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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	624
Number of Logic Elements/Cells	4992
Total RAM Bits	49152
Number of I/O	274
Number of Gates	257000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	356-LBGA
Supplier Device Package	356-BGA (35x35)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k100ebc356-1

Similar to the FLEX 10KE architecture, embedded gate arrays are the fastest-growing segment of the gate array market. As with standard gate arrays, embedded gate arrays implement general logic in a conventional “sea-of-gates” architecture. Additionally, embedded gate arrays have dedicated die areas for implementing large, specialized functions. By embedding functions in silicon, embedded gate arrays reduce die area and increase speed when compared to standard gate arrays. While embedded megafunctions typically cannot be customized, FLEX 10KE devices are programmable, providing the designer with full control over embedded megafunctions and general logic, while facilitating iterative design changes during debugging.

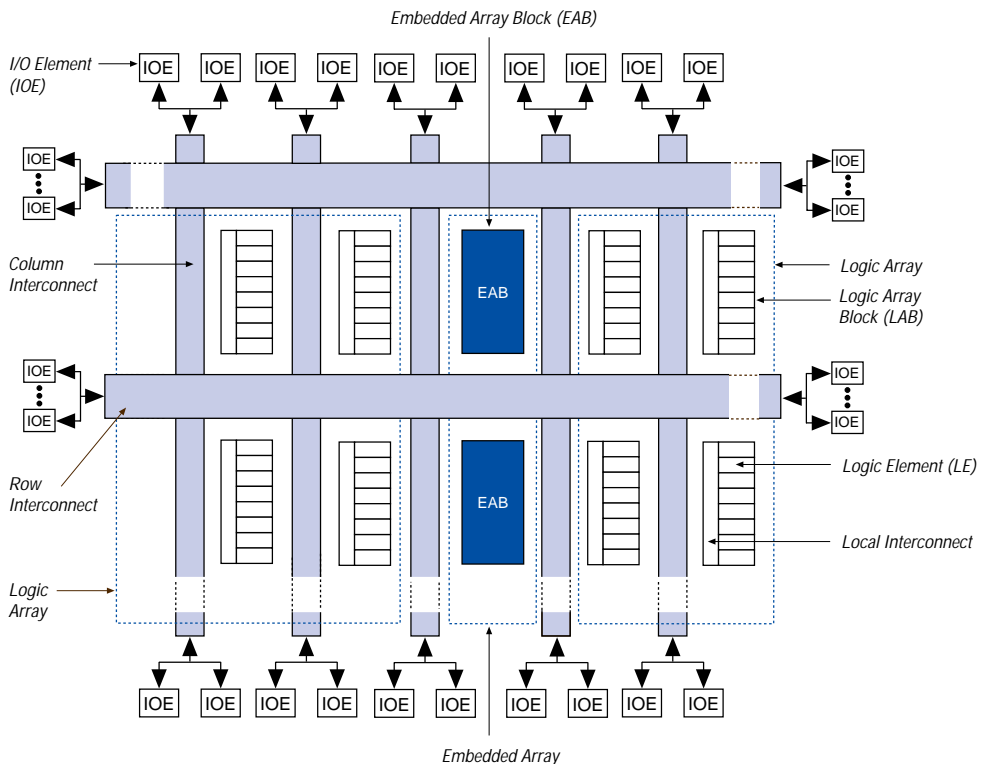
Each FLEX 10KE device contains an embedded array and a logic array. The embedded array is used to implement a variety of memory functions or complex logic functions, such as digital signal processing (DSP), wide data-path manipulation, microcontroller applications, and data-transformation functions. The logic array performs the same function as the sea-of-gates in the gate array and is used to implement general logic such as counters, adders, state machines, and multiplexers. The combination of embedded and logic arrays provides the high performance and high density of embedded gate arrays, enabling designers to implement an entire system on a single device.

FLEX 10KE devices are configured at system power-up with data stored in an Altera serial configuration device or provided by a system controller. Altera offers the EPC1, EPC2, and EPC16 configuration devices, which configure FLEX 10KE devices via a serial data stream. Configuration data can also be downloaded from system RAM or via the Altera BitBlaster™, ByteBlasterMV™, or MasterBlaster download cables. After a FLEX 10KE device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Because reconfiguration requires less than 85 ms, real-time changes can be made during system operation.

FLEX 10KE devices contain an interface that permits microprocessors to configure FLEX 10KE devices serially or in-parallel, and synchronously or asynchronously. The interface also enables microprocessors to treat a FLEX 10KE device as memory and configure it by writing to a virtual memory location, making it easy to reconfigure the device.

Figure 1 shows a block diagram of the FLEX 10KE architecture. Each group of LEs is combined into an LAB; groups of LABs are arranged into rows and columns. Each row also contains a single EAB. The LABs and EABs are interconnected by the FastTrack Interconnect routing structure. IOEs are located at the end of each row and column of the FastTrack Interconnect routing structure.

Figure 1. FLEX 10KE Device Block Diagram



FLEX 10KE devices provide six dedicated inputs that drive the flipflops' control inputs and ensure the efficient distribution of high-speed, low-skew (less than 1.5 ns) control signals. These signals use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect routing structure. Four of the dedicated inputs drive four global signals. These four global signals can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device.

The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock, clear, and preset control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the output of the LE.

The LE has two outputs that drive the interconnect: one drives the local interconnect and the other drives either the row or column FastTrack Interconnect routing structure. The two outputs can be controlled independently. For example, the LUT can drive one output while the register drives the other output. This feature, called register packing, can improve LE utilization because the register and the LUT can be used for unrelated functions.

The FLEX 10KE architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. The carry chain supports high-speed counters and adders and the cascade chain implements wide-input functions with minimum delay. Carry and cascade chains connect all LEs in a LAB as well as all LABs in the same row. Intensive use of carry and cascade chains can reduce routing flexibility. Therefore, the use of these chains should be limited to speed-critical portions of a design.

Carry Chain

The carry chain provides a very fast (as low as 0.2 ns) carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the FLEX 10KE architecture to implement high-speed counters, adders, and comparators of arbitrary width efficiently. Carry chain logic can be created automatically by the Altera Compiler during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of carry chains.

Carry chains longer than eight LEs are automatically implemented by linking LABs together. For enhanced fitting, a long carry chain skips alternate LABs in a row. A carry chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB. For example, the last LE of the first LAB in a row carries to the first LE of the third LAB in the row. The carry chain does not cross the EAB at the middle of the row. For instance, in the EPF10K50E device, the carry chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB.

LE Operating Modes

The FLEX 10KE LE can operate in the following four modes:

- Normal mode
- Arithmetic mode
- Up/down counter mode
- Clearable counter mode

Each of these modes uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. Three inputs to the LE provide clock, clear, and preset control for the register. The Altera software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions that use a specific LE operating mode for optimal performance.

The architecture provides a synchronous clock enable to the register in all four modes. The Altera software can set `DATA1` to enable the register synchronously, providing easy implementation of fully synchronous designs.

Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is ANDed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-states without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the Altera Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

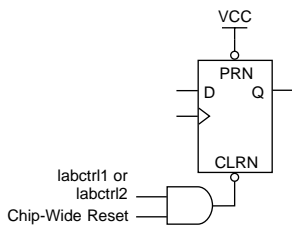
The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

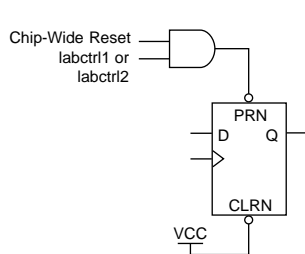
In addition to the six clear and preset modes, FLEX 10KE devices provide a chip-wide reset pin that can reset all registers in the device. Use of this feature is set during design entry. In any of the clear and preset modes, the chip-wide reset overrides all other signals. Registers with asynchronous presets may be preset when the chip-wide reset is asserted. Inversion can be used to implement the asynchronous preset. Figure 12 shows examples of how to setup the preset and clear inputs for the desired functionality.

Figure 12. FLEX 10KE LE Clear & Preset Modes

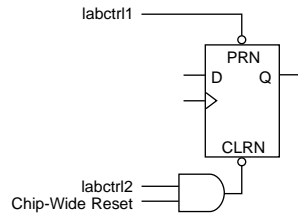
Asynchronous Clear



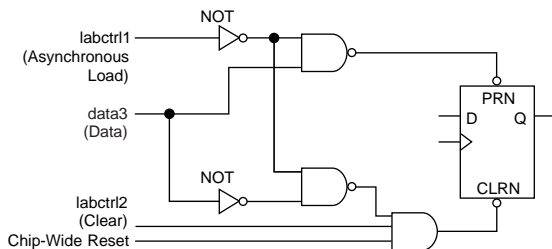
Asynchronous Preset



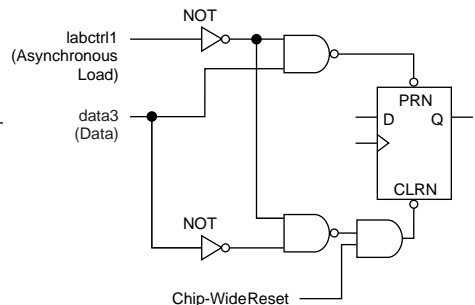
Asynchronous Preset & Clear



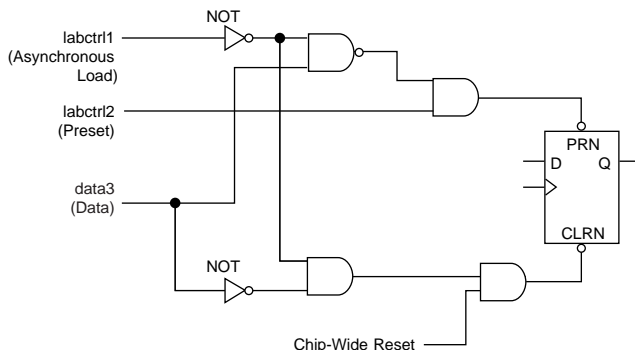
Asynchronous Load with Clear



Asynchronous Load without Clear or Preset



Asynchronous Load with Preset



Asynchronous Clear

The flipflop can be cleared by either LABCTRL1 or LABCTRL2. In this mode, the preset signal is tied to VCC to deactivate it.

Asynchronous Preset

An asynchronous preset is implemented as an asynchronous load, or with an asynchronous clear. If DATA3 is tied to VCC, asserting LABCTRL1 asynchronously loads a one into the register. Alternatively, the Altera software can provide preset control by using the clear and inverting the input and output of the register. Inversion control is available for the inputs to both LEs and IOEs. Therefore, if a register is preset by only one of the two LABCTRL signals, the DATA3 input is not needed and can be used for one of the LE operating modes.

Asynchronous Preset & Clear

When implementing asynchronous clear and preset, LABCTRL1 controls the preset and LABCTRL2 controls the clear. DATA3 is tied to VCC, so that asserting LABCTRL1 asynchronously loads a one into the register, effectively presetting the register. Asserting LABCTRL2 clears the register.

Asynchronous Load with Clear

When implementing an asynchronous load in conjunction with the clear, LABCTRL1 implements the asynchronous load of DATA3 by controlling the register preset and clear. LABCTRL2 implements the clear by controlling the register clear; LABCTRL2 does not have to feed the preset circuits.

Asynchronous Load with Preset

When implementing an asynchronous load in conjunction with preset, the Altera software provides preset control by using the clear and inverting the input and output of the register. Asserting LABCTRL2 presets the register, while asserting LABCTRL1 loads the register. The Altera software inverts the signal that drives DATA3 to account for the inversion of the register's output.

Asynchronous Load without Preset or Clear

When implementing an asynchronous load without preset or clear, LABCTRL1 implements the asynchronous load of DATA3 by controlling the register preset and clear.

Row-to-IOE Connections

When an IOE is used as an input signal, it can drive two separate row channels. The signal is accessible by all LEs within that row. When an IOE is used as an output, the signal is driven by a multiplexer that selects a signal from the row channels. Up to eight IOEs connect to each side of each row channel (see Figure 16).

Figure 16. FLEX 10KE Row-to-IOE Connections

The values for m and n are provided in Table 10.

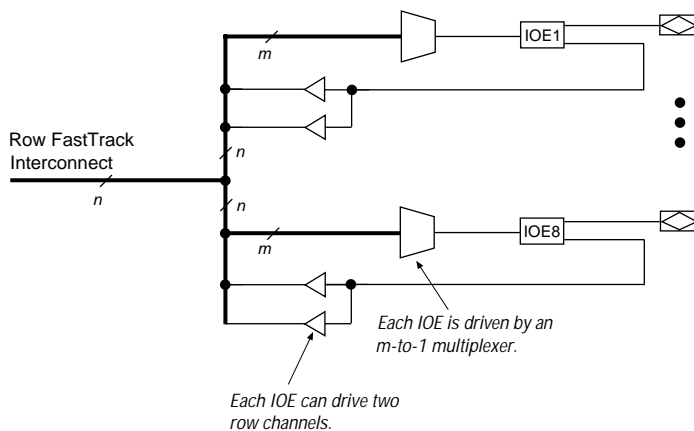


Table 10 lists the FLEX 10KE row-to-IOE interconnect resources.

Table 10. FLEX 10KE Row-to-IOE Interconnect Resources		
Device	Channels per Row (n)	Row Channels per Pin (m)
EPF10K30E	216	27
EPF10K50E EPF10K50S	216	27
EPF10K100E	312	39
EPF10K130E	312	39
EPF10K200E EPF10K200S	312	39

IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

All FLEX 10KE devices provide JTAG BST circuitry that complies with the IEEE Std. 1149.1-1990 specification. FLEX 10KE devices can also be configured using the JTAG pins through the BitBlaster or ByteBlasterMV download cable, or via hardware that uses the Jam™ STAPL programming and test language. JTAG boundary-scan testing can be performed before or after configuration, but not during configuration. FLEX 10KE devices support the JTAG instructions shown in [Table 15](#).

Table 15. FLEX 10KE JTAG Instructions

JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins.
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation.
USERCODE	Selects the user electronic signature (USERCODE) register and places it between the TDI and TDO pins, allowing the USERCODE to be serially shifted out of TDO.
IDCODE	Selects the IDCODE register and places it between TDI and TDO, allowing the IDCODE to be serially shifted out of TDO.
ICR Instructions	These instructions are used when configuring a FLEX 10KE device via JTAG ports with a BitBlaster or ByteBlasterMV download cable, or using a Jam File (.jam) or Jam Byte-Code File (.jbc) via an embedded processor.

The instruction register length of FLEX 10KE devices is 10 bits. The USERCODE register length in FLEX 10KE devices is 32 bits; 7 bits are determined by the user, and 25 bits are pre-determined. [Tables 16](#) and [17](#) show the boundary-scan register length and device IDCODE information for FLEX 10KE devices.

Table 16. FLEX 10KE Boundary-Scan Register Length

Device	Boundary-Scan Register Length
EPF10K30E	690
EPF10K50E EPF10K50S	798
EPF10K100E	1,050
EPF10K130E	1,308
EPF10K200E EPF10K200S	1,446

Figure 20 shows the timing requirements for the JTAG signals.

Figure 20. FLEX 10KE JTAG Waveforms

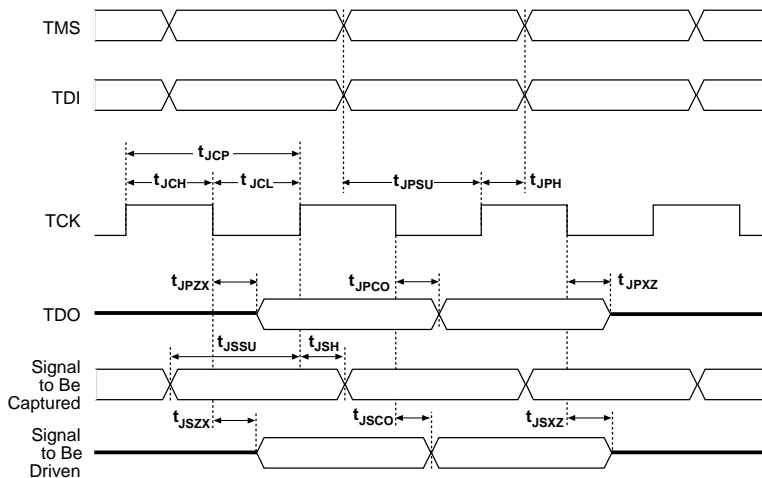


Table 18 shows the timing parameters and values for FLEX 10KE devices.

Table 18. FLEX 10KE JTAG Timing Parameters & Values

Symbol	Parameter	Min	Max	Unit
t_{JCP}	TCK clock period	100		ns
t_{JCH}	TCK clock high time	50		ns
t_{JCL}	TCK clock low time	50		ns
t_{JPSU}	JTAG port setup time	20		ns
t_{JPH}	JTAG port hold time	45		ns
t_{JPCO}	JTAG port clock to output		25	ns
t_{JPZX}	JTAG port high impedance to valid output		25	ns
t_{JPXZ}	JTAG port valid output to high impedance		25	ns
t_{JSSU}	Capture register setup time	20		ns
t_{JSH}	Capture register hold time	45		ns
t_{JSCO}	Update register clock to output		35	ns
t_{JSZX}	Update register high impedance to valid output		35	ns
t_{JSXZ}	Update register valid output to high impedance		35	ns

Figure 25. FLEX 10KE Device LE Timing Model

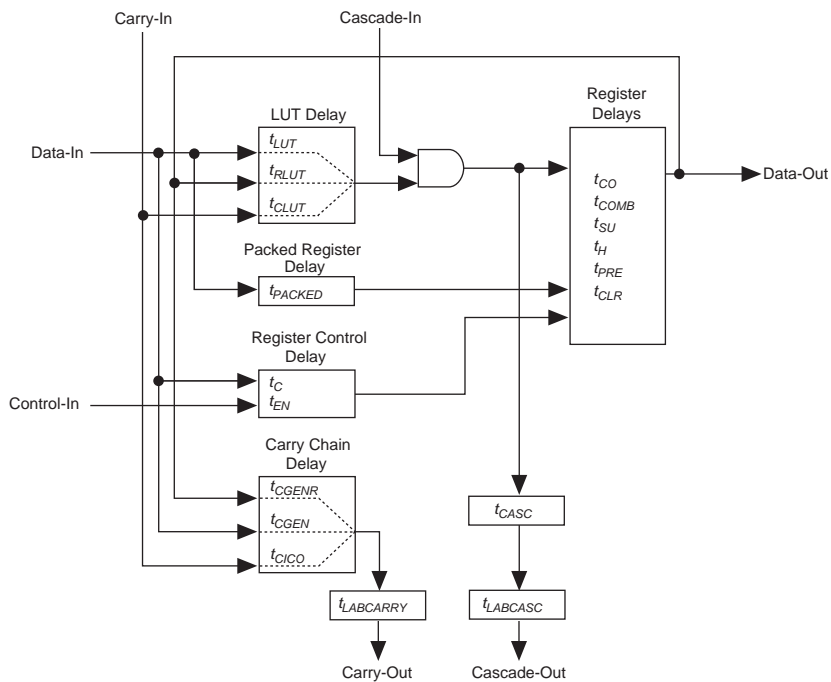
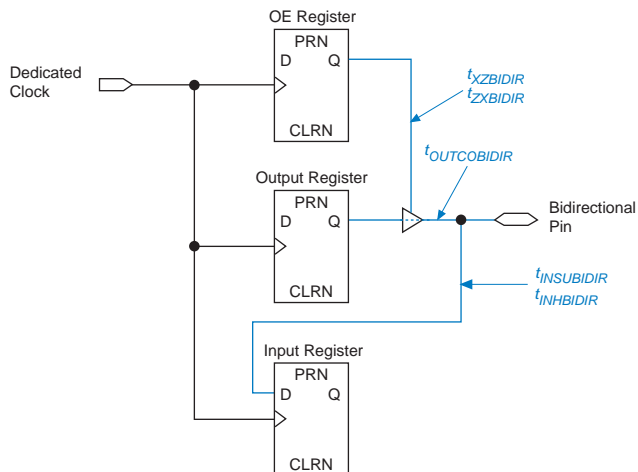


Figure 28. Synchronous Bidirectional Pin External Timing Model



Tables 24 through 28 describe the FLEX 10KE device internal timing parameters. Tables 29 through 30 describe the FLEX 10KE external timing parameters and their symbols.

Table 24. LE Timing Microparameters (Part 1 of 2) Note (1)

Symbol	Parameter	Condition
t_{LUT}	LUT delay for data-in	
t_{CLUT}	LUT delay for carry-in	
t_{RLUT}	LUT delay for LE register feedback	
t_{PACKED}	Data-in to packed register delay	
t_{EN}	LE register enable delay	
t_{CICO}	Carry-in to carry-out delay	
t_{CGEN}	Data-in to carry-out delay	
t_{CGENR}	LE register feedback to carry-out delay	
t_{CASC}	Cascade-in to cascade-out delay	
t_C	LE register control signal delay	
t_{CO}	LE register clock-to-output delay	
t_{COMB}	Combinatorial delay	
t_{SU}	LE register setup time for data and enable signals before clock; LE register recovery time after asynchronous clear, preset, or load	
t_H	LE register hold time for data and enable signals after clock	
t_{PRE}	LE register preset delay	

Table 26. EAB Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
$t_{EABDATA1}$	Data or address delay to EAB for combinatorial input	
$t_{EABDATA2}$	Data or address delay to EAB for registered input	
t_{EABWE1}	Write enable delay to EAB for combinatorial input	
t_{EABWE2}	Write enable delay to EAB for registered input	
t_{EABRE1}	Read enable delay to EAB for combinatorial input	
t_{EABRE2}	Read enable delay to EAB for registered input	
t_{EABCLK}	EAB register clock delay	
t_{EABCO}	EAB register clock-to-output delay	
$t_{EABYPASS}$	Bypass register delay	
t_{EABSU}	EAB register setup time before clock	
t_{EABH}	EAB register hold time after clock	
t_{EABCLR}	EAB register asynchronous clear time to output delay	
t_{AA}	Address access delay (including the read enable to output delay)	
t_{WP}	Write pulse width	
t_{RP}	Read pulse width	
t_{WDSU}	Data setup time before falling edge of write pulse	(5)
t_{WDH}	Data hold time after falling edge of write pulse	(5)
t_{WASU}	Address setup time before rising edge of write pulse	(5)
t_{WAH}	Address hold time after falling edge of write pulse	(5)
t_{RASU}	Address setup time with respect to the falling edge of the read enable	
t_{RAH}	Address hold time with respect to the falling edge of the read enable	
t_{WO}	Write enable to data output valid delay	
t_{DD}	Data-in to data-out valid delay	
t_{EABOUT}	Data-out delay	
t_{EABCH}	Clock high time	
t_{EABCL}	Clock low time	

Table 34. EPF10K30E Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		6.4		7.6		8.8	ns
$t_{EABRCOMB}$	6.4		7.6		8.8		ns
$t_{EABRCREG}$	4.4		5.1		6.0		ns
t_{EABWP}	2.5		2.9		3.3		ns
$t_{EABWCOMB}$	6.0		7.0		8.0		ns
$t_{EABWCREG}$	6.8		7.8		9.0		ns
t_{EABDD}		5.7		6.7		7.7	ns
$t_{EABDATACO}$		0.8		0.9		1.1	ns
$t_{EABDATASU}$	1.5		1.7		2.0		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	1.3		1.4		1.7		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	1.5		1.7		2.0		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	3.0		3.6		4.3		ns
t_{EABWAH}	0.5		0.5		0.4		ns
t_{EABWO}		5.1		6.0		6.8	ns

Table 41. EPF10K50E Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		6.4		7.6		10.2	ns
$t_{EABRCOMB}$	6.4		7.6		10.2		ns
$t_{EABRCREG}$	4.4		5.1		7.0		ns
t_{EABWP}	2.5		2.9		3.9		ns
$t_{EABWCOMB}$	6.0		7.0		9.5		ns
$t_{EABWCREG}$	6.8		7.8		10.6		ns
t_{EABDD}		5.7		6.7		9.0	ns
$t_{EABDATACO}$		0.8		0.9		1.3	ns
$t_{EABDATASU}$	1.5		1.7		2.3		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	1.3		1.4		2.0		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	1.5		1.7		2.3		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	3.0		3.6		4.8		ns
t_{EABWAH}	0.5		0.5		0.8		ns
t_{EABWO}		5.1		6.0		8.1	ns

Table 42. EPF10K50E Device Interconnect Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		3.5		4.3		5.6	ns
t_{DIN2LE}		2.1		2.5		3.4	ns
$t_{DIN2DATA}$		2.2		2.4		3.1	ns
$t_{DCLK2IOE}$		2.9		3.5		4.7	ns
$t_{DCLK2LE}$		2.1		2.5		3.4	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		1.1		1.1		1.5	ns
$t_{SAMECOLUMN}$		0.8		1.0		1.3	ns
$t_{DIFFROW}$		1.9		2.1		2.8	ns
$t_{TWOROWS}$		3.0		3.2		4.3	ns
$t_{LEPERIPH}$		3.1		3.3		3.7	ns
$t_{LABCARRY}$		0.1		0.1		0.2	ns
$t_{LABCASC}$		0.3		0.3		0.5	ns

Table 59. EPF10K200E Device LE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_H	0.9		1.1		1.5		ns
t_{PRE}		0.5		0.6		0.8	ns
t_{CLR}		0.5		0.6		0.8	ns
t_{CH}	2.0		2.5		3.0		ns
t_{CL}	2.0		2.5		3.0		ns

Table 60. EPF10K200E Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.6		1.9		2.6	ns
t_{IOC}		0.3		0.3		0.5	ns
t_{IOCO}		1.6		1.9		2.6	ns
t_{IOCOMB}		0.5		0.6		0.8	ns
t_{IOSU}	0.8		0.9		1.2		ns
t_{IOH}	0.7		0.8		1.1		ns
t_{IOCLR}		0.2		0.2		0.3	ns
t_{OD1}		0.6		0.7		0.9	ns
t_{OD2}		0.1		0.2		0.7	ns
t_{OD3}		2.5		3.0		3.9	ns
t_{XZ}		4.4		5.3		7.1	ns
t_{ZX1}		4.4		5.3		7.1	ns
t_{ZX2}		3.9		4.8		6.9	ns
t_{ZX3}		6.3		7.6		10.1	ns
t_{INREG}		4.8		5.7		7.7	ns
t_{IOFD}		1.5		1.8		2.4	ns
t_{INCOMB}		1.5		1.8		2.4	ns

Table 66. EPF10K50S Device LE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{CGENR}		0.1		0.1		0.1	ns
t_{CASC}		0.5		0.8		1.0	ns
t_C		0.5		0.6		0.8	ns
t_{CO}		0.6		0.6		0.7	ns
t_{COMB}		0.3		0.4		0.5	ns
t_{SU}	0.5		0.6		0.7		ns
t_H	0.5		0.6		0.8		ns
t_{PRE}		0.4		0.5		0.7	ns
t_{CLR}		0.8		1.0		1.2	ns
t_{CH}	2.0		2.5		3.0		ns
t_{CL}	2.0		2.5		3.0		ns

Table 67. EPF10K50S Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.3		1.3		1.9	ns
t_{IOC}		0.3		0.4		0.4	ns
t_{IOCO}		1.7		2.1		2.6	ns
t_{IOCOMB}		0.5		0.6		0.8	ns
t_{IOSU}	0.8		1.0		1.3		ns
t_{IOH}	0.4		0.5		0.6		ns
t_{IOCLR}		0.2		0.2		0.4	ns
t_{OD1}		1.2		1.2		1.9	ns
t_{OD2}		0.7		0.8		1.7	ns
t_{OD3}		2.7		3.0		4.3	ns
t_{XZ}		4.7		5.7		7.5	ns
t_{ZX1}		4.7		5.7		7.5	ns
t_{ZX2}		4.2		5.3		7.3	ns
t_{ZX3}		6.2		7.5		9.9	ns
t_{INREG}		3.5		4.2		5.6	ns
t_{IOFD}		1.1		1.3		1.8	ns
t_{INCOMB}		1.1		1.3		1.8	ns

Table 71. EPF10K50S External Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		8.0		9.5		12.5	ns
$t_{\text{INSU}}^{(2)}$	2.4		2.9		3.9		ns
$t_{\text{INH}}^{(2)}$	0.0		0.0		0.0		ns
$t_{\text{OUTCO}}^{(2)}$	2.0	4.3	2.0	5.2	2.0	7.3	ns
$t_{\text{INSU}}^{(3)}$	2.4		2.9				ns
$t_{\text{INH}}^{(3)}$	0.0		0.0				ns
$t_{\text{OUTCO}}^{(3)}$	0.5	3.3	0.5	4.1			ns
t_{PCISU}	2.4		2.9		—		ns
t_{PCIH}	0.0		0.0		—		ns
t_{PCICO}	2.0	6.0	2.0	7.7	—	—	ns

Table 72. EPF10K50S External Bidirectional Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{\text{INSUBIDIR}}^{(2)}$	2.7		3.2		4.3		ns
$t_{\text{INHBIDIR}}^{(2)}$	0.0		0.0		0.0		ns
$t_{\text{INHBIDIR}}^{(3)}$	0.0		0.0		—		ns
$t_{\text{INSUBIDIR}}^{(3)}$	3.7		4.2		—		ns
$t_{\text{OUTCOBIDIR}}^{(2)}$	2.0	4.5	2.0	5.2	2.0	7.3	ns
$t_{\text{XZBIDIR}}^{(2)}$		6.8		7.8		10.1	ns
$t_{\text{ZXBIDIR}}^{(2)}$		6.8		7.8		10.1	ns
$t_{\text{OUTCOBIDIR}}^{(3)}$	0.5	3.5	0.5	4.2	—	—	
$t_{\text{XZBIDIR}}^{(3)}$		6.8		8.4		—	ns
$t_{\text{ZXBIDIR}}^{(3)}$		6.8		8.4		—	ns

Notes to tables:

- (1) All timing parameters are described in [Tables 24 through 30](#).
- (2) This parameter is measured without use of the ClockLock or ClockBoost circuits.
- (3) This parameter is measured with use of the ClockLock or ClockBoost circuits

Table 73. EPF10K200S Device Internal & External Timing Parameters

Note (1)

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.7		0.8		1.2	ns
t_{CLUT}		0.4		0.5		0.6	ns
t_{RLUT}		0.5		0.7		0.9	ns
t_{PACKED}		0.4		0.5		0.7	ns
t_{EN}		0.6		0.5		0.6	ns
t_{CICO}		0.1		0.2		0.3	ns
t_{CGEN}		0.3		0.4		0.6	ns
t_{CGENR}		0.1		0.2		0.3	ns
t_{CASC}		0.7		0.8		1.2	ns
t_C		0.5		0.6		0.8	ns
t_{CO}		0.5		0.6		0.8	ns
t_{COMB}		0.3		0.6		0.8	ns
t_{SU}	0.4		0.6		0.7		ns
t_H	1.0		1.1		1.5		ns
t_{PRE}		0.4		0.6		0.8	ns
t_{CLR}		0.5		0.6		0.8	ns
t_{CH}	2.0		2.5		3.0		ns
t_{CL}	2.0		2.5		3.0		ns

Table 74. EPF10K200S Device IOE Timing Microparameters (Part 1 of 2)

Note (1)

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.8		1.9		2.6	ns
t_{IOC}		0.3		0.3		0.5	ns
t_{IOCO}		1.7		1.9		2.6	ns
t_{IOCOMB}		0.5		0.6		0.8	ns
t_{IOSU}	0.8		0.9		1.2		ns
t_{IOH}	0.4		0.8		1.1		ns
t_{IOCLR}		0.2		0.2		0.3	ns
t_{OD1}		1.3		0.7		0.9	ns
t_{OD2}		0.8		0.2		0.4	ns
t_{OD3}		2.9		3.0		3.9	ns
t_{XZ}		5.0		5.3		7.1	ns
t_{ZX1}		5.0		5.3		7.1	ns

During initialization, which occurs immediately after configuration, the device resets registers, enables I/O pins, and begins to operate as a logic device. The I/O pins are tri-stated during power-up, and before and during configuration. Together, the configuration and initialization processes are called *command mode*; normal device operation is called *user mode*.

SRAM configuration elements allow FLEX 10KE devices to be reconfigured in-circuit by loading new configuration data into the device. Real-time reconfiguration is performed by forcing the device into command mode with a device pin, loading different configuration data, reinitializing the device, and resuming user-mode operation. The entire reconfiguration process requires less than 85 ms and can be used to reconfigure an entire system dynamically. In-field upgrades can be performed by distributing new configuration files.

Before and during configuration, all I/O pins (except dedicated inputs, clock, or configuration pins) are pulled high by a weak pull-up resistor.

Programming Files

Despite being function- and pin-compatible, FLEX 10KE devices are not programming- or configuration file-compatible with FLEX 10K or FLEX 10KA devices. A design therefore must be recompiled before it is transferred from a FLEX 10K or FLEX 10KA device to an equivalent FLEX 10KE device. This recompilation should be performed both to create a new programming or configuration file and to check design timing in FLEX 10KE devices, which has different timing characteristics than FLEX 10K or FLEX 10KA devices.

FLEX 10KE devices are generally pin-compatible with equivalent FLEX 10KA devices. In some cases, FLEX 10KE devices have fewer I/O pins than the equivalent FLEX 10KA devices. [Table 81](#) shows which FLEX 10KE devices have fewer I/O pins than equivalent FLEX 10KA devices. However, power, ground, JTAG, and configuration pins are the same on FLEX 10KA and FLEX 10KE devices, enabling migration from a FLEX 10KA design to a FLEX 10KE design.