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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	624
Number of Logic Elements/Cells	4992
Total RAM Bits	49152
Number of I/O	147
Number of Gates	257000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k100eqc208-3n

Table 4. FLEX 10KE Package Sizes

Device	144-Pin TQFP	208-Pin PQFP	240-Pin PQFP RQFP	256-Pin FineLine BGA	356-Pin BGA	484-Pin FineLine BGA	599-Pin PGA	600-Pin BGA	672-Pin FineLine BGA
Pitch (mm)	0.50	0.50	0.50	1.0	1.27	1.0	—	1.27	1.0
Area (mm ²)	484	936	1,197	289	1,225	529	3,904	2,025	729
Length × width (mm × mm)	22 × 22	30.6 × 30.6	34.6 × 34.6	17 × 17	35 × 35	23 × 23	62.5 × 62.5	45 × 45	27 × 27

General Description

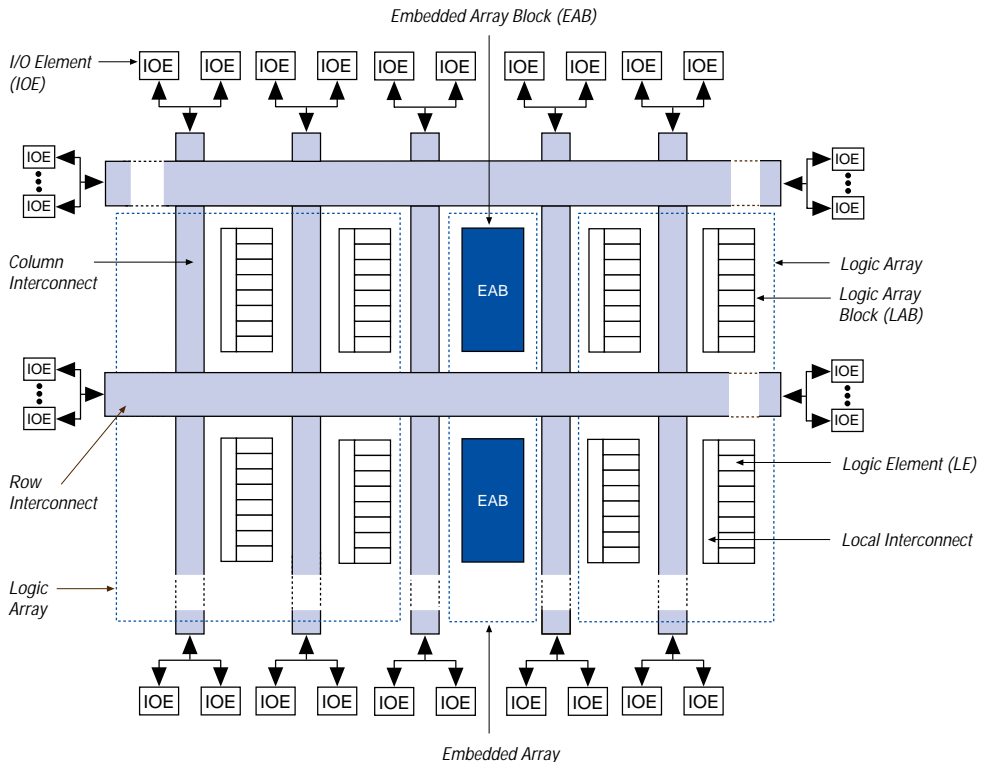
Altera FLEX 10KE devices are enhanced versions of FLEX 10K devices. Based on reconfigurable CMOS SRAM elements, the FLEX architecture incorporates all features necessary to implement common gate array megafunctions. With up to 200,000 typical gates, FLEX 10KE devices provide the density, speed, and features to integrate entire systems, including multiple 32-bit buses, into a single device.

The ability to reconfigure FLEX 10KE devices enables 100% testing prior to shipment and allows the designer to focus on simulation and design verification. FLEX 10KE reconfigurability eliminates inventory management for gate array designs and generation of test vectors for fault coverage.

Table 5 shows FLEX 10KE performance for some common designs. All performance values were obtained with Synopsys DesignWare or LPM functions. Special design techniques are not required to implement the applications; the designer simply infers or instantiates a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Figure 1 shows a block diagram of the FLEX 10KE architecture. Each group of LEs is combined into an LAB; groups of LABs are arranged into rows and columns. Each row also contains a single EAB. The LABs and EABs are interconnected by the FastTrack Interconnect routing structure. IOEs are located at the end of each row and column of the FastTrack Interconnect routing structure.

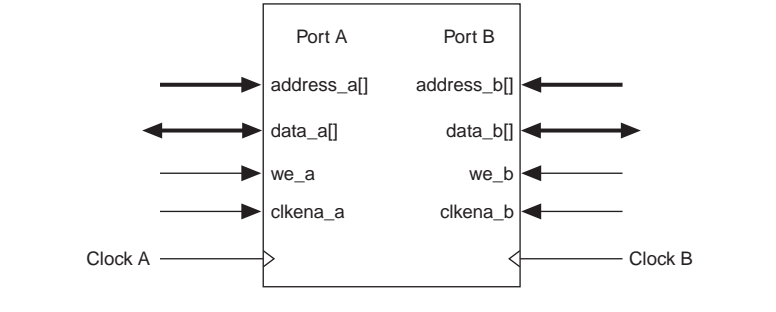
Figure 1. FLEX 10KE Device Block Diagram



FLEX 10KE devices provide six dedicated inputs that drive the flipflops' control inputs and ensure the efficient distribution of high-speed, low-skew (less than 1.5 ns) control signals. These signals use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect routing structure. Four of the dedicated inputs drive four global signals. These four global signals can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device.

The EAB can also use Altera megafunctions to implement dual-port RAM applications where both ports can read or write, as shown in [Figure 3](#).

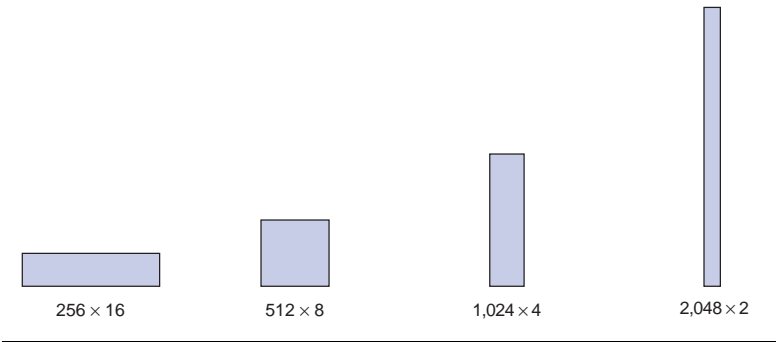
Figure 3. FLEX 10KE EAB in Dual-Port RAM Mode



The FLEX 10KE EAB can be used in a single-port mode, which is useful for backward-compatibility with FLEX 10K designs (see [Figure 4](#)).

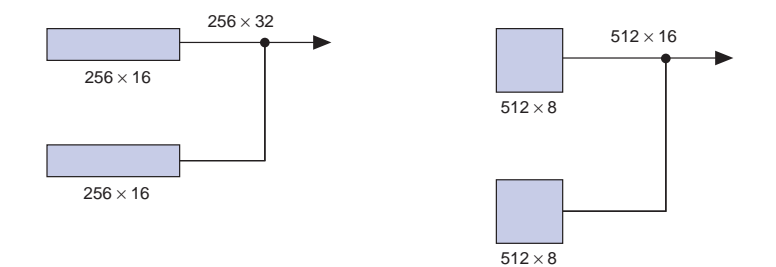
When used as RAM, each EAB can be configured in any of the following sizes: 256×16 , 512×8 , $1,024 \times 4$, or $2,048 \times 2$ (see [Figure 5](#)).

Figure 5. FLEX 10KE EAB Memory Configurations



Larger blocks of RAM are created by combining multiple EABs. For example, two 256×16 RAM blocks can be combined to form a 256×32 block; two 512×8 RAM blocks can be combined to form a 512×16 block (see [Figure 6](#)).

Figure 6. Examples of Combining FLEX 10KE EABs



If necessary, all EABs in a device can be cascaded to form a single RAM block. EABs can be cascaded to form RAM blocks of up to 2,048 words without impacting timing. The Altera software automatically combines EABs to meet a designer's RAM specifications.

EABs provide flexible options for driving and controlling clock signals. Different clocks and clock enables can be used for reading and writing to the EAB. Registers can be independently inserted on the data input, EAB output, write address, write enable signals, read address, and read enable signals. The global signals and the EAB local interconnect can drive write enable, read enable, and clock enable signals. The global signals, dedicated clock pins, and EAB local interconnect can drive the EAB clock signals. Because the LEs drive the EAB local interconnect, the LEs can control write enable, read enable, clear, clock, and clock enable signals.

An EAB is fed by a row interconnect and can drive out to row and column interconnects. Each EAB output can drive up to two row channels and up to two column channels; the unused row channel can be driven by other LEs. This feature increases the routing resources available for EAB outputs (see [Figures 2 and 4](#)). The column interconnect, which is adjacent to the EAB, has twice as many channels as other columns in the device.

Logic Array Block

An LAB consists of eight LEs, their associated carry and cascade chains, LAB control signals, and the LAB local interconnect. The LAB provides the coarse-grained structure to the FLEX 10KE architecture, facilitating efficient routing with optimum device utilization and high performance (see [Figure 7](#)).

Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is ANDed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-states without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the Altera Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

Asynchronous Clear

The flipflop can be cleared by either LABCTRL1 or LABCTRL2. In this mode, the preset signal is tied to VCC to deactivate it.

Asynchronous Preset

An asynchronous preset is implemented as an asynchronous load, or with an asynchronous clear. If DATA3 is tied to VCC, asserting LABCTRL1 asynchronously loads a one into the register. Alternatively, the Altera software can provide preset control by using the clear and inverting the input and output of the register. Inversion control is available for the inputs to both LEs and IOEs. Therefore, if a register is preset by only one of the two LABCTRL signals, the DATA3 input is not needed and can be used for one of the LE operating modes.

Asynchronous Preset & Clear

When implementing asynchronous clear and preset, LABCTRL1 controls the preset and LABCTRL2 controls the clear. DATA3 is tied to VCC, so that asserting LABCTRL1 asynchronously loads a one into the register, effectively presetting the register. Asserting LABCTRL2 clears the register.

Asynchronous Load with Clear

When implementing an asynchronous load in conjunction with the clear, LABCTRL1 implements the asynchronous load of DATA3 by controlling the register preset and clear. LABCTRL2 implements the clear by controlling the register clear; LABCTRL2 does not have to feed the preset circuits.

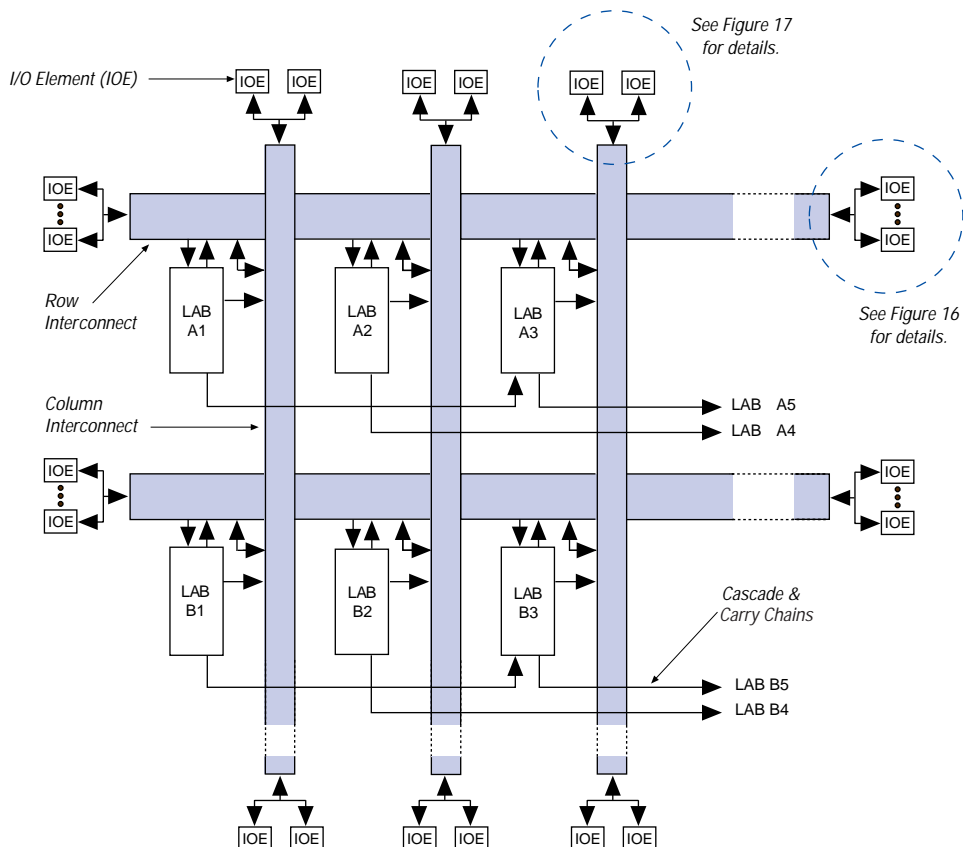
Asynchronous Load with Preset

When implementing an asynchronous load in conjunction with preset, the Altera software provides preset control by using the clear and inverting the input and output of the register. Asserting LABCTRL2 presets the register, while asserting LABCTRL1 loads the register. The Altera software inverts the signal that drives DATA3 to account for the inversion of the register's output.

Asynchronous Load without Preset or Clear

When implementing an asynchronous load without preset or clear, LABCTRL1 implements the asynchronous load of DATA3 by controlling the register preset and clear.

Figure 14. FLEX 10KE Interconnect Resources



I/O Element

An IOE contains a bidirectional I/O buffer and a register that can be used either as an input register for external data that requires a fast setup time, or as an output register for data that requires fast clock-to-output performance. In some cases, using an LE register for an input register will result in a faster setup time than using an IOE register. IOEs can be used as input, output, or bidirectional pins. For bidirectional registered I/O implementation, the output register should be in the IOE, and the data input and output enable registers should be LE registers placed adjacent to the bidirectional pin. The Altera Compiler uses the programmable inversion option to invert signals from the row and column interconnect automatically where appropriate. [Figure 15](#) shows the bidirectional I/O registers.

Table 9. Peripheral Bus Sources for EPF10K100E, EPF10K130E, EPF10K200E & EPF10K200S Devices

Peripheral Control Signal	EPF10K100E	EPF10K130E	EPF10K200E EPF10K200S
OE0	Row A	Row C	Row G
OE1	Row C	Row E	Row I
OE2	Row E	Row G	Row K
OE3	Row L	Row N	Row R
OE4	Row I	Row K	Row O
OE5	Row K	Row M	Row Q
CLKENA0/CLK0/GLOBAL0	Row F	Row H	Row L
CLKENA1/OE6/GLOBAL1	Row D	Row F	Row J
CLKENA2/CLR0	Row B	Row D	Row H
CLKENA3/OE7/GLOBAL2	Row H	Row J	Row N
CLKENA4/CLR1	Row J	Row L	Row P
CLKENA5/CLK1/GLOBAL3	Row G	Row I	Row M

Signals on the peripheral control bus can also drive the four global signals, referred to as GLOBAL0 through GLOBAL3 in [Tables 8 and 9](#). An internally generated signal can drive a global signal, providing the same low-skew, low-delay characteristics as a signal driven by an input pin. An LE drives the global signal by driving a row line that drives the peripheral bus, which then drives the global signal. This feature is ideal for internally generated clear or clock signals with high fan-out. However, internally driven global signals offer no advantage over the general-purpose interconnect for routing data signals. The dedicated input pin should be driven to a known logic state (such as ground) and not be allowed to float.

The chip-wide output enable pin is an active-high pin (DEV_OE) that can be used to tri-state all pins on the device. This option can be set in the Altera software. On EPF10K50E and EPF10K200E devices, the built-in I/O pin pull-up resistors (which are active during configuration) are active when the chip-wide output enable pin is asserted. The registers in the IOE can also be reset by the chip-wide reset pin.

Table 13. ClockLock & ClockBoost Parameters for -2 Speed-Grade Devices

Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_R	Input rise time				5	ns
t_F	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		75	MHz
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		37.5	MHz
f_{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software (1)				25,000 (2)	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock (3)				10	μ s
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock (4)	$t_{INCLKSTB} < 100$			250	ps
		$t_{INCLKSTB} < 50$			200 (4)	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Notes to tables:

- (1) To implement the ClockLock and ClockBoost circuitry with the MAX+PLUS II software, designers must specify the input frequency. The Altera software tunes the PLL in the ClockLock and ClockBoost circuitry to this frequency. The f_{CLKDEV} parameter specifies how much the incoming clock can differ from the specified frequency during device operation. Simulation does not reflect this parameter.
- (2) Twenty-five thousand parts per million (PPM) equates to 2.5% of input clock period.
- (3) During device configuration, the ClockLock and ClockBoost circuitry is configured before the rest of the device. If the incoming clock is supplied during configuration, the ClockLock and ClockBoost circuitry locks during configuration because the t_{LOCK} value is less than the time required for configuration.
- (4) The t_{JITTER} specification is measured under long-term observation. The maximum value for t_{JITTER} is 200 ps if $t_{INCLKSTB}$ is lower than 50 ps.

I/O Configuration

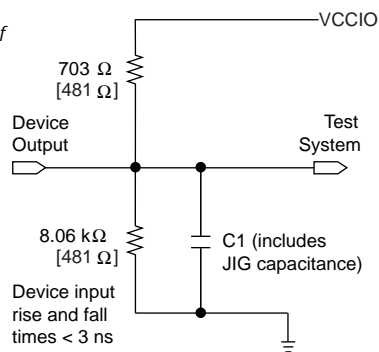
This section discusses the peripheral component interconnect (PCI) pull-up clamping diode option, slew-rate control, open-drain output option, and MultiVolt I/O interface for FLEX 10KE devices. The PCI pull-up clamping diode, slew-rate control, and open-drain output options are controlled pin-by-pin via Altera software logic options. The MultiVolt I/O interface is controlled by connecting V_{CCIO} to a different voltage than V_{CCINT} . Its effect can be simulated in the Altera software via the **Global Project Device Options** dialog box (Assign menu).

Generic Testing

Each FLEX 10KE device is functionally tested. Complete testing of each configurable static random access memory (SRAM) bit and all logic functionality ensures 100% yield. AC test measurements for FLEX 10KE devices are made under conditions equivalent to those shown in [Figure 21](#). Multiple test patterns can be used to configure devices during all stages of the production flow.

Figure 21. FLEX 10KE AC Test Conditions

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers in brackets are for 2.5-V devices or outputs. Numbers without brackets are for 3.3-V devices or outputs.



Operating Conditions

[Tables 19](#) through [23](#) provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for 2.5-V FLEX 10KE devices.

Table 19. FLEX 10KE 2.5-V Device Absolute Maximum Ratings *Note (1)*

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CCINT}	Supply voltage	With respect to ground (2)	–0.5	3.6	V
V_{CCIO}			–0.5	4.6	V
V_I	DC input voltage		–2.0	5.75	V
I_{OUT}	DC output current, per pin		–25	25	mA
T_{STG}	Storage temperature	No bias	–65	150	°C
T_{AMB}	Ambient temperature	Under bias	–65	135	°C
T_J	Junction temperature	PQFP, TQFP, BGA, and FineLine BGA packages, under bias		135	°C
		Ceramic PGA packages, under bias		150	°C

Table 28. Interconnect Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
$t_{DIN2IOE}$	Delay from dedicated input pin to IOE control input	(7)
t_{DIN2LE}	Delay from dedicated input pin to LE or EAB control input	(7)
$t_{DCLK2IOE}$	Delay from dedicated clock pin to IOE clock	(7)
$t_{DCLK2LE}$	Delay from dedicated clock pin to LE or EAB clock	(7)
$t_{DIN2DATA}$	Delay from dedicated input or clock to LE or EAB data	(7)
$t_{SAMELAB}$	Routing delay for an LE driving another LE in the same LAB	
$t_{SAMEROW}$	Routing delay for a row IOE, LE, or EAB driving a row IOE, LE, or EAB in the same row	(7)
$t_{SAMECOLUMN}$	Routing delay for an LE driving an IOE in the same column	(7)
$t_{DIFFROW}$	Routing delay for a column IOE, LE, or EAB driving an LE or EAB in a different row	(7)
$t_{TROWROWS}$	Routing delay for a row IOE or EAB driving an LE or EAB in a different row	(7)
$t_{LEPERIPH}$	Routing delay for an LE driving a control signal of an IOE via the peripheral control bus	(7)
$t_{LABCARRY}$	Routing delay for the carry-out signal of an LE driving the carry-in signal of a different LE in a different LAB	
$t_{LABCASC}$	Routing delay for the cascade-out signal of an LE driving the cascade-in signal of a different LE in a different LAB	

Table 29. External Timing Parameters

Symbol	Parameter	Conditions
t_{DRR}	Register-to-register delay via four LEs, three row interconnects, and four local interconnects	(8)
t_{INSU}	Setup time with global clock at IOE register	(9)
t_{INH}	Hold time with global clock at IOE register	(9)
t_{OUTCO}	Clock-to-output delay with global clock at IOE register	(9)
t_{PCISU}	Setup time with global clock for registers used in PCI designs	(9),(10)
t_{PCIH}	Hold time with global clock for registers used in PCI designs	(9),(10)
t_{PCICO}	Clock-to-output delay with global clock for registers used in PCI designs	(9),(10)

Table 37. EPF10K30E External Bidirectional Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{\text{INSUBIDIR}}$ (3)	2.8		3.9		5.2		ns
t_{INHBIDIR} (3)	0.0		0.0		0.0		ns
$t_{\text{INSUBIDIR}}$ (4)	3.8		4.9		—		ns
t_{INHBIDIR} (4)	0.0		0.0		—		ns
$t_{\text{OUTCOBIDIR}}$ (3)	2.0	4.9	2.0	5.9	2.0	7.6	ns
t_{XZBIDIR} (3)		6.1		7.5		9.7	ns
t_{ZXBIDIR} (3)		6.1		7.5		9.7	ns
$t_{\text{OUTCOBIDIR}}$ (4)	0.5	3.9	0.5	4.9	—	—	ns
t_{XZBIDIR} (4)		5.1		6.5		—	ns
t_{ZXBIDIR} (4)		5.1		6.5		—	ns

Notes to tables:

- (1) All timing parameters are described in Tables 24 through 30 in this data sheet.
- (2) These parameters are specified by characterization.
- (3) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
- (4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Tables 38 through 44 show EPF10K50E device internal and external timing parameters.

Table 38. EPF10K50E Device LE Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.6		0.9		1.3	ns
t_{CLUT}		0.5		0.6		0.8	ns
t_{RLUT}		0.7		0.8		1.1	ns
t_{PACKED}		0.4		0.5		0.6	ns
t_{EN}		0.6		0.7		0.9	ns
t_{CICO}		0.2		0.2		0.3	ns
t_{CGEN}		0.5		0.5		0.8	ns
t_{CGENR}		0.2		0.2		0.3	ns
t_{CASC}		0.8		1.0		1.4	ns
t_{C}		0.5		0.6		0.8	ns
t_{CO}		0.7		0.7		0.9	ns
t_{COMB}		0.5		0.6		0.8	ns
t_{SU}	0.7		0.7		0.8		ns

Table 40. EPF10K50E Device EAB Internal Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.7		2.0		2.7	ns
$t_{EABDATA1}$		0.6		0.7		0.9	ns
t_{EABWE1}		1.1		1.3		1.8	ns
t_{EABWE2}		0.4		0.4		0.6	ns
t_{EABRE1}		0.8		0.9		1.2	ns
t_{EABRE2}		0.4		0.4		0.6	ns
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		0.3		0.3		0.5	ns
$t_{EABYPASS}$		0.5		0.6		0.8	ns
t_{EABSU}	0.9		1.0		1.4		ns
t_{EABH}	0.4		0.4		0.6		ns
t_{EABCLR}	0.3		0.3		0.5		ns
t_{AA}		3.2		3.8		5.1	ns
t_{WP}	2.5		2.9		3.9		ns
t_{RP}	0.9		1.1		1.5		ns
t_{WDSU}	0.9		1.0		1.4		ns
t_{WDH}	0.1		0.1		0.2		ns
t_{WASU}	1.7		2.0		2.7		ns
t_{WAH}	1.8		2.1		2.9		ns
t_{RASU}	3.1		3.7		5.0		ns
t_{RAH}	0.2		0.2		0.3		ns
t_{WO}		2.5		2.9		3.9	ns
t_{DD}		2.5		2.9		3.9	ns
t_{EABOUT}		0.5		0.6		0.8	ns
t_{EABCH}	1.5		2.0		2.5		ns
t_{EABCL}	2.5		2.9		3.9		ns

Tables 52 through 58 show EPF10K130E device internal and external timing parameters.

Table 52. EPF10K130E Device LE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.6		0.9		1.3	ns
t_{CLUT}		0.6		0.8		1.0	ns
t_{RLUT}		0.7		0.9		0.2	ns
t_{PACKED}		0.3		0.5		0.6	ns
t_{EN}		0.2		0.3		0.4	ns
t_{CICO}		0.1		0.1		0.2	ns
t_{CGEN}		0.4		0.6		0.8	ns
t_{CGENR}		0.1		0.1		0.2	ns
t_{CASC}		0.6		0.9		1.2	ns
t_C		0.3		0.5		0.6	ns
t_{CO}		0.5		0.7		0.8	ns
t_{COMB}		0.3		0.5		0.6	ns
t_{SU}	0.5		0.7		0.8		ns
t_H	0.6		0.7		1.0		ns
t_{PRE}		0.9		1.2		1.6	ns
t_{CLR}		0.9		1.2		1.6	ns
t_{CH}	1.5		1.5		2.5		ns
t_{CL}	1.5		1.5		2.5		ns

Table 53. EPF10K130E Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.3		1.5		2.0	ns
t_{IOC}		0.0		0.0		0.0	ns
t_{IOCO}		0.6		0.8		1.0	ns
t_{IOCOMB}		0.6		0.8		1.0	ns
t_{IOSU}	1.0		1.2		1.6		ns
t_{IOH}	0.9		0.9		1.4		ns
t_{IOCLR}		0.6		0.8		1.0	ns
t_{OD1}		2.8		4.1		5.5	ns
t_{OD2}		2.8		4.1		5.5	ns

Table 61. EPF10K200E Device EAB Internal Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		2.0		2.4		3.2	ns
$t_{EABDATA1}$		0.4		0.5		0.6	ns
t_{EABWE1}		1.4		1.7		2.3	ns
t_{EABWE2}		0.0		0.0		0.0	ns
t_{EABRE1}		0		0		0	ns
t_{EABRE2}		0.4		0.5		0.6	ns
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		0.8		0.9		1.2	ns
$t_{EABYPASS}$		0.0		0.1		0.1	ns
t_{EABSU}	0.9		1.1		1.5		ns
t_{EABH}	0.4		0.5		0.6		ns
t_{EABCLR}	0.8		0.9		1.2		ns
t_{AA}		3.1		3.7		4.9	ns
t_{WP}	3.3		4.0		5.3		ns
t_{RP}	0.9		1.1		1.5		ns
t_{WDSU}	0.9		1.1		1.5		ns
t_{WDH}	0.1		0.1		0.1		ns
t_{WASU}	1.3		1.6		2.1		ns
t_{WAH}	2.1		2.5		3.3		ns
t_{RASU}	2.2		2.6		3.5		ns
t_{RAH}	0.1		0.1		0.2		ns
t_{WO}		2.0		2.4		3.2	ns
t_{DD}		2.0		2.4		3.2	ns
t_{EABOUT}		0.0		0.1		0.1	ns
t_{EABCH}	1.5		2.0		2.5		ns
t_{EABCL}	3.3		4.0		5.3		ns

Table 62. EPF10K200E Device EAB Internal Timing Macroparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		5.1		6.4		8.4	ns
$t_{EABRCOMB}$	5.1		6.4		8.4		ns
$t_{EABRCREG}$	4.8		5.7		7.6		ns
t_{EABWP}	3.3		4.0		5.3		ns

Table 68. EPF10K50S Device EAB Internal Microparameters *Note (1)*

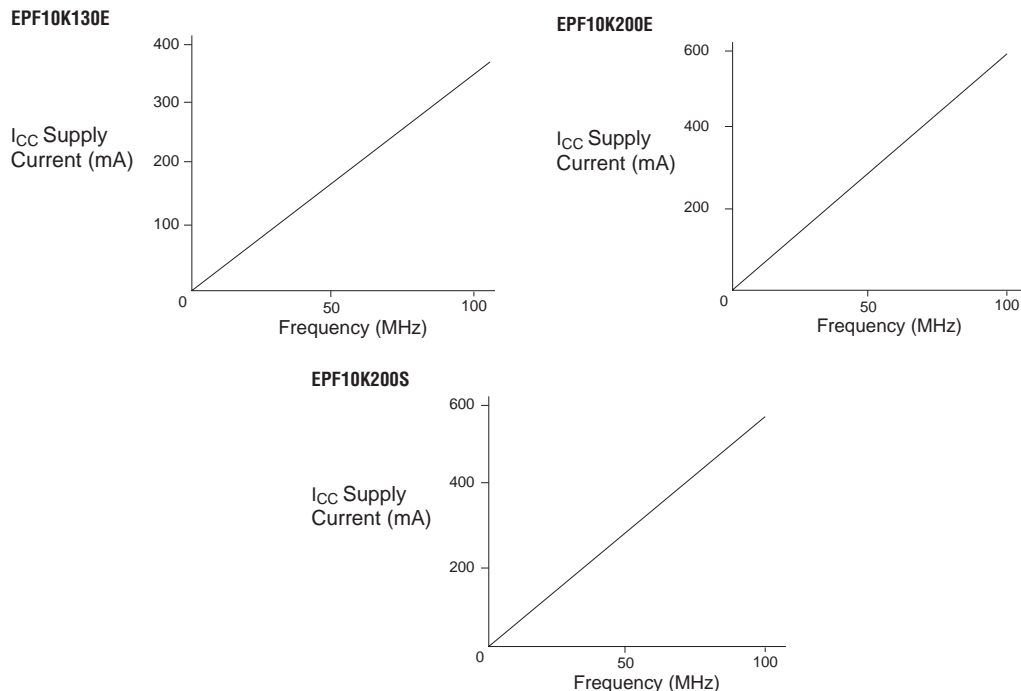
Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.7		2.4		3.2	ns
$t_{EABDATA2}$		0.4		0.6		0.8	ns
t_{EABWE1}		1.0		1.4		1.9	ns
t_{EABWE2}		0.0		0.0		0.0	ns
t_{EABRE1}		0.0		0.0		0.0	
t_{EABRE2}		0.4		0.6		0.8	
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		0.8		1.1		1.5	ns
$t_{EABYPASS}$		0.0		0.0		0.0	ns
t_{EABSU}	0.7		1.0		1.3		ns
t_{EABH}	0.4		0.6		0.8		ns
t_{EABCLR}	0.8		1.1		1.5		
t_{AA}		2.0		2.8		3.8	ns
t_{WP}	2.0		2.8		3.8		ns
t_{RP}	1.0		1.4		1.9		
t_{WDSU}	0.5		0.7		0.9		ns
t_{WDH}	0.1		0.1		0.2		ns
t_{WASU}	1.0		1.4		1.9		ns
t_{WAH}	1.5		2.1		2.9		ns
t_{RASU}	1.5		2.1		2.8		
t_{RAH}	0.1		0.1		0.2		
t_{WO}		2.1		2.9		4.0	ns
t_{DD}		2.1		2.9		4.0	ns
t_{EABOUT}		0.0		0.0		0.0	ns
t_{EABCH}	1.5		2.0		2.5		ns
t_{EABCL}	1.5		2.0		2.5		ns

Table 76. EPF10K200S Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		3.9		6.4		8.4	ns
$t_{EABRCOMB}$	3.9		6.4		8.4		ns
$t_{EABRCREG}$	3.6		5.7		7.6		ns
t_{EABWP}	2.1		4.0		5.3		ns
$t_{EABWCOMB}$	4.8		8.1		10.7		ns
$t_{EABWCREG}$	5.4		8.0		10.6		ns
t_{EABDD}		3.8		5.1		6.7	ns
$t_{EABDATACO}$		0.8		1.0		1.3	ns
$t_{EABDATASU}$	1.1		1.6		2.1		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	0.7		1.1		1.5		ns
t_{EABWEH}	0.4		0.5		0.6		ns
$t_{EABWDSU}$	1.2		1.8		2.4		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	1.9		3.6		4.7		ns
t_{EABWAH}	0.8		0.5		0.7		ns
t_{EABWO}		3.1		4.4		5.8	ns

Table 77. EPF10K200S Device Interconnect Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		4.4		4.8		5.5	ns
t_{DIN2LE}		0.6		0.6		0.9	ns
$t_{DIN2DATA}$		1.8		2.1		2.8	ns
$t_{DCLK2IOE}$		1.7		2.0		2.8	ns
$t_{DCLK2LE}$		0.6		0.6		0.9	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		3.0		4.6		5.7	ns
$t_{SAMECOLUMN}$		3.5		4.9		6.4	ns
$t_{DIFFROW}$		6.5		9.5		12.1	ns
$t_{TWOROWS}$		9.5		14.1		17.8	ns
$t_{LEPERIPH}$		5.5		6.2		7.2	ns
$t_{LABCARRY}$		0.3		0.1		0.2	ns

Figure 31. FLEX 10KE $I_{CCACTIVE}$ vs. Operating Frequency (Part 2 of 2)

Configuration & Operation

The FLEX 10KE architecture supports several configuration schemes. This section summarizes the device operating modes and available device configuration schemes.

Operating Modes

The FLEX 10KE architecture uses SRAM configuration elements that require configuration data to be loaded every time the circuit powers up. The process of physically loading the SRAM data into the device is called *configuration*. Before configuration, as V_{CC} rises, the device initiates a Power-On Reset (POR). This POR event clears the device and prepares it for configuration. The FLEX 10KE POR time does not exceed 50 μ s.

When configuring with a configuration device, refer to the respective configuration device data sheet for POR timing information.



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