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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	832
Number of Logic Elements/Cells	6656
Total RAM Bits	65536
Number of I/O	274
Number of Gates	342000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	356-LBGA
Supplier Device Package	356-BGA (35x35)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k130ebc356-1

Similar to the FLEX 10KE architecture, embedded gate arrays are the fastest-growing segment of the gate array market. As with standard gate arrays, embedded gate arrays implement general logic in a conventional “sea-of-gates” architecture. Additionally, embedded gate arrays have dedicated die areas for implementing large, specialized functions. By embedding functions in silicon, embedded gate arrays reduce die area and increase speed when compared to standard gate arrays. While embedded megafunctions typically cannot be customized, FLEX 10KE devices are programmable, providing the designer with full control over embedded megafunctions and general logic, while facilitating iterative design changes during debugging.

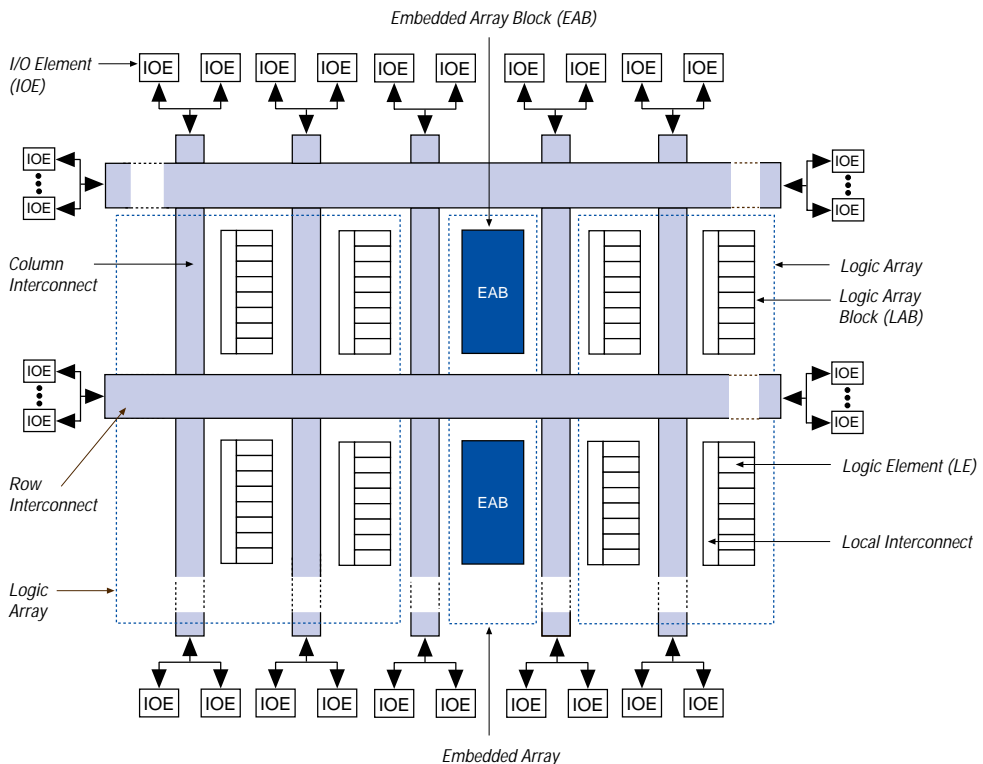
Each FLEX 10KE device contains an embedded array and a logic array. The embedded array is used to implement a variety of memory functions or complex logic functions, such as digital signal processing (DSP), wide data-path manipulation, microcontroller applications, and data-transformation functions. The logic array performs the same function as the sea-of-gates in the gate array and is used to implement general logic such as counters, adders, state machines, and multiplexers. The combination of embedded and logic arrays provides the high performance and high density of embedded gate arrays, enabling designers to implement an entire system on a single device.

FLEX 10KE devices are configured at system power-up with data stored in an Altera serial configuration device or provided by a system controller. Altera offers the EPC1, EPC2, and EPC16 configuration devices, which configure FLEX 10KE devices via a serial data stream. Configuration data can also be downloaded from system RAM or via the Altera BitBlaster™, ByteBlasterMV™, or MasterBlaster download cables. After a FLEX 10KE device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Because reconfiguration requires less than 85 ms, real-time changes can be made during system operation.

FLEX 10KE devices contain an interface that permits microprocessors to configure FLEX 10KE devices serially or in-parallel, and synchronously or asynchronously. The interface also enables microprocessors to treat a FLEX 10KE device as memory and configure it by writing to a virtual memory location, making it easy to reconfigure the device.

Figure 1 shows a block diagram of the FLEX 10KE architecture. Each group of LEs is combined into an LAB; groups of LABs are arranged into rows and columns. Each row also contains a single EAB. The LABs and EABs are interconnected by the FastTrack Interconnect routing structure. IOEs are located at the end of each row and column of the FastTrack Interconnect routing structure.

Figure 1. FLEX 10KE Device Block Diagram



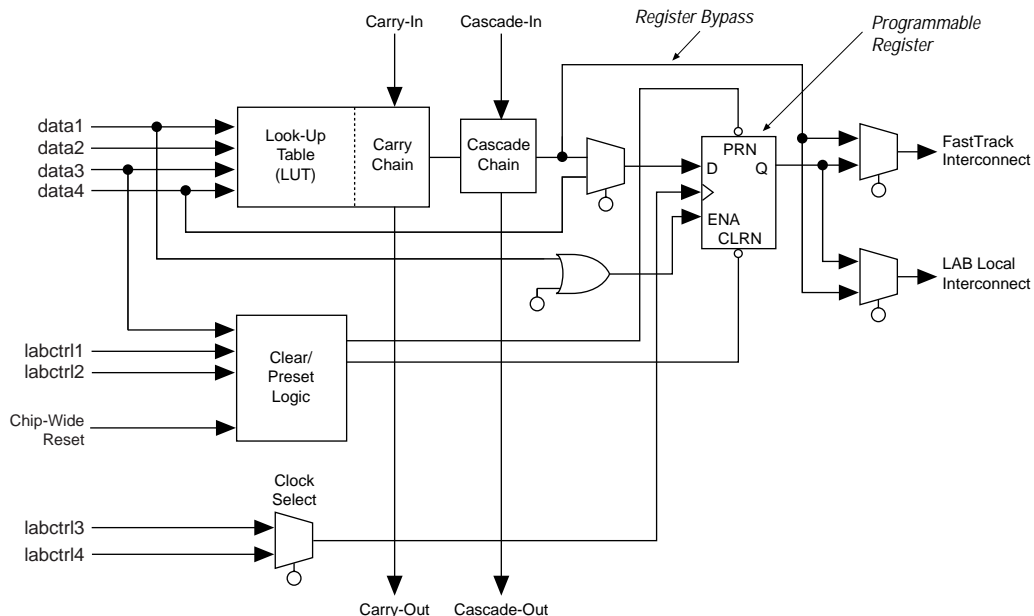
FLEX 10KE devices provide six dedicated inputs that drive the flipflops' control inputs and ensure the efficient distribution of high-speed, low-skew (less than 1.5 ns) control signals. These signals use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect routing structure. Four of the dedicated inputs drive four global signals. These four global signals can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device.

Each LAB provides four control signals with programmable inversion that can be used in all eight LEs. Two of these signals can be used as clocks, the other two can be used for clear/preset control. The LAB clocks can be driven by the dedicated clock input pins, global signals, I/O signals, or internal signals via the LAB local interconnect. The LAB preset and clear control signals can be driven by the global signals, I/O signals, or internal signals via the LAB local interconnect. The global control signals are typically used for global clock, clear, or preset signals because they provide asynchronous control with very low skew across the device. If logic is required on a control signal, it can be generated in one or more LE in any LAB and driven into the local interconnect of the target LAB. In addition, the global control signals can be generated from LE outputs.

Logic Element

The LE, the smallest unit of logic in the FLEX 10KE architecture, has a compact size that provides efficient logic utilization. Each LE contains a four-input LUT, which is a function generator that can quickly compute any function of four variables. In addition, each LE contains a programmable flipflop with a synchronous clock enable, a carry chain, and a cascade chain. Each LE drives both the local and the FastTrack Interconnect routing structure (see [Figure 8](#)).

Figure 8. FLEX 10KE Logic Element



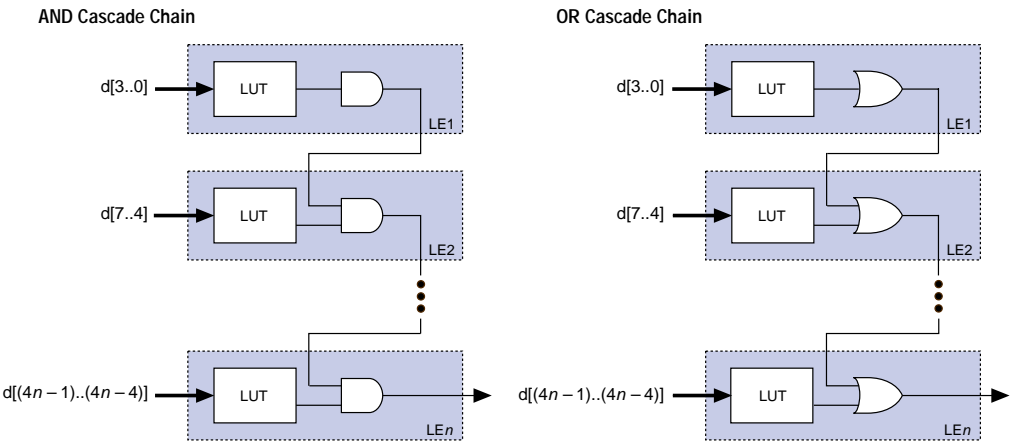
Cascade Chain

With the cascade chain, the FLEX 10KE architecture can implement functions that have a very wide fan-in. Adjacent LUTs can be used to compute portions of the function in parallel; the cascade chain serially connects the intermediate values. The cascade chain can use a logical AND or logical OR (via De Morgan's inversion) to connect the outputs of adjacent LEs. An a delay as low as 0.6 ns per LE, each additional LE provides four more inputs to the effective width of a function. Cascade chain logic can be created automatically by the Altera Compiler during design processing, or manually by the designer during design entry.

Cascade chains longer than eight bits are implemented automatically by linking several LABs together. For easier routing, a long cascade chain skips every other LAB in a row. A cascade chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB (e.g., the last LE of the first LAB in a row cascades to the first LE of the third LAB). The cascade chain does not cross the center of the row (e.g., in the EPF10K50E device, the cascade chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB). This break is due to the EAB's placement in the middle of the row.

Figure 10 shows how the cascade function can connect adjacent LEs to form functions with a wide fan-in. These examples show functions of $4n$ variables implemented with n LEs. The LE delay is 0.9 ns; the cascade chain delay is 0.6 ns. With the cascade chain, 2.7 ns are needed to decode a 16-bit address.

Figure 10. FLEX 10KE Cascade Chain Operation



LE Operating Modes

The FLEX 10KE LE can operate in the following four modes:

- Normal mode
- Arithmetic mode
- Up/down counter mode
- Clearable counter mode

Each of these modes uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. Three inputs to the LE provide clock, clear, and preset control for the register. The Altera software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions that use a specific LE operating mode for optimal performance.

The architecture provides a synchronous clock enable to the register in all four modes. The Altera software can set `DATA1` to enable the register synchronously, providing easy implementation of fully synchronous designs.

Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is ANDed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-states without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

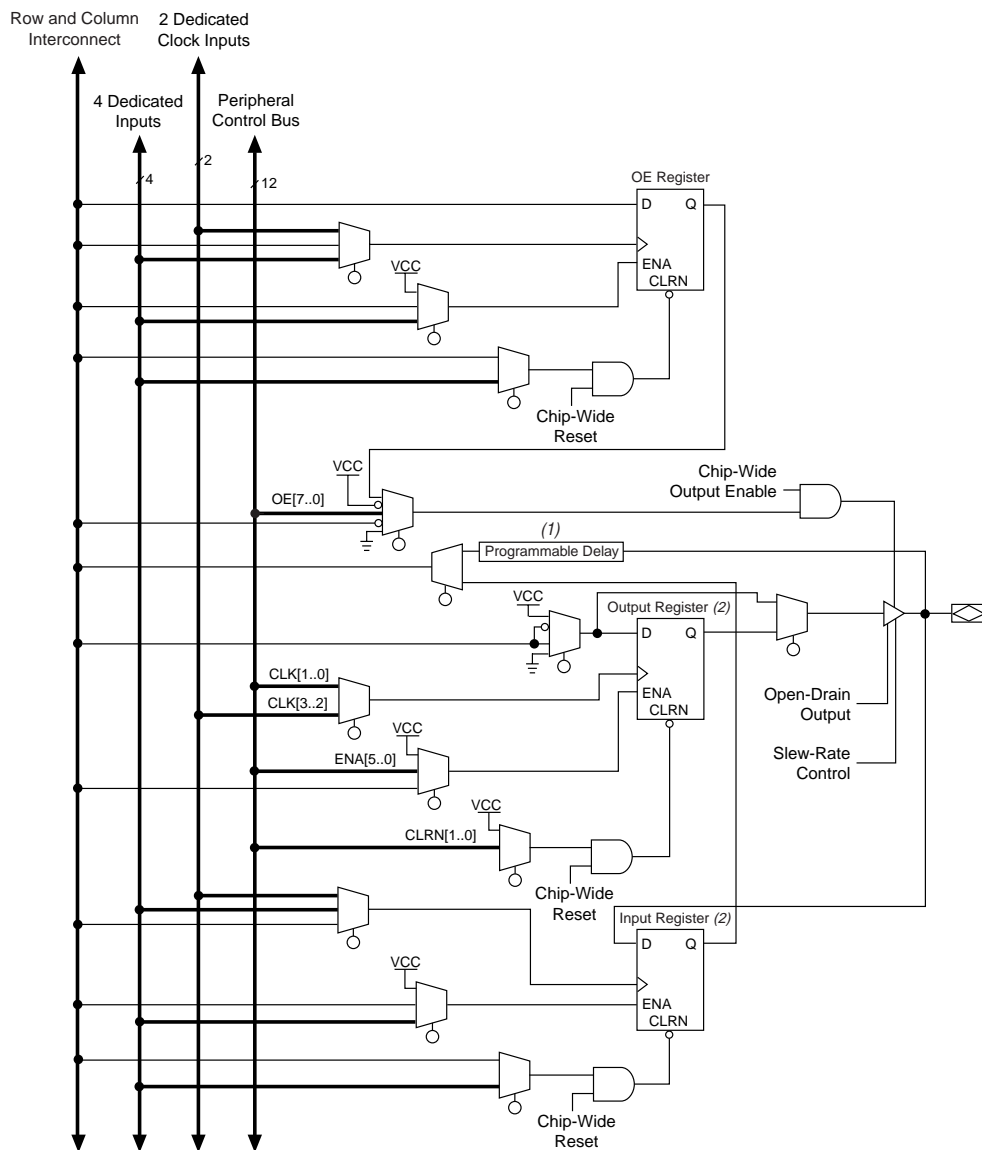
Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the Altera Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

Figure 15. FLEX 10KE Bidirectional I/O Registers

**Note:**

- (1) All FLEX 10KE devices (except the EPF10K50E and EPF10K200E devices) have a programmable input delay buffer on the input path.

When dedicated inputs drive non-inverted and inverted peripheral clears, clock enables, and output enables, two signals on the peripheral control bus will be used.

Tables 8 and 9 list the sources for each peripheral control signal, and show how the output enable, clock enable, clock, and clear signals share 12 peripheral control signals. The tables also show the rows that can drive global signals.

Table 8. Peripheral Bus Sources for EPF10K30E, EPF10K50E & EPF10K50S Devices

Peripheral Control Signal	EPF10K30E	EPF10K50E EPF10K50S
OE0	Row A	Row A
OE1	Row B	Row B
OE2	Row C	Row D
OE3	Row D	Row F
OE4	Row E	Row H
OE5	Row F	Row J
CLKENA0/CLK0/GLOBAL0	Row A	Row A
CLKENA1/OE6/GLOBAL1	Row B	Row C
CLKENA2/CLR0	Row C	Row E
CLKENA3/OE7/GLOBAL2	Row D	Row G
CLKENA4/CLR1	Row E	Row I
CLKENA5/CLK1/GLOBAL3	Row F	Row J

ClockLock & ClockBoost Features

To support high-speed designs, FLEX 10KE devices offer optional ClockLock and ClockBoost circuitry containing a phase-locked loop (PLL) used to increase design speed and reduce resource usage. The ClockLock circuitry uses a synchronizing PLL that reduces the clock delay and skew within a device. This reduction minimizes clock-to-output and setup times while maintaining zero hold times. The ClockBoost circuitry, which provides a clock multiplier, allows the designer to enhance device area efficiency by resource sharing within the device. The ClockBoost feature allows the designer to distribute a low-speed clock and multiply that clock on-device. Combined, the ClockLock and ClockBoost features provide significant improvements in system performance and bandwidth.

All FLEX 10KE devices, except EPF10K50E and EPF10K200E devices, support ClockLock and ClockBoost circuitry. EPF10K50S and EPF10K200S devices support this circuitry. Devices that support ClockLock and ClockBoost circuitry are distinguished with an "X" suffix in the ordering code; for instance, the EPF10K200SFC672-1X device supports this circuit.

The ClockLock and ClockBoost features in FLEX 10KE devices are enabled through the Altera software. External devices are not required to use these features. The output of the ClockLock and ClockBoost circuits is not available at any of the device pins.

The ClockLock and ClockBoost circuitry locks onto the rising edge of the incoming clock. The circuit output can drive the clock inputs of registers only; the generated clock cannot be gated or inverted.

The dedicated clock pin (`GCLK1`) supplies the clock to the ClockLock and ClockBoost circuitry. When the dedicated clock pin is driving the ClockLock or ClockBoost circuitry, it cannot drive elsewhere in the device.

For designs that require both a multiplied and non-multiplied clock, the clock trace on the board can be connected to the `GCLK1` pin. In the Altera software, the `GCLK1` pin can feed both the ClockLock and ClockBoost circuitry in the FLEX 10KE device. However, when both circuits are used, the other clock pin cannot be used.

Tables 12 and 13 summarize the ClockLock and ClockBoost parameters for -1 and -2 speed-grade devices, respectively.

Table 12. ClockLock & ClockBoost Parameters for -1 Speed-Grade Devices

Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_R	Input rise time				5	ns
t_F	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		180	MHz
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		90	MHz
f_{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software (1)				25,000 (2)	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock (3)				10	μs
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock (4)	$t_{INCLKSTB} < 100$			250	ps
		$t_{INCLKSTB} < 50$			200 (4)	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Table 13. ClockLock & ClockBoost Parameters for -2 Speed-Grade Devices

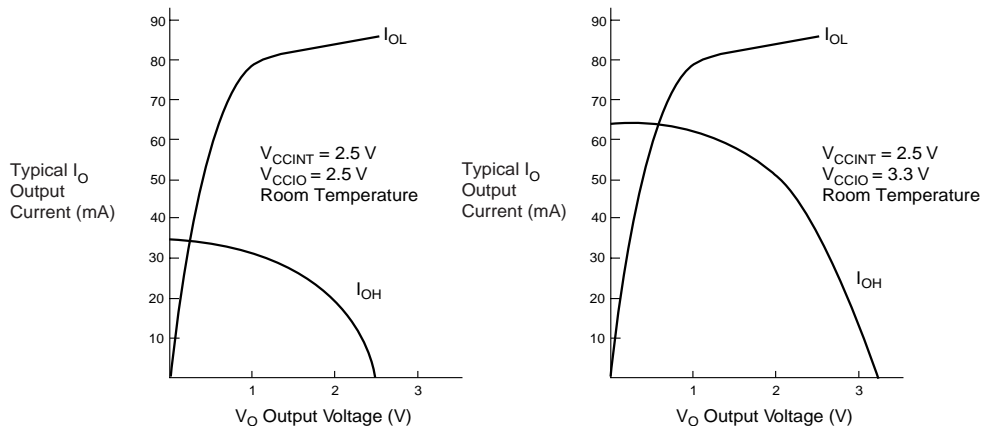
Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_R	Input rise time				5	ns
t_F	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		75	MHz
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		37.5	MHz
f_{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software (1)				25,000 (2)	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock (3)				10	μs
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock (4)	$t_{INCLKSTB} < 100$			250	ps
		$t_{INCLKSTB} < 50$			200 (4)	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Notes to tables:

- (1) To implement the ClockLock and ClockBoost circuitry with the MAX+PLUS II software, designers must specify the input frequency. The Altera software tunes the PLL in the ClockLock and ClockBoost circuitry to this frequency. The f_{CLKDEV} parameter specifies how much the incoming clock can differ from the specified frequency during device operation. Simulation does not reflect this parameter.
- (2) Twenty-five thousand parts per million (PPM) equates to 2.5% of input clock period.
- (3) During device configuration, the ClockLock and ClockBoost circuitry is configured before the rest of the device. If the incoming clock is supplied during configuration, the ClockLock and ClockBoost circuitry locks during configuration because the t_{LOCK} value is less than the time required for configuration.
- (4) The t_{JITTER} specification is measured under long-term observation. The maximum value for t_{JITTER} is 200 ps if $t_{INCLKSTB}$ is lower than 50 ps.

I/O Configuration

This section discusses the peripheral component interconnect (PCI) pull-up clamping diode option, slew-rate control, open-drain output option, and MultiVolt I/O interface for FLEX 10KE devices. The PCI pull-up clamping diode, slew-rate control, and open-drain output options are controlled pin-by-pin via Altera software logic options. The MultiVolt I/O interface is controlled by connecting V_{CCIO} to a different voltage than V_{CCINT} . Its effect can be simulated in the Altera software via the **Global Project Device Options** dialog box (Assign menu).

Figure 23. Output Drive Characteristics of FLEX 10KE Devices *Note (1)***Note:**

(1) These are transient (AC) currents.

Timing Model

The continuous, high-performance FastTrack Interconnect routing resources ensure predictable performance and accurate simulation and timing analysis. This predictable performance contrasts with that of FPGAs, which use a segmented connection scheme and therefore have unpredictable performance.

Device performance can be estimated by following the signal path from a source, through the interconnect, to the destination. For example, the registered performance between two LEs on the same row can be calculated by adding the following parameters:

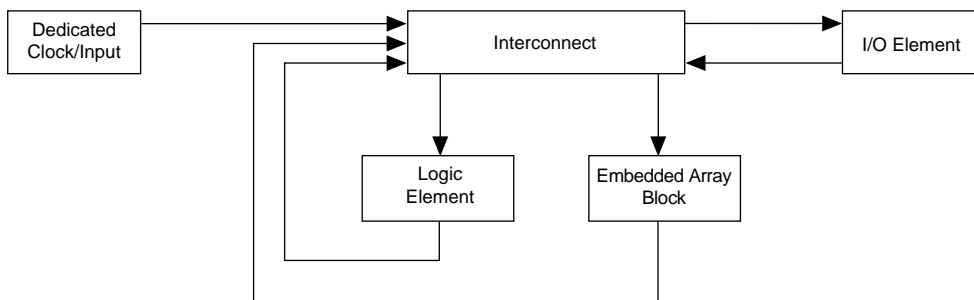
- LE register clock-to-output delay (t_{CO})
- Interconnect delay ($t_{SAMEROW}$)
- LE look-up table delay (t_{LUT})
- LE register setup time (t_{SU})

The routing delay depends on the placement of the source and destination LEs. A more complex registered path may involve multiple combinatorial LEs between the source and destination LEs.

Timing simulation and delay prediction are available with the Altera Simulator and Timing Analyzer, or with industry-standard EDA tools. The Simulator offers both pre-synthesis functional simulation to evaluate logic design accuracy and post-synthesis timing simulation with 0.1-ns resolution. The Timing Analyzer provides point-to-point timing delay information, setup and hold time analysis, and device-wide performance analysis.

Figure 24 shows the overall timing model, which maps the possible paths to and from the various elements of the FLEX 10KE device.

Figure 24. FLEX 10KE Device Timing Model



Figures 25 through 28 show the delays that correspond to various paths and functions within the LE, IOE, EAB, and bidirectional timing models.

Table 35. EPF10K30E Device Interconnect Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		1.8		2.4		2.9	ns
t_{DIN2LE}		1.5		1.8		2.4	ns
$t_{DIN2DATA}$		1.5		1.8		2.2	ns
$t_{DCLK2IOE}$		2.2		2.6		3.0	ns
$t_{DCLK2LE}$		1.5		1.8		2.4	ns
$t_{SAMELAB}$		0.1		0.2		0.3	ns
$t_{SAMEROW}$		2.0		2.4		2.7	ns
$t_{SAMECOLUMN}$		0.7		1.0		0.8	ns
$t_{DIFFROW}$		2.7		3.4		3.5	ns
$t_{TWOROWS}$		4.7		5.8		6.2	ns
$t_{LEPERIPH}$		2.7		3.4		3.8	ns
$t_{LABCARRY}$		0.3		0.4		0.5	ns
$t_{LABCASC}$		0.8		0.8		1.1	ns

Table 36. EPF10K30E External Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DDR}		8.0		9.5		12.5	ns
t_{INSU} (3)	2.1		2.5		3.9		ns
t_{INH} (3)	0.0		0.0		0.0		ns
t_{OUTCO} (3)	2.0	4.9	2.0	5.9	2.0	7.6	ns
t_{INSU} (4)	1.1		1.5		—		ns
t_{INH} (4)	0.0		0.0		—		ns
t_{OUTCO} (4)	0.5	3.9	0.5	4.9	—	—	ns
t_{PCISU}	3.0		4.2		—		ns
t_{PCIH}	0.0		0.0		—		ns
t_{PCICO}	2.0	6.0	2.0	7.5	—	—	ns

Table 38. EPF10K50E Device LE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_H	0.9		1.0		1.4		ns
t_{PRE}		0.5		0.6		0.8	ns
t_{CLR}		0.5		0.6		0.8	ns
t_{CH}	2.0		2.5		3.0		ns
t_{CL}	2.0		2.5		3.0		ns

Table 39. EPF10K50E Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		2.2		2.4		3.3	ns
t_{IOC}		0.3		0.3		0.5	ns
t_{IOCO}		1.0		1.0		1.4	ns
t_{IOCOMB}		0.0		0.0		0.2	ns
t_{IOSU}	1.0		1.2		1.7		ns
t_{IOH}	0.3		0.3		0.5		ns
t_{IOCLR}		0.9		1.0		1.4	ns
t_{OD1}		0.8		0.9		1.2	ns
t_{OD2}		0.3		0.4		0.7	ns
t_{OD3}		3.0		3.5		3.5	ns
t_{XZ}		1.4		1.7		2.3	ns
t_{ZX1}		1.4		1.7		2.3	ns
t_{ZX2}		0.9		1.2		1.8	ns
t_{ZX3}		3.6		4.3		4.6	ns
t_{INREG}		4.9		5.8		7.8	ns
t_{IOFD}		2.8		3.3		4.5	ns
t_{INCOMB}		2.8		3.3		4.5	ns

Table 58. EPF10K130E External Bidirectional Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{\text{INSUBIDIR}}$ (3)	2.2		2.4		3.2		ns
t_{INHBIDIR} (3)	0.0		0.0		0.0		ns
$t_{\text{INSUBIDIR}}$ (4)	2.8		3.0		—		ns
t_{INHBIDIR} (4)	0.0		0.0		—		ns
$t_{\text{OUTCOBIDIR}}$ (3)	2.0	5.0	2.0	7.0	2.0	9.2	ns
t_{XZBIDIR} (3)		5.6		8.1		10.8	ns
t_{XZBIDIR} (3)		5.6		8.1		10.8	ns
$t_{\text{OUTCOBIDIR}}$ (4)	0.5	4.0	0.5	6.0	—	—	ns
t_{XZBIDIR} (4)		4.6		7.1		—	ns
t_{XZBIDIR} (4)		4.6		7.1		—	ns

Notes to tables:

- (1) All timing parameters are described in Tables 24 through 30 in this data sheet.
- (2) These parameters are specified by characterization.
- (3) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
- (4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Tables 59 through 65 show EPF10K200E device internal and external timing parameters.

Table 59. EPF10K200E Device LE Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.7		0.8		1.2	ns
t_{CLUT}		0.4		0.5		0.6	ns
t_{RLUT}		0.6		0.7		0.9	ns
t_{PACKED}		0.3		0.5		0.7	ns
t_{EN}		0.4		0.5		0.6	ns
t_{CICO}		0.2		0.2		0.3	ns
t_{CGEN}		0.4		0.4		0.6	ns
t_{CGENR}		0.2		0.2		0.3	ns
t_{CASC}		0.7		0.8		1.2	ns
t_{C}		0.5		0.6		0.8	ns
t_{CO}		0.5		0.6		0.8	ns
t_{COMB}		0.4		0.6		0.8	ns
t_{SU}	0.4		0.6		0.7		ns

Table 59. EPF10K200E Device LE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_H	0.9		1.1		1.5		ns
t_{PRE}		0.5		0.6		0.8	ns
t_{CLR}		0.5		0.6		0.8	ns
t_{CH}	2.0		2.5		3.0		ns
t_{CL}	2.0		2.5		3.0		ns

Table 60. EPF10K200E Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.6		1.9		2.6	ns
t_{IOC}		0.3		0.3		0.5	ns
t_{IOCO}		1.6		1.9		2.6	ns
t_{IOCOMB}		0.5		0.6		0.8	ns
t_{IOSU}	0.8		0.9		1.2		ns
t_{IOH}	0.7		0.8		1.1		ns
t_{IOCLR}		0.2		0.2		0.3	ns
t_{OD1}		0.6		0.7		0.9	ns
t_{OD2}		0.1		0.2		0.7	ns
t_{OD3}		2.5		3.0		3.9	ns
t_{XZ}		4.4		5.3		7.1	ns
t_{ZX1}		4.4		5.3		7.1	ns
t_{ZX2}		3.9		4.8		6.9	ns
t_{ZX3}		6.3		7.6		10.1	ns
t_{INREG}		4.8		5.7		7.7	ns
$t_{I OFD}$		1.5		1.8		2.4	ns
t_{INCOMB}		1.5		1.8		2.4	ns

Table 76. EPF10K200S Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		3.9		6.4		8.4	ns
$t_{EABRCOMB}$	3.9		6.4		8.4		ns
$t_{EABRCREG}$	3.6		5.7		7.6		ns
t_{EABWP}	2.1		4.0		5.3		ns
$t_{EABWCOMB}$	4.8		8.1		10.7		ns
$t_{EABWCREG}$	5.4		8.0		10.6		ns
t_{EABDD}		3.8		5.1		6.7	ns
$t_{EABDATACO}$		0.8		1.0		1.3	ns
$t_{EABDATASU}$	1.1		1.6		2.1		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	0.7		1.1		1.5		ns
t_{EABWEH}	0.4		0.5		0.6		ns
$t_{EABWDSU}$	1.2		1.8		2.4		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	1.9		3.6		4.7		ns
t_{EABWAH}	0.8		0.5		0.7		ns
t_{EABWO}		3.1		4.4		5.8	ns

Table 77. EPF10K200S Device Interconnect Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		4.4		4.8		5.5	ns
t_{DIN2LE}		0.6		0.6		0.9	ns
$t_{DIN2DATA}$		1.8		2.1		2.8	ns
$t_{DCLK2IOE}$		1.7		2.0		2.8	ns
$t_{DCLK2LE}$		0.6		0.6		0.9	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		3.0		4.6		5.7	ns
$t_{SAMECOLUMN}$		3.5		4.9		6.4	ns
$t_{DIFFROW}$		6.5		9.5		12.1	ns
$t_{TWOROWS}$		9.5		14.1		17.8	ns
$t_{LEPERIPH}$		5.5		6.2		7.2	ns
$t_{LABCARRY}$		0.3		0.1		0.2	ns

Table 77. EPF10K200S Device Interconnect Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{LABCASC}$		0.5		1.0		1.4	ns

Table 78. EPF10K200S External Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		9.0		12.0		16.0	ns
$t_{INSU}^{(2)}$	3.1		3.7		4.7		ns
$t_{INH}^{(2)}$	0.0		0.0		0.0		ns
$t_{OUTCO}^{(2)}$	2.0	3.7	2.0	4.4	2.0	6.3	ns
$t_{INSU}^{(3)}$	2.1		2.7		—		ns
$t_{INH}^{(3)}$	0.0		0.0		—		ns
$t_{OUTCO}^{(3)}$	0.5	2.7	0.5	3.4	—	—	ns
t_{PCISU}	3.0		4.2		—		ns
t_{PCIH}	0.0		0.0		—		ns
t_{PCICO}	2.0	6.0	2.0	8.9	—	—	ns

Table 79. EPF10K200S External Bidirectional Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{INSUBIDIR}^{(2)}$	2.3		3.4		4.4		ns
$t_{INHBIDIR}^{(2)}$	0.0		0.0		0.0		ns
$t_{INSUBIDIR}^{(3)}$	3.3		4.4		—		ns
$t_{INHBIDIR}^{(3)}$	0.0		0.0		—		ns
$t_{OUTCOBIDIR}^{(2)}$	2.0	3.7	2.0	4.4	2.0	6.3	ns
$t_{XZBIDIR}^{(2)}$		6.9		7.6		9.2	ns
$t_{ZXBIDIR}^{(2)}$		5.9		6.6		—	ns
$t_{OUTCOBIDIR}^{(3)}$	0.5	2.7	0.5	3.4	—	—	ns
$t_{XZBIDIR}^{(3)}$		6.9		7.6		9.2	ns
$t_{ZXBIDIR}^{(3)}$		5.9		6.6		—	ns

Notes to tables:

- (1) All timing parameters are described in Tables 24 through 30 in this data sheet.
 (2) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
 (3) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Additionally, the Altera software offers several features that help plan for future device migration by preventing the use of conflicting I/O pins.

Table 81. I/O Counts for FLEX 10KA & FLEX 10KE Devices

FLEX 10KA		FLEX 10KE	
Device	I/O Count	Device	I/O Count
EPF10K30AF256	191	EPF10K30EF256	176
EPF10K30AF484	246	EPF10K30EF484	220
EPF10K50VB356	274	EPF10K50SB356	220
EPF10K50VF484	291	EPF10K50EF484	254
EPF10K50VF484	291	EPF10K50SF484	254
EPF10K100AF484	369	EPF10K100EF484	338

Configuration Schemes

The configuration data for a FLEX 10KE device can be loaded with one of five configuration schemes (see [Table 82](#)), chosen on the basis of the target application. An EPC1, EPC2, or EPC16 configuration device, intelligent controller, or the JTAG port can be used to control the configuration of a FLEX 10KE device, allowing automatic configuration on system power-up.

Multiple FLEX 10KE devices can be configured in any of the five configuration schemes by connecting the configuration enable (\overline{nCE}) and configuration enable output (\overline{nCEO}) pins on each device. Additional FLEX 10K, FLEX 10KA, FLEX 10KE, and FLEX 6000 devices can be configured in the same serial chain.

Table 82. Data Sources for FLEX 10KE Configuration

Configuration Scheme	Data Source
Configuration device	EPC1, EPC2, or EPC16 configuration device
Passive serial (PS)	BitBlaster, ByteBlasterMV, or MasterBlaster download cables, or serial data source
Passive parallel asynchronous (PPA)	Parallel data source
Passive parallel synchronous (PPS)	Parallel data source
JTAG	BitBlaster or ByteBlasterMV download cables, or microprocessor with a Jam STAPL file or JBC file