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Intel - EPF10K50EFC484-3N Datasheet



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Understanding <u>Embedded - FPGAs (Field</u> <u>Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Detuns	
Product Status	Obsolete
Number of LABs/CLBs	360
Number of Logic Elements/Cells	2880
Total RAM Bits	40960
Number of I/O	220
Number of Gates	199000
Voltage - Supply	2.3V ~ 2.7V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	484-BBGA
Supplier Device Package	484-FBGA (23x23)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k50efc484-3n

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

Table 4. FLEX	(10KE Pa	ckage Sizes							
Device	144- Pin TQFP	208-Pin PQFP	240-Pin PQFP RQFP	256-Pin FineLine BGA	356- Pin BGA	484-Pin FineLine BGA	599-Pin PGA	600- Pin BGA	672-Pin FineLine BGA
Pitch (mm)	0.50	0.50	0.50	1.0	1.27	1.0	-	1.27	1.0
Area (mm ²)	484	936	1,197	289	1,225	529	3,904	2,025	729
$\begin{array}{l} \text{Length} \times \text{width} \\ \text{(mm} \times \text{mm)} \end{array}$	22 × 22	30.6×30.6	34.6×34.6	17 × 17	35×35	23 × 23	62.5 × 62.5	45×45	27 × 27

General Description

Altera FLEX 10KE devices are enhanced versions of FLEX 10K devices. Based on reconfigurable CMOS SRAM elements, the FLEX architecture incorporates all features necessary to implement common gate array megafunctions. With up to 200,000 typical gates, FLEX 10KE devices provide the density, speed, and features to integrate entire systems, including multiple 32-bit buses, into a single device.

The ability to reconfigure FLEX 10KE devices enables 100% testing prior to shipment and allows the designer to focus on simulation and design verification. FLEX 10KE reconfigurability eliminates inventory management for gate array designs and generation of test vectors for fault coverage.

Table 5 shows FLEX 10KE performance for some common designs. All performance values were obtained with Synopsys DesignWare or LPM functions. Special design techniques are not required to implement the applications; the designer simply infers or instantiates a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Figure 1 shows a block diagram of the FLEX 10KE architecture. Each group of LEs is combined into an LAB; groups of LABs are arranged into rows and columns. Each row also contains a single EAB. The LABs and EABs are interconnected by the FastTrack Interconnect routing structure. IOEs are located at the end of each row and column of the FastTrack Interconnect routing structure.



FLEX 10KE devices provide six dedicated inputs that drive the flipflops' control inputs and ensure the efficient distribution of high-speed, low-skew (less than 1.5 ns) control signals. These signals use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect routing structure. Four of the dedicated inputs drive four global signals. These four global signals can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device.

Figure 7. FLEX 10KE LAB



Notes:

- (1) EPF10K30E, EPF10K50E, and EPF10K50S devices have 22 inputs to the LAB local interconnect channel from the row; EPF10K100E, EPF10K130E, EPF10K200E, and EPF10K200S devices have 26.
- (2) EPF10K30E, EPF10K50E, and EPF10K50S devices have 30 LAB local interconnect channels; EPF10K100E, EPF10K130E, EPF10K200E, and EPF10K200S devices have 34.

The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock, clear, and preset control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the output of the LE.

The LE has two outputs that drive the interconnect: one drives the local interconnect and the other drives either the row or column FastTrack Interconnect routing structure. The two outputs can be controlled independently. For example, the LUT can drive one output while the register drives the other output. This feature, called register packing, can improve LE utilization because the register and the LUT can be used for unrelated functions.

The FLEX 10KE architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. The carry chain supports high-speed counters and adders and the cascade chain implements wide-input functions with minimum delay. Carry and cascade chains connect all LEs in a LAB as well as all LABs in the same row. Intensive use of carry and cascade chains can reduce routing flexibility. Therefore, the use of these chains should be limited to speed-critical portions of a design.

Carry Chain

The carry chain provides a very fast (as low as 0.2 ns) carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the FLEX 10KE architecture to implement high-speed counters, adders, and comparators of arbitrary width efficiently. Carry chain logic can be created automatically by the Altera Compiler during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of carry chains.

Carry chains longer than eight LEs are automatically implemented by linking LABs together. For enhanced fitting, a long carry chain skips alternate LABs in a row. A carry chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from oddnumbered LAB to odd-numbered LAB. For example, the last LE of the first LAB in a row carries to the first LE of the third LAB in the row. The carry chain does not cross the EAB at the middle of the row. For instance, in the EPF10K50E device, the carry chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB.





I/O Element

An IOE contains a bidirectional I/O buffer and a register that can be used either as an input register for external data that requires a fast setup time, or as an output register for data that requires fast clock-to-output performance. In some cases, using an LE register for an input register will result in a faster setup time than using an IOE register. IOEs can be used as input, output, or bidirectional pins. For bidirectional registered I/O implementation, the output register should be in the IOE, and the data input and output enable registers should be LE registers placed adjacent to the bidirectional pin. The Altera Compiler uses the programmable inversion option to invert signals from the row and column interconnect automatically where appropriate. Figure 15 shows the bidirectional I/O registers. When dedicated inputs drive non-inverted and inverted peripheral clears, clock enables, and output enables, two signals on the peripheral control bus will be used.

Tables 8 and 9 list the sources for each peripheral control signal, and show how the output enable, clock enable, clock, and clear signals share 12 peripheral control signals. The tables also show the rows that can drive global signals.

Table 8. Peripheral Bus Sources for EPF10K30E, E	PF10K50E & EPF10K50S Devi	ices
Peripheral Control Signal	EPF10K30E	EPF10K50E EPF10K50S
OEO	Row A	Row A
OE1	Row B	Row B
OE2	Row C	Row D
OE3	Row D	Row F
OE4	Row E	Row H
OE5	Row F	Row J
CLKENA0/CLK0/GLOBAL0	Row A	Row A
CLKENA1/OE6/GLOBAL1	Row B	Row C
CLKENA2/CLR0	Row C	Row E
CLKENA3/OE7/GLOBAL2	Row D	Row G
CLKENA4/CLR1	Row E	Row I
CLKENA5/CLK1/GLOBAL3	Row F	Row J

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Peripheral Control Signal	EPF10K100E	EPF10K130E	EPF10K200E EPF10K200S
OE0	Row A	Row C	Row G
OE1	Row C	Row E	Row I
OE2	Row E	Row G	Row K
OE 3	Row L	Row N	Row R
OE4	Row I	Row K	Row O
OE5	Row K	Row M	Row Q
CLKENA0/CLK0/GLOBAL0	Row F	Row H	Row L
CLKENA1/OE6/GLOBAL1	Row D	Row F	Row J
CLKENA2/CLR0	Row B	Row D	Row H
CLKENA3/OE7/GLOBAL2	Row H	Row J	Row N
CLKENA4/CLR1	Row J	Row L	Row P
CLKENA5/CLK1/GLOBAL3	Row G	Row I	Row M

Signals on the peripheral control bus can also drive the four global signals, referred to as GLOBAL0 through GLOBAL3 in Tables 8 and 9. An internally generated signal can drive a global signal, providing the same low-skew, low-delay characteristics as a signal driven by an input pin. An LE drives the global signal by driving a row line that drives the peripheral bus, which then drives the global signal. This feature is ideal for internally generated clear or clock signals with high fan-out. However, internally driven global signals offer no advantage over the general-purpose interconnect for routing data signals. The dedicated input pin should be driven to a known logic state (such as ground) and not be allowed to float.

The chip-wide output enable pin is an active-high pin (DEV_OE) that can be used to tri-state all pins on the device. This option can be set in the Altera software. On EPF10K50E and EPF10K200E devices, the built-in I/O pin pull-up resistors (which are active during configuration) are active when the chip-wide output enable pin is asserted. The registers in the IOE can also be reset by the chip-wide reset pin.

ClockLock & ClockBoost Features

To support high-speed designs, FLEX 10KE devices offer optional ClockLock and ClockBoost circuitry containing a phase-locked loop (PLL) used to increase design speed and reduce resource usage. The ClockLock circuitry uses a synchronizing PLL that reduces the clock delay and skew within a device. This reduction minimizes clock-to-output and setup times while maintaining zero hold times. The ClockBoost circuitry, which provides a clock multiplier, allows the designer to enhance device area efficiency by resource sharing within the device. The ClockBoost feature allows the designer to distribute a low-speed clock and multiply that clock on-device. Combined, the ClockLock and ClockBoost features provide significant improvements in system performance and bandwidth.

All FLEX 10KE devices, except EPF10K50E and EPF10K200E devices, support ClockLock and ClockBoost circuitry. EPF10K50S and EPF10K200S devices support this circuitry. Devices that support Clock-Lock and ClockBoost circuitry are distinguished with an "X" suffix in the ordering code; for instance, the EPF10K200SFC672-1X device supports this circuit.

The ClockLock and ClockBoost features in FLEX 10KE devices are enabled through the Altera software. External devices are not required to use these features. The output of the ClockLock and ClockBoost circuits is not available at any of the device pins.

The ClockLock and ClockBoost circuitry locks onto the rising edge of the incoming clock. The circuit output can drive the clock inputs of registers only; the generated clock cannot be gated or inverted.

The dedicated clock pin (GCLK1) supplies the clock to the ClockLock and ClockBoost circuitry. When the dedicated clock pin is driving the ClockLock or ClockBoost circuitry, it cannot drive elsewhere in the device.

For designs that require both a multiplied and non-multiplied clock, the clock trace on the board can be connected to the GCLK1 pin. In the Altera software, the GCLK1 pin can feed both the ClockLock and ClockBoost circuitry in the FLEX 10KE device. However, when both circuits are used, the other clock pin cannot be used.

Symbol	Parameter	Condition	Min	Тур	Max	Unit
t _R	Input rise time				5	ns
t _F	Input fall time				5	ns
t _{INDUTY}	Input duty cycle		40		60	%
f _{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		75	MHz
f _{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		37.5	MHz
f _{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software (1)				25,000 (2)	PPM
t _{INCLKSTB}	Input clock stability (measured between adjacent clocks)				100	ps
t _{LOCK}	Time required for ClockLock or ClockBoost to acquire lock (3)				10	μs
t _{JITTER}	Jitter on ClockLock or ClockBoost-	$t_{INCLKSTB} < 100$			250	ps
	generated clock (4)	$t_{INCLKSTB} < 50$			200 (4)	ps
toutduty	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Notes to tables:

- (1) To implement the ClockLock and ClockBoost circuitry with the MAX+PLUS II software, designers must specify the input frequency. The Altera software tunes the PLL in the ClockLock and ClockBoost circuitry to this frequency. The f_{CLKDEV} parameter specifies how much the incoming clock can differ from the specified frequency during device operation. Simulation does not reflect this parameter.
- (2) Twenty-five thousand parts per million (PPM) equates to 2.5% of input clock period.
- (3) During device configuration, the ClockLock and ClockBoost circuitry is configured before the rest of the device. If the incoming clock is supplied during configuration, the ClockLock and ClockBoost circuitry locks during configuration because the t_{LOCK} value is less than the time required for configuration.
- (4) The t_{ITTER} specification is measured under long-term observation. The maximum value for t_{ITTER} is 200 ps if t_{INCLKSTB} is lower than 50 ps.

I/O Configuration

This section discusses the peripheral component interconnect (PCI) pull-up clamping diode option, slew-rate control, open-drain output option, and MultiVolt I/O interface for FLEX 10KE devices. The PCI pull-up clamping diode, slew-rate control, and open-drain output options are controlled pin-by-pin via Altera software logic options. The MultiVolt I/O interface is controlled by connecting V_{CCIO} to a different voltage than V_{CCINT} . Its effect can be simulated in the Altera software via the **Global Project Device Options** dialog box (Assign menu).

PCI Pull-Up Clamping Diode Option

FLEX 10KE devices have a pull-up clamping diode on every I/O, dedicated input, and dedicated clock pin. PCI clamping diodes clamp the signal to the $V_{\rm CCIO}$ value and are required for 3.3-V PCI compliance. Clamping diodes can also be used to limit overshoot in other systems.

Clamping diodes are controlled on a pin-by-pin basis. When V_{CCIO} is 3.3 V, a pin that has the clamping diode option turned on can be driven by a 2.5-V or 3.3-V signal, but not a 5.0-V signal. When V_{CCIO} is 2.5 V, a pin that has the clamping diode option turned on can be driven by a 2.5-V signal, but not a 3.3-V or 5.0-V signal. Additionally, a clamping diode can be activated for a subset of pins, which would allow a device to bridge between a 3.3-V PCI bus and a 5.0-V device.

Slew-Rate Control

The output buffer in each IOE has an adjustable output slew rate that can be configured for low-noise or high-speed performance. A slower slew rate reduces system noise and adds a maximum delay of 4.3 ns. The fast slew rate should be used for speed-critical outputs in systems that are adequately protected against noise. Designers can specify the slew rate pin-by-pin or assign a default slew rate to all pins on a device-wide basis. The slow slew rate setting affects the falling edge of the output.

Open-Drain Output Option

FLEX 10KE devices provide an optional open-drain output (electrically equivalent to open-collector output) for each I/O pin. This open-drain output enables the device to provide system-level control signals (e.g., interrupt and write enable signals) that can be asserted by any of several devices. It can also provide an additional wired-OR plane.

MultiVolt I/O Interface

The FLEX 10KE device architecture supports the MultiVolt I/O interface feature, which allows FLEX 10KE devices in all packages to interface with systems of differing supply voltages. These devices have one set of V_{CC} pins for internal operation and input buffers (VCCINT), and another set for I/O output drivers (VCCIO).

IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

All FLEX 10KE devices provide JTAG BST circuitry that complies with the IEEE Std. 1149.1-1990 specification. FLEX 10KE devices can also be configured using the JTAG pins through the BitBlaster or ByteBlasterMV download cable, or via hardware that uses the Jam[™] STAPL programming and test language. JTAG boundary-scan testing can be performed before or after configuration, but not during configuration. FLEX 10KE devices support the JTAG instructions shown in Table 15.

Table 15. FLEX 10KE	JTAG Instructions
JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins.
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation.
USERCODE	Selects the user electronic signature (USERCODE) register and places it between the TDI and TDO pins, allowing the USERCODE to be serially shifted out of TDO.
IDCODE	Selects the IDCODE register and places it between TDI and TDO, allowing the IDCODE to be serially shifted out of TDO.
ICR Instructions	These instructions are used when configuring a FLEX 10KE device via JTAG ports with a BitBlaster or ByteBlasterMV download cable, or using a Jam File (.jam) or Jam Byte-Code File (.jbc) via an embedded processor.

The instruction register length of FLEX 10KE devices is 10 bits. The USERCODE register length in FLEX 10KE devices is 32 bits; 7 bits are determined by the user, and 25 bits are pre-determined. Tables 16 and 17 show the boundary-scan register length and device IDCODE information for FLEX 10KE devices.

Table 16. FLEX 10KE Boundary	Scan Register Length
Device	Boundary-Scan Register Length
EPF10K30E	690
EPF10K50E	798
EPF10K50S	
EPF10K100E	1,050
EPF10K130E	1,308
EPF10K200E EPF10K200S	1,446

Table 17. 32-	Bit IDCOD	E for FLEX 10KE Devices	Note (1)				
Device		IDCODE (32 Bits)					
	Version (4 Bits)	Part Number (16 Bits)	Manufacturer's Identity (11 Bits)	1 (1 Bit) (2)			
EPF10K30E	0001	0001 0000 0011 0000	00001101110	1			
EPF10K50E EPF10K50S	0001	0001 0000 0101 0000	00001101110	1			
EPF10K100E	0010	0000 0001 0000 0000	00001101110	1			
EPF10K130E	0001	0000 0001 0011 0000	00001101110	1			
EPF10K200E EPF10K200S	0001	0000 0010 0000 0000	00001101110	1			

Notes:

(1) The most significant bit (MSB) is on the left.

(2) The least significant bit (LSB) for all JTAG IDCODEs is 1.

FLEX 10KE devices include weak pull-up resistors on the JTAG pins.



For more information, see the following documents:

- Application Note 39 (IEEE Std. 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices)
- BitBlaster Serial Download Cable Data Sheet
- ByteBlasterMV Parallel Port Download Cable Data Sheet
- Jam Programming & Test Language Specification

Figure 20 shows the timing requirements for the JTAG signals.



Figure 20. FLEX 10KE JTAG Waveforms

Table 18 shows the timing parameters and values for FLEX 10KE devices.

Sumbol	Parameter	Min	Max	Unit
Symbol	Parameter	IVIIII	IVIAX	Unit
t _{JCP}	TCK clock period	100		ns
t _{JCH}	TCK clock high time	50		ns
t _{JCL}	TCK clock low time	50		ns
t _{JPSU}	JTAG port setup time	20		ns
t _{JPH}	JTAG port hold time	45		ns
t _{JPCO}	JTAG port clock to output		25	ns
t _{JPZX}	JTAG port high impedance to valid output		25	ns
t _{JPXZ}	JTAG port valid output to high impedance		25	ns
t _{JSSU}	Capture register setup time	20		ns
t _{JSH}	Capture register hold time	45		ns
t _{JSCO}	Update register clock to output		35	ns
t _{JSZX}	Update register high impedance to valid output		35	ns
t _{JSXZ}	Update register valid output to high impedance		35	ns

Figure 25. FLEX 10KE Device LE Timing Model





Figure 28. Synchronous Bidirectional Pin External Timing Model

Tables 24 through 28 describe the FLEX 10KE device internal timing parameters. Tables 29 through 30 describe the FLEX 10KE external timing parameters and their symbols.

Symbol	Parameter	Condition		
t _{LUT}	LUT delay for data-in			
t _{CLUT}	LUT delay for carry-in			
t _{RLUT}	LUT delay for LE register feedback			
t _{PACKED}	Data-in to packed register delay			
t _{EN}	LE register enable delay			
t _{CICO}	Carry-in to carry-out delay			
t _{CGEN}	Data-in to carry-out delay			
t _{CGENR}	LE register feedback to carry-out delay			
t _{CASC}	Cascade-in to cascade-out delay			
t _C	LE register control signal delay			
t _{CO}	LE register clock-to-output delay			
t _{COMB}	Combinatorial delay			
t _{SU}	LE register setup time for data and enable signals before clock; LE register			
	recovery time after asynchronous clear, preset, or load			
t _H	LE register hold time for data and enable signals after clock			
t _{PRE}	LE register preset delay			

FLEX 10KE Embedded Programmable Logic Devices Data Sheet

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{EABAA}		6.4		7.6		8.8	ns
t _{EABRCOMB}	6.4		7.6		8.8		ns
t _{EABRCREG}	4.4		5.1		6.0		ns
t _{EABWP}	2.5		2.9		3.3		ns
t _{EABWCOMB}	6.0		7.0		8.0		ns
t _{EABWCREG}	6.8		7.8		9.0		ns
t _{EABDD}		5.7		6.7		7.7	ns
t _{EABDATACO}		0.8		0.9		1.1	ns
t _{EABDATASU}	1.5		1.7		2.0		ns
t _{EABDATAH}	0.0		0.0		0.0		ns
t _{EABWESU}	1.3		1.4		1.7		ns
t _{EABWEH}	0.0		0.0		0.0		ns
t _{EABWDSU}	1.5		1.7		2.0		ns
t _{EABWDH}	0.0		0.0		0.0		ns
t _{EABWASU}	3.0		3.6		4.3		ns
t _{EABWAH}	0.5		0.5		0.4		ns
t _{EABWO}		5.1		6.0		6.8	ns

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		1.7		2.4		3.2	ns
t _{EABDATA2}		0.4		0.6		0.8	ns
t _{EABWE1}		1.0		1.4		1.9	ns
t _{EABWE2}		0.0		0.0		0.0	ns
t _{EABRE1}		0.0		0.0		0.0	
t _{EABRE2}		0.4		0.6		0.8	
t _{EABCLK}		0.0		0.0		0.0	ns
t _{EABCO}		0.8		1.1		1.5	ns
t _{EABBYPASS}		0.0		0.0		0.0	ns
t _{EABSU}	0.7		1.0		1.3		ns
t _{EABH}	0.4		0.6		0.8		ns
t _{EABCLR}	0.8		1.1		1.5		
t _{AA}		2.0		2.8		3.8	ns
t _{WP}	2.0		2.8		3.8		ns
t _{RP}	1.0		1.4		1.9		
t _{WDSU}	0.5		0.7		0.9		ns
t _{WDH}	0.1		0.1		0.2		ns
t _{WASU}	1.0		1.4		1.9		ns
t _{WAH}	1.5		2.1		2.9		ns
t _{RASU}	1.5		2.1		2.8		
t _{RAH}	0.1		0.1		0.2		
t _{WO}		2.1		2.9		4.0	ns
t _{DD}		2.1		2.9		4.0	ns
t _{EABOUT}		0.0		0.0		0.0	ns
t _{EABCH}	1.5		2.0		2.5		ns
t _{EABCL}	1.5		2.0		2.5	İ	ns

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{EABAA}		3.7		5.2		7.0	ns
t _{EABRCCOMB}	3.7		5.2		7.0		ns
t _{EABRCREG}	3.5		4.9		6.6		ns
t _{EABWP}	2.0		2.8		3.8		ns
t _{EABWCCOMB}	4.5		6.3		8.6		ns
t _{EABWCREG}	5.6		7.8		10.6		ns
t _{EABDD}		3.8		5.3		7.2	ns
t _{EABDATACO}		0.8		1.1		1.5	ns
t _{EABDATASU}	1.1		1.6		2.1		ns
t _{EABDATAH}	0.0		0.0		0.0		ns
t _{EABWESU}	0.7		1.0		1.3		ns
t _{EABWEH}	0.4		0.6		0.8		ns
t _{EABWDSU}	1.2		1.7		2.2		ns
t _{EABWDH}	0.0		0.0		0.0		ns
t _{EABWASU}	1.6		2.3		3.0		ns
t _{EABWAH}	0.9		1.2		1.8		ns
t _{EABWO}		3.1		4.3		5.9	ns

Table 70. EPF10	K50S Device	Interconnec	t Timing Mi	croparamete	e rs Note	(1)	
Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{DIN2IOE}		3.1		3.7		4.6	ns
t _{DIN2LE}		1.7		2.1		2.7	ns
t _{DIN2DATA}		2.7		3.1		5.1	ns
t _{DCLK2IOE}		1.6		1.9		2.6	ns
t _{DCLK2LE}		1.7		2.1		2.7	ns
t _{SAMELAB}		0.1		0.1		0.2	ns
t _{SAMEROW}		1.5		1.7		2.4	ns
t _{SAMECOLUMN}		1.0		1.3		2.1	ns
t _{DIFFROW}		2.5		3.0		4.5	ns
t _{TWOROWS}		4.0		4.7		6.9	ns
t _{LEPERIPH}		2.6		2.9		3.4	ns
t _{LABCARRY}		0.1		0.2		0.2	ns
t _{LABCASC}		0.8		1.0		1.3	ns

During initialization, which occurs immediately after configuration, the device resets registers, enables I/O pins, and begins to operate as a logic device. The I/O pins are tri-stated during power-up, and before and during configuration. Together, the configuration and initialization processes are called *command mode*; normal device operation is called *user mode*.

SRAM configuration elements allow FLEX 10KE devices to be reconfigured in-circuit by loading new configuration data into the device. Real-time reconfiguration is performed by forcing the device into command mode with a device pin, loading different configuration data, reinitializing the device, and resuming user-mode operation. The entire reconfiguration process requires less than 85 ms and can be used to reconfigure an entire system dynamically. In-field upgrades can be performed by distributing new configuration files.

Before and during configuration, all I/O pins (except dedicated inputs, clock, or configuration pins) are pulled high by a weak pull-up resistor.

Programming Files

Despite being function- and pin-compatible, FLEX 10KE devices are not programming- or configuration file-compatible with FLEX 10K or FLEX 10KA devices. A design therefore must be recompiled before it is transferred from a FLEX 10K or FLEX 10KA device to an equivalent FLEX 10KE device. This recompilation should be performed both to create a new programming or configuration file and to check design timing in FLEX 10KE devices, which has different timing characteristics than FLEX 10K or FLEX 10KA devices.

FLEX 10KE devices are generally pin-compatible with equivalent FLEX 10KA devices. In some cases, FLEX 10KE devices have fewer I/O pins than the equivalent FLEX 10KA devices. Table 81 shows which FLEX 10KE devices have fewer I/O pins than equivalent FLEX 10KA devices. However, power, ground, JTAG, and configuration pins are the same on FLEX 10KA and FLEX 10KE devices, enabling migration from a FLEX 10KA design to a FLEX 10KE design.



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100