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Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	216
Number of Logic Elements/Cells	1728
Total RAM Bits	24576
Number of I/O	176
Number of Gates	119000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	256-BGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k30efi256-2

Table 2. FLEX 10KE Device Features

Feature	EPF10K100E (2)	EPF10K130E	EPF10K200E EPF10K200S
Typical gates (1)	100,000	130,000	200,000
Maximum system gates	257,000	342,000	513,000
Logic elements (LEs)	4,992	6,656	9,984
EABs	12	16	24
Total RAM bits	49,152	65,536	98,304
Maximum user I/O pins	338	413	470

Note to tables:

- (1) The embedded IEEE Std. 1149.1 JTAG circuitry adds up to 31,250 gates in addition to the listed typical or maximum system gates.
- (2) New EPF10K100B designs should use EPF10K100E devices.

...and More Features

- Fabricated on an advanced process and operate with a 2.5-V internal supply voltage
- In-circuit reconfigurability (ICR) via external configuration devices, intelligent controller, or JTAG port
- ClockLock™ and ClockBoost™ options for reduced clock delay / skew and clock multiplication
- Built-in low-skew clock distribution trees
- 100% functional testing of all devices; test vectors or scan chains are not required
- Pull-up on I/O pins before and during configuration
- Flexible interconnect
 - FastTrack® Interconnect continuous routing structure for fast, predictable interconnect delays
 - Dedicated carry chain that implements arithmetic functions such as fast adders, counters, and comparators (automatically used by software tools and megafunctions)
 - Dedicated cascade chain that implements high-speed, high-fan-in logic functions (automatically used by software tools and megafunctions)
 - Tri-state emulation that implements internal tri-state buses
 - Up to six global clock signals and four global clear signals
- Powerful I/O pins
 - Individual tri-state output enable control for each pin
 - Open-drain option on each I/O pin
 - Programmable output slew-rate control to reduce switching noise
 - Clamp to V_{CCIO} user-selectable on a pin-by-pin basis
 - Supports hot-socketing

Table 4. FLEX 10KE Package Sizes

Device	144-Pin TQFP	208-Pin PQFP	240-Pin PQFP RQFP	256-Pin FineLine BGA	356-Pin BGA	484-Pin FineLine BGA	599-Pin PGA	600-Pin BGA	672-Pin FineLine BGA
Pitch (mm)	0.50	0.50	0.50	1.0	1.27	1.0	—	1.27	1.0
Area (mm ²)	484	936	1,197	289	1,225	529	3,904	2,025	729
Length × width (mm × mm)	22 × 22	30.6 × 30.6	34.6 × 34.6	17 × 17	35 × 35	23 × 23	62.5 × 62.5	45 × 45	27 × 27

General Description

Altera FLEX 10KE devices are enhanced versions of FLEX 10K devices. Based on reconfigurable CMOS SRAM elements, the FLEX architecture incorporates all features necessary to implement common gate array megafunctions. With up to 200,000 typical gates, FLEX 10KE devices provide the density, speed, and features to integrate entire systems, including multiple 32-bit buses, into a single device.

The ability to reconfigure FLEX 10KE devices enables 100% testing prior to shipment and allows the designer to focus on simulation and design verification. FLEX 10KE reconfigurability eliminates inventory management for gate array designs and generation of test vectors for fault coverage.

Table 5 shows FLEX 10KE performance for some common designs. All performance values were obtained with Synopsys DesignWare or LPM functions. Special design techniques are not required to implement the applications; the designer simply infers or instantiates a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Table 5. FLEX 10KE Performance

Application	Resources Used		Performance			Units
	LEs	EABs	-1 Speed Grade	-2 Speed Grade	-3 Speed Grade	
16-bit loadable counter	16	0	285	250	200	MHz
16-bit accumulator	16	0	285	250	200	MHz
16-to-1 multiplexer (1)	10	0	3.5	4.9	7.0	ns
16-bit multiplier with 3-stage pipeline (2)	592	0	156	131	93	MHz
256 × 16 RAM read cycle speed (2)	0	1	196	154	118	MHz
256 × 16 RAM write cycle speed (2)	0	1	185	143	106	MHz

Notes:

- (1) This application uses combinatorial inputs and outputs.
 (2) This application uses registered inputs and outputs.

Table 6 shows FLEX 10KE performance for more complex designs. These designs are available as Altera MegaCore® functions.

Table 6. FLEX 10KE Performance for Complex Designs

Application	LEs Used	Performance			Units
		-1 Speed Grade	-2 Speed Grade	-3 Speed Grade	
8-bit, 16-tap parallel finite impulse response (FIR) filter	597	192	156	116	MSPS
8-bit, 512-point fast Fourier transform (FFT) function	1,854	23.4	28.7	38.9	μs (1)
		113	92	68	MHz
a16450 universal asynchronous receiver/transmitter (UART)	342	36	28	20.5	MHz

Note:

- (1) These values are for calculation time. Calculation time = number of clocks required / $f_{\max}$. Number of clocks required = ceiling $[\log_2 (\text{points})/2] \times [\text{points} + 14 + \text{ceiling}]$

EABs provide flexible options for driving and controlling clock signals. Different clocks and clock enables can be used for reading and writing to the EAB. Registers can be independently inserted on the data input, EAB output, write address, write enable signals, read address, and read enable signals. The global signals and the EAB local interconnect can drive write enable, read enable, and clock enable signals. The global signals, dedicated clock pins, and EAB local interconnect can drive the EAB clock signals. Because the LEs drive the EAB local interconnect, the LEs can control write enable, read enable, clear, clock, and clock enable signals.

An EAB is fed by a row interconnect and can drive out to row and column interconnects. Each EAB output can drive up to two row channels and up to two column channels; the unused row channel can be driven by other LEs. This feature increases the routing resources available for EAB outputs (see [Figures 2 and 4](#)). The column interconnect, which is adjacent to the EAB, has twice as many channels as other columns in the device.

Logic Array Block

An LAB consists of eight LEs, their associated carry and cascade chains, LAB control signals, and the LAB local interconnect. The LAB provides the coarse-grained structure to the FLEX 10KE architecture, facilitating efficient routing with optimum device utilization and high performance (see [Figure 7](#)).

The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock, clear, and preset control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the output of the LE.

The LE has two outputs that drive the interconnect: one drives the local interconnect and the other drives either the row or column FastTrack Interconnect routing structure. The two outputs can be controlled independently. For example, the LUT can drive one output while the register drives the other output. This feature, called register packing, can improve LE utilization because the register and the LUT can be used for unrelated functions.

The FLEX 10KE architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. The carry chain supports high-speed counters and adders and the cascade chain implements wide-input functions with minimum delay. Carry and cascade chains connect all LEs in a LAB as well as all LABs in the same row. Intensive use of carry and cascade chains can reduce routing flexibility. Therefore, the use of these chains should be limited to speed-critical portions of a design.

Carry Chain

The carry chain provides a very fast (as low as 0.2 ns) carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the FLEX 10KE architecture to implement high-speed counters, adders, and comparators of arbitrary width efficiently. Carry chain logic can be created automatically by the Altera Compiler during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of carry chains.

Carry chains longer than eight LEs are automatically implemented by linking LABs together. For enhanced fitting, a long carry chain skips alternate LABs in a row. A carry chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB. For example, the last LE of the first LAB in a row carries to the first LE of the third LAB in the row. The carry chain does not cross the EAB at the middle of the row. For instance, in the EPF10K50E device, the carry chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB.

When dedicated inputs drive non-inverted and inverted peripheral clears, clock enables, and output enables, two signals on the peripheral control bus will be used.

Tables 8 and 9 list the sources for each peripheral control signal, and show how the output enable, clock enable, clock, and clear signals share 12 peripheral control signals. The tables also show the rows that can drive global signals.

Table 8. Peripheral Bus Sources for EPF10K30E, EPF10K50E & EPF10K50S Devices

Peripheral Control Signal	EPF10K30E	EPF10K50E EPF10K50S
OE0	Row A	Row A
OE1	Row B	Row B
OE2	Row C	Row D
OE3	Row D	Row F
OE4	Row E	Row H
OE5	Row F	Row J
CLKENA0/CLK0/GLOBAL0	Row A	Row A
CLKENA1/OE6/GLOBAL1	Row B	Row C
CLKENA2/CLR0	Row C	Row E
CLKENA3/OE7/GLOBAL2	Row D	Row G
CLKENA4/CLR1	Row E	Row I
CLKENA5/CLK1/GLOBAL3	Row F	Row J

Table 9. Peripheral Bus Sources for EPF10K100E, EPF10K130E, EPF10K200E & EPF10K200S Devices

Peripheral Control Signal	EPF10K100E	EPF10K130E	EPF10K200E EPF10K200S
OE0	Row A	Row C	Row G
OE1	Row C	Row E	Row I
OE2	Row E	Row G	Row K
OE3	Row L	Row N	Row R
OE4	Row I	Row K	Row O
OE5	Row K	Row M	Row Q
CLKENA0/CLK0/GLOBAL0	Row F	Row H	Row L
CLKENA1/OE6/GLOBAL1	Row D	Row F	Row J
CLKENA2/CLR0	Row B	Row D	Row H
CLKENA3/OE7/GLOBAL2	Row H	Row J	Row N
CLKENA4/CLR1	Row J	Row L	Row P
CLKENA5/CLK1/GLOBAL3	Row G	Row I	Row M

Signals on the peripheral control bus can also drive the four global signals, referred to as GLOBAL0 through GLOBAL3 in [Tables 8 and 9](#). An internally generated signal can drive a global signal, providing the same low-skew, low-delay characteristics as a signal driven by an input pin. An LE drives the global signal by driving a row line that drives the peripheral bus, which then drives the global signal. This feature is ideal for internally generated clear or clock signals with high fan-out. However, internally driven global signals offer no advantage over the general-purpose interconnect for routing data signals. The dedicated input pin should be driven to a known logic state (such as ground) and not be allowed to float.

The chip-wide output enable pin is an active-high pin (DEV_OE) that can be used to tri-state all pins on the device. This option can be set in the Altera software. On EPF10K50E and EPF10K200E devices, the built-in I/O pin pull-up resistors (which are active during configuration) are active when the chip-wide output enable pin is asserted. The registers in the IOE can also be reset by the chip-wide reset pin.

Table 13. ClockLock & ClockBoost Parameters for -2 Speed-Grade Devices

Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_R	Input rise time				5	ns
t_F	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		75	MHz
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		37.5	MHz
f_{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software (1)				25,000 (2)	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock (3)				10	μs
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock (4)	$t_{INCLKSTB} < 100$			250	ps
		$t_{INCLKSTB} < 50$			200 (4)	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Notes to tables:

- (1) To implement the ClockLock and ClockBoost circuitry with the MAX+PLUS II software, designers must specify the input frequency. The Altera software tunes the PLL in the ClockLock and ClockBoost circuitry to this frequency. The f_{CLKDEV} parameter specifies how much the incoming clock can differ from the specified frequency during device operation. Simulation does not reflect this parameter.
- (2) Twenty-five thousand parts per million (PPM) equates to 2.5% of input clock period.
- (3) During device configuration, the ClockLock and ClockBoost circuitry is configured before the rest of the device. If the incoming clock is supplied during configuration, the ClockLock and ClockBoost circuitry locks during configuration because the t_{LOCK} value is less than the time required for configuration.
- (4) The t_{JITTER} specification is measured under long-term observation. The maximum value for t_{JITTER} is 200 ps if $t_{INCLKSTB}$ is lower than 50 ps.

I/O Configuration

This section discusses the peripheral component interconnect (PCI) pull-up clamping diode option, slew-rate control, open-drain output option, and MultiVolt I/O interface for FLEX 10KE devices. The PCI pull-up clamping diode, slew-rate control, and open-drain output options are controlled pin-by-pin via Altera software logic options. The MultiVolt I/O interface is controlled by connecting V_{CCIO} to a different voltage than V_{CCINT} . Its effect can be simulated in the Altera software via the **Global Project Device Options** dialog box (Assign menu).

The VCCINT pins must always be connected to a 2.5-V power supply. With a 2.5-V V_{CCINT} level, input voltages are compatible with 2.5-V, 3.3-V, and 5.0-V inputs. The VCCIO pins can be connected to either a 2.5-V or 3.3-V power supply, depending on the output requirements. When the VCCIO pins are connected to a 2.5-V power supply, the output levels are compatible with 2.5-V systems. When the VCCIO pins are connected to a 3.3-V power supply, the output high is at 3.3 V and is therefore compatible with 3.3-V or 5.0-V systems. Devices operating with V_{CCIO} levels higher than 3.0 V achieve a faster timing delay of t_{OD2} instead of t_{OD1} .

Table 14 summarizes FLEX 10KE MultiVolt I/O support.

Table 14. FLEX 10KE MultiVolt I/O Support						
V_{CCIO} (V)	Input Signal (V)			Output Signal (V)		
	2.5	3.3	5.0	2.5	3.3	5.0
2.5	✓	✓(1)	✓(1)	✓		
3.3	✓	✓	✓(1)	✓(2)	✓	✓

Notes:

- (1) The PCI clamping diode must be disabled to drive an input with voltages higher than V_{CCIO} .
- (2) When $V_{CCIO} = 3.3$ V, a FLEX 10KE device can drive a 2.5-V device that has 3.3-V tolerant inputs.

Open-drain output pins on FLEX 10KE devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a V_{IH} of 3.5 V. When the open-drain pin is active, it will drive low. When the pin is inactive, the trace will be pulled up to 5.0 V by the resistor. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The I_{OL} current specification should be considered when selecting a pull-up resistor.

Power Sequencing & Hot-Socketing

Because FLEX 10KE devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The V_{CCIO} and V_{CCINT} power planes can be powered in any order.

Signals can be driven into FLEX 10KE devices before and during power up without damaging the device. Additionally, FLEX 10KE devices do not drive out during power up. Once operating conditions are reached, FLEX 10KE devices operate as specified by the user.

Figure 20 shows the timing requirements for the JTAG signals.

Figure 20. FLEX 10KE JTAG Waveforms

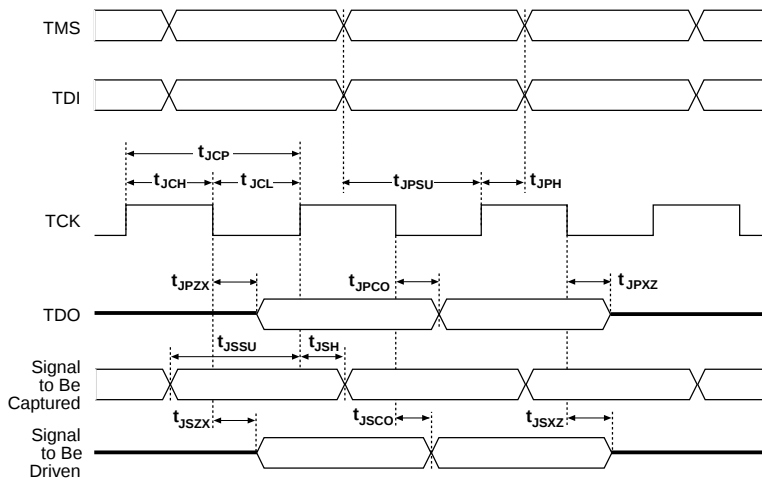


Table 18 shows the timing parameters and values for FLEX 10KE devices.

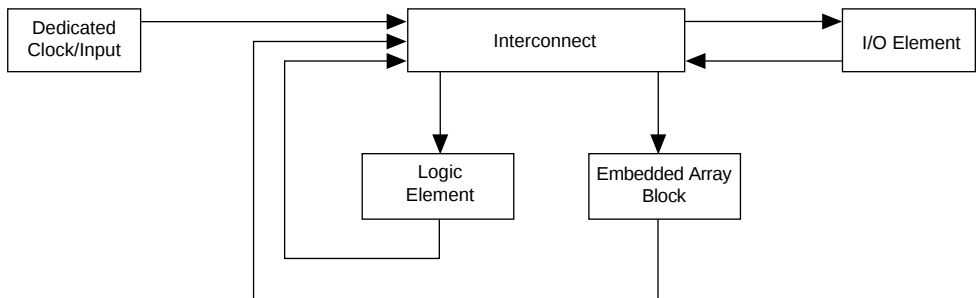
Table 18. FLEX 10KE JTAG Timing Parameters & Values

Symbol	Parameter	Min	Max	Unit
t_{JCP}	TCK clock period	100		ns
t_{JCH}	TCK clock high time	50		ns
t_{JCL}	TCK clock low time	50		ns
t_{JPSU}	JTAG port setup time	20		ns
t_{JPH}	JTAG port hold time	45		ns
t_{JPCO}	JTAG port clock to output		25	ns
t_{JPZX}	JTAG port high impedance to valid output		25	ns
t_{JPXZ}	JTAG port valid output to high impedance		25	ns
t_{JSSU}	Capture register setup time	20		ns
t_{JSH}	Capture register hold time	45		ns
t_{JSCO}	Update register clock to output		35	ns
t_{JSZX}	Update register high impedance to valid output		35	ns
t_{JSXZ}	Update register valid output to high impedance		35	ns

Timing simulation and delay prediction are available with the Altera Simulator and Timing Analyzer, or with industry-standard EDA tools. The Simulator offers both pre-synthesis functional simulation to evaluate logic design accuracy and post-synthesis timing simulation with 0.1-ns resolution. The Timing Analyzer provides point-to-point timing delay information, setup and hold time analysis, and device-wide performance analysis.

Figure 24 shows the overall timing model, which maps the possible paths to and from the various elements of the FLEX 10KE device.

Figure 24. FLEX 10KE Device Timing Model



Figures 25 through 28 show the delays that correspond to various paths and functions within the LE, IOE, EAB, and bidirectional timing models.

Table 28. Interconnect Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
$t_{DIN2IOE}$	Delay from dedicated input pin to IOE control input	(7)
t_{DIN2LE}	Delay from dedicated input pin to LE or EAB control input	(7)
$t_{DCLK2IOE}$	Delay from dedicated clock pin to IOE clock	(7)
$t_{DCLK2LE}$	Delay from dedicated clock pin to LE or EAB clock	(7)
$t_{DIN2DATA}$	Delay from dedicated input or clock to LE or EAB data	(7)
$t_{SAMELAB}$	Routing delay for an LE driving another LE in the same LAB	
$t_{SAMEROW}$	Routing delay for a row IOE, LE, or EAB driving a row IOE, LE, or EAB in the same row	(7)
$t_{SAMECOLUMN}$	Routing delay for an LE driving an IOE in the same column	(7)
$t_{DIFFROW}$	Routing delay for a column IOE, LE, or EAB driving an LE or EAB in a different row	(7)
$t_{TROWROWS}$	Routing delay for a row IOE or EAB driving an LE or EAB in a different row	(7)
$t_{LEPERIPH}$	Routing delay for an LE driving a control signal of an IOE via the peripheral control bus	(7)
$t_{LABCARRY}$	Routing delay for the carry-out signal of an LE driving the carry-in signal of a different LE in a different LAB	
$t_{LABCASC}$	Routing delay for the cascade-out signal of an LE driving the cascade-in signal of a different LE in a different LAB	

Table 29. External Timing Parameters

Symbol	Parameter	Conditions
t_{DRR}	Register-to-register delay via four LEs, three row interconnects, and four local interconnects	(8)
t_{INSU}	Setup time with global clock at IOE register	(9)
t_{INH}	Hold time with global clock at IOE register	(9)
t_{OUTCO}	Clock-to-output delay with global clock at IOE register	(9)
t_{PCISU}	Setup time with global clock for registers used in PCI designs	(9),(10)
t_{PCIH}	Hold time with global clock for registers used in PCI designs	(9),(10)
t_{PCICO}	Clock-to-output delay with global clock for registers used in PCI designs	(9),(10)

Table 31. EPF10K30E Device LE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{CGENR}		0.1		0.1		0.2	ns
t_{CASC}		0.6		0.8		1.0	ns
t_C		0.0		0.0		0.0	ns
t_{CO}		0.3		0.4		0.5	ns
t_{COMB}		0.4		0.4		0.6	ns
t_{SU}	0.4		0.6		0.6		ns
t_H	0.7		1.0		1.3		ns
t_{PRE}		0.8		0.9		1.2	ns
t_{CLR}		0.8		0.9		1.2	ns
t_{CH}	2.0		2.5		2.5		ns
t_{CL}	2.0		2.5		2.5		ns

Table 32. EPF10K30E Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		2.4		2.8		3.8	ns
t_{IOC}		0.3		0.4		0.5	ns
t_{IOCO}		1.0		1.1		1.6	ns
t_{IOCOMB}		0.0		0.0		0.0	ns
t_{IOSU}	1.2		1.4		1.9		ns
t_{IOH}	0.3		0.4		0.5		ns
t_{IOCLR}		1.0		1.1		1.6	ns
t_{OD1}		1.9		2.3		3.0	ns
t_{OD2}		1.4		1.8		2.5	ns
t_{OD3}		4.4		5.2		7.0	ns
t_{XZ}		2.7		3.1		4.3	ns
t_{ZX1}		2.7		3.1		4.3	ns
t_{ZX2}		2.2		2.6		3.8	ns
t_{ZX3}		5.2		6.0		8.3	ns
t_{INREG}		3.4		4.1		5.5	ns
t_{IOFD}		0.8		1.3		2.4	ns
t_{INCOMB}		0.8		1.3		2.4	ns

Table 35. EPF10K30E Device Interconnect Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		1.8		2.4		2.9	ns
t_{DIN2LE}		1.5		1.8		2.4	ns
$t_{DIN2DATA}$		1.5		1.8		2.2	ns
$t_{DCLK2IOE}$		2.2		2.6		3.0	ns
$t_{DCLK2LE}$		1.5		1.8		2.4	ns
$t_{SAMELAB}$		0.1		0.2		0.3	ns
$t_{SAMEROW}$		2.0		2.4		2.7	ns
$t_{SAMECOLUMN}$		0.7		1.0		0.8	ns
$t_{DIFFROW}$		2.7		3.4		3.5	ns
$t_{TWOROWS}$		4.7		5.8		6.2	ns
$t_{LEPERIPH}$		2.7		3.4		3.8	ns
$t_{LABCARRY}$		0.3		0.4		0.5	ns
$t_{LABCASC}$		0.8		0.8		1.1	ns

Table 36. EPF10K30E External Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DDR}		8.0		9.5		12.5	ns
$t_{INSU}^{(3)}$	2.1		2.5		3.9		ns
$t_{INH}^{(3)}$	0.0		0.0		0.0		ns
$t_{OUTCO}^{(3)}$	2.0	4.9	2.0	5.9	2.0	7.6	ns
$t_{INSU}^{(4)}$	1.1		1.5		—		ns
$t_{INH}^{(4)}$	0.0		0.0		—		ns
$t_{OUTCO}^{(4)}$	0.5	3.9	0.5	4.9	—	—	ns
t_{PCISU}	3.0		4.2		—		ns
t_{PCIH}	0.0		0.0		—		ns
t_{PCICO}	2.0	6.0	2.0	7.5	—	—	ns

Table 37. EPF10K30E External Bidirectional Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{\text{INSUBIDIR}}$ (3)	2.8		3.9		5.2		ns
t_{INHBIDIR} (3)	0.0		0.0		0.0		ns
$t_{\text{INSUBIDIR}}$ (4)	3.8		4.9		—		ns
t_{INHBIDIR} (4)	0.0		0.0		—		ns
$t_{\text{OUTCOBIDIR}}$ (3)	2.0	4.9	2.0	5.9	2.0	7.6	ns
t_{XZBIDIR} (3)		6.1		7.5		9.7	ns
t_{ZXBIDIR} (3)		6.1		7.5		9.7	ns
$t_{\text{OUTCOBIDIR}}$ (4)	0.5	3.9	0.5	4.9	—	—	ns
t_{XZBIDIR} (4)		5.1		6.5		—	ns
t_{ZXBIDIR} (4)		5.1		6.5		—	ns

Notes to tables:

- (1) All timing parameters are described in Tables 24 through 30 in this data sheet.
- (2) These parameters are specified by characterization.
- (3) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
- (4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Tables 38 through 44 show EPF10K50E device internal and external timing parameters.

Table 38. EPF10K50E Device LE Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.6		0.9		1.3	ns
t_{CLUT}		0.5		0.6		0.8	ns
t_{RLUT}		0.7		0.8		1.1	ns
t_{PACKED}		0.4		0.5		0.6	ns
t_{EN}		0.6		0.7		0.9	ns
t_{CICO}		0.2		0.2		0.3	ns
t_{CGEN}		0.5		0.5		0.8	ns
t_{CGENR}		0.2		0.2		0.3	ns
t_{CASC}		0.8		1.0		1.4	ns
t_{C}		0.5		0.6		0.8	ns
t_{CO}		0.7		0.7		0.9	ns
t_{COMB}		0.5		0.6		0.8	ns
t_{SU}	0.7		0.7		0.8		ns

Table 40. EPF10K50E Device EAB Internal Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.7		2.0		2.7	ns
$t_{EABDATA1}$		0.6		0.7		0.9	ns
t_{EABWE1}		1.1		1.3		1.8	ns
t_{EABWE2}		0.4		0.4		0.6	ns
t_{EABRE1}		0.8		0.9		1.2	ns
t_{EABRE2}		0.4		0.4		0.6	ns
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		0.3		0.3		0.5	ns
$t_{EABYPASS}$		0.5		0.6		0.8	ns
t_{EABSU}	0.9		1.0		1.4		ns
t_{EABH}	0.4		0.4		0.6		ns
t_{EABCLR}	0.3		0.3		0.5		ns
t_{AA}		3.2		3.8		5.1	ns
t_{WP}	2.5		2.9		3.9		ns
t_{RP}	0.9		1.1		1.5		ns
t_{WDSU}	0.9		1.0		1.4		ns
t_{WDH}	0.1		0.1		0.2		ns
t_{WASU}	1.7		2.0		2.7		ns
t_{WAH}	1.8		2.1		2.9		ns
t_{RASU}	3.1		3.7		5.0		ns
t_{RAH}	0.2		0.2		0.3		ns
t_{WO}		2.5		2.9		3.9	ns
t_{DD}		2.5		2.9		3.9	ns
t_{EABOUT}		0.5		0.6		0.8	ns
t_{EABCH}	1.5		2.0		2.5		ns
t_{EABCL}	2.5		2.9		3.9		ns

Table 41. EPF10K50E Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		6.4		7.6		10.2	ns
$t_{EABRCOMB}$	6.4		7.6		10.2		ns
$t_{EABRCREG}$	4.4		5.1		7.0		ns
t_{EABWP}	2.5		2.9		3.9		ns
$t_{EABWCOMB}$	6.0		7.0		9.5		ns
$t_{EABWCREG}$	6.8		7.8		10.6		ns
t_{EABDD}		5.7		6.7		9.0	ns
$t_{EABDATACO}$		0.8		0.9		1.3	ns
$t_{EABDATASU}$	1.5		1.7		2.3		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	1.3		1.4		2.0		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	1.5		1.7		2.3		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	3.0		3.6		4.8		ns
t_{EABWAH}	0.5		0.5		0.8		ns
t_{EABWO}		5.1		6.0		8.1	ns

Table 42. EPF10K50E Device Interconnect Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		3.5		4.3		5.6	ns
t_{DIN2LE}		2.1		2.5		3.4	ns
$t_{DIN2DATA}$		2.2		2.4		3.1	ns
$t_{DCLK2IOE}$		2.9		3.5		4.7	ns
$t_{DCLK2LE}$		2.1		2.5		3.4	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		1.1		1.1		1.5	ns
$t_{SAMECOLUMN}$		0.8		1.0		1.3	ns
$t_{DIFFROW}$		1.9		2.1		2.8	ns
$t_{TROWROWS}$		3.0		3.2		4.3	ns
$t_{LEPERIPH}$		3.1		3.3		3.7	ns
$t_{LABCARRY}$		0.1		0.1		0.2	ns
$t_{LABCASC}$		0.3		0.3		0.5	ns

Table 47. EPF10K100E Device EAB Internal Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.5		2.0		2.6	ns
$t_{EABDATA1}$		0.0		0.0		0.0	ns
t_{EABWE1}		1.5		2.0		2.6	ns
t_{EABWE2}		0.3		0.4		0.5	ns
t_{EABRE1}		0.3		0.4		0.5	ns
t_{EABRE2}		0.0		0.0		0.0	ns
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		0.3		0.4		0.5	ns
$t_{EABYPASS}$		0.1		0.1		0.2	ns
t_{EABSU}	0.8		1.0		1.4		ns
t_{EABH}	0.1		0.1		0.2		ns
t_{EABCLR}	0.3		0.4		0.5		ns
t_{AA}		4.0		5.1		6.6	ns
t_{WP}	2.7		3.5		4.7		ns
t_{RP}	1.0		1.3		1.7		ns
t_{WDSU}	1.0		1.3		1.7		ns
t_{WDH}	0.2		0.2		0.3		ns
t_{WASU}	1.6		2.1		2.8		ns
t_{WAH}	1.6		2.1		2.8		ns
t_{RASU}	3.0		3.9		5.2		ns
t_{RAH}	0.1		0.1		0.2		ns
t_{WO}		1.5		2.0		2.6	ns
t_{DD}		1.5		2.0		2.6	ns
t_{EABOUT}		0.2		0.3		0.3	ns
t_{EABCH}	1.5		2.0		2.5		ns
t_{EABCL}	2.7		3.5		4.7		ns

Table 48. EPF10K100E Device EAB Internal Timing Macroparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		5.9		7.6		9.9	ns
$t_{EABRCOMB}$	5.9		7.6		9.9		ns
$t_{EABRCREG}$	5.1		6.5		8.5		ns
t_{EABWP}	2.7		3.5		4.7		ns

Table 76. EPF10K200S Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		3.9		6.4		8.4	ns
$t_{EABRCOMB}$	3.9		6.4		8.4		ns
$t_{EABRCREG}$	3.6		5.7		7.6		ns
t_{EABWP}	2.1		4.0		5.3		ns
$t_{EABWCOMB}$	4.8		8.1		10.7		ns
$t_{EABWCREG}$	5.4		8.0		10.6		ns
t_{EABDD}		3.8		5.1		6.7	ns
$t_{EABDATAO}$		0.8		1.0		1.3	ns
$t_{EABDATASU}$	1.1		1.6		2.1		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	0.7		1.1		1.5		ns
t_{EABWEH}	0.4		0.5		0.6		ns
$t_{EABWDSU}$	1.2		1.8		2.4		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	1.9		3.6		4.7		ns
t_{EABWAH}	0.8		0.5		0.7		ns
t_{EABWO}		3.1		4.4		5.8	ns

Table 77. EPF10K200S Device Interconnect Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		4.4		4.8		5.5	ns
t_{DIN2LE}		0.6		0.6		0.9	ns
$t_{DIN2DATA}$		1.8		2.1		2.8	ns
$t_{DCLK2IOE}$		1.7		2.0		2.8	ns
$t_{DCLK2LE}$		0.6		0.6		0.9	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		3.0		4.6		5.7	ns
$t_{SAMECOLUMN}$		3.5		4.9		6.4	ns
$t_{DIFFROW}$		6.5		9.5		12.1	ns
$t_{TROWROWS}$		9.5		14.1		17.8	ns
$t_{LEPERIPH}$		5.5		6.2		7.2	ns
$t_{LABCARRY}$		0.3		0.1		0.2	ns

Power Consumption

The supply power (P) for FLEX 10KE devices can be calculated with the following equation:

$$P = P_{INT} + P_{IO} = (I_{CCSTANDBY} + I_{CCACTIVE}) \times V_{CC} + P_{IO}$$

The $I_{CCACTIVE}$ value depends on the switching frequency and the application logic. This value is calculated based on the amount of current that each LE typically consumes. The P_{IO} value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in [Application Note 74 \(Evaluating Power for Altera Devices\)](#).

Compared to the rest of the device, the embedded array consumes a negligible amount of power. Therefore, the embedded array can be ignored when calculating supply current.

The $I_{CCACTIVE}$ value can be calculated with the following equation:

$$I_{CCACTIVE} = K \times f_{MAX} \times N \times \text{tog}_{LC} \times \frac{\mu A}{MHz \times LE}$$

Where:

- f_{MAX} = Maximum operating frequency in MHz
- N = Total number of LEs used in the device
- tog_{LC} = Average percent of LEs toggling at each clock (typically 12.5%)
- K = Constant

[Table 80](#) provides the constant (K) values for FLEX 10KE devices.

Table 80. FLEX 10KE K Constant Values	
Device	K Value
EPF10K30E	4.5
EPF10K50E	4.8
EPF10K50S	4.5
EPF10K100E	4.5
EPF10K130E	4.6
EPF10K200E	4.8
EPF10K200S	4.6

This calculation provides an I_{CC} estimate based on typical conditions with no output load. The actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.