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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Active
Number of LABs/CLBs	624
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	191
Number of Gates	-
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	256-BGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=epf10k100efc256-2x

Table 2. FLEX 10KE Device Features

Feature	EPF10K100E (2)	EPF10K130E	EPF10K200E EPF10K200S
Typical gates (1)	100,000	130,000	200,000
Maximum system gates	257,000	342,000	513,000
Logic elements (LEs)	4,992	6,656	9,984
EABs	12	16	24
Total RAM bits	49,152	65,536	98,304
Maximum user I/O pins	338	413	470

Note to tables:

- (1) The embedded IEEE Std. 1149.1 JTAG circuitry adds up to 31,250 gates in addition to the listed typical or maximum system gates.
- (2) New EPF10K100B designs should use EPF10K100E devices.

...and More Features

- Fabricated on an advanced process and operate with a 2.5-V internal supply voltage
- In-circuit reconfigurability (ICR) via external configuration devices, intelligent controller, or JTAG port
- ClockLock™ and ClockBoost™ options for reduced clock delay/skew and clock multiplication
- Built-in low-skew clock distribution trees
- 100% functional testing of all devices; test vectors or scan chains are not required
- Pull-up on I/O pins before and during configuration
- Flexible interconnect
 - FastTrack® Interconnect continuous routing structure for fast, predictable interconnect delays
 - Dedicated carry chain that implements arithmetic functions such as fast adders, counters, and comparators (automatically used by software tools and megafunctions)
 - Dedicated cascade chain that implements high-speed, high-fan-in logic functions (automatically used by software tools and megafunctions)
 - Tri-state emulation that implements internal tri-state buses
 - Up to six global clock signals and four global clear signals
- Powerful I/O pins
 - Individual tri-state output enable control for each pin
 - Open-drain option on each I/O pin
 - Programmable output slew-rate control to reduce switching noise
 - Clamp to V_{CCIO} user-selectable on a pin-by-pin basis
 - Supports hot-socketing



For more information on FLEX device configuration, see the following documents:

- [Configuration Devices for APEX & FLEX Devices Data Sheet](#)
- [BitBlaster Serial Download Cable Data Sheet](#)
- [ByteBlasterMV Parallel Port Download Cable Data Sheet](#)
- [MasterBlaster Download Cable Data Sheet](#)
- [Application Note 116 \(Configuring APEX 20K, FLEX 10K, & FLEX 6000 Devices\)](#)

FLEX 10KE devices are supported by the Altera development systems, which are integrated packages that offer schematic, text (including AHDL), and waveform design entry, compilation and logic synthesis, full simulation and worst-case timing analysis, and device configuration. The Altera software provides EDIF 2.0.0 and 3.0.0, LPM, VHDL, Verilog HDL, and other interfaces for additional design entry and simulation support from other industry-standard PC- and UNIX workstation-based EDA tools.

The Altera software works easily with common gate array EDA tools for synthesis and simulation. For example, the Altera software can generate Verilog HDL files for simulation with tools such as Cadence Verilog-XL. Additionally, the Altera software contains EDA libraries that use device-specific features such as carry chains, which are used for fast counter and arithmetic functions. For instance, the Synopsys Design Compiler library supplied with the Altera development system includes DesignWare functions that are optimized for the FLEX 10KE architecture.

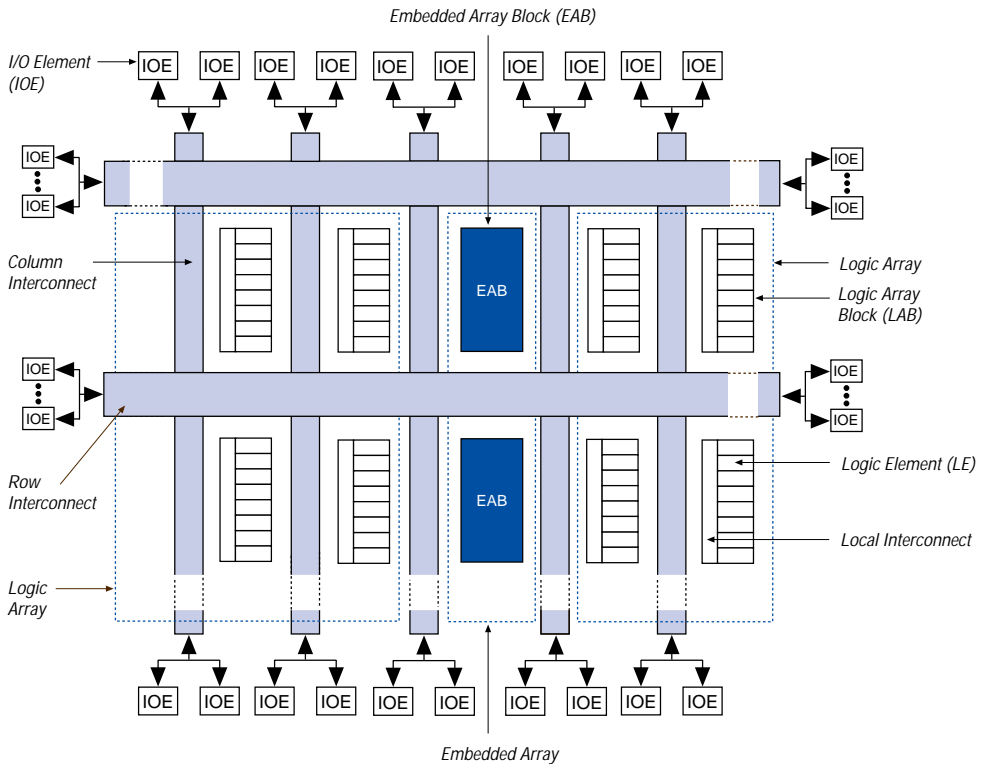
The Altera development system runs on Windows-based PCs and Sun SPARCstation, and HP 9000 Series 700/800.



See the [MAX+PLUS II Programmable Logic Development System & Software Data Sheet](#) and the [Quartus Programmable Logic Development System & Software Data Sheet](#) for more information.

Figure 1 shows a block diagram of the FLEX 10KE architecture. Each group of LEs is combined into an LAB; groups of LABs are arranged in rows and columns. Each row also contains a single EAB. The LABs and EABs are interconnected by the FastTrack Interconnect routing structure. IOEs are located at the end of each row and column of the FastTrack Interconnect routing structure.

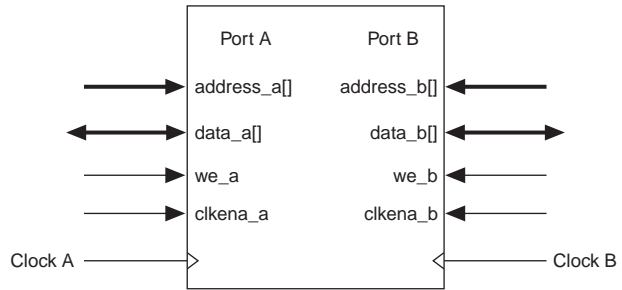
Figure 1. FLEX 10KE Device Block Diagram



FLEX 10KE devices provide six dedicated inputs that drive the flipflops' control inputs and ensure the efficient distribution of high-speed, low-skew (less than 1.5 ns) control signals. These signals use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect routing structure. Four of the dedicated inputs drive four global signals. These four global signals can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device.

The EAB can also use Altera megafunctions to implement dual-port RAM applications where both ports can read or write, as shown in [Figure 3](#).

Figure 3. FLEX 10KE EAB in Dual-Port RAM Mode



The FLEX 10KE EAB can be used in a single-port mode, which is useful for backward-compatibility with FLEX 10K designs (see [Figure 4](#)).

EABs provide flexible options for driving and controlling clock signals. Different clocks and clock enables can be used for reading and writing to the EAB. Registers can be independently inserted on the data input, EAB output, write address, write enable signals, read address, and read enable signals. The global signals and the EAB local interconnect can drive write enable, read enable, and clock enable signals. The global signals, dedicated clock pins, and EAB local interconnect can drive the EAB clock signals. Because the LEs drive the EAB local interconnect, the LEs can control write enable, read enable, clear, clock, and clock enable signals.

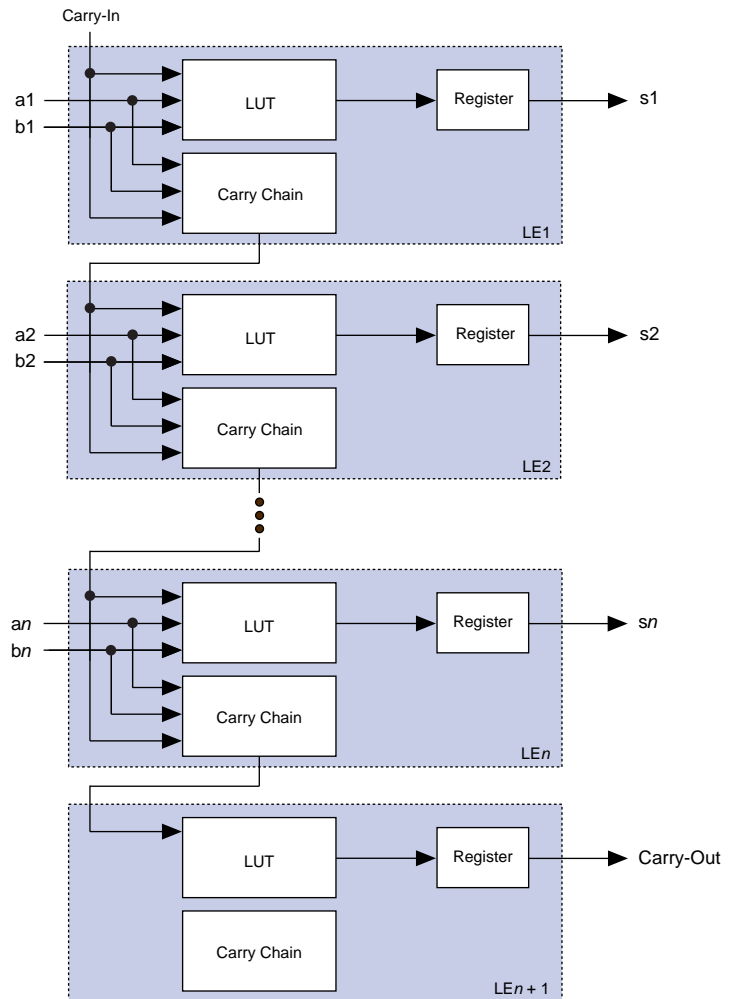
An EAB is fed by a row interconnect and can drive out to row and column interconnects. Each EAB output can drive up to two row channels and up to two column channels; the unused row channel can be driven by other LEs. This feature increases the routing resources available for EAB outputs (see [Figures 2 and 4](#)). The column interconnect, which is adjacent to the EAB, has twice as many channels as other columns in the device.

Logic Array Block

An LAB consists of eight LEs, their associated carry and cascade chains, LAB control signals, and the LAB local interconnect. The LAB provides the coarse-grained structure to the FLEX 10KE architecture, facilitating efficient routing with optimum device utilization and high performance (see [Figure 7](#)).

Figure 9 shows how an n -bit full adder can be implemented in $n + 1$ LEs with the carry chain. One portion of the LUT generates the sum of two bits using the input signals and the carry-in signal; the sum is routed to the output of the LE. The register can be bypassed for simple adders or used for an accumulator function. Another portion of the LUT and the carry chain logic generates the carry-out signal, which is routed directly to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it can be used as a general-purpose signal.

Figure 9. FLEX 10KE Carry Chain Operation (n -Bit Full Adder)



LE Operating Modes

The FLEX 10KE LE can operate in the following four modes:

- Normal mode
- Arithmetic mode
- Up/down counter mode
- Clearable counter mode

Each of these modes uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. Three inputs to the LE provide clock, clear, and preset control for the register. The Altera software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions that use a specific LE operating mode for optimal performance.

The architecture provides a synchronous clock enable to the register in all four modes. The Altera software can set `DATA1` to enable the register synchronously, providing easy implementation of fully synchronous designs.

Normal Mode

The normal mode is suitable for general logic applications and wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a four-input LUT. The Altera Compiler automatically selects the carry-in or the `DATA3` signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. Either the register or the LUT can be used to drive both the local interconnect and the FastTrack Interconnect routing structure at the same time.

The LUT and the register in the LE can be used independently (register packing). To support register packing, the LE has two outputs; one drives the local interconnect, and the other drives the FastTrack Interconnect routing structure. The `DATA4` signal can drive the register directly, allowing the LUT to compute a function that is independent of the registered signal; a three-input function can be computed in the LUT, and a fourth independent signal can be registered. Alternatively, a four-input function can be generated, and one of the inputs to this function can be used to drive the register. The register in a packed LE can still use the clock enable, clear, and preset signals in the LE. In a packed LE, the register can drive the FastTrack Interconnect routing structure while the LUT drives the local interconnect, or vice versa.

Arithmetic Mode

The arithmetic mode offers 2 three-input LUTs that are ideal for implementing adders, accumulators, and comparators. One LUT computes a three-input function; the other generates a carry output. As shown in [Figure 11](#) on [page 22](#), the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, in an adder, this output is the sum of three signals: `a`, `b`, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

Up/Down Counter Mode

The up/down counter mode offers counter enable, clock enable, synchronous up/down control, and data loading options. These control signals are generated by the data inputs from the LAB local interconnect, the carry-in signal, and output feedback from the programmable register. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading. Data can also be loaded asynchronously with the clear and preset register control signals without using the LUT resources.

Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is ANDed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-states without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the Altera Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

FastTrack Interconnect Routing Structure

In the FLEX 10KE architecture, connections between LEs, EABs, and device I/O pins are provided by the FastTrack Interconnect routing structure, which is a series of continuous horizontal and vertical routing channels that traverses the device. This global routing structure provides predictable performance, even in complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.

The FastTrack Interconnect routing structure consists of row and column interconnect channels that span the entire device. Each row of LABs is served by a dedicated row interconnect. The row interconnect can drive I/O pins and feed other LABs in the row. The column interconnect routes signals between rows and can drive I/O pins.

Row channels drive into the LAB or EAB local interconnect. The row signal is buffered at every LAB or EAB to reduce the effect of fan-out on delay. A row channel can be driven by an LE or by one of three column channels. These four signals feed dual 4-to-1 multiplexers that connect to two specific row channels. These multiplexers, which are connected to each LE, allow column channels to drive row channels even when all eight LEs in a LAB drive the row interconnect.

Each column of LABs or EABs is served by a dedicated column interconnect. The column interconnect that serves the EABs has twice as many channels as other column interconnects. The column interconnect can then drive I/O pins or another row's interconnect to route the signals to other LABs or EABs in the device. A signal from the column interconnect, which can be either the output of a LE or an input from an I/O pin, must be routed to the row interconnect before it can enter a LAB or EAB. Each row channel that is driven by an IOE or EAB can drive one specific column channel.

Access to row and column channels can be switched between LEs in adjacent pairs of LABs. For example, a LE in one LAB can drive the row and column channels normally driven by a particular LE in the adjacent LAB in the same row, and vice versa. This flexibility enables routing resources to be used more efficiently (see [Figure 13](#)).

On all FLEX 10KE devices (except EPF10K50E and EPF10K200E devices), the input path from the I/O pad to the FastTrack Interconnect has a programmable delay element that can be used to guarantee a zero hold time. EPF10K50S and EPF10K200S devices also support this feature. Depending on the placement of the IOE relative to what it is driving, the designer may choose to turn on the programmable delay to ensure a zero hold time or turn it off to minimize setup time. This feature is used to reduce setup time for complex pin-to-register paths (e.g., PCI designs).

Each IOE selects the clock, clear, clock enable, and output enable controls from a network of I/O control signals called the peripheral control bus. The peripheral control bus uses high-speed drivers to minimize signal skew across the device and provides up to 12 peripheral control signals that can be allocated as follows:

- Up to eight output enable signals
- Up to six clock enable signals
- Up to two clock signals
- Up to two clear signals

If more than six clock enable or eight output enable signals are required, each IOE on the device can be controlled by clock enable and output enable signals driven by specific LEs. In addition to the two clock signals available on the peripheral control bus, each IOE can use one of two dedicated clock pins. Each peripheral control signal can be driven by any of the dedicated input pins or the first LE of each LAB in a particular row. In addition, a LE in a different row can drive a column interconnect, which causes a row interconnect to drive the peripheral control signal. The chip-wide reset signal resets all IOE registers, overriding any other control signals.

When a dedicated clock pin drives IOE registers, it can be inverted for all IOEs in the device. All IOEs must use the same sense of the clock. For example, if any IOE uses the inverted clock, all IOEs must use the inverted clock and no IOE can use the non-inverted clock. However, LEs can still use the true or complement of the clock on a LAB-by-LAB basis.

The incoming signal may be inverted at the dedicated clock pin and will drive all IOEs. For the true and complement of a clock to be used to drive IOEs, drive it into both global clock pins. One global clock pin will supply the true, and the other will supply the complement.

When the true and complement of a dedicated input drives IOE clocks, two signals on the peripheral control bus are consumed, one for each sense of the clock.

IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

All FLEX 10KE devices provide JTAG BST circuitry that complies with the IEEE Std. 1149.1-1990 specification. FLEX 10KE devices can also be configured using the JTAG pins through the BitBlaster or ByteBlasterMV download cable, or via hardware that uses the Jam™ STAPL programming and test language. JTAG boundary-scan testing can be performed before or after configuration, but not during configuration. FLEX 10KE devices support the JTAG instructions shown in [Table 15](#).

Table 15. FLEX 10KE JTAG Instructions

JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins.
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation.
USERCODE	Selects the user electronic signature (USERCODE) register and places it between the TDI and TDO pins, allowing the USERCODE to be serially shifted out of TDO.
IDCODE	Selects the IDCODE register and places it between TDI and TDO, allowing the IDCODE to be serially shifted out of TDO.
ICR Instructions	These instructions are used when configuring a FLEX 10KE device via JTAG ports with a BitBlaster or ByteBlasterMV download cable, or using a Jam File (.jam) or Jam Byte-Code File (.jbc) via an embedded processor.

The instruction register length of FLEX 10KE devices is 10 bits. The USERCODE register length in FLEX 10KE devices is 32 bits; 7 bits are determined by the user, and 25 bits are pre-determined. [Tables 16](#) and [17](#) show the boundary-scan register length and device IDCODE information for FLEX 10KE devices.

Table 16. FLEX 10KE Boundary-Scan Register Length

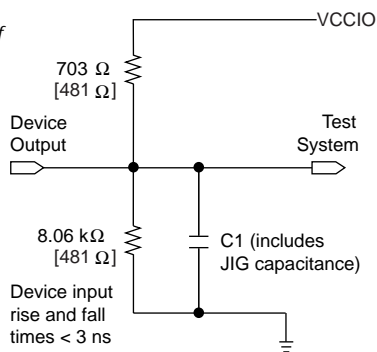
Device	Boundary-Scan Register Length
EPF10K30E	690
EPF10K50E EPF10K50S	798
EPF10K100E	1,050
EPF10K130E	1,308
EPF10K200E EPF10K200S	1,446

Generic Testing

Each FLEX 10KE device is functionally tested. Complete testing of each configurable static random access memory (SRAM) bit and all logic functionality ensures 100% yield. AC test measurements for FLEX 10KE devices are made under conditions equivalent to those shown in **Figure 21**. Multiple test patterns can be used to configure devices during all stages of the production flow.

Figure 21. FLEX 10KE AC Test Conditions

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers in brackets are for 2.5-V devices or outputs. Numbers without brackets are for 3.3-V devices or outputs.



Operating Conditions

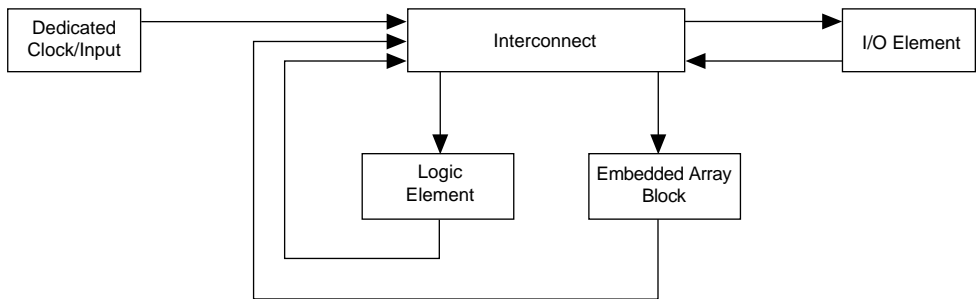
Tables 19 through 23 provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for 2.5-V FLEX 10KE devices.

Symbol	Parameter	Conditions	Min	Max	Unit
V _{CCINT}	Supply voltage	With respect to ground (2)	-0.5	3.6	V
V _{CCIO}			-0.5	4.6	V
V _I	DC input voltage		-2.0	5.75	V
I _{OUT}	DC output current, per pin		-25	25	mA
T _{STG}	Storage temperature	No bias	-65	150	° C
T _{AMB}	Ambient temperature	Under bias	-65	135	° C
T _J	Junction temperature	PQFP, TQFP, BGA, and FineLine BGA packages, under bias		135	° C
		Ceramic PGA packages, under bias		150	° C

Timing simulation and delay prediction are available with the Altera Simulator and Timing Analyzer, or with industry-standard EDA tools. The Simulator offers both pre-synthesis functional simulation to evaluate logic design accuracy and post-synthesis timing simulation with 0.1-ns resolution. The Timing Analyzer provides point-to-point timing delay information, setup and hold time analysis, and device-wide performance analysis.

Figure 24 shows the overall timing model, which maps the possible paths to and from the various elements of the FLEX 10KE device.

Figure 24. FLEX 10KE Device Timing Model



Figures 25 through 28 show the delays that correspond to various paths and functions within the LE, IOE, EAB, and bidirectional timing models.

Figure 26. FLEX 10KE Device IOE Timing Model

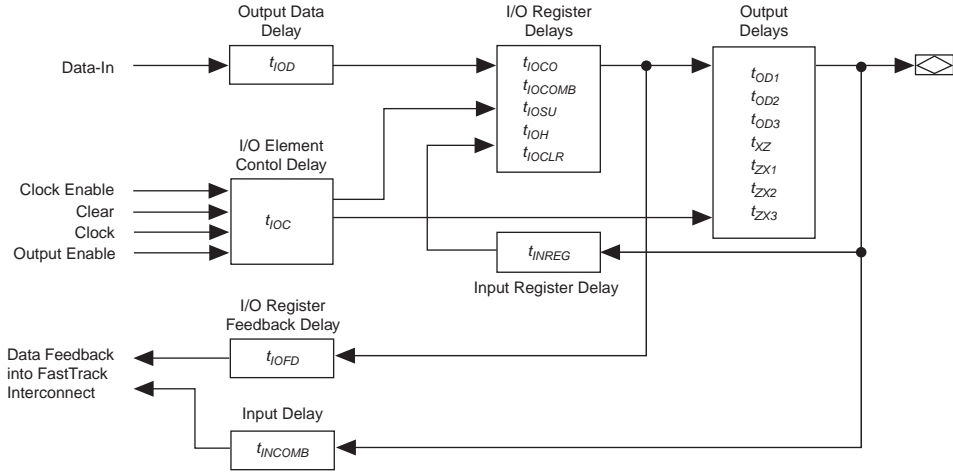


Figure 27. FLEX 10KE Device EAB Timing Model

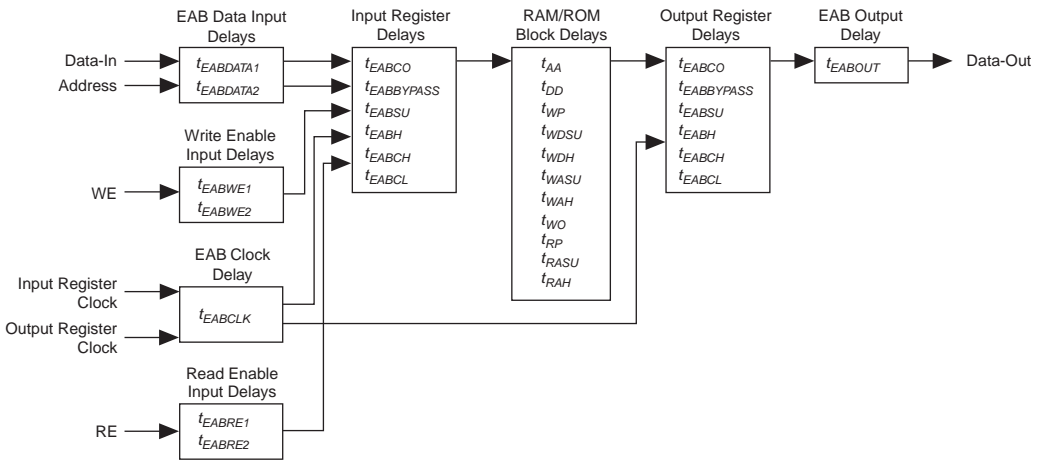


Table 37. EPF10K30E External Bidirectional Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{\text{INSUBIDIR}}$ (3)	2.8		3.9		5.2		ns
t_{INHBIDIR} (3)	0.0		0.0		0.0		ns
$t_{\text{INSUBIDIR}}$ (4)	3.8		4.9		–		ns
t_{INHBIDIR} (4)	0.0		0.0		–		ns
$t_{\text{OUTCOBIDIR}}$ (3)	2.0	4.9	2.0	5.9	2.0	7.6	ns
t_{XZBIDIR} (3)		6.1		7.5		9.7	ns
t_{ZXBIDIR} (3)		6.1		7.5		9.7	ns
$t_{\text{OUTCOBIDIR}}$ (4)	0.5	3.9	0.5	4.9	–	–	ns
t_{XZBIDIR} (4)		5.1		6.5		–	ns
t_{ZXBIDIR} (4)		5.1		6.5		–	ns

Notes to tables:

- (1) All timing parameters are described in Tables 24 through 30 in this data sheet.
- (2) These parameters are specified by characterization.
- (3) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
- (4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Tables 38 through 44 show EPF10K50E device internal and external timing parameters.

Table 38. EPF10K50E Device LE Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.6		0.9		1.3	ns
t_{CLUT}		0.5		0.6		0.8	ns
t_{RLUT}		0.7		0.8		1.1	ns
t_{PACKED}		0.4		0.5		0.6	ns
t_{EN}		0.6		0.7		0.9	ns
t_{CICO}		0.2		0.2		0.3	ns
t_{CGEN}		0.5		0.5		0.8	ns
t_{CGENR}		0.2		0.2		0.3	ns
t_{CASC}		0.8		1.0		1.4	ns
t_{C}		0.5		0.6		0.8	ns
t_{CO}		0.7		0.7		0.9	ns
t_{COMB}		0.5		0.6		0.8	ns
t_{SU}	0.7		0.7		0.8		ns

Table 43. EPF10K50E External Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		8.5		10.0		13.5	ns
t_{INSU}	2.7		3.2		4.3		ns
t_{INH}	0.0		0.0		0.0		ns
t_{OUTCO}	2.0	4.5	2.0	5.2	2.0	7.3	ns
t_{PCISU}	3.0		4.2		-		ns
t_{PCIH}	0.0		0.0		-		ns
t_{PCICO}	2.0	6.0	2.0	7.7	-	-	ns

Table 44. EPF10K50E External Bidirectional Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{\text{INSUBIDIR}}$	2.7		3.2		4.3		ns
t_{INHBIDIR}	0.0		0.0		0.0		ns
$t_{\text{OUTCOBIDIR}}$	2.0	4.5	2.0	5.2	2.0	7.3	ns
t_{XZBIDIR}		6.8		7.8		10.1	ns
t_{ZXBIDIR}		6.8		7.8		10.1	ns

Notes to tables:

- (1) All timing parameters are described in [Tables 24](#) through [30](#) in this data sheet.
- (2) These parameters are specified by characterization.

[Tables 45](#) through [51](#) show EPF10K100E device internal and external timing parameters.

Table 45. EPF10K100E Device LE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.7		1.0		1.5	ns
t_{CLUT}		0.5		0.7		0.9	ns
t_{RLUT}		0.6		0.8		1.1	ns
t_{PACKED}		0.3		0.4		0.5	ns
t_{EN}		0.2		0.3		0.3	ns
t_{CICO}		0.1		0.1		0.2	ns
t_{CGEN}		0.4		0.5		0.7	ns

Table 68. EPF10K50S Device EAB Internal Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.7		2.4		3.2	ns
$t_{EABDATA2}$		0.4		0.6		0.8	ns
t_{EABWE1}		1.0		1.4		1.9	ns
t_{EABWE2}		0.0		0.0		0.0	ns
t_{EABRE1}		0.0		0.0		0.0	
t_{EABRE2}		0.4		0.6		0.8	
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		0.8		1.1		1.5	ns
$t_{EABYPASS}$		0.0		0.0		0.0	ns
t_{EABSU}	0.7		1.0		1.3		ns
t_{EABH}	0.4		0.6		0.8		ns
t_{EABCLR}	0.8		1.1		1.5		
t_{AA}		2.0		2.8		3.8	ns
t_{WP}	2.0		2.8		3.8		ns
t_{RP}	1.0		1.4		1.9		
t_{WDSU}	0.5		0.7		0.9		ns
t_{WDH}	0.1		0.1		0.2		ns
t_{WASU}	1.0		1.4		1.9		ns
t_{WAH}	1.5		2.1		2.9		ns
t_{RASU}	1.5		2.1		2.8		
t_{RAH}	0.1		0.1		0.2		
t_{WO}		2.1		2.9		4.0	ns
t_{DD}		2.1		2.9		4.0	ns
t_{EABOUT}		0.0		0.0		0.0	ns
t_{EABCH}	1.5		2.0		2.5		ns
t_{EABCL}	1.5		2.0		2.5		ns

Table 71. EPF10K50S External Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		8.0		9.5		12.5	ns
$t_{\text{INSU}}^{(2)}$	2.4		2.9		3.9		ns
$t_{\text{INH}}^{(2)}$	0.0		0.0		0.0		ns
$t_{\text{OUTCO}}^{(2)}$	2.0	4.3	2.0	5.2	2.0	7.3	ns
$t_{\text{INSU}}^{(3)}$	2.4		2.9				ns
$t_{\text{INH}}^{(3)}$	0.0		0.0				ns
$t_{\text{OUTCO}}^{(3)}$	0.5	3.3	0.5	4.1			ns
t_{PCISU}	2.4		2.9		–		ns
t_{PCIH}	0.0		0.0		–		ns
t_{PCICO}	2.0	6.0	2.0	7.7	–	–	ns

Table 72. EPF10K50S External Bidirectional Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{\text{INSUBIDIR}}^{(2)}$	2.7		3.2		4.3		ns
$t_{\text{INHBIDIR}}^{(2)}$	0.0		0.0		0.0		ns
$t_{\text{INHBIDIR}}^{(3)}$	0.0		0.0		–		ns
$t_{\text{INSUBIDIR}}^{(3)}$	3.7		4.2		–		ns
$t_{\text{OUTCOBIDIR}}^{(2)}$	2.0	4.5	2.0	5.2	2.0	7.3	ns
$t_{\text{XZBIDIR}}^{(2)}$		6.8		7.8		10.1	ns
$t_{\text{ZXBIDIR}}^{(2)}$		6.8		7.8		10.1	ns
$t_{\text{OUTCOBIDIR}}^{(3)}$	0.5	3.5	0.5	4.2	–	–	
$t_{\text{XZBIDIR}}^{(3)}$		6.8		8.4		–	ns
$t_{\text{ZXBIDIR}}^{(3)}$		6.8		8.4		–	ns

Notes to tables:

- (1) All timing parameters are described in [Tables 24](#) through [30](#).
- (2) This parameter is measured without use of the ClockLock or ClockBoost circuits.
- (3) This parameter is measured with use of the ClockLock or ClockBoost circuits

Table 73. EPF10K200S Device Internal & External Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.7		0.8		1.2	ns
t_{CLUT}		0.4		0.5		0.6	ns
t_{RLUT}		0.5		0.7		0.9	ns
t_{PACKED}		0.4		0.5		0.7	ns
t_{EN}		0.6		0.5		0.6	ns
t_{CICO}		0.1		0.2		0.3	ns
t_{CGEN}		0.3		0.4		0.6	ns
t_{CGENR}		0.1		0.2		0.3	ns
t_{CASC}		0.7		0.8		1.2	ns
t_C		0.5		0.6		0.8	ns
t_{CO}		0.5		0.6		0.8	ns
t_{COMB}		0.3		0.6		0.8	ns
t_{SU}	0.4		0.6		0.7		ns
t_H	1.0		1.1		1.5		ns
t_{PRE}		0.4		0.6		0.8	ns
t_{CLR}		0.5		0.6		0.8	ns
t_{CH}	2.0		2.5		3.0		ns
t_{CL}	2.0		2.5		3.0		ns

Table 74. EPF10K200S Device IOE Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.8		1.9		2.6	ns
t_{IOC}		0.3		0.3		0.5	ns
t_{IOCO}		1.7		1.9		2.6	ns
t_{IOCOMB}		0.5		0.6		0.8	ns
t_{IOSU}	0.8		0.9		1.2		ns
t_{IOH}	0.4		0.8		1.1		ns
t_{IOCLR}		0.2		0.2		0.3	ns
t_{OD1}		1.3		0.7		0.9	ns
t_{OD2}		0.8		0.2		0.4	ns
t_{OD3}		2.9		3.0		3.9	ns
t_{XZ}		5.0		5.3		7.1	ns
t_{ZX1}		5.0		5.3		7.1	ns