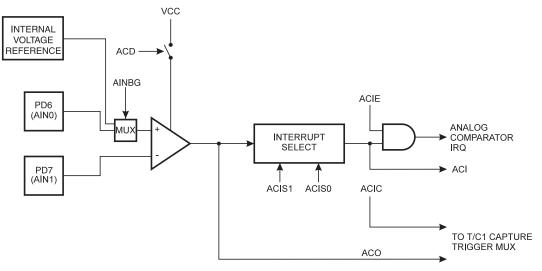
Bit	15	14	13	12	11	10	9	8	_
\$03 (\$23)	-	-	-	-	MSB			LSB	UBRRHI
\$09 (\$29)	MSB							LSB	UBRR
	7	6	5	4	3	2	1	0	-
Read/Write	R	R	R	R	R/W	R/W	R/W	R/W	
	R/W								
Initial value	0	0	0	0	0	0	0	0	
	0	0	0	0	0	0	0	0	

This is a 12-bit register which contains the UART Baud Rate according to the equation on the previous page. The UBRRHI contains the 4 most significant bits, and the UBRR contains the 8 least significant bits of the UART Baud Rate.

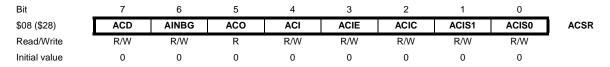
# **Analog Comparator**

The analog comparator compares the input values on the positive input PD6 (AIN0) and negative input PD7 (AIN1). When the voltage on the positive input PD6 (AIN0) is higher than the voltage on the negative input PD7 (AIN1), the Analog Comparator Output, ACO is set (one). The comparator's output can be set to trigger the Timer/Counter1 Input Capture function. In addition, the comparator can trigger a separate interrupt, exclusive to the Analog Comparator. The user can select Interrupt triggering on comparator output rise, fall or toggle. A block diagram of the comparator and its surrounding logic is shown in Figure 43.





#### Analog Comparator Control And Status Register - ACSR

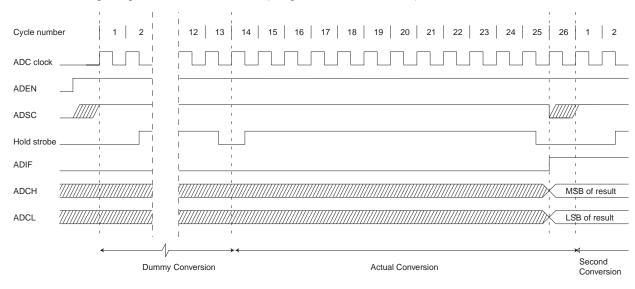


#### • Bit 7 - ACD: Analog Comparator Disable

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When this bit is set(one), the power to the analog comparator is switched off. This bit can be set at any time to turn off the analog comparator. When changing the ACD bit, the Analog Comparator Interrupt must be disabled by clearing the ACIE bit in ACSR. Otherwise an interrupt can occur when the bit is changed.





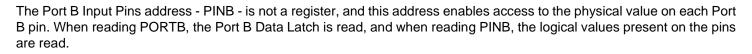
#### Figure 46. ADC Timing Diagram, First Conversion (Single Conversion Mode)

#### Table 21. ADC Conversion Time

Condition	Sample Cycle Number	Result Ready (cycle number)	Total Conversion Time (cycles)	Total Conversion Time (μs)
1st Conversion, Free Run	14	25	25	125 - 500
1st Conversion, Single	14	25	26	130 - 520
Free Run Conversion	2	13	13	65 - 260
Single Conversion	2	13	14	70 - 280

#### Figure 47. ADC Timing Diagram, Single Conversion

Cycle num	ber   1   2   3   4   5   6   7   8   9   10   11   12   13   14	1   2
ADC clock		
ADSC		1
Hold strobe		1 1
ADIF		   
ADCH		MSB of result
ADCL		LSB of result
	← One Conversion	× Next Conversion



#### Port B As General Digital I/O

All 6 pins in Port B have equal functionality when used as digital I/O pins.

PBn, General I/O pin: The DDBn bit in the DDRB register selects the direction of this pin, if DDBn is set (one), PBn is configured as an output pin. If DDBn is cleared (zero), PBn is configured as an input pin. If PORTBn is set (one) when the pin configured as an input pin, the MOS pull up resistor is activated. To switch the pull up resistor off, the PORTBn has to be cleared (zero) or the pin has to be configured as an output pin. The port pins are tristated when a reset condition becomes active, even if the clock is not running.

Table 24. DDB	n Effects on	Port B Pins
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DDBn	PORTBn	I/O	Pull Up	Comment
0	0	Input	No	Tri-state (Hi-Z)
0	1	Input	Yes	PBn will source current if ext. pulled low.
1	0	Output	No	Push-Pull Zero Output
1	1	Output	No	Push-Pull One Output

Note: n: 5...0, pin number.

#### **Alternate Functions Of Port B**

The alternate pin configuration is as follows:

#### • SCK - Port B, Bit 5

SCK: Master clock output, slave clock input pin for SPI channel. When the SPI is enabled as a slave, this pin is configured as an input regardless of the setting of DDB5. When the SPI is enabled as a master, the data direction of this pin is controlled by DDB5. When the pin is forced to be an input, the pull-up can still be controlled by the PORTB5 bit. See the description of the SPI port for further details.

#### • MISO - Port B, Bit 4

MISO: Master data input, slave data output pin for SPI channel. When the SPI is enabled as a master, this pin is configured as an input regardless of the setting of DDB4. When the SPI is enabled as a slave, the data direction of this pin is controlled by DDB4. When the pin is forced to be an input, the pull-up can still be controlled by the PORTB4 bit. See the description of the SPI port for further details.

#### • MOSI - Port B, Bit 3

MOSI: SPI Master data output, slave data input for SPI channel. When the SPI is enabled as a slave, this pin is configured as an input regardless of the setting of DDB3. When the SPI is enabled as a master, the data direction of this pin is controlled by DDB3. When the pin is forced to be an input, the pull-up can still be controlled by the PORTB3 bit. See the description of the SPI port for further details.

#### • SS - Port B, Bit 2

S: Slave port select input. When the SPI is enabled as a slave, this pin is configured as an input regardless of the setting of DDB2. As a slave, the SPI is activated when this pin is driven low. When the SPI is enabled as a master, the data direction of this pin is controlled by DDB2. When the pin is forced to be an input, the pull-up can still be controlled by the PORTB2 bit. See the description of the SPI port for further details.

#### • OC1 - Port B, Bit1

OC1, Output compare match output: PB1 pin can serve as an external output for the Timer/Counter1 output compare. The pin has to be configured as an output (DDB1 set (one)) to serve this function. See the timer description on how to enable this function. The OC1 pin is also the output pin for the PWM mode timer function.

#### • ICP- Port B, Bit0

ICP, Input Capture Pin: PB0 pin can serve as an external input for the Timer/Counter1 input capture. The pin has to be configured as an input (DDB0 cleared (zero)) to serve this function. See the timer description on how to enable this function.

Figure 54. Port B Schematic Diagram (Pin PB4)

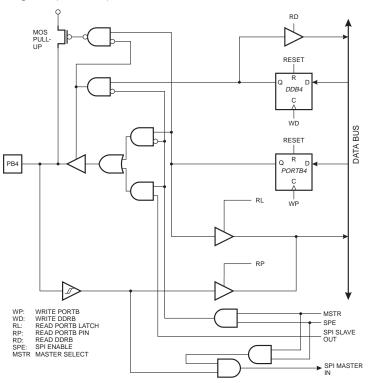
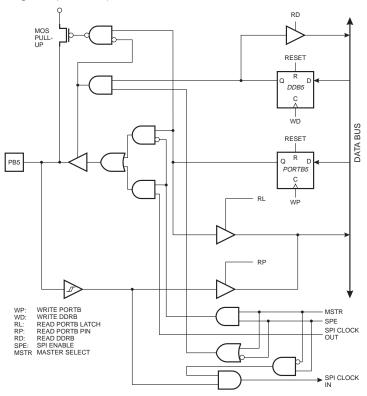


Figure 55. Port B Schematic Diagram (Pin PB5)







#### **Port D Schematics**

Note that all port pins are synchronized. The synchronization latches are however, not shown in the figures.



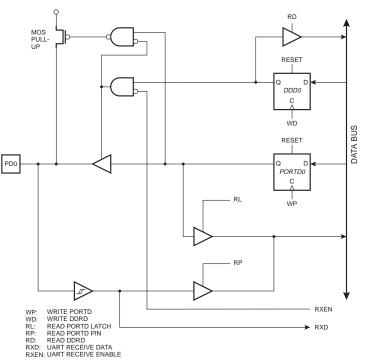
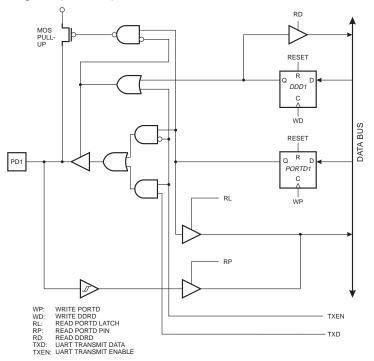


Figure 58. Port D Schematic Diagram (Pin PD1)



# AT90S/LS2333 and AT90S/LS4433

#### Table 30. Pin Name Mapping

Signal Name in Programming Mode	Pin Name	I/O	Function
RDY/BSY	PD1	0	0: Device is busy programming, 1: Device is ready for new command
ŌĒ	PD2	I	Output Enable (Active low)
WR	PD3	I	Write Pulse (Active low)
BS	PD4	Ι	Byte Select ('0' selects low byte, '1' selects high byte)
XA0	PD5	Ι	XTAL Action Bit 0
XA1	PD6	I	XTAL Action Bit 1
DATA	PC1-0, PB5-0	I/O	Bidirectional Databus (Output when OE is low)

#### Table 31. XA1 and XA0 Coding

XA1	XA0	Action when XTAL1 is Pulsed
0	0	Load Flash or EEPROM Address (High or low address byte determined by BS)
0	1	Load Data (High or Low data byte for Flash determined by BS)
1	0	Load Command
1	1	No Action, Idle

#### Table 32. Command Byte Bit Coding

Command Byte	Command Executed
1000 0000	Chip Erase
0100 0000	Write Fuse Bits
0010 0000	Write Lock Bits
0001 0000	Write Flash
0001 0001	Write EEPROM
0000 1000	Read Signature Bytes
0000 0100	Read Fuse and Lock Bits
0000 0010	Read Flash
0000 0011	Read EEPROM

#### **Enter Programming Mode**

The following algorithm puts the device in parallel programming mode:

- 1. Apply supply voltage according to Table 29, between  $V_{CC}$  and GND.
- 2. Set the RESET and BS pin to '0' and wait at least 100 ns.
- 3. Apply 11.5 12.5V to RESET. Any activity on BS within 100 ns after +12V has been applied to RESET, will cause the device to fail entering programming mode.

#### Chip Erase

The Chip Erase command will erase the Flash and EEPROM memories, and the Lock bits. The Lock bits are not reset until the Flash and EEPROM have been completely erased. The Fuse bits are not changed. Chip Erase must be performed before the Flash or EEPROM is reprogrammed.

Load Command "Chip Erase"



For the EEPROM, an auto-erase cycle is provided within the self-timed write instruction and there is no need to first execute the Chip Erase instruction. The Chip Erase instruction turns the content of every memory location in both the Program and EEPROM arrays into \$FF.

The Program and EEPROM memory arrays have separate address spaces:

0000 to \$03FF/\$07FF (AT90S2333/AT90S4433) for Program memory and \$0000 to \$007F/\$00FF (AT90S2333/AT90S4433) for EEPROM memory.

Either an external system clock is supplied at pin XTAL1 or a crystal needs to be connected across pins XTAL1 and XTAL2. The minimum low and high periods for the serial clock (SCK) input are defined as follows:

Low:> 2 XTAL1 clock cycles

High:> 2 XTAL1 clock cycles

#### **Serial Programming Algorithm**

When writing serial data to the AT90S2333/AT90S4433, data is clocked on the rising edge of CLK.

When reading data from the AT90S2333/AT90S4433, data is clocked on the falling edge of CLK. See Figure 67, Figure 68 and Table 36 for details.

To program and verify the AT90S2333/AT90S4433 in the serial programming mode, the following sequence is recommended (See four byte instruction formats in Table 35):

1. Power-up sequence:

Apply power between  $V_{CC}$  and GND while RESET and SCK are set to '0'. If a crystal is not connected across pins XTAL1 and XTAL2, apply a clock signal to the XTAL1 pin. In some systems, the programmer can not guarantee that SCK is held low during power-up. In this case, RESET must be given a positive pulse of at least two XTAL1 cycles duration after SCK has been set to '0'.

- 2. Wait for at least 20 ms and enable serial programming by sending the Programming Enable serial instruction to pin MOSI/PB3.
- 3. The serial programming instructions will not work if the communication is out of syncronization. When in sync, the second byte (\$53) will echo back when issuing the third byte of the Programming Enable instruction. Wheter the echo is correct or not, all 4 bytes of the instruction must be transmitted. If the \$53 did not echo back, give SCK a positive pulse and issue a new Programming Enable instruction. If the \$53 is not seen within 32 attempts, there is no functional device connected.
- 4. If a Chip Erase is performed (must be done to erase the Flash), wait t<sub>WD\_ERASE</sub> after the instruction, give RESET a positive pulse, and start over from Step 2. See Table 37 on page 82 for t<sub>WD\_ERASE</sub> value.
- 5. The Flash or EEPROM array is programmed one byte at a time by supplying the address and data together with the appropriate Write instruction. An EEPROM memory location is first automatically erased before new data is written. Use Data Polling to detect when the next byte in the Flash or EEPROM can be written. If polling is not used, wait t<sub>WD\_PROG</sub> before transmitting the next instruction. In an erased device, no \$FFs in the data file(s) needs to be programmed. See Table 38 on page 82 for t<sub>WD\_PROG</sub> value.
- 6. Any memory location can be verified by using the Read instruction which returns the content at the selected address at serial output MISO/PB4.
- 7. At the end of the programming session, RESET can be set high to commence normal operation.
- 8. Power-off sequence (if needed):

Set XTAL1 to '0' (if a crystal is not used). Set  $\overline{\text{RESET}}$  to '1'. Turn V<sub>CC</sub> power off





# **DC Characteristics**

$T_A = -40^{\circ}C$ to $85^{\circ}C$ , $V_{CC} = 2.7V$ to $6.0V$	(unless otherwise noted)	(Continued)
$T_A = -40$ C to 65 C, $v_{CC} = 2.7$ v to 0.0 v		(Continueu)

Symbol	Parameter	Condition	Min	Тур	Мах	Units
V <sub>ACIO</sub>	Analog Comparator Input Offset Voltage	$V_{CC} = 5V$			40	mV
I <sub>ACLK</sub>	Analog Comparator Input Leakage A	$V_{CC} = 5V$ $V_{in} = V_{CC}/2$	-50		50	nA
t <sub>ACPD</sub>	Analog Comparator Propagation Delay	$V_{CC} = 2.7V$ $V_{CC} = 4.0V$		750 500		ns

Notes: 1. "Max" means the highest value where the pin is guaranteed to be read as low (logical zero).

2. "Min" means the lowest value where the pin is guaranteed to be read as high (logical one).

3. Although each I/O port can sink more than the test conditions (20mA at Vcc = 5V, 10mA at Vcc = 3V) under steady state conditions (non-transient), the following must be observed:

1] The sum of all IOL, for all ports, should not exceed 300 mA.

2] The sum of all IOL, for port C0-C5, should not exceed 100 mA.

3] The sum of all IOL, for ports B0-B5, D0-D7 and XTAL2, should not exceed 200 mA.

If IOL exceeds the test condition, VOL may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test condition.

4. Although each I/O port can source more than the test conditions (3mA at Vcc = 5V, 1.5mA at Vcc = 3V) under steady state conditions (non-transient), the following must be observed:

1] The sum of all IOL, for all ports, should not exceed 300 mA.

2] The sum of all IOL, for port C0-C5 , should not exceed 100 mA.

3] The sum of all IOL, for ports B0-B5, D0-D7 and XTAL2, should not exceed 200 mA.

If IOH exceeds the test condition, VOH may exceed the related specification. Pins are not guaranteed to source current greater than the listed test condition.

5. Minimum  $V_{CC}$  for Power Down is 2V.



# **Typical Characteristics**

The following charts show typical behavior. These data are characterized, but not tested. All current consumption measurements are performed with all I/O pins configured as inputs and with internal pull-ups enabled. A sine wave generator with rail to rail output is used as clock source.

The power consumption in power-down mode is independent of clock selection.

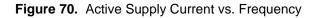
The current consumption is a function of several factors such as: operating voltage, operating frequency, loading of I/O pins, switching rate of I/O pins, code executed and ambient temperature. The dominating factors are operating voltage and frequency.

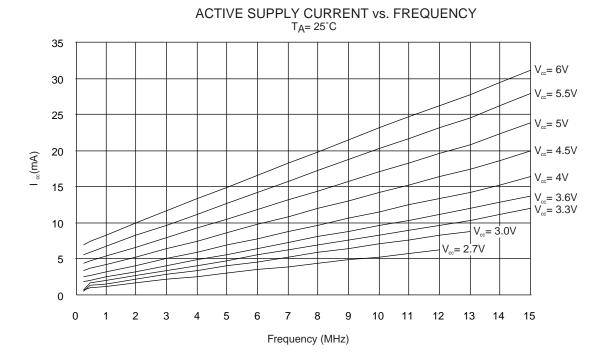
The current drawn from capacitive loaded pins may be estimated (for one pin) as  $C_L^*V_{CC}^*f$  where  $C_L$  = load capacitance,  $V_{CC}$  = operating voltage and f = average switching frequency of I/O pin.

The parts are characterized at frequencies higher than test limits. Parts are not guaranted to function properly at frequencies higher than the ordering code indicates.

The difference between current consumption in Power Down mode with Watchdog timer enabled and Power Down mode with Watchdog timer disabled represents the differential current drawn by the watchdog timer.

The difference between Power Down mode with Brown Out Detector enabled and Power Down mode with Watchdog timer disabled represents the differential current drawn by the brown out detector.





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Figure 73. Idle Supply Current vs.  $V_{CC}$ 

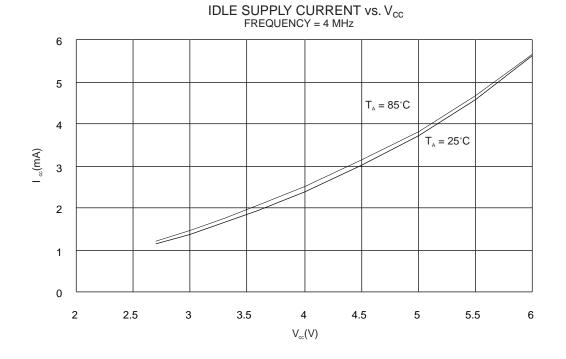


Figure 74. Power Down Supply Current vs. V<sub>CC</sub>

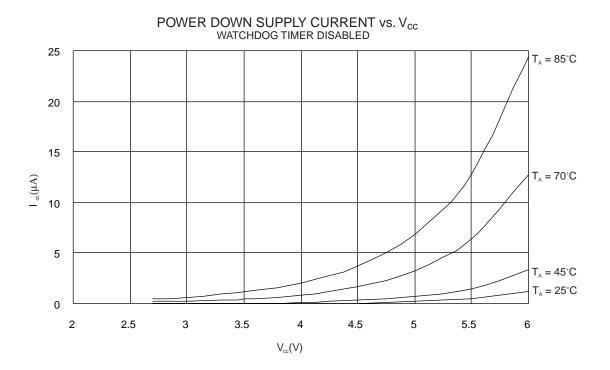


Figure 75. Power Down Supply Current vs. V<sub>CC</sub>

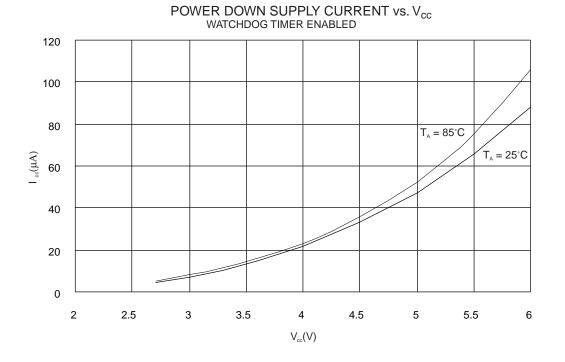
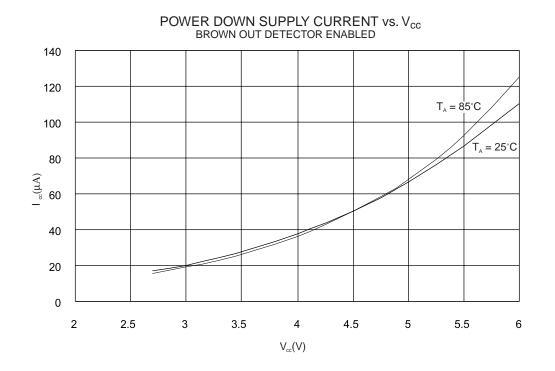


Figure 76. Power Down Supply Current vs. V<sub>CC</sub>







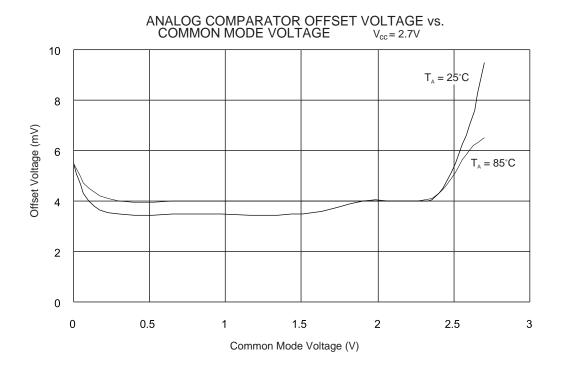
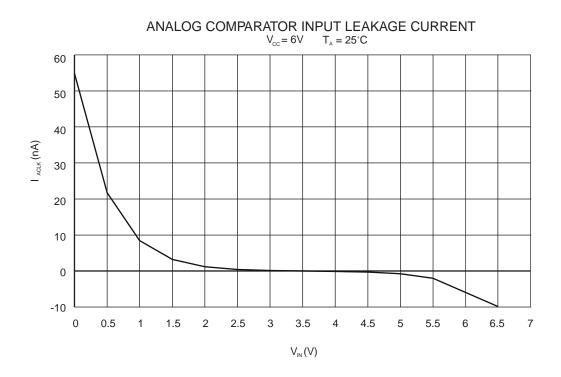
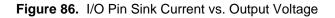


Figure 80. Analog Comparator Input Leakage Current





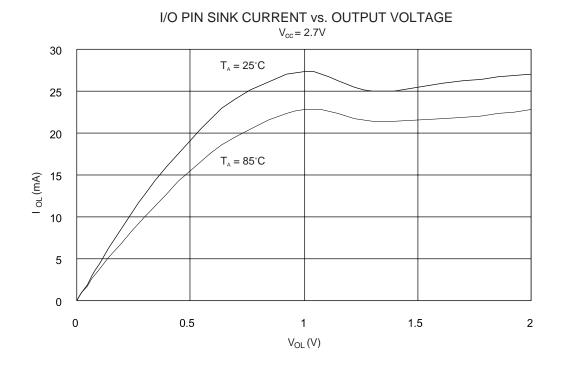


Figure 87. I/O Pin Source Current vs. Output Voltage

