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Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	624
Number of Logic Elements/Cells	4992
Total RAM Bits	49152
Number of I/O	191
Number of Gates	257000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	256-BGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k100efc256-3

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Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

Table 2. FLEX 10KE Device Features			
Feature	EPF10K100E (2)	EPF10K130E	EPF10K200E EPF10K200S
Typical gates (1)	100,000	130,000	200,000
Maximum system gates	257,000	342,000	513,000
Logic elements (LEs)	4,992	6,656	9,984
EABs	12	16	24
Total RAM bits	49,152	65,536	98,304
Maximum user I/O pins	338	413	470

Note to tables:

- (1) The embedded IEEE Std. 1149.1 JTAG circuitry adds up to 31,250 gates in addition to the listed typical or maximum system gates.
- (2) New EPF10K100B designs should use EPF10K100E devices.

...and More Features

- Fabricated on an advanced process and operate with a 2.5-V internal supply voltage
- In-circuit reconfigurability (ICR) via external configuration devices, intelligent controller, or JTAG port
- ClockLockTM and ClockBoostTM options for reduced clock delay/skew and clock multiplication
- Built-in low-skew clock distribution trees
- 100% functional testing of all devices; test vectors or scan chains are not required
- Pull-up on I/O pins before and during configuration

■ Flexible interconnect

- FastTrack® Interconnect continuous routing structure for fast, predictable interconnect delays
- Dedicated carry chain that implements arithmetic functions such as fast adders, counters, and comparators (automatically used by software tools and megafunctions)
- Dedicated cascade chain that implements high-speed, high-fan-in logic functions (automatically used by software tools and megafunctions)
- Tri-state emulation that implements internal tri-state buses
- Up to six global clock signals and four global clear signals

■ Powerful I/O pins

- Individual tri-state output enable control for each pin
- Open-drain option on each I/O pin
- Programmable output slew-rate control to reduce switching noise
- Clamp to V_{CCIO} user-selectable on a pin-by-pin basis
- Supports hot-socketing

Figure 1 shows a block diagram of the FLEX 10KE architecture. Each group of LEs is combined into an LAB; groups of LABs are arranged into rows and columns. Each row also contains a single EAB. The LABs and EABs are interconnected by the FastTrack Interconnect routing structure. IOEs are located at the end of each row and column of the FastTrack Interconnect routing structure.

Embedded Array Block (EAB) I/O Element IOE IOE IOE IOE IOE IOE IOE IOE IOE (IOE) IOE Column Logic Array Interconnect EAB Logic Array Block (LAB) IOE Logic Element (LE) Row EAB Interconnect Local Interconnect Logic Array

Figure 1. FLEX 10KE Device Block Diagram

IOE

IOE

IOE

IOE

IOE

IOE

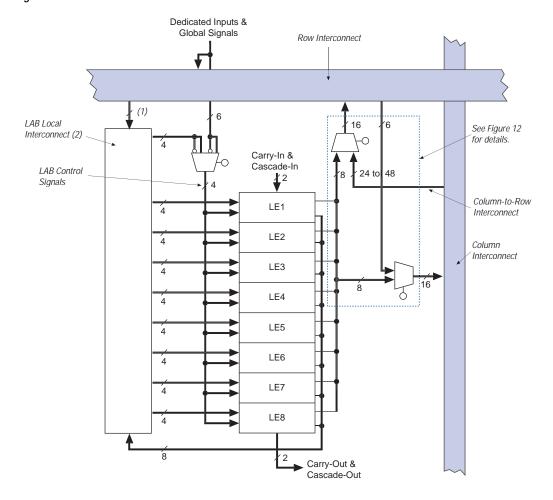
Embedded Array

FLEX 10KE devices provide six dedicated inputs that drive the flipflops' control inputs and ensure the efficient distribution of high-speed, low-skew (less than 1.5 ns) control signals. These signals use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect routing structure. Four of the dedicated inputs drive four global signals. These four global signals can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device.

IOE

IOE

Figure 7. FLEX 10KE LAB



Notes:

- (1) EPF10K30E, EPF10K50E, and EPF10K50S devices have 22 inputs to the LAB local interconnect channel from the row; EPF10K100E, EPF10K130E, EPF10K200E, and EPF10K200S devices have 26.
- (2) EPF10K30E, EPF10K50E, and EPF10K50S devices have 30 LAB local interconnect channels; EPF10K100E, EPF10K130E, EPF10K200E, and EPF10K200S devices have 34.

Figure 9 shows how an n-bit full adder can be implemented in n+1 LEs with the carry chain. One portion of the LUT generates the sum of two bits using the input signals and the carry-in signal; the sum is routed to the output of the LE. The register can be bypassed for simple adders or used for an accumulator function. Another portion of the LUT and the carry chain logic generates the carry-out signal, which is routed directly to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it can be used as a general-purpose signal.

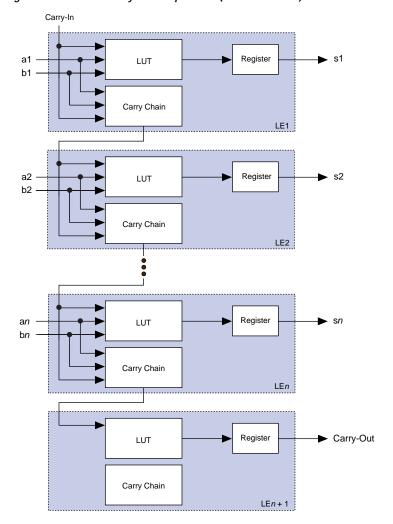


Figure 9. FLEX 10KE Carry Chain Operation (n-Bit Full Adder)

LE Operating Modes

The FLEX 10KE LE can operate in the following four modes:

- Normal mode
- Arithmetic mode
- Up/down counter mode
- Clearable counter mode

Each of these modes uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. Three inputs to the LE provide clock, clear, and preset control for the register. The Altera software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions that use a specific LE operating mode for optimal performance.

The architecture provides a synchronous clock enable to the register in all four modes. The Altera software can set DATA1 to enable the register synchronously, providing easy implementation of fully synchronous designs.

Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is ANDed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-states without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the Altera Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

For improved routing, the row interconnect consists of a combination of full-length and half-length channels. The full-length channels connect to all LABs in a row; the half-length channels connect to the LABs in half of the row. The EAB can be driven by the half-length channels in the left half of the row and by the full-length channels. The EAB drives out to the full-length channels. In addition to providing a predictable, row-wide interconnect, this architecture provides increased routing resources. Two neighboring LABs can be connected using a half-row channel, thereby saving the other half of the channel for the other half of the row.

Table 7 summarizes the FastTrack Interconnect routing structure resources available in each FLEX 10KE device.

Table 7. FLEX 1	Table 7. FLEX 10KE FastTrack Interconnect Resources									
Device	Rows	Channels per Row	Columns	Channels per Column						
EPF10K30E	6	216	36	24						
EPF10K50E EPF10K50S	10	216	36	24						
EPF10K100E	12	312	52	24						
EPF10K130E	16	312	52	32						
EPF10K200E EPF10K200S	24	312	52	48						

In addition to general-purpose I/O pins, FLEX 10KE devices have six dedicated input pins that provide low-skew signal distribution across the device. These six inputs can be used for global clock, clear, preset, and peripheral output enable and clock enable control signals. These signals are available as control signals for all LABs and IOEs in the device. The dedicated inputs can also be used as general-purpose data inputs because they can feed the local interconnect of each LAB in the device.

Figure 14 shows the interconnection of adjacent LABs and EABs, with row, column, and local interconnects, as well as the associated cascade and carry chains. Each LAB is labeled according to its location: a letter represents the row and a number represents the column. For example, LAB B3 is in row B, column 3.

Tables 12 and 13 summarize the ClockLock and ClockBoost parameters for -1 and -2 speed-grade devices, respectively.

Table 12.	. ClockLock & ClockBoost Param	eters for -1 Speed-C	Grade Device	es		
Symbol	Parameter	Condition	Min	Тур	Max	Unit
t_R	Input rise time				5	ns
t _F	Input fall time				5	ns
t _{INDUTY}	Input duty cycle		40		60	%
f _{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		180	MHz
f _{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		90	MHz
f _{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software (1)				25,000 (2)	PPM
t _{INCLKSTB}	Input clock stability (measured between adjacent clocks)				100	ps
t _{LOCK}	Time required for ClockLock or ClockBoost to acquire lock (3)				10	μs
t _{JITTER}	Jitter on ClockLock or ClockBoost-	$t_{INCLKSTB} < 100$			250	ps
	generated clock (4)	$t_{INCLKSTB} < 50$			200 (4)	ps
t _{OUTDUTY}	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Symbol	Parameter	Condition	Min	Тур	Max	Unit
t_R	Input rise time				5	ns
t _F	Input fall time				5	ns
t _{INDUTY}	Input duty cycle		40		60	%
f _{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		75	MHz
f _{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		37.5	MHz
f _{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software (1)				25,000 (2)	PPM
t _{INCLKSTB}	Input clock stability (measured between adjacent clocks)				100	ps
t _{LOCK}	Time required for ClockLock or ClockBoost to acquire lock (3)				10	μs
t _{JITTER}	Jitter on ClockLock or ClockBoost-	$t_{INCLKSTB} < 100$			250	ps
	generated clock (4)	$t_{INCLKSTB} < 50$			200 (4)	ps
t _{OUTDUTY}	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Notes to tables:

- (1) To implement the ClockLock and ClockBoost circuitry with the MAX+PLUS II software, designers must specify the input frequency. The Altera software tunes the PLL in the ClockLock and ClockBoost circuitry to this frequency. The f_{CLKDEV} parameter specifies how much the incoming clock can differ from the specified frequency during device operation. Simulation does not reflect this parameter.
- (2) Twenty-five thousand parts per million (PPM) equates to 2.5% of input clock period.
- (3) During device configuration, the ClockLock and ClockBoost circuitry is configured before the rest of the device. If the incoming clock is supplied during configuration, the ClockLock and ClockBoost circuitry locks during configuration because the t_{LOCK} value is less than the time required for configuration.
- (4) The t_{IITTER} specification is measured under long-term observation. The maximum value for t_{IITTER} is 200 ps if $t_{INCLKSTB}$ is lower than 50 ps.

I/O Configuration

This section discusses the peripheral component interconnect (PCI) pull-up clamping diode option, slew-rate control, open-drain output option, and MultiVolt I/O interface for FLEX 10KE devices. The PCI pull-up clamping diode, slew-rate control, and open-drain output options are controlled pin-by-pin via Altera software logic options. The MultiVolt I/O interface is controlled by connecting $V_{\rm CCIO}$ to a different voltage than $V_{\rm CCINT}.$ Its effect can be simulated in the Altera software via the **Global Project Device Options** dialog box (Assign menu).

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{IH}	High-level input voltage		1.7, 0.5 × V _{CCIO} (8)		5.75	V
V _{IL}	Low-level input voltage		-0.5		0.8, 0.3 × V _{CCIO} (8)	V
V _{OH}	3.3-V high-level TTL output voltage	$I_{OH} = -8 \text{ mA DC},$ $V_{CCIO} = 3.00 \text{ V } (9)$	2.4			V
	3.3-V high-level CMOS output voltage	$I_{OH} = -0.1 \text{ mA DC},$ $V_{CCIO} = 3.00 \text{ V } (9)$	V _{CCIO} – 0.2			V
	3.3-V high-level PCI output voltage	$I_{OH} = -0.5 \text{ mA DC},$ $V_{CCIO} = 3.00 \text{ to } 3.60 \text{ V } (9)$	0.9 × V _{CCIO}			V
	2.5-V high-level output voltage	$I_{OH} = -0.1 \text{ mA DC},$ $V_{CCIO} = 2.30 \text{ V } (9)$	2.1			V
		$I_{OH} = -1 \text{ mA DC},$ $V_{CCIO} = 2.30 \text{ V } (9)$	2.0			V
		$I_{OH} = -2 \text{ mA DC},$ $V_{CCIO} = 2.30 \text{ V } (9)$	1.7			V
V _{OL}	3.3-V low-level TTL output voltage	I _{OL} = 12 mA DC, V _{CCIO} = 3.00 V (10)			0.45	V
	3.3-V low-level CMOS output voltage	I _{OL} = 0.1 mA DC, V _{CCIO} = 3.00 V (10)			0.2	V
	3.3-V low-level PCI output voltage	I _{OL} = 1.5 mA DC, V _{CCIO} = 3.00 to 3.60 V (10)			0.1 × V _{CCIO}	V
	2.5-V low-level output voltage	I _{OL} = 0.1 mA DC, V _{CCIO} = 2.30 V (10)			0.2	V
		I _{OL} = 1 mA DC, V _{CCIO} = 2.30 V (10)			0.4	V
		I _{OL} = 2 mA DC, V _{CCIO} = 2.30 V (10)			0.7	V
I _I	Input pin leakage current	$V_I = V_{CCIOmax}$ to 0 V (11)	-10		10	μA
I _{OZ}	Tri-stated I/O pin leakage current	$V_O = V_{CCIOmax}$ to 0 V (11)	-10		10	μA
I _{CC0}	V _{CC} supply current (standby)	V _I = ground, no load, no toggling inputs		5		mA
		V _I = ground, no load, no toggling inputs (12)		10		mA
R _{CONF}	Value of I/O pin pull-	V _{CCIO} = 3.0 V (13)	20		50	k¾
	up resistor before and during configuration	V _{CCIO} = 2.3 V (13)	30		80	k¾

Table 2	3. FLEX 10KE Device Capacit	ance Note (14)			
Symbol	Parameter	Conditions	Min	Max	Unit
C _{IN}	Input capacitance	V _{IN} = 0 V, f = 1.0 MHz		10	pF
C _{INCLK}	Input capacitance on dedicated clock pin	V _{IN} = 0 V, f = 1.0 MHz		12	pF
C _{OUT}	Output capacitance	V _{OUT} = 0 V, f = 1.0 MHz		10	pF

Notes to tables:

- (1) See the Operating Requirements for Altera Devices Data Sheet.
- (2) Minimum DC input voltage is -0.5 V. During transitions, the inputs may undershoot to -2.0 V for input currents less than 100 mA and periods shorter than 20 ns.
- (3) Numbers in parentheses are for industrial-temperature-range devices.
- (4) Maximum V_{CC} rise time is 100 ms, and V_{CC} must rise monotonically.
- (5) All pins, including dedicated inputs, clock, I/O, and JTAG pins, may be driven before V_{CCINT} and V_{CCIO} are powered.
- (6) Typical values are for $T_A = 25^{\circ}$ C, $V_{CCINT} = 2.5$ V, and $V_{CCIO} = 2.5$ V or 3.3 V.
- (7) These values are specified under the FLEX 10KE Recommended Operating Conditions shown in Tables 20 and 21.
- (8) The FLEX 10KE input buffers are compatible with 2.5-V, 3.3-V (LVTTL and LVCMOS), and 5.0-V TTL and CMOS signals. Additionally, the input buffers are 3.3-V PCI compliant when V_{CCIO} and V_{CCINT} meet the relationship shown in Figure 22.
- (9) The I_{OH} parameter refers to high-level TTL, PCI, or CMOS output current.
- (10) The I_{OL} parameter refers to low-level TTL, PCI, or CMOS output current. This parameter applies to open-drain pins as well as output pins.
- (11) This value is specified for normal device operation. The value may vary during power-up.
- (12) This parameter applies to -1 speed-grade commercial-temperature devices and -2 speed-grade-industrial temperature devices.
- (13) Pin pull-up resistance values will be lower if the pin is driven higher than V_{CCIO} by an external source.
- (14) Capacitance is sample-tested only.

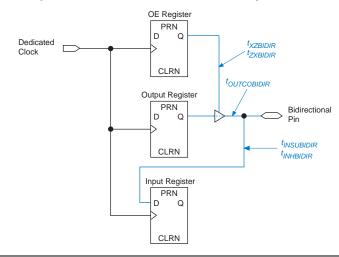


Figure 28. Synchronous Bidirectional Pin External Timing Model

Tables 24 through 28 describe the FLEX 10KE device internal timing parameters. Tables 29 through 30 describe the FLEX 10KE external timing parameters and their symbols.

Symbol	Parameter	Condition
t _{LUT}	LUT delay for data-in	
t _{CLUT}	LUT delay for carry-in	
t _{RLUT}	LUT delay for LE register feedback	
t _{PACKED}	Data-in to packed register delay	
t _{EN}	LE register enable delay	
t _{CICO}	Carry-in to carry-out delay	
t _{CGEN}	Data-in to carry-out delay	
t _{CGENR}	LE register feedback to carry-out delay	
t _{CASC}	Cascade-in to cascade-out delay	
t_{C}	LE register control signal delay	
t _{CO}	LE register clock-to-output delay	
t _{COMB}	Combinatorial delay	
t _{SU}	LE register setup time for data and enable signals before clock; LE register recovery time after asynchronous clear, preset, or load	
t_H	LE register hold time for data and enable signals after clock	
t _{PRE}	LE register preset delay	

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		1.7		2.0		2.3	ns
t _{EABDATA1}		0.6		0.7		0.8	ns
t _{EABWE1}		1.1		1.3		1.4	ns
t _{EABWE2}		0.4		0.4		0.5	ns
t _{EABRE1}		0.8		0.9		1.0	ns
t _{EABRE2}		0.4		0.4		0.5	ns
t _{EABCLK}		0.0		0.0		0.0	ns
t _{EABCO}		0.3		0.3		0.4	ns
t _{EABBYPASS}		0.5		0.6		0.7	ns
t _{EABSU}	0.9		1.0		1.2		ns
t _{EABH}	0.4		0.4		0.5		ns
t _{EABCLR}	0.3		0.3		0.3		ns
t_{AA}		3.2		3.8		4.4	ns
t_{WP}	2.5		2.9		3.3		ns
t_{RP}	0.9		1.1		1.2		ns
t _{WDSU}	0.9		1.0		1.1		ns
t _{WDH}	0.1		0.1		0.1		ns
t _{WASU}	1.7		2.0		2.3		ns
t _{WAH}	1.8		2.1		2.4		ns
t _{RASU}	3.1		3.7		4.2		ns
t _{RAH}	0.2		0.2		0.2		ns
t _{WO}		2.5		2.9		3.3	ns
t _{DD}		2.5		2.9		3.3	ns
t _{EABOUT}		0.5		0.6		0.7	ns
t _{EABCH}	1.5		2.0		2.3		ns
t _{EABCL}	2.5		2.9		3.3		ns

Table 38. EPF10K	50E Device	LE Timing M	licroparame	ters (Part 2	of 2) No	te (1)	
Symbol	-1 Speed Grade		-2 Spee	d Grade	-3 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_H	0.9		1.0		1.4		ns
t _{PRE}		0.5		0.6		0.8	ns
t _{CLR}		0.5		0.6		0.8	ns
t _{CH}	2.0		2.5		3.0		ns
t_{CL}	2.0		2.5		3.0		ns

Symbol	-1 Spee	d Grade	-2 Spee	ed Grade	-3 Spee	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		2.2		2.4		3.3	ns
t _{IOC}		0.3		0.3		0.5	ns
t _{IOCO}		1.0		1.0		1.4	ns
t_{IOCOMB}		0.0		0.0		0.2	ns
t _{IOSU}	1.0		1.2		1.7		ns
t_{IOH}	0.3		0.3		0.5		ns
t_{IOCLR}		0.9		1.0		1.4	ns
t_{OD1}		0.8		0.9		1.2	ns
t_{OD2}		0.3		0.4		0.7	ns
t_{OD3}		3.0		3.5		3.5	ns
t_{XZ}		1.4		1.7		2.3	ns
t_{ZX1}		1.4		1.7		2.3	ns
t_{ZX2}		0.9		1.2		1.8	ns
t _{ZX3}		3.6		4.3		4.6	ns
t _{INREG}		4.9		5.8		7.8	ns
t _{IOFD}		2.8		3.3		4.5	ns
t _{INCOMB}		2.8		3.3		4.5	ns

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		1.5		2.0		2.6	ns
t _{EABDATA1}		0.0		0.0		0.0	ns
t _{EABWE1}		1.5		2.0		2.6	ns
t _{EABWE2}		0.3		0.4		0.5	ns
t _{EABRE1}		0.3		0.4		0.5	ns
t _{EABRE2}		0.0		0.0		0.0	ns
t _{EABCLK}		0.0		0.0		0.0	ns
t _{EABCO}		0.3		0.4		0.5	ns
t _{EABBYPASS}		0.1		0.1		0.2	ns
t _{EABSU}	0.8		1.0		1.4		ns
t _{EABH}	0.1		0.1		0.2		ns
t _{EABCLR}	0.3		0.4		0.5		ns
t_{AA}		4.0		5.1		6.6	ns
t_{WP}	2.7		3.5		4.7		ns
t_{RP}	1.0		1.3		1.7		ns
t _{WDSU}	1.0		1.3		1.7		ns
t _{WDH}	0.2		0.2		0.3		ns
t _{WASU}	1.6		2.1		2.8		ns
t _{WAH}	1.6		2.1		2.8		ns
t _{RASU}	3.0		3.9		5.2		ns
t _{RAH}	0.1		0.1		0.2		ns
t _{WO}		1.5		2.0		2.6	ns
t _{DD}		1.5		2.0		2.6	ns
t _{EABOUT}		0.2		0.3		0.3	ns
t _{EABCH}	1.5		2.0		2.5		ns
t _{EABCL}	2.7		3.5		4.7		ns

Table 48. EPF10	K100E Device	EAB Intern	al Timing M	acroparame	ters (Part 1	of 2) No	ote (1)
Symbol	-1 Spee	d Grade	-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{EABAA}		5.9		7.6		9.9	ns
t _{EABRCOMB}	5.9		7.6		9.9		ns
t _{EABRCREG}	5.1		6.5		8.5		ns
t _{EABWP}	2.7		3.5		4.7		ns

Table 59. EPF10K	200E Device	LE Timing	Microparam	eters (Part	2 of 2) N	ote (1)	
Symbol	-1 Spee	ed Grade -2 Speed Grade		-3 Speed Grade		Unit	
	Min	Max	Min	Max	Min	Max	
t_H	0.9		1.1		1.5		ns
t _{PRE}		0.5		0.6		0.8	ns
t _{CLR}		0.5		0.6		0.8	ns
t _{CH}	2.0		2.5		3.0		ns
t_{CL}	2.0		2.5		3.0		ns

Table 60. EPF10I	K200E Device	e IOE Timing	Micropara	meters No	ote (1)		
Symbol	-1 Speed Grade		-2 Speed Grade		-3 Spec	ed Grade	Unit
	Min	Max Min	Min	Max	Min	Max	
t_{IOD}		1.6		1.9		2.6	ns
t_{IOC}		0.3		0.3		0.5	ns
t _{IOCO}		1.6		1.9		2.6	ns
t _{IOCOMB}		0.5		0.6		0.8	ns
t _{IOSU}	0.8		0.9		1.2		ns
t _{IOH}	0.7		0.8		1.1		ns
t _{IOCLR}		0.2		0.2		0.3	ns
t _{OD1}		0.6		0.7		0.9	ns
t _{OD2}		0.1		0.2		0.7	ns
t _{OD3}		2.5		3.0		3.9	ns
t_{XZ}		4.4		5.3		7.1	ns
t _{ZX1}		4.4		5.3		7.1	ns
t_{ZX2}		3.9		4.8		6.9	ns
t_{ZX3}		6.3		7.6		10.1	ns
t _{INREG}		4.8		5.7		7.7	ns
t _{IOFD}		1.5		1.8		2.4	ns
t _{INCOMB}		1.5		1.8		2.4	ns

Symbol	-1 Spee	d Grade	-2 Spee	d Grade	-3 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		2.0		2.4		3.2	ns
t _{EABDATA1}		0.4		0.5		0.6	ns
EABWE1		1.4		1.7		2.3	ns
t _{EABWE2}		0.0		0.0		0.0	ns
t _{EABRE1}		0		0		0	ns
t _{EABRE2}		0.4		0.5		0.6	ns
t _{EABCLK}		0.0		0.0		0.0	ns
t _{EABCO}		0.8		0.9		1.2	ns
t _{EABBYPASS}		0.0		0.1		0.1	ns
t _{EABSU}	0.9		1.1		1.5		ns
t _{EABH}	0.4		0.5		0.6		ns
t _{EABCLR}	0.8		0.9		1.2		ns
t _{AA}		3.1		3.7		4.9	ns
t_{WP}	3.3		4.0		5.3		ns
t _{RP}	0.9		1.1		1.5		ns
twosu	0.9		1.1		1.5		ns
t _{WDH}	0.1		0.1		0.1		ns
^t wasu	1.3		1.6		2.1		ns
t _{WAH}	2.1		2.5		3.3		ns
t _{RASU}	2.2		2.6		3.5		ns
t_{RAH}	0.1		0.1		0.2		ns
^t wo		2.0		2.4		3.2	ns
t _{DD}		2.0		2.4		3.2	ns
t _{EABOUT}		0.0		0.1		0.1	ns
t _{EABCH}	1.5		2.0		2.5		ns
EABCL	3.3		4.0		5.3		ns

Table 62. EPF10K	200E Device	EAB Interna	al Timing Ma	acroparame	ters (Part 1	of 2) No	te (1)
Symbol	-1 Spee	d Grade	-2 Spee	d Grade	Unit		
	Min	Max	Min	Max	Min	Max	
t _{EABAA}		5.1		6.4		8.4	ns
t _{EABRCOMB}	5.1		6.4		8.4		ns
t _{EABRCREG}	4.8		5.7		7.6		ns
t _{EABWP}	3.3		4.0		5.3		ns

Table 64. EPF10	K200E Extern	al Timing Pa	arameters	Notes (1),	(2)		
Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{DRR}		10.0		12.0		16.0	ns
t _{INSU}	2.8		3.4		4.4		ns
t _{INH}	0.0		0.0		0.0		ns
t _{OUTCO}	2.0	4.5	2.0	5.3	2.0	7.8	ns
t _{PCISU}	3.0		6.2		-		ns
t _{PCIH}	0.0		0.0		-		ns
t _{PCICO}	2.0	6.0	2.0	8.9	-	-	ns

Table 65. EPF10K	200E Extern	al Bidirectio	nal Timing	Parameters	Notes (1)), (2)	
Symbol	-1 Speed Grade -2		-2 Spee	-2 Speed Grade		d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{INSUBIDIR}	3.0		4.0		5.5		ns
t _{INHBIDIR}	0.0		0.0		0.0		ns
t _{OUTCOBIDIR}	2.0	4.5	2.0	5.3	2.0	7.8	ns
t _{XZBIDIR}		8.1		9.5		13.0	ns
t _{ZXBIDIR}		8.1		9.5		13.0	ns

Notes to tables:

- (1) All timing parameters are described in Tables 24 through 30 in this data sheet.
- (2) These parameters are specified by characterization.

Tables 66 through 79 show EPF10K50S and EPF10K200S device external timing parameters.

Table 66. EPF10k	K50S Device	LE Timing N	1icroparame	ters (Part 1	of 2) No	ote (1)	
Symbol	-1 Speed Grade		-2 Spee	-2 Speed Grade		d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.6		0.8		1.1	ns
t _{CLUT}		0.5		0.6		0.8	ns
t _{RLUT}		0.6		0.7		0.9	ns
t _{PACKED}		0.2		0.3		0.4	ns
t_{EN}		0.6		0.7		0.9	ns
t _{CICO}		0.1		0.1		0.1	ns
t _{CGEN}		0.4		0.5		0.6	ns

Symbol	-1 Spee	d Grade	-2 Spee	d Grade	-3 Spee	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		1.7		2.4		3.2	ns
t _{EABDATA2}		0.4		0.6		0.8	ns
t _{EABWE1}		1.0		1.4		1.9	ns
t _{EABWE2}		0.0		0.0		0.0	ns
t _{EABRE1}		0.0		0.0		0.0	
t _{EABRE2}		0.4		0.6		0.8	
t _{EABCLK}		0.0		0.0		0.0	ns
t _{EABCO}		0.8		1.1		1.5	ns
t _{EABBYPASS}		0.0		0.0		0.0	ns
t _{EABSU}	0.7		1.0		1.3		ns
t _{EABH}	0.4		0.6		0.8		ns
t _{EABCLR}	0.8		1.1		1.5		
t_{AA}		2.0		2.8		3.8	ns
t_{WP}	2.0		2.8		3.8		ns
t_{RP}	1.0		1.4		1.9		
t _{WDSU}	0.5		0.7		0.9		ns
t_{WDH}	0.1		0.1		0.2		ns
t _{WASU}	1.0		1.4		1.9		ns
t _{WAH}	1.5		2.1		2.9		ns
t _{RASU}	1.5		2.1		2.8		
t _{RAH}	0.1		0.1		0.2		
t_{WO}		2.1		2.9		4.0	ns
t_{DD}		2.1		2.9		4.0	ns
t _{EABOUT}		0.0		0.0		0.0	ns
t _{EABCH}	1.5		2.0		2.5		ns
t _{EABCL}	1.5		2.0		2.5		ns

To better reflect actual designs, the power model (and the constant K in the power calculation equations) for continuous interconnect FLEX devices assumes that LEs drive FastTrack Interconnect channels. In contrast, the power model of segmented FPGAs assumes that all LEs drive only one short interconnect segment. This assumption may lead to inaccurate results when compared to measured power consumption for actual designs in segmented FPGAs.

Figure 31 shows the relationship between the current and operating frequency of FLEX 10KE devices.

Figure 31. FLEX 10KE I_{CCACTIVE} vs. Operating Frequency (Part 1 of 2)

