Intel - EPF10K100EQI208-2 Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	624
Number of Logic Elements/Cells	4992
Total RAM Bits	49152
Number of I/O	147
Number of Gates	257000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k100eqi208-2

- Software design support and automatic place-and-route provided by Altera's development systems for Windows-based PCs and Sun SPARCstation, and HP 9000 Series 700/800
- Flexible package options
  - Available in a variety of packages with 144 to 672 pins, including the innovative FineLine BGA™ packages (see Tables 3 and 4)
  - SameFrame™ pin-out compatibility between FLEX 10KA and FLEX 10KE devices across a range of device densities and pin counts
- Additional design entry and simulation support provided by EDIF 2 0 0 and 3 0 0 netlist files, library of parameterized modules (LPM), DesignWare components, Verilog HDL, VHDL, and other interfaces to popular EDA tools from manufacturers such as Cadence, Exemplar Logic, Mentor Graphics, OrCAD, Synopsys, Synplicity, VeriBest, and Viewlogic

Table 3. FLEX 10KE Package Options & I/O Pin Count *Notes (1), (2)*

Device	144-Pin TQFP	208-Pin PQFP	240-Pin PQFP RQFP	256-Pin FineLine BGA	356-Pin BGA	484-Pin FineLine BGA	599-Pin PGA	600-Pin BGA	672-Pin FineLine BGA
EPF10K30E	102	147		176		220			220 *(3)*
EPF10K50E	102	147	189	191		254			254 *(3)*
EPF10K50S	102	147	189	191	220	254			254 *(3)*
EPF10K100E		147	189	191	274	338			338 *(3)*
EPF10K130E			186		274	369		424	413
EPF10K200E							470	470	470
EPF10K200S			182		274	369	470	470	470

Notes:

(1) FLEX 10KE device package types include thin quad flat pack (TQFP), plastic quad flat pack (PQFP), power quad flat pack (RQFP), pin-grid array (PGA), and ball-grid array (BGA) packages.

(2) Devices in the same package are pin-compatible, although some devices have more I/O pins than others. When planning device migration, use the I/O pins that are common to all devices.

(3) This option is supported with a 484-pin FineLine BGA package. By using SameFrame pin migration, all FineLine BGA packages are pin-compatible. For example, a board can be designed to support 256-pin, 484-pin, and 672-pin FineLine BGA packages. The Altera software automatically avoids conflicting pins when future migration is set.

Table 5. FLEX 10KE Performance

Application	Resources Used		Performance			Units
	LEs	EABs	-1 Speed Grade	-2 Speed Grade	-3 Speed Grade	
16-bit loadable counter	16	0	285	250	200	MHz
16-bit accumulator	16	0	285	250	200	MHz
16-to-1 multiplexer *(1)*	10	0	3.5	4.9	7.0	ns
16-bit multiplier with 3-stage pipeline *(2)*	592	0	156	131	93	MHz
256 × 16 RAM read cycle speed *(2)*	0	1	196	154	118	MHz
256 × 16 RAM write cycle speed *(2)*	0	1	185	143	106	MHz

Notes:
(1) This application uses combinatorial inputs and outputs.
(2) This application uses registered inputs and outputs.

Table 6 shows FLEX 10KE performance for more complex designs. These designs are available as Altera MegaCore® functions.

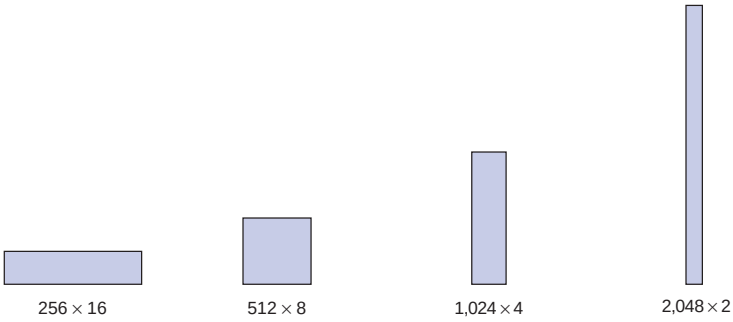
Table 6. FLEX 10KE Performance for Complex Designs

Application	LEs Used	Performance			Units
		-1 Speed Grade	-2 Speed Grade	-3 Speed Grade	
8-bit, 16-tap parallel finite impulse response (FIR) filter	597	192	156	116	MSPS
8-bit, 512-point fast Fourier transform (FFT) function	1,854	23.4	28.7	38.9	µs *(1)*
		113	92	68	MHz
a16450 universal asynchronous receiver/transmitter (UART)	342	36	28	20.5	MHz

Note:
(1) These values are for calculation time. Calculation time = number of clocks required / f_{max}. Number of clocks required = ceiling [log 2 (points) / 2] × [points +14 + ceiling]

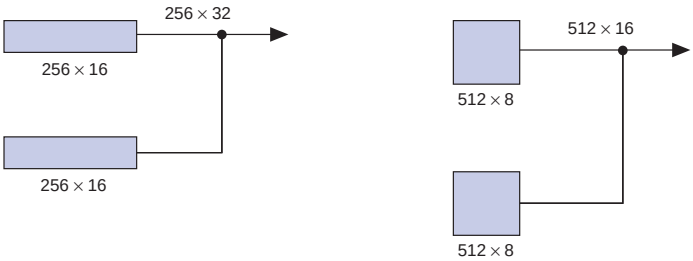
When used as RAM, each EAB can be configured in any of the following sizes: 256 × 16, 512 × 8, 1,024 × 4, or 2,048 × 2 (see Figure 5).

Figure 5. FLEX 10KE EAB Memory Configurations



Larger blocks of RAM are created by combining multiple EABs. For example, two 256 × 16 RAM blocks can be combined to form a 256 × 32 block; two 512 × 8 RAM blocks can be combined to form a 512 × 16 block (see Figure 6).

Figure 6. Examples of Combining FLEX 10KE EABs



If necessary, all EABs in a device can be cascaded to form a single RAM block. EABs can be cascaded to form RAM blocks of up to 2,048 words without impacting timing. The Altera software automatically combines EABs to meet a designer's RAM specifications.

Normal Mode

The normal mode is suitable for general logic applications and wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a four-input LUT. The Altera Compiler automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. Either the register or the LUT can be used to drive both the local interconnect and the FastTrack Interconnect routing structure at the same time.

The LUT and the register in the LE can be used independently (register packing). To support register packing, the LE has two outputs; one drives the local interconnect, and the other drives the FastTrack Interconnect routing structure. The DATA4 signal can drive the register directly, allowing the LUT to compute a function that is independent of the registered signal; a three-input function can be computed in the LUT, and a fourth independent signal can be registered. Alternatively, a four-input function can be generated, and one of the inputs to this function can be used to drive the register. The register in a packed LE can still use the clock enable, clear, and preset signals in the LE. In a packed LE, the register can drive the FastTrack Interconnect routing structure while the LUT drives the local interconnect, or vice versa.

Arithmetic Mode

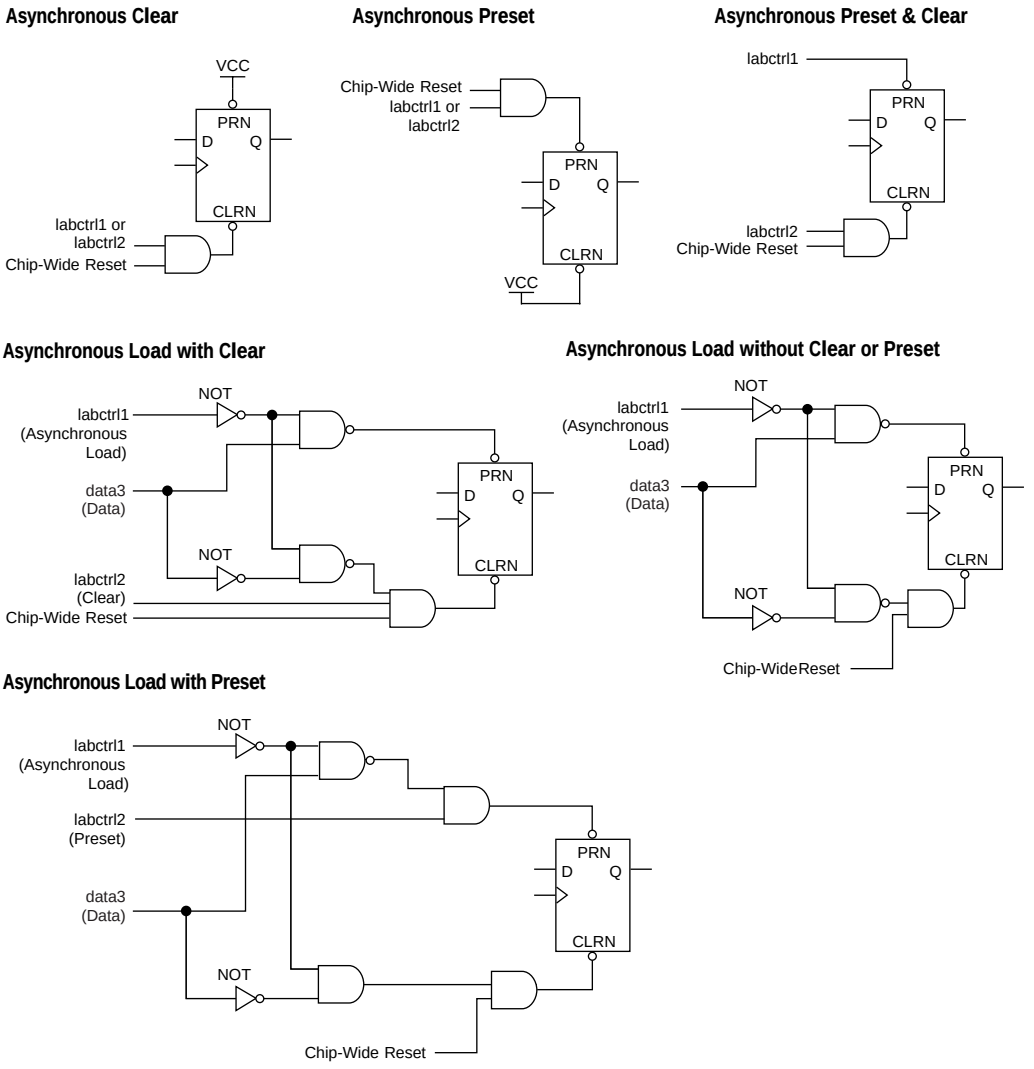
The arithmetic mode offers 2 three-input LUTs that are ideal for implementing adders, accumulators, and comparators. One LUT computes a three-input function; the other generates a carry output. As shown in Figure 11 on page 22, the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, in an adder, this output is the sum of three signals: a, b, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

Up/Down Counter Mode

The up / down counter mode offers counter enable, clock enable, synchronous up / down control, and data loading options. These control signals are generated by the data inputs from the LAB local interconnect, the carry-in signal, and output feedback from the programmable register. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading. Data can also be loaded asynchronously with the clear and preset register control signals without using the LUT resources.

In addition to the six clear and preset modes, FLEX 10KE devices provide a chip-wide reset pin that can reset all registers in the device. Use of this feature is set during design entry. In any of the clear and preset modes, the chip-wide reset overrides all other signals. Registers with asynchronous presets may be preset when the chip-wide reset is asserted. Inversion can be used to implement the asynchronous preset. Figure 12 shows examples of how to setup the preset and clear inputs for the desired functionality.

Figure 12. FLEX 10KE LE Clear & Preset Modes



FastTrack Interconnect Routing Structure

In the FLEX 10KE architecture, connections between LEs, EABs, and device I/O pins are provided by the FastTrack Interconnect routing structure, which is a series of continuous horizontal and vertical routing channels that traverses the device. This global routing structure provides predictable performance, even in complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.

The FastTrack Interconnect routing structure consists of row and column interconnect channels that span the entire device. Each row of LABs is served by a dedicated row interconnect. The row interconnect can drive I/O pins and feed other LABs in the row. The column interconnect routes signals between rows and can drive I/O pins.

Row channels drive into the LAB or EAB local interconnect. The row signal is buffered at every LAB or EAB to reduce the effect of fan-out on delay. A row channel can be driven by an LE or by one of three column channels. These four signals feed dual 4-to-1 multiplexers that connect to two specific row channels. These multiplexers, which are connected to each LE, allow column channels to drive row channels even when all eight LEs in a LAB drive the row interconnect.

Each column of LABs or EABs is served by a dedicated column interconnect. The column interconnect that serves the EABs has twice as many channels as other column interconnects. The column interconnect can then drive I/O pins or another row's interconnect to route the signals to other LABs or EABs in the device. A signal from the column interconnect, which can be either the output of a LE or an input from an I/O pin, must be routed to the row interconnect before it can enter a LAB or EAB. Each row channel that is driven by an IOE or EAB can drive one specific column channel.

Access to row and column channels can be switched between LEs in adjacent pairs of LABs. For example, a LE in one LAB can drive the row and column channels normally driven by a particular LE in the adjacent LAB in the same row, and vice versa. This flexibility enables routing resources to be used more efficiently (see Figure 13).

Column-to-IOE Connections

When an IOE is used as an input, it can drive up to two separate column channels. When an IOE is used as an output, the signal is driven by a multiplexer that selects a signal from the column channels. Two IOEs connect to each side of the column channels. Each IOE can be driven by column channels via a multiplexer. The set of column channels is different for each IOE (see Figure 17).

Figure 17. FLEX 10KE Column-to-IOE Connections

The values for m and n are provided in Table 11.

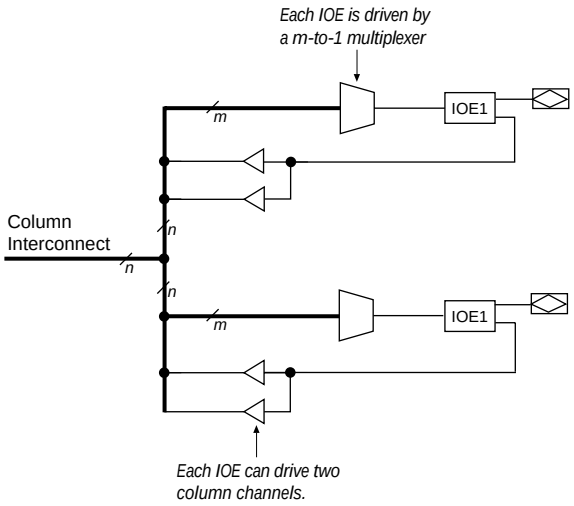


Table 11 lists the FLEX 10KE column-to-IOE interconnect resources.

Table 11. FLEX 10KE Column-to-IOE Interconnect Resources

Device	Channels per Column *(n)*	Column Channels per Pin *(m)*
EPF10K30E	24	16
EPF10K50E EPF10K50S	24	16
EPF10K100E	24	16
EPF10K130E	32	24
EPF10K200E EPF10K200S	48	40

ClockLock & ClockBoost Features

To support high-speed designs, FLEX 10KE devices offer optional ClockLock and ClockBoost circuitry containing a phase-locked loop (PLL) used to increase design speed and reduce resource usage. The ClockLock circuitry uses a synchronizing PLL that reduces the clock delay and skew within a device. This reduction minimizes clock-to-output and setup times while maintaining zero hold times. The ClockBoost circuitry, which provides a clock multiplier, allows the designer to enhance device area efficiency by resource sharing within the device. The ClockBoost feature allows the designer to distribute a low-speed clock and multiply that clock on-device. Combined, the ClockLock and ClockBoost features provide significant improvements in system performance and bandwidth.

All FLEX 10KE devices, except EPF10K50E and EPF10K200E devices, support ClockLock and ClockBoost circuitry. EPF10K50S and EPF10K200S devices support this circuitry. Devices that support ClockLock and ClockBoost circuitry are distinguished with an "X" suffix in the ordering code; for instance, the EPF10K200SFC672-1X device supports this circuit.

The ClockLock and ClockBoost features in FLEX 10KE devices are enabled through the Altera software. External devices are not required to use these features. The output of the ClockLock and ClockBoost circuits is not available at any of the device pins.

The ClockLock and ClockBoost circuitry locks onto the rising edge of the incoming clock. The circuit output can drive the clock inputs of registers only; the generated clock cannot be gated or inverted.

The dedicated clock pin (GCLK1) supplies the clock to the ClockLock and ClockBoost circuitry. When the dedicated clock pin is driving the ClockLock or ClockBoost circuitry, it cannot drive elsewhere in the device.

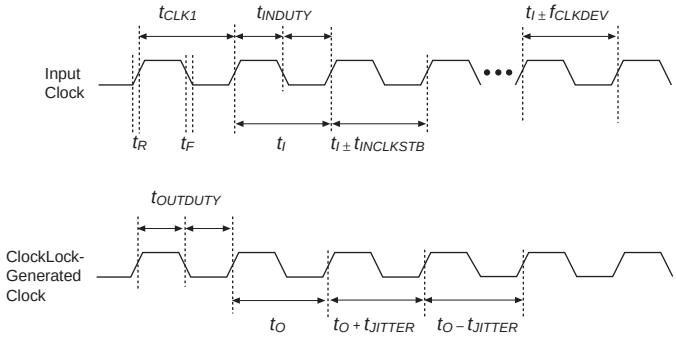
For designs that require both a multiplied and non-multiplied clock, the clock trace on the board can be connected to the GCLK1 pin. In the Altera software, the GCLK1 pin can feed both the ClockLock and ClockBoost circuitry in the FLEX 10KE device. However, when both circuits are used, the other clock pin cannot be used.

ClockLock & ClockBoost Timing Parameters

For the ClockLock and ClockBoost circuitry to function properly, the incoming clock must meet certain requirements. If these specifications are not met, the circuitry may not lock onto the incoming clock, which generates an erroneous clock within the device. The clock generated by the ClockLock and ClockBoost circuitry must also meet certain specifications. If the incoming clock meets these requirements during configuration, the ClockLock and ClockBoost circuitry will lock onto the clock during configuration. The circuit will be ready for use immediately after configuration. Figure 19 shows the incoming and generated clock specifications.

Figure 19. Specifications for Incoming & Generated Clocks

The t_I parameter refers to the nominal input clock period; the t_O parameter refers to the nominal output clock period.



Tables 12 and 13 summarize the ClockLock and ClockBoost parameters for -1 and -2 speed-grade devices, respectively.

Table 12. ClockLock & ClockBoost Parameters for -1 Speed-Grade Devices

Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_R	Input rise time				5	ns
t_F	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		180	MHz
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		90	MHz
f_{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software *(1)*				25,000 *(2)*	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock *(3)*				10	µs
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock *(4)*	$t_{INCLKSTB}$ < 100			250	ps
		$t_{INCLKSTB}$ < 50			200 *(4)*	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Table 13. ClockLock & ClockBoost Parameters for -2 Speed-Grade Devices

Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_R	Input rise time				5	ns
t_F	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		75	MHz
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		37.5	MHz
f_{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software *(1)*				25,000 *(2)*	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock *(3)*				10	µs
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock *(4)*	$t_{INCLKSTB}$ < 100			250	ps
		$t_{INCLKSTB}$ < 50			200 *(4)*	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Notes to tables:

(1) To implement the ClockLock and ClockBoost circuitry with the MAX+PLUS II software, designers must specify the input frequency. The Altera software tunes the PLL in the ClockLock and ClockBoost circuitry to this frequency. The f_{CLKDEV} parameter specifies how much the incoming clock can differ from the specified frequency during device operation. Simulation does not reflect this parameter.

(2) Twenty-five thousand parts per million (PPM) equates to 2.5% of input clock period.

(3) During device configuration, the ClockLock and ClockBoost circuitry is configured before the rest of the device. If the incoming clock is supplied during configuration, the ClockLock and ClockBoost circuitry locks during configuration because the t_{LOCK} value is less than the time required for configuration.

(4) The t_{JITTER} specification is measured under long-term observation. The maximum value for t_{JITTER} is 200 ps if $t_{INCLKSTB}$ is lower than 50 ps.

I/O Configuration

This section discusses the peripheral component interconnect (PCI) pull-up clamping diode option, slew-rate control, open-drain output option, and MultiVolt I/O interface for FLEX 10KE devices. The PCI pull-up clamping diode, slew-rate control, and open-drain output options are controlled pin-by-pin via Altera software logic options. The MultiVolt I/O interface is controlled by connecting V_{CCIO} to a different voltage than V_{CCINT}. Its effect can be simulated in the Altera software via the **Global Project Device Options** dialog box (Assign menu).

The VCCINT pins must always be connected to a 2.5-V power supply. With a 2.5-V V_{CCINT} level, input voltages are compatible with 2.5-V, 3.3-V, and 5.0-V inputs. The VCCIO pins can be connected to either a 2.5-V or 3.3-V power supply, depending on the output requirements. When the VCCIO pins are connected to a 2.5-V power supply, the output levels are compatible with 2.5-V systems. When the VCCIO pins are connected to a 3.3-V power supply, the output high is at 3.3 V and is therefore compatible with 3.3-V or 5.0-V systems. Devices operating with V_{CCIO} levels higher than 3.0 V achieve a faster timing delay of t_{OD2} instead of t_{OD1}.

Table 14 summarizes FLEX 10KE MultiVolt I/O support.

Table 14. FLEX 10KE MultiVolt I/O Support

V_{CCIO} (V)	Input Signal (V)			Output Signal (V)		
	2.5	3.3	5.0	2.5	3.3	5.0
2.5	✓	✓ *(1)*	✓ *(1)*	✓		
3.3	✓	✓	✓ *(1)*	✓ *(2)*	✓	✓

Notes:

(1) The PCI clamping diode must be disabled to drive an input with voltages higher than V_{CCIO}.

(2) When V_{CCIO} = 3.3 V, a FLEX 10KE device can drive a 2.5-V device that has 3.3-V tolerant inputs.

Open-drain output pins on FLEX 10KE devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a V_{IH} of 3.5 V. When the open-drain pin is active, it will drive low. When the pin is inactive, the trace will be pulled up to 5.0 V by the resistor. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The I_{OL} current specification should be considered when selecting a pull-up resistor.

Power Sequencing & Hot-Socketing

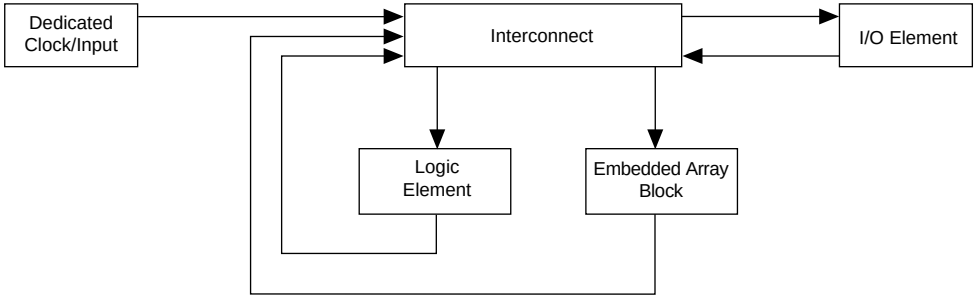
Because FLEX 10KE devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The V_{CCIO} and V_{CCINT} power planes can be powered in any order.

Signals can be driven into FLEX 10KE devices before and during power up without damaging the device. Additionally, FLEX 10KE devices do not drive out during power up. Once operating conditions are reached, FLEX 10KE devices operate as specified by the user.

Timing simulation and delay prediction are available with the Altera Simulator and Timing Analyzer, or with industry-standard EDA tools. The Simulator offers both pre-synthesis functional simulation to evaluate logic design accuracy and post-synthesis timing simulation with 0.1-ns resolution. The Timing Analyzer provides point-to-point timing delay information, setup and hold time analysis, and device-wide performance analysis.

Figure 24 shows the overall timing model, which maps the possible paths to and from the various elements of the FLEX 10KE device.

Figure 24. FLEX 10KE Device Timing Model



Figures 25 through 28 show the delays that correspond to various paths and functions within the LE, IOE, EAB, and bidirectional timing models.

Table 26. EAB Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
$t_{EABDATA1}$	Data or address delay to EAB for combinatorial input	
$t_{EABDATA2}$	Data or address delay to EAB for registered input	
t_{EABWE1}	Write enable delay to EAB for combinatorial input	
t_{EABWE2}	Write enable delay to EAB for registered input	
t_{EABRE1}	Read enable delay to EAB for combinatorial input	
t_{EABRE2}	Read enable delay to EAB for registered input	
t_{EABCLK}	EAB register clock delay	
t_{EABCO}	EAB register clock-to-output delay	
$t_{EABBYPASS}$	Bypass register delay	
t_{EABSU}	EAB register setup time before clock	
t_{EABH}	EAB register hold time after clock	
t_{EABCLR}	EAB register asynchronous clear time to output delay	
t_{AA}	Address access delay (including the read enable to output delay)	
t_{WP}	Write pulse width	
t_{RP}	Read pulse width	
t_{WDSU}	Data setup time before falling edge of write pulse	*(5)*
t_{WDH}	Data hold time after falling edge of write pulse	*(5)*
t_{WASU}	Address setup time before rising edge of write pulse	*(5)*
t_{WAH}	Address hold time after falling edge of write pulse	*(5)*
t_{RASU}	Address setup time with respect to the falling edge of the read enable	
t_{RAH}	Address hold time with respect to the falling edge of the read enable	
t_{WO}	Write enable to data output valid delay	
t_{DD}	Data-in to data-out valid delay	
t_{EABOUT}	Data-out delay	
t_{EABCH}	Clock high time	
t_{EABCL}	Clock low time	

Table 33. EPF10K30E Device EAB Internal Microparameters *Note (1)*							
Symbol	**-1 Speed Grade**		**-2 Speed Grade**		**-3 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	**Min**	**Max**	
$t_{EABDATA1}$		1.7		2.0		2.3	ns
$t_{EABDATA1}$		0.6		0.7		0.8	ns
t_{EABWE1}		1.1		1.3		1.4	ns
t_{EABWE2}		0.4		0.4		0.5	ns
t_{EABRE1}		0.8		0.9		1.0	ns
t_{EABRE2}		0.4		0.4		0.5	ns
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		0.3		0.3		0.4	ns
$t_{EABBYPASS}$		0.5		0.6		0.7	ns
t_{EABSU}	0.9		1.0		1.2		ns
t_{EABH}	0.4		0.4		0.5		ns
t_{EABCLR}	0.3		0.3		0.3		ns
t_{AA}		3.2		3.8		4.4	ns
t_{WP}	2.5		2.9		3.3		ns
t_{RP}	0.9		1.1		1.2		ns
t_{WDSU}	0.9		1.0		1.1		ns
t_{WDH}	0.1		0.1		0.1		ns
t_{WASU}	1.7		2.0		2.3		ns
t_{WAH}	1.8		2.1		2.4		ns
t_{RASU}	3.1		3.7		4.2		ns
t_{RAH}	0.2		0.2		0.2		ns
t_{WO}		2.5		2.9		3.3	ns
t_{DD}		2.5		2.9		3.3	ns
t_{EABOUT}		0.5		0.6		0.7	ns
t_{EABCH}	1.5		2.0		2.3		ns
t_{EABCL}	2.5		2.9		3.3		ns

Table 34. EPF10K30E Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		6.4		7.6		8.8	ns
$t_{EABRCOMB}$	6.4		7.6		8.8		ns
$t_{EABRCREG}$	4.4		5.1		6.0		ns
t_{EABWP}	2.5		2.9		3.3		ns
$t_{EABWCOMB}$	6.0		7.0		8.0		ns
$t_{EABWCREG}$	6.8		7.8		9.0		ns
t_{EABDD}		5.7		6.7		7.7	ns
$t_{EABDATACO}$		0.8		0.9		1.1	ns
$t_{EABDATASU}$	1.5		1.7		2.0		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	1.3		1.4		1.7		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	1.5		1.7		2.0		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	3.0		3.6		4.3		ns
t_{EABWAH}	0.5		0.5		0.4		ns
t_{EABWO}		5.1		6.0		6.8	ns

Table 41. EPF10K50E Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		6.4		7.6		10.2	ns
$t_{EABRCOMB}$	6.4		7.6		10.2		ns
$t_{EABRCREG}$	4.4		5.1		7.0		ns
t_{EABWP}	2.5		2.9		3.9		ns
$t_{EABWCOMB}$	6.0		7.0		9.5		ns
$t_{EABWCREG}$	6.8		7.8		10.6		ns
t_{EABDD}		5.7		6.7		9.0	ns
$t_{EABDATACO}$		0.8		0.9		1.3	ns
$t_{EABDATASU}$	1.5		1.7		2.3		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	1.3		1.4		2.0		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	1.5		1.7		2.3		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	3.0		3.6		4.8		ns
t_{EABWAH}	0.5		0.5		0.8		ns
t_{EABWO}		5.1		6.0		8.1	ns

Table 42. EPF10K50E Device Interconnect Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		3.5		4.3		5.6	ns
t_{DIN2LE}		2.1		2.5		3.4	ns
$t_{DIN2DATA}$		2.2		2.4		3.1	ns
$t_{DCLK2IOE}$		2.9		3.5		4.7	ns
$t_{DCLK2LE}$		2.1		2.5		3.4	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		1.1		1.1		1.5	ns
$t_{SAMECOLUMN}$		0.8		1.0		1.3	ns
$t_{DIFFROW}$		1.9		2.1		2.8	ns
$t_{TWOROWS}$		3.0		3.2		4.3	ns
$t_{LEPERIPH}$		3.1		3.3		3.7	ns
$t_{LABCARRY}$		0.1		0.1		0.2	ns
$t_{LABCASC}$		0.3		0.3		0.5	ns

Tables 52 through 58 show EPF10K130E device internal and external timing parameters.

Table 52. EPF10K130E Device LE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.6		0.9		1.3	ns
t_{CLUT}		0.6		0.8		1.0	ns
t_{RLUT}		0.7		0.9		0.2	ns
t_{PACKED}		0.3		0.5		0.6	ns
t_{EN}		0.2		0.3		0.4	ns
t_{CICO}		0.1		0.1		0.2	ns
t_{CGEN}		0.4		0.6		0.8	ns
t_{CGENR}		0.1		0.1		0.2	ns
t_{CASC}		0.6		0.9		1.2	ns
t_{C}		0.3		0.5		0.6	ns
t_{CO}		0.5		0.7		0.8	ns
t_{COMB}		0.3		0.5		0.6	ns
t_{SU}	0.5		0.7		0.8		ns
t_{H}	0.6		0.7		1.0		ns
t_{PRE}		0.9		1.2		1.6	ns
t_{CLR}		0.9		1.2		1.6	ns
t_{CH}	1.5		1.5		2.5		ns
t_{CL}	1.5		1.5		2.5		ns

Table 53. EPF10K130E Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.3		1.5		2.0	ns
t_{IOC}		0.0		0.0		0.0	ns
t_{IOCO}		0.6		0.8		1.0	ns
t_{IOCOMB}		0.6		0.8		1.0	ns
t_{IOSU}	1.0		1.2		1.6		ns
t_{IOH}	0.9		0.9		1.4		ns
t_{IOCLR}		0.6		0.8		1.0	ns
t_{OD1}		2.8		4.1		5.5	ns
t_{OD2}		2.8		4.1		5.5	ns

Table 58. EPF10K130E External Bidirectional Timing Parameters Notes (1), (2)							
Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{INSUBIDIR}$ (3)	2.2		2.4		3.2		ns
$t_{INHBIDIR}$ (3)	0.0		0.0		0.0		ns
$t_{INSUBIDIR}$ (4)	2.8		3.0		–		ns
$t_{INHBIDIR}$ (4)	0.0		0.0		–		ns
$t_{OUTCOBIDIR}$ (3)	2.0	5.0	2.0	7.0	2.0	9.2	ns
$t_{XZBIDIR}$ (3)		5.6		8.1		10.8	ns
$t_{ZXBIDIR}$ (3)		5.6		8.1		10.8	ns
$t_{OUTCOBIDIR}$ (4)	0.5	4.0	0.5	6.0	–	–	ns
$t_{XZBIDIR}$ (4)		4.6		7.1		–	ns
$t_{ZXBIDIR}$ (4)		4.6		7.1		–	ns

Notes to tables:

(1) All timing parameters are described in Tables 24 through 30 in this data sheet.
(2) These parameters are specified by characterization.
(3) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
(4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Tables 59 through 65 show EPF10K200E device internal and external timing parameters.

Table 59. EPF10K200E Device LE Timing Microparameters (Part 1 of 2) Note (1)							
Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.7		0.8		1.2	ns
t_{CLUT}		0.4		0.5		0.6	ns
t_{RLUT}		0.6		0.7		0.9	ns
t_{PACKED}		0.3		0.5		0.7	ns
t_{EN}		0.4		0.5		0.6	ns
t_{CICO}		0.2		0.2		0.3	ns
t_{CGEN}		0.4		0.4		0.6	ns
t_{CGENR}		0.2		0.2		0.3	ns
t_{CASC}		0.7		0.8		1.2	ns
t_{C}		0.5		0.6		0.8	ns
t_{CO}		0.5		0.6		0.8	ns
t_{COMB}		0.4		0.6		0.8	ns
t_{SU}	0.4		0.6		0.7		ns

Table 73. EPF10K200S Device Internal & External Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.7		0.8		1.2	ns
t_{CLUT}		0.4		0.5		0.6	ns
t_{RLUT}		0.5		0.7		0.9	ns
t_{PACKED}		0.4		0.5		0.7	ns
t_{EN}		0.6		0.5		0.6	ns
t_{CICO}		0.1		0.2		0.3	ns
t_{CGEN}		0.3		0.4		0.6	ns
t_{CGENR}		0.1		0.2		0.3	ns
t_{CASC}		0.7		0.8		1.2	ns
t_{C}		0.5		0.6		0.8	ns
t_{CO}		0.5		0.6		0.8	ns
t_{COMB}		0.3		0.6		0.8	ns
t_{SU}	0.4		0.6		0.7		ns
t_{H}	1.0		1.1		1.5		ns
t_{PRE}		0.4		0.6		0.8	ns
t_{CLR}		0.5		0.6		0.8	ns
t_{CH}	2.0		2.5		3.0		ns
t_{CL}	2.0		2.5		3.0		ns

Table 74. EPF10K200S Device IOE Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.8		1.9		2.6	ns
t_{IOC}		0.3		0.3		0.5	ns
t_{IOCO}		1.7		1.9		2.6	ns
t_{IOCOMB}		0.5		0.6		0.8	ns
t_{IOSU}	0.8		0.9		1.2		ns
t_{IOH}	0.4		0.8		1.1		ns
t_{IOCLR}		0.2		0.2		0.3	ns
t_{OD1}		1.3		0.7		0.9	ns
t_{OD2}		0.8		0.2		0.4	ns
t_{OD3}		2.9		3.0		3.9	ns
t_{XZ}		5.0		5.3		7.1	ns
t_{ZX1}		5.0		5.3		7.1	ns