Intel - EPF10K200SRC240-1N Datasheet

Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	1248
Number of Logic Elements/Cells	9984
Total RAM Bits	98304
Number of I/O	182
Number of Gates	513000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	240-BFQFP Exposed Pad
Supplier Device Package	240-RQFP (32x32)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k200src240-1n

Table 4. FLEX 10KE Package Sizes

Device	144-Pin TQFP	208-Pin PQFP	240-Pin PQFP RQFP	256-Pin FineLine BGA	356-Pin BGA	484-Pin FineLine BGA	599-Pin PGA	600-Pin BGA	672-Pin FineLine BGA
Pitch (mm)	0.50	0.50	0.50	1.0	1.27	1.0	–	1.27	1.0
Area (mm^2)	484	936	1,197	289	1,225	529	3,904	2,025	729
Length × width (mm × mm)	22 × 22	30.6 × 30.6	34.6 × 34.6	17 × 17	35 × 35	23 × 23	62.5 × 62.5	45 × 45	27 × 27

General Description

Altera FLEX 10KE devices are enhanced versions of FLEX 10K devices. Based on reconfigurable CMOS SRAM elements, the FLEX architecture incorporates all features necessary to implement common gate array megafunctions. With up to 200,000 typical gates, FLEX 10KE devices provide the density, speed, and features to integrate entire systems, including multiple 32-bit buses, into a single device.

The ability to reconfigure FLEX 10KE devices enables 100% testing prior to shipment and allows the designer to focus on simulation and design verification. FLEX 10KE reconfigurability eliminates inventory management for gate array designs and generation of test vectors for fault coverage.

Table 5 shows FLEX 10KE performance for some common designs. All performance values were obtained with Synopsys DesignWare or LPM functions. Special design techniques are not required to implement the applications; the designer simply infers or instantiates a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Similar to the FLEX 10KE architecture, embedded gate arrays are the fastest-growing segment of the gate array market. As with standard gate arrays, embedded gate arrays implement general logic in a conventional "sea-of-gates" architecture. Additionally, embedded gate arrays have dedicated die areas for implementing large, specialized functions. By embedding functions in silicon, embedded gate arrays reduce die area and increase speed when compared to standard gate arrays. While embedded megafunctions typically cannot be customized, FLEX 10KE devices are programmable, providing the designer with full control over embedded megafunctions and general logic, while facilitating iterative design changes during debugging.

Each FLEX 10KE device contains an embedded array and a logic array. The embedded array is used to implement a variety of memory functions or complex logic functions, such as digital signal processing (DSP), wide data-path manipulation, microcontroller applications, and data-transformation functions. The logic array performs the same function as the sea-of-gates in the gate array and is used to implement general logic such as counters, adders, state machines, and multiplexers. The combination of embedded and logic arrays provides the high performance and high density of embedded gate arrays, enabling designers to implement an entire system on a single device.

FLEX 10KE devices are configured at system power-up with data stored in an Altera serial configuration device or provided by a system controller. Altera offers the EPC1, EPC2, and EPC16 configuration devices, which configure FLEX 10KE devices via a serial data stream. Configuration data can also be downloaded from system RAM or via the Altera BitBlaster™, ByteBlasterMV™, or MasterBlaster download cables. After a FLEX 10KE device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Because reconfiguration requires less than 85 ms, real-time changes can be made during system operation.

FLEX 10KE devices contain an interface that permits microprocessors to configure FLEX 10KE devices serially or in-parallel, and synchronously or asynchronously. The interface also enables microprocessors to treat a FLEX 10KE device as memory and configure it by writing to a virtual memory location, making it easy to reconfigure the device.

For more information on FLEX device configuration, see the following documents:

- *Configuration Devices for APEX & FLEX Devices Data Sheet*
- *BitBlaster Serial Download Cable Data Sheet*
- *ByteBlasterMV Parallel Port Download Cable Data Sheet*
- *MasterBlaster Download Cable Data Sheet*
- *Application Note 116 (Configuring APEX 20K, FLEX 10K, & FLEX 6000 Devices)*

FLEX 10KE devices are supported by the Altera development systems, which are integrated packages that offer schematic, text (including AHDL), and waveform design entry, compilation and logic synthesis, full simulation and worst-case timing analysis, and device configuration. The Altera software provides EDIF 2 0 0 and 3 0 0, LPM, VHDL, Verilog HDL, and other interfaces for additional design entry and simulation support from other industry-standard PC- and UNIX workstation-based EDA tools.

The Altera software works easily with common gate array EDA tools for synthesis and simulation. For example, the Altera software can generate Verilog HDL files for simulation with tools such as Cadence Verilog-XL. Additionally, the Altera software contains EDA libraries that use device-specific features such as carry chains, which are used for fast counter and arithmetic functions. For instance, the Synopsys Design Compiler library supplied with the Altera development system includes DesignWare functions that are optimized for the FLEX 10KE architecture.

The Altera development system runs on Windows-based PCs and Sun SPARCstation, and HP 9000 Series 700/800.

See the *MAX+PLUS II Programmable Logic Development System & Software Data Sheet* and the *Quartus Programmable Logic Development System & Software Data Sheet* for more information.

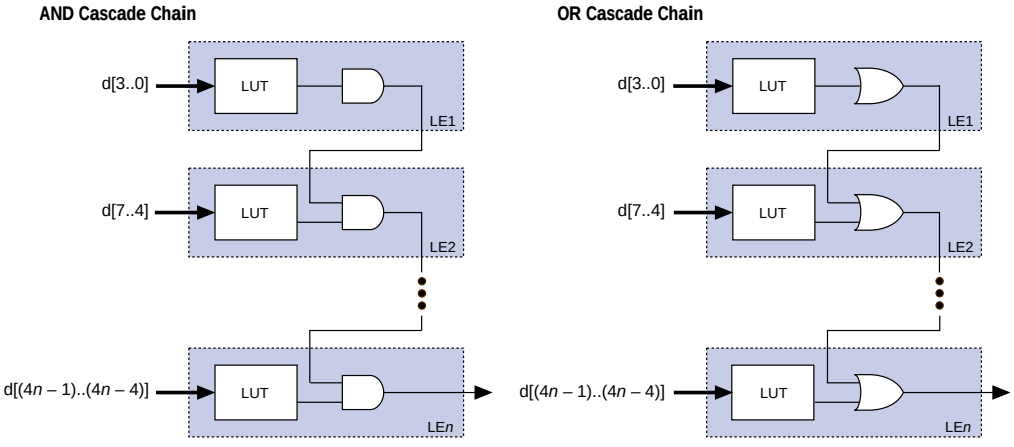
Cascade Chain

With the cascade chain, the FLEX 10KE architecture can implement functions that have a very wide fan-in. Adjacent LUTs can be used to compute portions of the function in parallel; the cascade chain serially connects the intermediate values. The cascade chain can use a logical AND or logical OR (via De Morgan's inversion) to connect the outputs of adjacent LEs. An a delay as low as 0.6 ns per LE, each additional LE provides four more inputs to the effective width of a function. Cascade chain logic can be created automatically by the Altera Compiler during design processing, or manually by the designer during design entry.

Cascade chains longer than eight bits are implemented automatically by linking several LABs together. For easier routing, a long cascade chain skips every other LAB in a row. A cascade chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB (e.g., the last LE of the first LAB in a row cascades to the first LE of the third LAB). The cascade chain does not cross the center of the row (e.g., in the EPF10K50E device, the cascade chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB). This break is due to the EAB's placement in the middle of the row.

Figure 10 shows how the cascade function can connect adjacent LEs to form functions with a wide fan-in. These examples show functions of 4*n* variables implemented with *n* LEs. The LE delay is 0.9 ns; the cascade chain delay is 0.6 ns. With the cascade chain, 2.7 ns are needed to decode a 16-bit address.

Figure 10. FLEX 10KE Cascade Chain Operation



Normal Mode

The normal mode is suitable for general logic applications and wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a four-input LUT. The Altera Compiler automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. Either the register or the LUT can be used to drive both the local interconnect and the FastTrack Interconnect routing structure at the same time.

The LUT and the register in the LE can be used independently (register packing). To support register packing, the LE has two outputs; one drives the local interconnect, and the other drives the FastTrack Interconnect routing structure. The DATA4 signal can drive the register directly, allowing the LUT to compute a function that is independent of the registered signal; a three-input function can be computed in the LUT, and a fourth independent signal can be registered. Alternatively, a four-input function can be generated, and one of the inputs to this function can be used to drive the register. The register in a packed LE can still use the clock enable, clear, and preset signals in the LE. In a packed LE, the register can drive the FastTrack Interconnect routing structure while the LUT drives the local interconnect, or vice versa.

Arithmetic Mode

The arithmetic mode offers 2 three-input LUTs that are ideal for implementing adders, accumulators, and comparators. One LUT computes a three-input function; the other generates a carry output. As shown in Figure 11 on page 22, the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, in an adder, this output is the sum of three signals: a, b, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

Up/Down Counter Mode

The up / down counter mode offers counter enable, clock enable, synchronous up / down control, and data loading options. These control signals are generated by the data inputs from the LAB local interconnect, the carry-in signal, and output feedback from the programmable register. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading. Data can also be loaded asynchronously with the clear and preset register control signals without using the LUT resources.

Clearable Counter Mode

The clearable counter mode is similar to the up / down counter mode, but supports a synchronous clear instead of the up / down control. The clear function is substituted for the cascade-in signal in the up / down counter mode. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is ANDed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-states without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

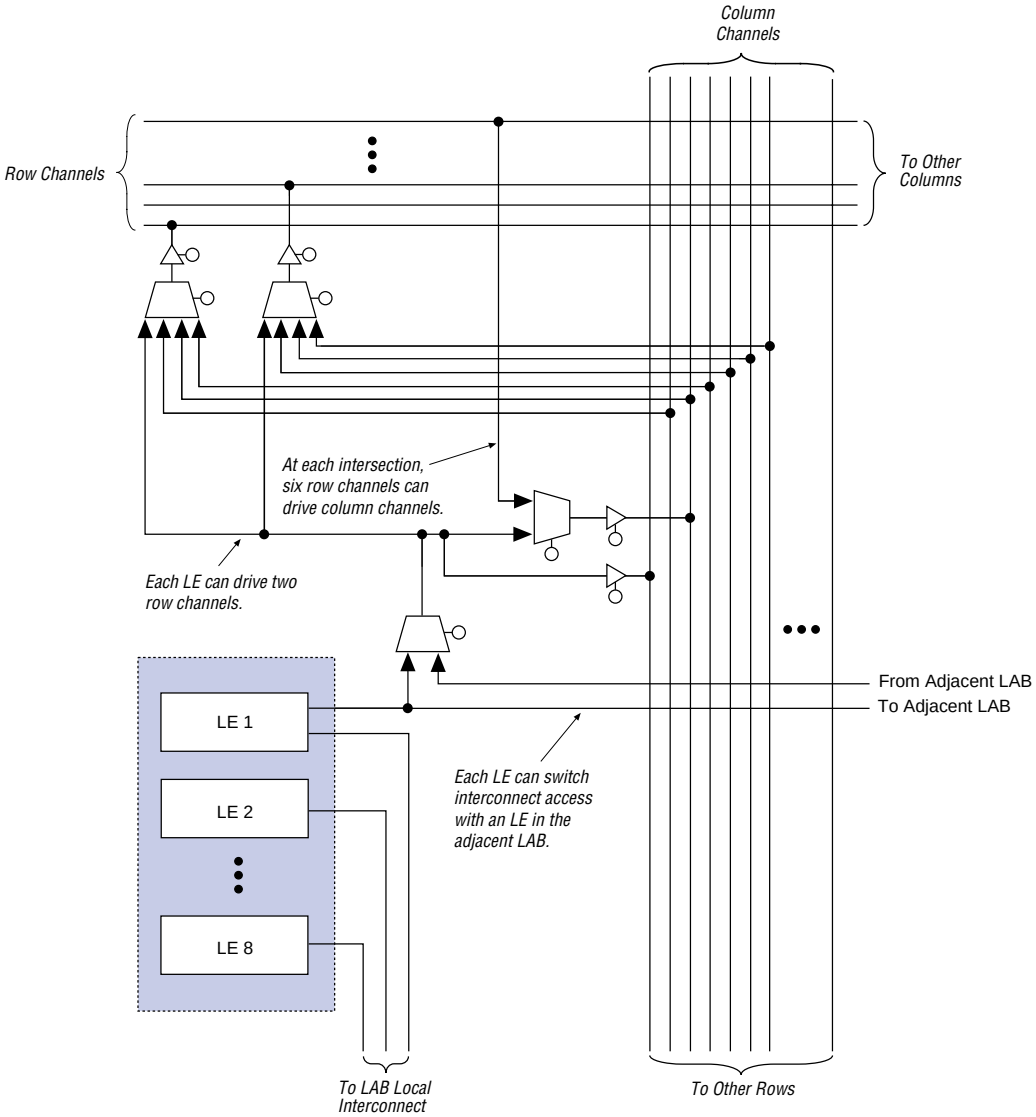
Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the Altera Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

Figure 13. FLEX 10KE LAB Connections to Row & Column Interconnect

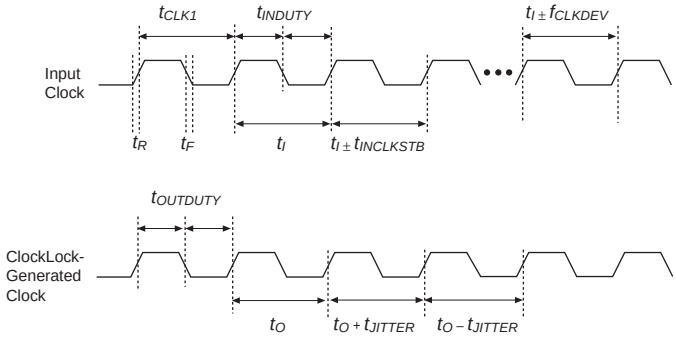


ClockLock & ClockBoost Timing Parameters

For the ClockLock and ClockBoost circuitry to function properly, the incoming clock must meet certain requirements. If these specifications are not met, the circuitry may not lock onto the incoming clock, which generates an erroneous clock within the device. The clock generated by the ClockLock and ClockBoost circuitry must also meet certain specifications. If the incoming clock meets these requirements during configuration, the ClockLock and ClockBoost circuitry will lock onto the clock during configuration. The circuit will be ready for use immediately after configuration. Figure 19 shows the incoming and generated clock specifications.

Figure 19. Specifications for Incoming & Generated Clocks

The t_I parameter refers to the nominal input clock period; the t_O parameter refers to the nominal output clock period.



Tables 12 and 13 summarize the ClockLock and ClockBoost parameters for -1 and -2 speed-grade devices, respectively.

Table 12. ClockLock & ClockBoost Parameters for -1 Speed-Grade Devices

Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_R	Input rise time				5	ns
t_F	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		180	MHz
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		90	MHz
f_{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software *(1)*				25,000 *(2)*	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock *(3)*				10	µs
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock *(4)*	$t_{INCLKSTB}$ < 100			250	ps
		$t_{INCLKSTB}$ < 50			200 *(4)*	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Table 22. FLEX 10KE 2.5-V Device DC Operating Conditions *Notes (6), (7)*

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{IH}	High-level input voltage		1.7, 0.5 × V_{CCIO} *(8)*		5.75	V
V_{IL}	Low-level input voltage		–0.5		0.8, 0.3 × V_{CCIO} *(8)*	V
V_{OH}	3.3-V high-level TTL output voltage	I_{OH} = –8 mA DC, V_{CCIO} = 3.00 V *(9)*	2.4			V
	3.3-V high-level CMOS output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 3.00 V *(9)*	V_{CCIO} – 0.2			V
	3.3-V high-level PCI output voltage	I_{OH} = –0.5 mA DC, V_{CCIO} = 3.00 to 3.60 V *(9)*	0.9 × V_{CCIO}			V
	2.5-V high-level output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 2.30 V *(9)*	2.1			V
		I_{OH} = –1 mA DC, V_{CCIO} = 2.30 V *(9)*	2.0			V
		I_{OH} = –2 mA DC, V_{CCIO} = 2.30 V *(9)*	1.7			V
V_{OL}	3.3-V low-level TTL output voltage	I_{OL} = 12 mA DC, V_{CCIO} = 3.00 V *(10)*			0.45	V
	3.3-V low-level CMOS output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 3.00 V *(10)*			0.2	V
	3.3-V low-level PCI output voltage	I_{OL} = 1.5 mA DC, V_{CCIO} = 3.00 to 3.60 V *(10)*			0.1 × V_{CCIO}	V
	2.5-V low-level output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 2.30 V *(10)*			0.2	V
		I_{OL} = 1 mA DC, V_{CCIO} = 2.30 V *(10)*			0.4	V
		I_{OL} = 2 mA DC, V_{CCIO} = 2.30 V *(10)*			0.7	V
I_I	Input pin leakage current	V_I = $V_{CCIOmax}$ to 0 V *(11)*	–10		10	µA
I_{OZ}	Tri-stated I/O pin leakage current	V_O = $V_{CCIOmax}$ to 0 V *(11)*	–10		10	µA
I_{CC0}	V_{CC} supply current (standby)	V_I = ground, no load, no toggling inputs		5		mA
		V_I = ground, no load, no toggling inputs *(12)*		10		mA
R_{CONF}	Value of I/O pin pull-up resistor before and during configuration	V_{CCIO} = 3.0 V *(13)*	20		50	k¾
		V_{CCIO} = 2.3 V *(13)*	30		80	k¾

Figure 25. FLEX 10KE Device LE Timing Model

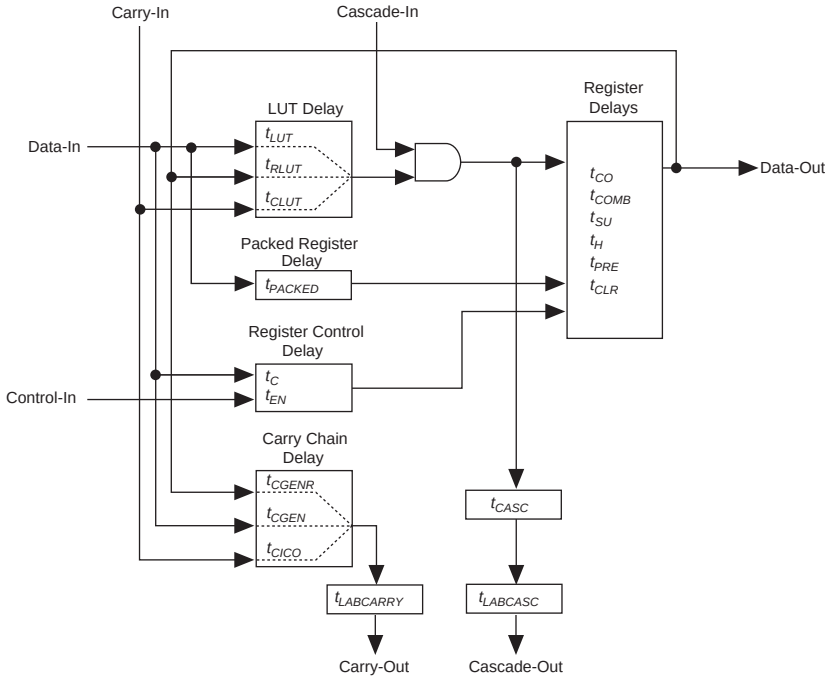


Figure 26. FLEX 10KE Device IOE Timing Model

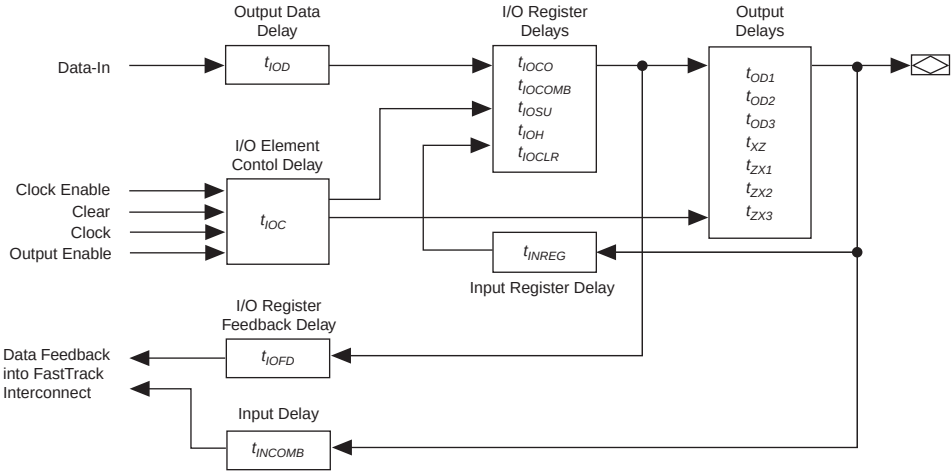
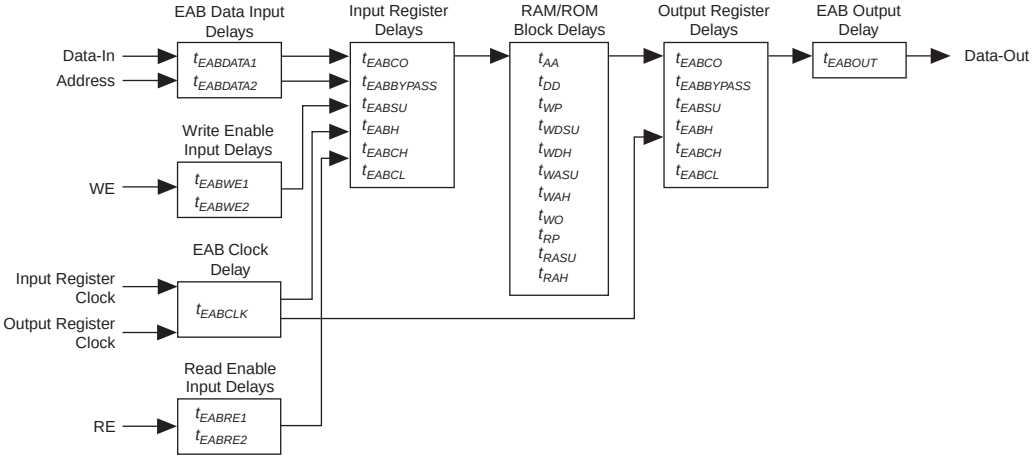


Figure 27. FLEX 10KE Device EAB Timing Model



Figures 29 and 30 show the asynchronous and synchronous timing waveforms, respectively, or the EAB macroparameters in Tables 26 and 27.

Figure 29. EAB Asynchronous Timing Waveforms

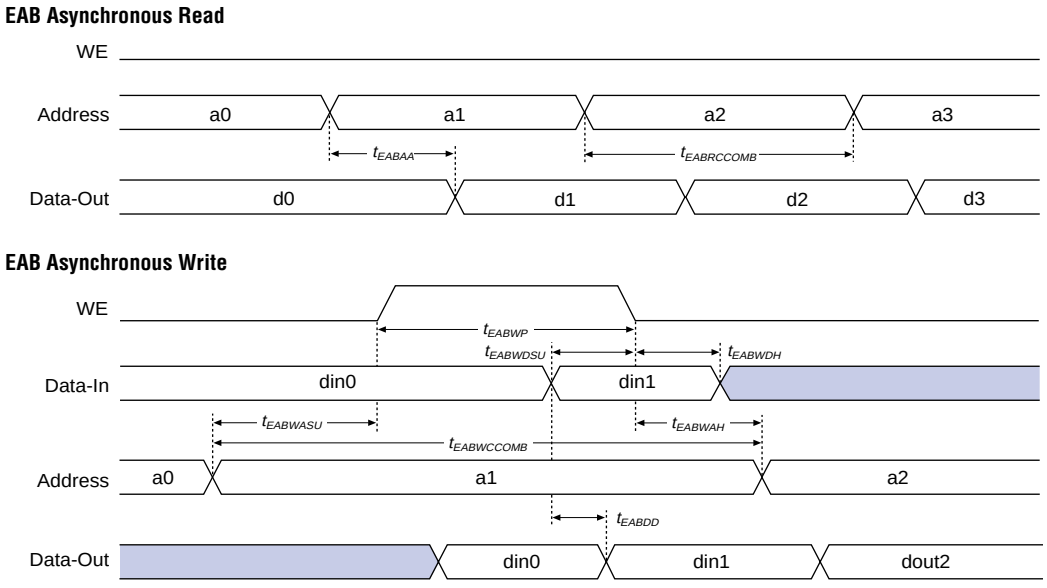


Table 35. EPF10K30E Device Interconnect Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		1.8		2.4		2.9	ns
t_{DIN2LE}		1.5		1.8		2.4	ns
$t_{DIN2DATA}$		1.5		1.8		2.2	ns
$t_{DCLK2IOE}$		2.2		2.6		3.0	ns
$t_{DCLK2LE}$		1.5		1.8		2.4	ns
$t_{SAMELAB}$		0.1		0.2		0.3	ns
$t_{SAMEROW}$		2.0		2.4		2.7	ns
$t_{SAMECOLUMN}$		0.7		1.0		0.8	ns
$t_{DIFFROW}$		2.7		3.4		3.5	ns
$t_{TWOROWS}$		4.7		5.8		6.2	ns
$t_{LEPERIPH}$		2.7		3.4		3.8	ns
$t_{LABCARRY}$		0.3		0.4		0.5	ns
$t_{LABCASC}$		0.8		0.8		1.1	ns

Table 36. EPF10K30E External Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		8.0		9.5		12.5	ns
t_{INSU} *(3)*	2.1		2.5		3.9		ns
t_{INH} *(3)*	0.0		0.0		0.0		ns
t_{OUTCO} *(3)*	2.0	4.9	2.0	5.9	2.0	7.6	ns
t_{INSU} *(4)*	1.1		1.5		–		ns
t_{INH} *(4)*	0.0		0.0		–		ns
t_{OUTCO} *(4)*	0.5	3.9	0.5	4.9	–	–	ns
t_{PCISU}	3.0		4.2		–		ns
t_{PCIH}	0.0		0.0		–		ns
t_{PCICO}	2.0	6.0	2.0	7.5	–	–	ns

Table 50. EPF10K100E External Timing Parameters *Notes (1), (2)*							
Symbol	**-1 Speed Grade**		**-2 Speed Grade**		**-3 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	**Min**	**Max**	
t_{DRR}		9.0		12.0		16.0	ns
t_{INSU} *(3)*	2.0		2.5		3.3		ns
t_{INH} *(3)*	0.0		0.0		0.0		ns
t_{OUTCO} *(3)*	2.0	5.2	2.0	6.9	2.0	9.1	ns
t_{INSU} *(4)*	2.0		2.2		–		ns
t_{INH} *(4)*	0.0		0.0		–		ns
t_{OUTCO} *(4)*	0.5	3.0	0.5	4.6	–	–	ns
t_{PCISU}	3.0		6.2		–		ns
t_{PCIH}	0.0		0.0		–		ns
t_{PCICO}	2.0	6.0	2.0	6.9	–	–	ns

Table 51. EPF10K100E External Bidirectional Timing Parameters *Notes (1), (2)*							
Symbol	**-1 Speed Grade**		**-2 Speed Grade**		**-3 Speed Grade**		**Unit**
	Min	**Max**	**Min**	**Max**	**Min**	**Max**	
$t_{INSUBIDIR}$ *(3)*	1.7		2.5		3.3		ns
$t_{INHBIDIR}$ *(3)*	0.0		0.0		0.0		ns
$t_{INSUBIDIR}$ *(4)*	2.0		2.8		–		ns
$t_{INHBIDIR}$ *(4)*	0.0		0.0		–		ns
$t_{OUTCOBIDIR}$ *(3)*	2.0	5.2	2.0	6.9	2.0	9.1	ns
$t_{XZBIDIR}$ *(3)*		5.6		7.5		10.1	ns
$t_{ZXBIDIR}$ *(3)*		5.6		7.5		10.1	ns
$t_{OUTCOBIDIR}$ *(4)*	0.5	3.0	0.5	4.6	–	–	ns
$t_{XZBIDIR}$ *(4)*		4.6		6.5		–	ns
$t_{ZXBIDIR}$ *(4)*		4.6		6.5		–	ns

Notes to tables:

(1) All timing parameters are described in Tables 24 through 30 in this data sheet.
(2) These parameters are specified by characterization.
(3) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
(4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Table 54. EPF10K130E Device EAB Internal Microparameters *(Part 2 of 2)* *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DD}		1.5		2.0		2.6	ns
t_{EABOUT}		0.2		0.3		0.3	ns
t_{EABCH}	1.5		2.0		2.5		ns
t_{EABCL}	2.7		3.5		4.7		ns

Table 55. EPF10K130E Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		5.9		7.5		9.9	ns
$t_{EABRCOMB}$	5.9		7.5		9.9		ns
$t_{EABRCREG}$	5.1		6.4		8.5		ns
t_{EABWP}	2.7		3.5		4.7		ns
$t_{EABWCOMB}$	5.9		7.7		10.3		ns
$t_{EABWCREG}$	5.4		7.0		9.4		ns
t_{EABDD}		3.4		4.5		5.9	ns
$t_{EABDATACO}$		0.5		0.7		0.8	ns
$t_{EABDATASU}$	0.8		1.0		1.4		ns
$t_{EABDATAH}$	0.1		0.1		0.2		ns
$t_{EABWESU}$	1.1		1.4		1.9		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	1.0		1.3		1.7		ns
t_{EABWDH}	0.2		0.2		0.3		ns
$t_{EABWASU}$	4.1		5.1		6.8		ns
t_{EABWAH}	0.0		0.0		0.0		ns
t_{EABWO}		3.4		4.5		5.9	ns

Table 66. EPF10K50S Device LE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{CGENR}		0.1		0.1		0.1	ns
t_{CASC}		0.5		0.8		1.0	ns
t_{C}		0.5		0.6		0.8	ns
t_{CO}		0.6		0.6		0.7	ns
t_{COMB}		0.3		0.4		0.5	ns
t_{SU}	0.5		0.6		0.7		ns
t_{H}	0.5		0.6		0.8		ns
t_{PRE}		0.4		0.5		0.7	ns
t_{CLR}		0.8		1.0		1.2	ns
t_{CH}	2.0		2.5		3.0		ns
t_{CL}	2.0		2.5		3.0		ns

Table 67. EPF10K50S Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.3		1.3		1.9	ns
t_{IOC}		0.3		0.4		0.4	ns
t_{IOCO}		1.7		2.1		2.6	ns
t_{IOCOMB}		0.5		0.6		0.8	ns
t_{IOSU}	0.8		1.0		1.3		ns
t_{IOH}	0.4		0.5		0.6		ns
t_{IOCLR}		0.2		0.2		0.4	ns
t_{OD1}		1.2		1.2		1.9	ns
t_{OD2}		0.7		0.8		1.7	ns
t_{OD3}		2.7		3.0		4.3	ns
t_{XZ}		4.7		5.7		7.5	ns
t_{ZX1}		4.7		5.7		7.5	ns
t_{ZX2}		4.2		5.3		7.3	ns
t_{ZX3}		6.2		7.5		9.9	ns
t_{INREG}		3.5		4.2		5.6	ns
t_{IOFD}		1.1		1.3		1.8	ns
t_{INCOMB}		1.1		1.3		1.8	ns

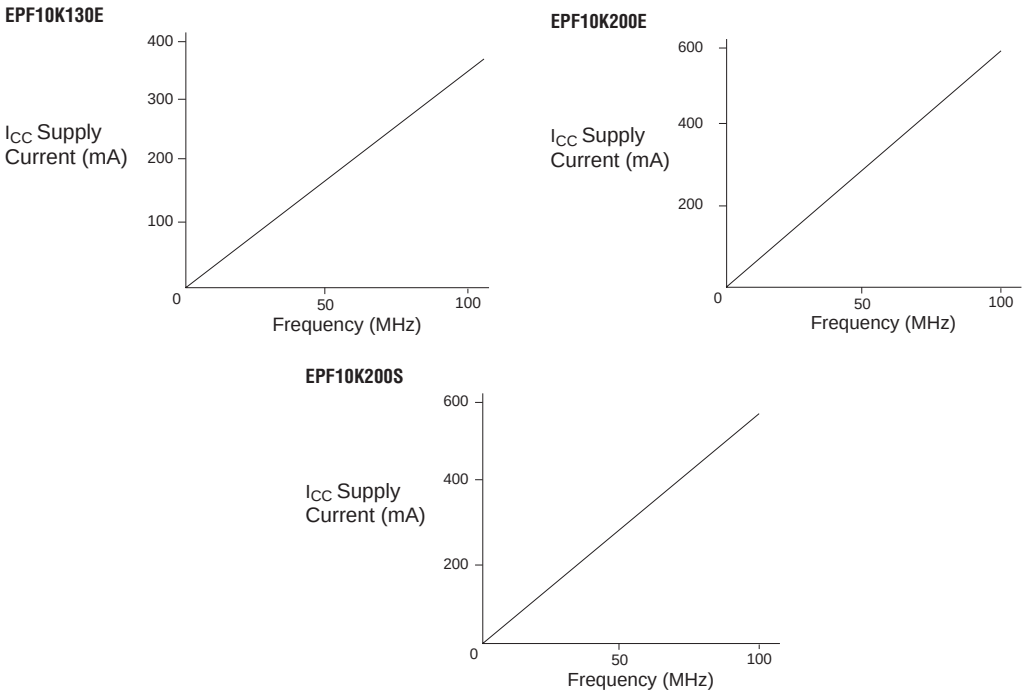
Table 69. EPF10K50S Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		3.7		5.2		7.0	ns
$t_{EABRCCOMB}$	3.7		5.2		7.0		ns
$t_{EABRCREG}$	3.5		4.9		6.6		ns
t_{EABWP}	2.0		2.8		3.8		ns
$t_{EABWCCOMB}$	4.5		6.3		8.6		ns
$t_{EABWCREG}$	5.6		7.8		10.6		ns
t_{EABDD}		3.8		5.3		7.2	ns
$t_{EABDATACO}$		0.8		1.1		1.5	ns
$t_{EABDATASU}$	1.1		1.6		2.1		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	0.7		1.0		1.3		ns
t_{EABWEH}	0.4		0.6		0.8		ns
$t_{EABWDSU}$	1.2		1.7		2.2		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	1.6		2.3		3.0		ns
t_{EABWAH}	0.9		1.2		1.8		ns
t_{EABWO}		3.1		4.3		5.9	ns

Table 70. EPF10K50S Device Interconnect Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		3.1		3.7		4.6	ns
t_{DIN2LE}		1.7		2.1		2.7	ns
$t_{DIN2DATA}$		2.7		3.1		5.1	ns
$t_{DCLK2IOE}$		1.6		1.9		2.6	ns
$t_{DCLK2LE}$		1.7		2.1		2.7	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		1.5		1.7		2.4	ns
$t_{SAMECOLUMN}$		1.0		1.3		2.1	ns
$t_{DIFFROW}$		2.5		3.0		4.5	ns
$t_{TWOROWS}$		4.0		4.7		6.9	ns
$t_{LEPERIPH}$		2.6		2.9		3.4	ns
$t_{LABCARRY}$		0.1		0.2		0.2	ns
$t_{LABCASC}$		0.8		1.0		1.3	ns

Figure 31. FLEX 10KE $I_{CCACTIVE}$ vs. Operating Frequency (Part 2 of 2)



Configuration & Operation

The FLEX 10KE architecture supports several configuration schemes. This section summarizes the device operating modes and available device configuration schemes.

Operating Modes

The FLEX 10KE architecture uses SRAM configuration elements that require configuration data to be loaded every time the circuit powers up. The process of physically loading the SRAM data into the device is called *configuration*. Before configuration, as V_{CC} rises, the device initiates a Power-On Reset (POR). This POR event clears the device and prepares it for configuration. The FLEX 10KE POR time does not exceed 50 µs.

When configuring with a configuration device, refer to the respective configuration device data sheet for POR timing information.

101 Innovation Drive
San Jose, CA 95134
(408) 544-7000
http://www.altera.com
Applications Hotline:
(800) 800-EPLD
Literature Services:
lit_req@altera.com