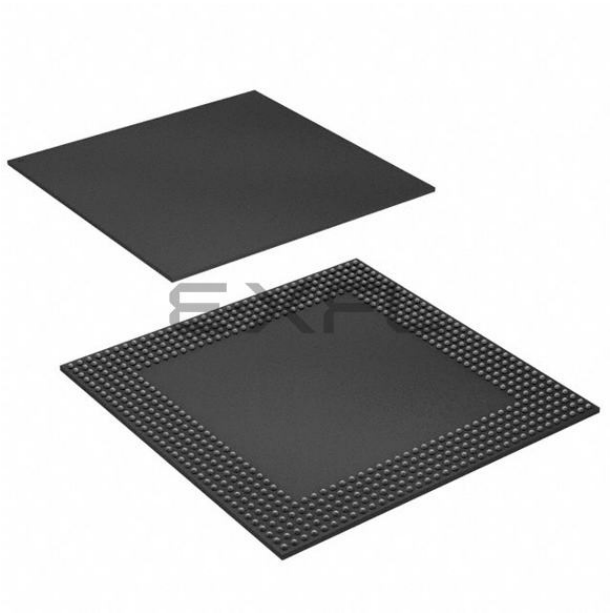

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	1248
Number of Logic Elements/Cells	9984
Total RAM Bits	98304
Number of I/O	470
Number of Gates	513000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	600-BGA
Supplier Device Package	600-BGA (45x45)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k200sbc600-3

Table 2. FLEX 10KE Device Features

Feature	EPF10K100E *(2)*	EPF10K130E	EPF10K200E EPF10K200S
Typical gates *(1)*	100,000	130,000	200,000
Maximum system gates	257,000	342,000	513,000
Logic elements (LEs)	4,992	6,656	9,984
EABs	12	16	24
Total RAM bits	49,152	65,536	98,304
Maximum user I/O pins	338	413	470

Note to tables:

(1) The embedded IEEE Std. 1149.1 JTAG circuitry adds up to 31,250 gates in addition to the listed typical or maximum system gates.

(2) New EPF10K100B designs should use EPF10K100E devices.

…and More Features

- Fabricated on an advanced process and operate with a 2.5-V internal supply voltage
- In-circuit reconfigurability (ICR) via external configuration devices, intelligent controller, or JTAG port
- ClockLock™ and ClockBoost™ options for reduced clock delay/skew and clock multiplication
- Built-in low-skew clock distribution trees
- 100% functional testing of all devices; test vectors or scan chains are not required
- Pull-up on I/O pins before and during configuration

■ Flexible interconnect

- FastTrack® Interconnect continuous routing structure for fast, predictable interconnect delays
- Dedicated carry chain that implements arithmetic functions such as fast adders, counters, and comparators (automatically used by software tools and megafunctions)
- Dedicated cascade chain that implements high-speed, high-fan-in logic functions (automatically used by software tools and megafunctions)
- Tri-state emulation that implements internal tri-state buses
- Up to six global clock signals and four global clear signals

■ Powerful I/O pins

- Individual tri-state output enable control for each pin
- Open-drain option on each I/O pin
- Programmable output slew-rate control to reduce switching noise
- Clamp to V_{CCIO} user-selectable on a pin-by-pin basis
- Supports hot-socketing

Similar to the FLEX 10KE architecture, embedded gate arrays are the fastest-growing segment of the gate array market. As with standard gate arrays, embedded gate arrays implement general logic in a conventional "sea-of-gates" architecture. Additionally, embedded gate arrays have dedicated die areas for implementing large, specialized functions. By embedding functions in silicon, embedded gate arrays reduce die area and increase speed when compared to standard gate arrays. While embedded megafunctions typically cannot be customized, FLEX 10KE devices are programmable, providing the designer with full control over embedded megafunctions and general logic, while facilitating iterative design changes during debugging.

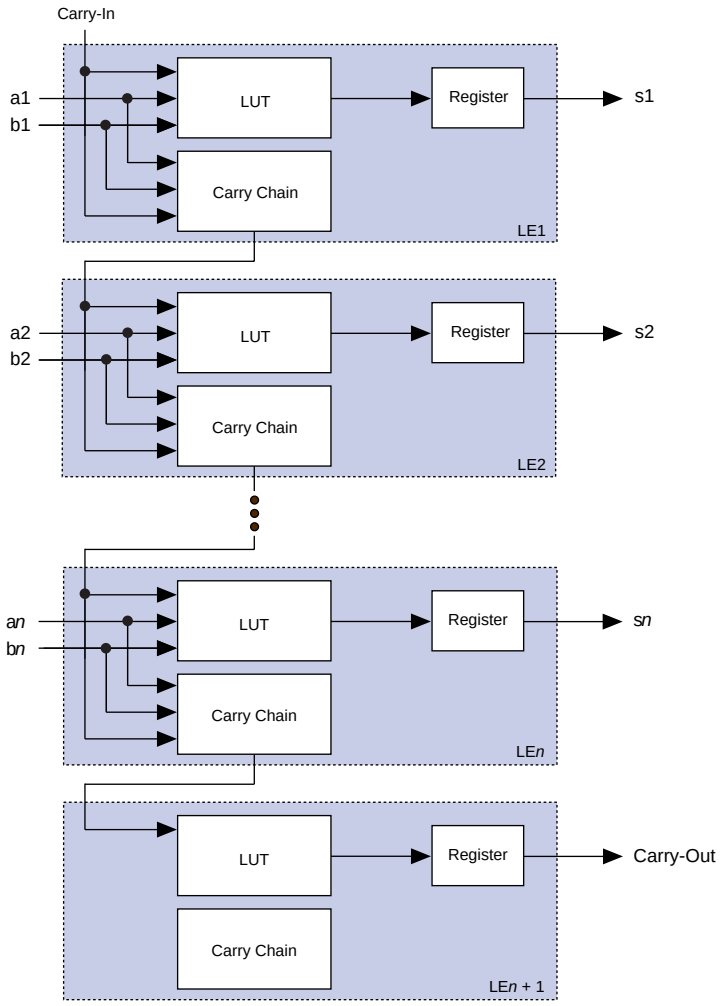
Each FLEX 10KE device contains an embedded array and a logic array. The embedded array is used to implement a variety of memory functions or complex logic functions, such as digital signal processing (DSP), wide data-path manipulation, microcontroller applications, and data-transformation functions. The logic array performs the same function as the sea-of-gates in the gate array and is used to implement general logic such as counters, adders, state machines, and multiplexers. The combination of embedded and logic arrays provides the high performance and high density of embedded gate arrays, enabling designers to implement an entire system on a single device.

FLEX 10KE devices are configured at system power-up with data stored in an Altera serial configuration device or provided by a system controller. Altera offers the EPC1, EPC2, and EPC16 configuration devices, which configure FLEX 10KE devices via a serial data stream. Configuration data can also be downloaded from system RAM or via the Altera BitBlaster™, ByteBlasterMV™, or MasterBlaster download cables. After a FLEX 10KE device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Because reconfiguration requires less than 85 ms, real-time changes can be made during system operation.

FLEX 10KE devices contain an interface that permits microprocessors to configure FLEX 10KE devices serially or in-parallel, and synchronously or asynchronously. The interface also enables microprocessors to treat a FLEX 10KE device as memory and configure it by writing to a virtual memory location, making it easy to reconfigure the device.

Figure 9 shows how an *n*-bit full adder can be implemented in *n* + 1 LEs with the carry chain. One portion of the LUT generates the sum of two bits using the input signals and the carry-in signal; the sum is routed to the output of the LE. The register can be bypassed for simple adders or used for an accumulator function. Another portion of the LUT and the carry chain logic generates the carry-out signal, which is routed directly to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it can be used as a general-purpose signal.

Figure 9. FLEX 10KE Carry Chain Operation (n-Bit Full Adder)



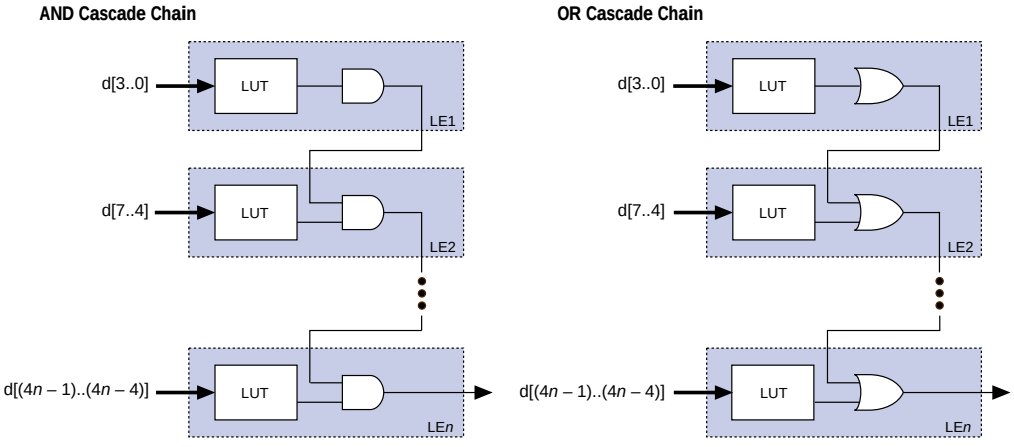
Cascade Chain

With the cascade chain, the FLEX 10KE architecture can implement functions that have a very wide fan-in. Adjacent LUTs can be used to compute portions of the function in parallel; the cascade chain serially connects the intermediate values. The cascade chain can use a logical AND or logical OR (via De Morgan's inversion) to connect the outputs of adjacent LEs. An a delay as low as 0.6 ns per LE, each additional LE provides four more inputs to the effective width of a function. Cascade chain logic can be created automatically by the Altera Compiler during design processing, or manually by the designer during design entry.

Cascade chains longer than eight bits are implemented automatically by linking several LABs together. For easier routing, a long cascade chain skips every other LAB in a row. A cascade chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB (e.g., the last LE of the first LAB in a row cascades to the first LE of the third LAB). The cascade chain does not cross the center of the row (e.g., in the EPF10K50E device, the cascade chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB). This break is due to the EAB's placement in the middle of the row.

Figure 10 shows how the cascade function can connect adjacent LEs to form functions with a wide fan-in. These examples show functions of 4*n* variables implemented with *n* LEs. The LE delay is 0.9 ns; the cascade chain delay is 0.6 ns. With the cascade chain, 2.7 ns are needed to decode a 16-bit address.

Figure 10. FLEX 10KE Cascade Chain Operation



LE Operating Modes

The FLEX 10KE LE can operate in the following four modes:

- Normal mode
- Arithmetic mode
- Up / down counter mode
- Clearable counter mode

Each of these modes uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. Three inputs to the LE provide clock, clear, and preset control for the register. The Altera software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions that use a specific LE operating mode for optimal performance.

The architecture provides a synchronous clock enable to the register in all four modes. The Altera software can set DATA1 to enable the register synchronously, providing easy implementation of fully synchronous designs.

Table 13. ClockLock & ClockBoost Parameters for -2 Speed-Grade Devices

Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_R	Input rise time				5	ns
t_F	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		75	MHz
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		37.5	MHz
f_{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software *(1)*				25,000 *(2)*	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock *(3)*				10	µs
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock *(4)*	$t_{INCLKSTB}$ < 100			250	ps
		$t_{INCLKSTB}$ < 50			200 *(4)*	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Notes to tables:

(1) To implement the ClockLock and ClockBoost circuitry with the MAX+PLUS II software, designers must specify the input frequency. The Altera software tunes the PLL in the ClockLock and ClockBoost circuitry to this frequency. The f_{CLKDEV} parameter specifies how much the incoming clock can differ from the specified frequency during device operation. Simulation does not reflect this parameter.

(2) Twenty-five thousand parts per million (PPM) equates to 2.5% of input clock period.

(3) During device configuration, the ClockLock and ClockBoost circuitry is configured before the rest of the device. If the incoming clock is supplied during configuration, the ClockLock and ClockBoost circuitry locks during configuration because the t_{LOCK} value is less than the time required for configuration.

(4) The t_{JITTER} specification is measured under long-term observation. The maximum value for t_{JITTER} is 200 ps if $t_{INCLKSTB}$ is lower than 50 ps.

I/O Configuration

This section discusses the peripheral component interconnect (PCI) pull-up clamping diode option, slew-rate control, open-drain output option, and MultiVolt I/O interface for FLEX 10KE devices. The PCI pull-up clamping diode, slew-rate control, and open-drain output options are controlled pin-by-pin via Altera software logic options. The MultiVolt I/O interface is controlled by connecting V_{CCIO} to a different voltage than V_{CCINT}. Its effect can be simulated in the Altera software via the **Global Project Device Options** dialog box (Assign menu).

The VCCINT pins must always be connected to a 2.5-V power supply. With a 2.5-V V_{CCINT} level, input voltages are compatible with 2.5-V, 3.3-V, and 5.0-V inputs. The VCCIO pins can be connected to either a 2.5-V or 3.3-V power supply, depending on the output requirements. When the VCCIO pins are connected to a 2.5-V power supply, the output levels are compatible with 2.5-V systems. When the VCCIO pins are connected to a 3.3-V power supply, the output high is at 3.3 V and is therefore compatible with 3.3-V or 5.0-V systems. Devices operating with V_{CCIO} levels higher than 3.0 V achieve a faster timing delay of t_{OD2} instead of t_{OD1}.

Table 14 summarizes FLEX 10KE MultiVolt I/O support.

Table 14. FLEX 10KE MultiVolt I/O Support

V_{CCIO} (V)	Input Signal (V)			Output Signal (V)		
	2.5	3.3	5.0	2.5	3.3	5.0
2.5	✓	✓ *(1)*	✓ *(1)*	✓		
3.3	✓	✓	✓ *(1)*	✓ *(2)*	✓	✓

Notes:

(1) The PCI clamping diode must be disabled to drive an input with voltages higher than V_{CCIO}.

(2) When V_{CCIO} = 3.3 V, a FLEX 10KE device can drive a 2.5-V device that has 3.3-V tolerant inputs.

Open-drain output pins on FLEX 10KE devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a V_{IH} of 3.5 V. When the open-drain pin is active, it will drive low. When the pin is inactive, the trace will be pulled up to 5.0 V by the resistor. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The I_{OL} current specification should be considered when selecting a pull-up resistor.

Power Sequencing & Hot-Socketing

Because FLEX 10KE devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The V_{CCIO} and V_{CCINT} power planes can be powered in any order.

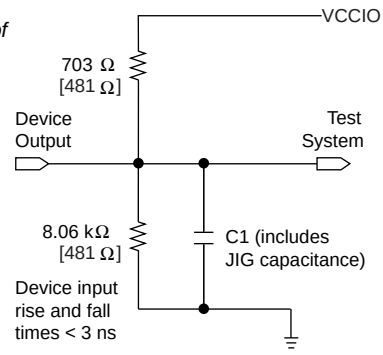
Signals can be driven into FLEX 10KE devices before and during power up without damaging the device. Additionally, FLEX 10KE devices do not drive out during power up. Once operating conditions are reached, FLEX 10KE devices operate as specified by the user.

Generic Testing

Each FLEX 10KE device is functionally tested. Complete testing of each configurable static random access memory (SRAM) bit and all logic functionality ensures 100% yield. AC test measurements for FLEX 10KE devices are made under conditions equivalent to those shown in Figure 21. Multiple test patterns can be used to configure devices during all stages of the production flow.

Figure 21. FLEX 10KE AC Test Conditions

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers in brackets are for 2.5-V devices or outputs. Numbers without brackets are for 3.3-V. devices or outputs.



Operating Conditions

Tables 19 through 23 provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for 2.5-V FLEX 10KE devices.

Table 19. FLEX 10KE 2.5-V Device Absolute Maximum Ratings *Note (1)*

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CCINT}	Supply voltage	With respect to ground *(2)*	–0.5	3.6	V
V_{CCIO}			–0.5	4.6	V
V_I	DC input voltage		–2.0	5.75	V
I_{OUT}	DC output current, per pin		–25	25	mA
T_{STG}	Storage temperature	No bias	–65	150	° C
T_{AMB}	Ambient temperature	Under bias	–65	135	° C
T_J	Junction temperature	PQFP, TQFP, BGA, and FineLine BGA packages, under bias		135	° C
		Ceramic PGA packages, under bias		150	° C

Table 23. FLEX 10KE Device Capacitance *Note (14)*

Symbol	Parameter	Conditions	Min	Max	Unit
C_{IN}	Input capacitance	V_{IN} = 0 V, f = 1.0 MHz		10	pF
C_{INCLK}	Input capacitance on dedicated clock pin	V_{IN} = 0 V, f = 1.0 MHz		12	pF
C_{OUT}	Output capacitance	V_{OUT} = 0 V, f = 1.0 MHz		10	pF

Notes to tables:

(1) See the *Operating Requirements for Altera Devices Data Sheet*.
(2) Minimum DC input voltage is –0.5 V. During transitions, the inputs may undershoot to –2.0 V for input currents less than 100 mA and periods shorter than 20 ns.
(3) Numbers in parentheses are for industrial-temperature-range devices.
(4) Maximum V_{CC} rise time is 100 ms, and V_{CC} must rise monotonically.
(5) All pins, including dedicated inputs, clock, I/O, and JTAG pins, may be driven before V_{CCINT} and V_{CCIO} are powered.
(6) Typical values are for T_A = 25° C, V_{CCINT} = 2.5 V, and V_{CCIO} = 2.5 V or 3.3 V.
(7) These values are specified under the FLEX 10KE Recommended Operating Conditions shown in Tables 20 and 21.
(8) The FLEX 10KE input buffers are compatible with 2.5-V, 3.3-V (LVTTL and LVCMOS), and 5.0-V TTL and CMOS signals. Additionally, the input buffers are 3.3-V PCI compliant when V_{CCIO} and V_{CCINT} meet the relationship shown in Figure 22.
(9) The I_{OH} parameter refers to high-level TTL, PCI, or CMOS output current.
(10) The I_{OL} parameter refers to low-level TTL, PCI, or CMOS output current. This parameter applies to open-drain pins as well as output pins.
(11) This value is specified for normal device operation. The value may vary during power-up.
(12) This parameter applies to -1 speed-grade commercial-temperature devices and -2 speed-grade-industrial temperature devices.
(13) Pin pull-up resistance values will be lower if the pin is driven higher than V_{CCIO} by an external source.
(14) Capacitance is sample-tested only.

Figure 22 shows the required relationship between V_{CCIO} and V_{CCINT} for 3.3-V PCI compliance.

Figure 22. Relationship between V_{CCIO} & V_{CCINT} for 3.3-V PCI Compliance

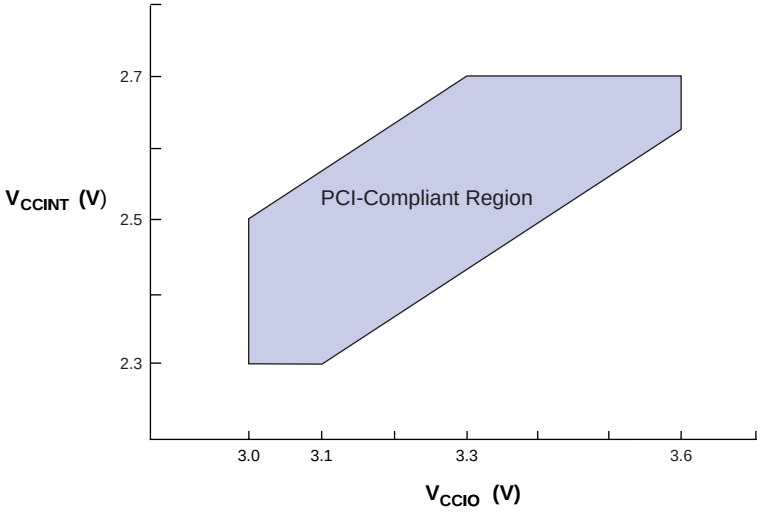


Figure 23 shows the typical output drive characteristics of FLEX 10KE devices with 3.3-V and 2.5-V V_{CCIO}. The output driver is compliant to the 3.3-V ***PCI Local Bus Specification, Revision 2.2*** (when VCCIO pins are connected to 3.3 V). FLEX 10KE devices with a -1 speed grade also comply with the drive strength requirements of the ***PCI Local Bus Specification, Revision 2.2*** (when VCCINT pins are powered with a minimum supply of 2.375 V, and VCCIO pins are connected to 3.3 V). Therefore, these devices can be used in open 5.0-V PCI systems.

Figure 26. FLEX 10KE Device IOE Timing Model

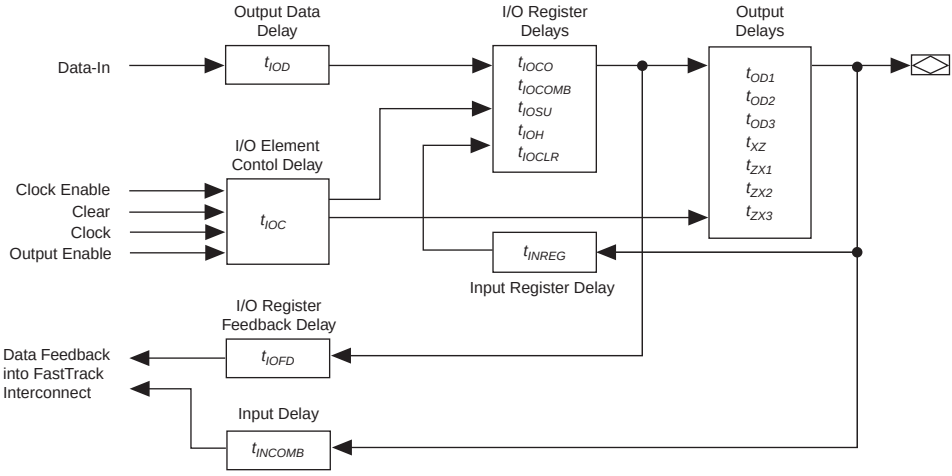


Figure 27. FLEX 10KE Device EAB Timing Model

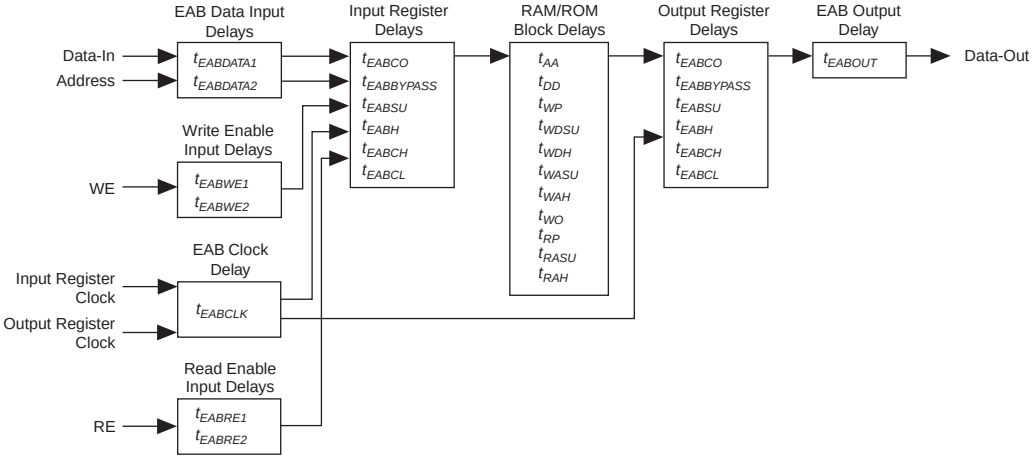


Table 26. EAB Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
$t_{EABDATA1}$	Data or address delay to EAB for combinatorial input	
$t_{EABDATA2}$	Data or address delay to EAB for registered input	
t_{EABWE1}	Write enable delay to EAB for combinatorial input	
t_{EABWE2}	Write enable delay to EAB for registered input	
t_{EABRE1}	Read enable delay to EAB for combinatorial input	
t_{EABRE2}	Read enable delay to EAB for registered input	
t_{EABCLK}	EAB register clock delay	
t_{EABCO}	EAB register clock-to-output delay	
$t_{EABBYPASS}$	Bypass register delay	
t_{EABSU}	EAB register setup time before clock	
t_{EABH}	EAB register hold time after clock	
t_{EABCLR}	EAB register asynchronous clear time to output delay	
t_{AA}	Address access delay (including the read enable to output delay)	
t_{WP}	Write pulse width	
t_{RP}	Read pulse width	
t_{WDSU}	Data setup time before falling edge of write pulse	*(5)*
t_{WDH}	Data hold time after falling edge of write pulse	*(5)*
t_{WASU}	Address setup time before rising edge of write pulse	*(5)*
t_{WAH}	Address hold time after falling edge of write pulse	*(5)*
t_{RASU}	Address setup time with respect to the falling edge of the read enable	
t_{RAH}	Address hold time with respect to the falling edge of the read enable	
t_{WO}	Write enable to data output valid delay	
t_{DD}	Data-in to data-out valid delay	
t_{EABOUT}	Data-out delay	
t_{EABCH}	Clock high time	
t_{EABCL}	Clock low time	

Table 37. EPF10K30E External Bidirectional Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{INSUBIDIR}$ (3)	2.8		3.9		5.2		ns
$t_{INHBIDIR}$ (3)	0.0		0.0		0.0		ns
$t_{INSUBIDIR}$ (4)	3.8		4.9		–		ns
$t_{INHBIDIR}$ (4)	0.0		0.0		–		ns
$t_{OUTCOBIDIR}$ (3)	2.0	4.9	2.0	5.9	2.0	7.6	ns
$t_{XZBIDIR}$ (3)		6.1		7.5		9.7	ns
$t_{ZXBIDIR}$ (3)		6.1		7.5		9.7	ns
$t_{OUTCOBIDIR}$ (4)	0.5	3.9	0.5	4.9	–	–	ns
$t_{XZBIDIR}$ (4)		5.1		6.5		–	ns
$t_{ZXBIDIR}$ (4)		5.1		6.5		–	ns

Notes to tables:

(1) All timing parameters are described in Tables 24 through 30 in this data sheet.
(2) These parameters are specified by characterization.
(3) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
(4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Tables 38 through 44 show EPF10K50E device internal and external timing parameters.

Table 38. EPF10K50E Device LE Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.6		0.9		1.3	ns
t_{CLUT}		0.5		0.6		0.8	ns
t_{RLUT}		0.7		0.8		1.1	ns
t_{PACKED}		0.4		0.5		0.6	ns
t_{EN}		0.6		0.7		0.9	ns
t_{CICO}		0.2		0.2		0.3	ns
t_{CGEN}		0.5		0.5		0.8	ns
t_{CGENR}		0.2		0.2		0.3	ns
t_{CASC}		0.8		1.0		1.4	ns
t_{C}		0.5		0.6		0.8	ns
t_{CO}		0.7		0.7		0.9	ns
t_{COMB}		0.5		0.6		0.8	ns
t_{SU}	0.7		0.7		0.8		ns

Tables 52 through 58 show EPF10K130E device internal and external timing parameters.

Table 52. EPF10K130E Device LE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.6		0.9		1.3	ns
t_{CLUT}		0.6		0.8		1.0	ns
t_{RLUT}		0.7		0.9		0.2	ns
t_{PACKED}		0.3		0.5		0.6	ns
t_{EN}		0.2		0.3		0.4	ns
t_{CICO}		0.1		0.1		0.2	ns
t_{CGEN}		0.4		0.6		0.8	ns
t_{CGENR}		0.1		0.1		0.2	ns
t_{CASC}		0.6		0.9		1.2	ns
t_{C}		0.3		0.5		0.6	ns
t_{CO}		0.5		0.7		0.8	ns
t_{COMB}		0.3		0.5		0.6	ns
t_{SU}	0.5		0.7		0.8		ns
t_{H}	0.6		0.7		1.0		ns
t_{PRE}		0.9		1.2		1.6	ns
t_{CLR}		0.9		1.2		1.6	ns
t_{CH}	1.5		1.5		2.5		ns
t_{CL}	1.5		1.5		2.5		ns

Table 53. EPF10K130E Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.3		1.5		2.0	ns
t_{IOC}		0.0		0.0		0.0	ns
t_{IOCO}		0.6		0.8		1.0	ns
t_{IOCOMB}		0.6		0.8		1.0	ns
t_{IOSU}	1.0		1.2		1.6		ns
t_{IOH}	0.9		0.9		1.4		ns
t_{IOCLR}		0.6		0.8		1.0	ns
t_{OD1}		2.8		4.1		5.5	ns
t_{OD2}		2.8		4.1		5.5	ns

Table 54. EPF10K130E Device EAB Internal Microparameters *(Part 2 of 2)* *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DD}		1.5		2.0		2.6	ns
t_{EABOUT}		0.2		0.3		0.3	ns
t_{EABCH}	1.5		2.0		2.5		ns
t_{EABCL}	2.7		3.5		4.7		ns

Table 55. EPF10K130E Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		5.9		7.5		9.9	ns
$t_{EABRCOMB}$	5.9		7.5		9.9		ns
$t_{EABRCREG}$	5.1		6.4		8.5		ns
t_{EABWP}	2.7		3.5		4.7		ns
$t_{EABWCOMB}$	5.9		7.7		10.3		ns
$t_{EABWCREG}$	5.4		7.0		9.4		ns
t_{EABDD}		3.4		4.5		5.9	ns
$t_{EABDATACO}$		0.5		0.7		0.8	ns
$t_{EABDATASU}$	0.8		1.0		1.4		ns
$t_{EABDATAH}$	0.1		0.1		0.2		ns
$t_{EABWESU}$	1.1		1.4		1.9		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	1.0		1.3		1.7		ns
t_{EABWDH}	0.2		0.2		0.3		ns
$t_{EABWASU}$	4.1		5.1		6.8		ns
t_{EABWAH}	0.0		0.0		0.0		ns
t_{EABWO}		3.4		4.5		5.9	ns

Table 56. EPF10K130E Device Interconnect Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		2.8		3.5		4.4	ns
t_{DIN2LE}		0.7		1.2		1.6	ns
$t_{DIN2DATA}$		1.6		1.9		2.2	ns
$t_{DCLK2IOE}$		1.6		2.1		2.7	ns
$t_{DCLK2LE}$		0.7		1.2		1.6	ns
$t_{SAMELAB}$		0.1		0.2		0.2	ns
$t_{SAMEROW}$		1.9		3.4		5.1	ns
$t_{SAMECOLUMN}$		0.9		2.6		4.4	ns
$t_{DIFFROW}$		2.8		6.0		9.5	ns
$t_{TWOROWS}$		4.7		9.4		14.6	ns
$t_{LEPERIPH}$		3.1		4.7		6.9	ns
$t_{LABCARRY}$		0.6		0.8		1.0	ns
$t_{LABCASC}$		0.9		1.2		1.6	ns

Table 57. EPF10K130E External Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		9.0		12.0		16.0	ns
t_{INSU} *(3)*	1.9		2.1		3.0		ns
t_{INH} *(3)*	0.0		0.0		0.0		ns
t_{OUTCO} *(3)*	2.0	5.0	2.0	7.0	2.0	9.2	ns
t_{INSU} *(4)*	0.9		1.1		–		ns
t_{INH} *(4)*	0.0		0.0		–		ns
t_{OUTCO} *(4)*	0.5	4.0	0.5	6.0	–	–	ns
t_{PCISU}	3.0		6.2		–		ns
t_{PCIH}	0.0		0.0		–		ns
t_{PCICO}	2.0	6.0	2.0	6.9	–	–	ns

Table 74. EPF10K200S Device IOE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{ZX2}		4.5		4.8		6.6	ns
t_{ZX3}		6.6		7.6		10.1	ns
t_{INREG}		3.7		5.7		7.7	ns
t_{IOFD}		1.8		3.4		4.0	ns
t_{INCOMB}		1.8		3.4		4.0	ns

Table 75. EPF10K200S Device EAB Internal Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.8		2.4		3.2	ns
$t_{EABDATA1}$		0.4		0.5		0.6	ns
t_{EABWE1}		1.1		1.7		2.3	ns
t_{EABWE2}		0.0		0.0		0.0	ns
t_{EABRE1}		0		0		0	ns
t_{EABRE2}		0.4		0.5		0.6	ns
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		0.8		0.9		1.2	ns
$t_{EABBYPASS}$		0.0		0.1		0.1	ns
t_{EABSU}	0.7		1.1		1.5		ns
t_{EABH}	0.4		0.5		0.6		ns
t_{EABCLR}	0.8		0.9		1.2		ns
t_{AA}		2.1		3.7		4.9	ns
t_{WP}	2.1		4.0		5.3		ns
t_{RP}	1.1		1.1		1.5		ns
t_{WDSU}	0.5		1.1		1.5		ns
t_{WDH}	0.1		0.1		0.1		ns
t_{WASU}	1.1		1.6		2.1		ns
t_{WAH}	1.6		2.5		3.3		ns
t_{RASU}	1.6		2.6		3.5		ns
t_{RAH}	0.1		0.1		0.2		ns
t_{WO}		2.0		2.4		3.2	ns
t_{DD}		2.0		2.4		3.2	ns
t_{EABOUT}		0.0		0.1		0.1	ns
t_{EABCH}	1.5		2.0		2.5		ns
t_{EABCL}	2.1		2.8		3.8		ns

Power Consumption

The supply power (P) for FLEX 10KE devices can be calculated with the following equation:

$$P = P_{INT} + P_{IO} = (I_{CCSTANDBY} + I_{CCACTIVE}) \times V_{CC} + P_{IO}$$

The $I_{CCACTIVE}$ value depends on the switching frequency and the application logic. This value is calculated based on the amount of current that each LE typically consumes. The P_{IO} value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

Compared to the rest of the device, the embedded array consumes a negligible amount of power. Therefore, the embedded array can be ignored when calculating supply current.

The $I_{CCACTIVE}$ value can be calculated with the following equation:

$$I_{CCACTIVE} = K \times \mathbf{f_{MAX}} \times N \times \mathbf{tog_{LC}} \times \frac{\mu A}{MHz \times LE}$$

Where:

$\mathbf{f_{MAX}}$ = Maximum operating frequency in MHz
N = Total number of LEs used in the device
$\mathbf{tog_{LC}}$ = Average percent of LEs toggling at each clock (typically 12.5%)
K = Constant

Table 80 provides the constant (K) values for FLEX 10KE devices.

Table 80. FLEX 10KE K Constant Values

Device	K Value
EPF10K30E	4.5
EPF10K50E	4.8
EPF10K50S	4.5
EPF10K100E	4.5
EPF10K130E	4.6
EPF10K200E	4.8
EPF10K200S	4.6

This calculation provides an I_{CC} estimate based on typical conditions with no output load. The actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

To better reflect actual designs, the power model (and the constant K in the power calculation equations) for continuous interconnect FLEX devices assumes that LEs drive FastTrack Interconnect channels. In contrast, the power model of segmented FPGAs assumes that all LEs drive only one short interconnect segment. This assumption may lead to inaccurate results when compared to measured power consumption for actual designs in segmented FPGAs.

Figure 31 shows the relationship between the current and operating frequency of FLEX 10KE devices.

Figure 31. FLEX 10KE $I_{CCACTIVE}$ vs. Operating Frequency (Part 1 of 2)

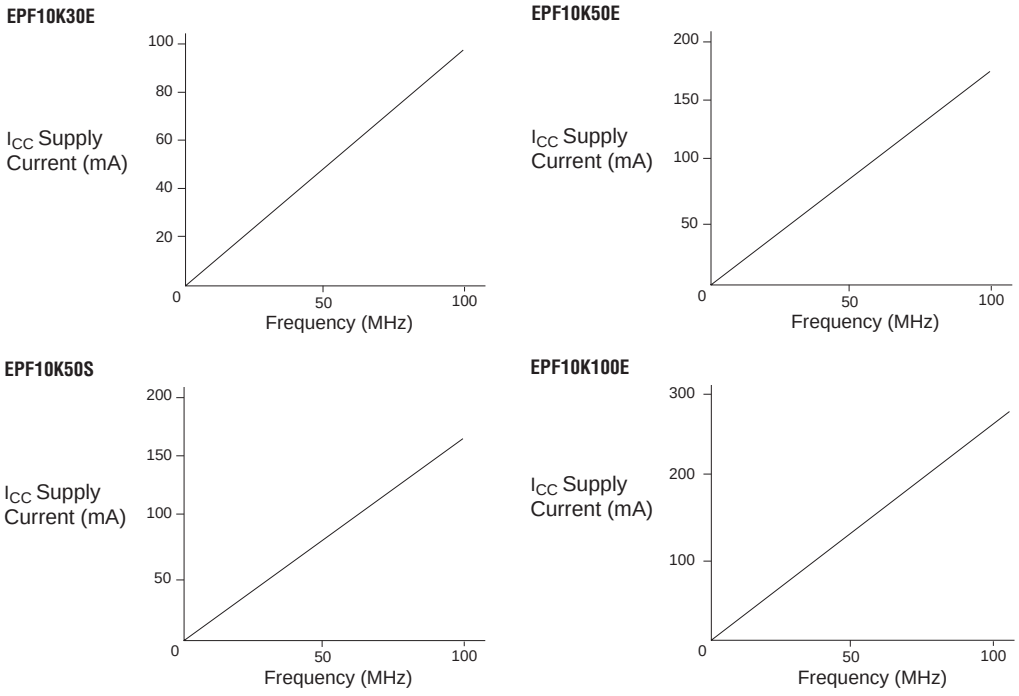
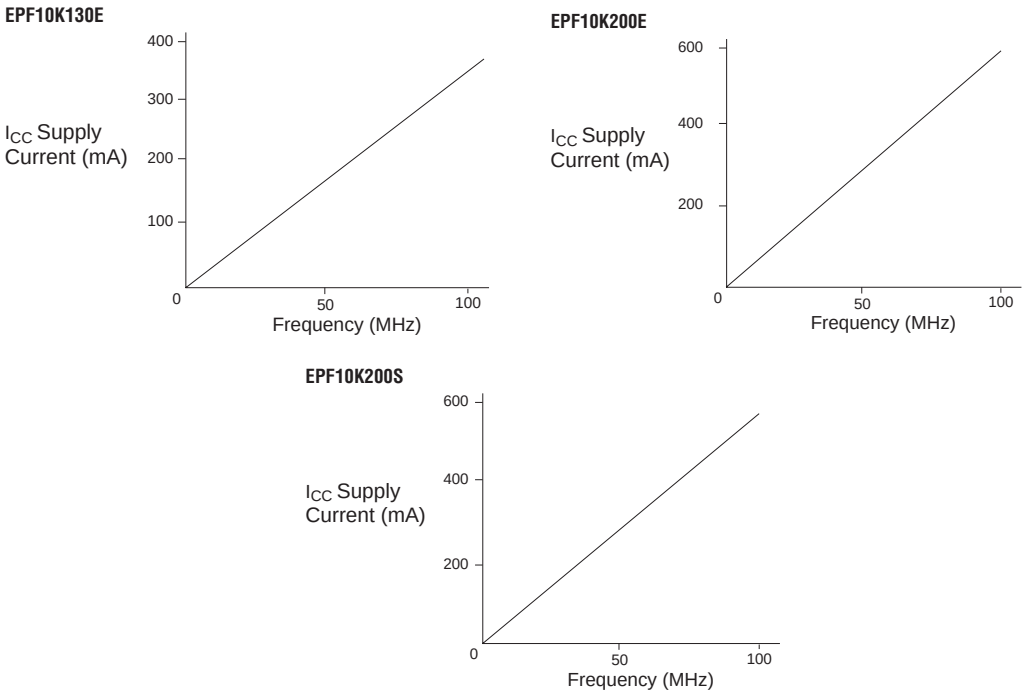


Figure 31. FLEX 10KE $I_{CCACTIVE}$ vs. Operating Frequency (Part 2 of 2)



Configuration & Operation

The FLEX 10KE architecture supports several configuration schemes. This section summarizes the device operating modes and available device configuration schemes.

Operating Modes

The FLEX 10KE architecture uses SRAM configuration elements that require configuration data to be loaded every time the circuit powers up. The process of physically loading the SRAM data into the device is called *configuration*. Before configuration, as V_{CC} rises, the device initiates a Power-On Reset (POR). This POR event clears the device and prepares it for configuration. The FLEX 10KE POR time does not exceed 50 µs.

When configuring with a configuration device, refer to the respective configuration device data sheet for POR timing information.