Intel - EPF10K50EQC240-2 Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	360
Number of Logic Elements/Cells	2880
Total RAM Bits	40960
Number of I/O	189
Number of Gates	199000
Voltage - Supply	2.3V ~ 2.7V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	240-BFQFP
Supplier Device Package	240-PQFP (32x32)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k50eqc240-2

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Functional Description

Each FLEX 10KE device contains an enhanced embedded array to implement memory and specialized logic functions, and a logic array to implement general logic.

The embedded array consists of a series of EABs. When implementing memory functions, each EAB provides 4,096 bits, which can be used to create RAM, ROM, dual-port RAM, or first-in first-out (FIFO) functions. When implementing logic, each EAB can contribute 100 to 600 gates towards complex logic functions, such as multipliers, microcontrollers, state machines, and DSP functions. EABs can be used independently, or multiple EABs can be combined to implement larger functions.

The logic array consists of logic array blocks (LABs). Each LAB contains eight LEs and a local interconnect. An LE consists of a four-input look-up table (LUT), a programmable flipflop, and dedicated signal paths for carry and cascade functions. The eight LEs can be used to create medium-sized blocks of logic—such as 8-bit counters, address decoders, or state machines—or combined across LABs to create larger logic blocks. Each LAB represents about 96 usable gates of logic.

Signal interconnections within FLEX 10KE devices (as well as to and from device pins) are provided by the FastTrack Interconnect routing structure, which is a series of fast, continuous row and column channels that run the entire length and width of the device.

Each I/O pin is fed by an I/O element (IOE) located at the end of each row and column of the FastTrack Interconnect routing structure. Each IOE contains a bidirectional I/O buffer and a flipflop that can be used as either an output or input register to feed input, output, or bidirectional signals. When used with a dedicated clock pin, these registers provide exceptional performance. As inputs, they provide setup times as low as 0.9 ns and hold times of 0 ns. As outputs, these registers provide clock-to-output times as low as 3.0 ns. IOEs provide a variety of features, such as JTAG BST support, slew-rate control, tri-state buffers, and open-drain outputs.

The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock, clear, and preset control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the output of the LE.

The LE has two outputs that drive the interconnect: one drives the local interconnect and the other drives either the row or column FastTrack Interconnect routing structure. The two outputs can be controlled independently. For example, the LUT can drive one output while the register drives the other output. This feature, called register packing, can improve LE utilization because the register and the LUT can be used for unrelated functions.

The FLEX 10KE architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. The carry chain supports high-speed counters and adders and the cascade chain implements wide-input functions with minimum delay. Carry and cascade chains connect all LEs in a LAB as well as all LABs in the same row. Intensive use of carry and cascade chains can reduce routing flexibility. Therefore, the use of these chains should be limited to speed-critical portions of a design.

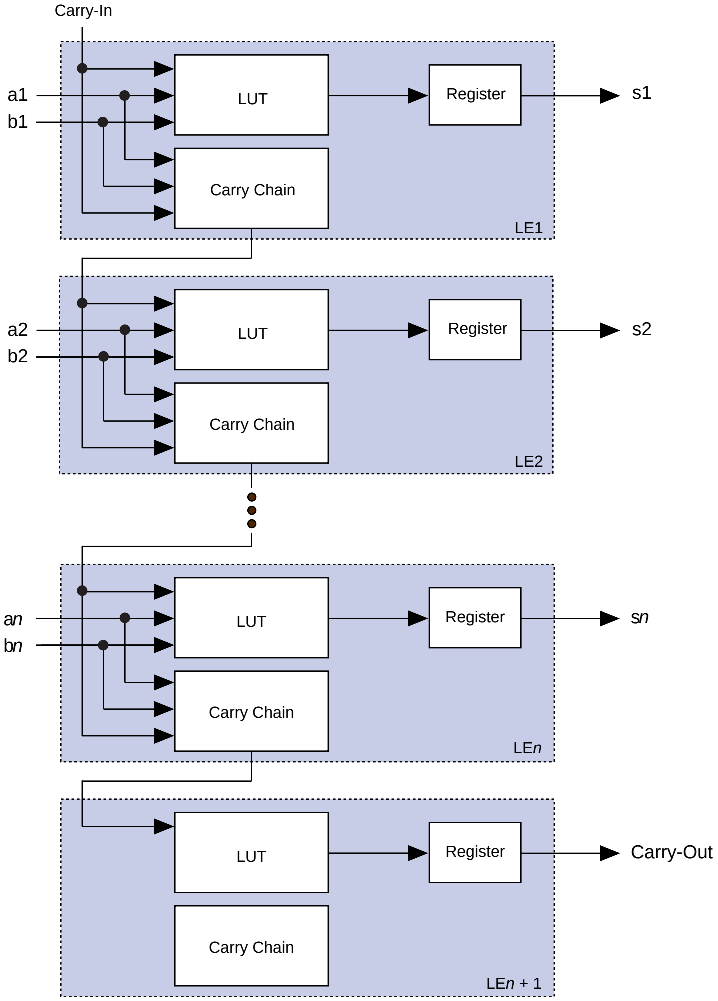
Carry Chain

The carry chain provides a very fast (as low as 0.2 ns) carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the FLEX 10KE architecture to implement high-speed counters, adders, and comparators of arbitrary width efficiently. Carry chain logic can be created automatically by the Altera Compiler during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of carry chains.

Carry chains longer than eight LEs are automatically implemented by linking LABs together. For enhanced fitting, a long carry chain skips alternate LABs in a row. A carry chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB. For example, the last LE of the first LAB in a row carries to the first LE of the third LAB in the row. The carry chain does not cross the EAB at the middle of the row. For instance, in the EPF10K50E device, the carry chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB.

Figure 9 shows how an *n*-bit full adder can be implemented in *n* + 1 LEs with the carry chain. One portion of the LUT generates the sum of two bits using the input signals and the carry-in signal; the sum is routed to the output of the LE. The register can be bypassed for simple adders or used for an accumulator function. Another portion of the LUT and the carry chain logic generates the carry-out signal, which is routed directly to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it can be used as a general-purpose signal.

Figure 9. FLEX 10KE Carry Chain Operation (n-Bit Full Adder)



Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is `AND`ed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-states without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (`OE`) signals select which signal drives the bus. However, if multiple `OE` signals are active, contending signals can be driven onto the bus. Conversely, if no `OE` signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the `DATA3`, `LABCTRL1`, and `LABCTRL2` inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either `LABCTRL1` or `LABCTRL2` can control the asynchronous clear. Alternatively, the register can be set up so that `LABCTRL1` implements an asynchronous load. The data to be loaded is driven to `DATA3`; when `LABCTRL1` is asserted, `DATA3` is loaded into the register.

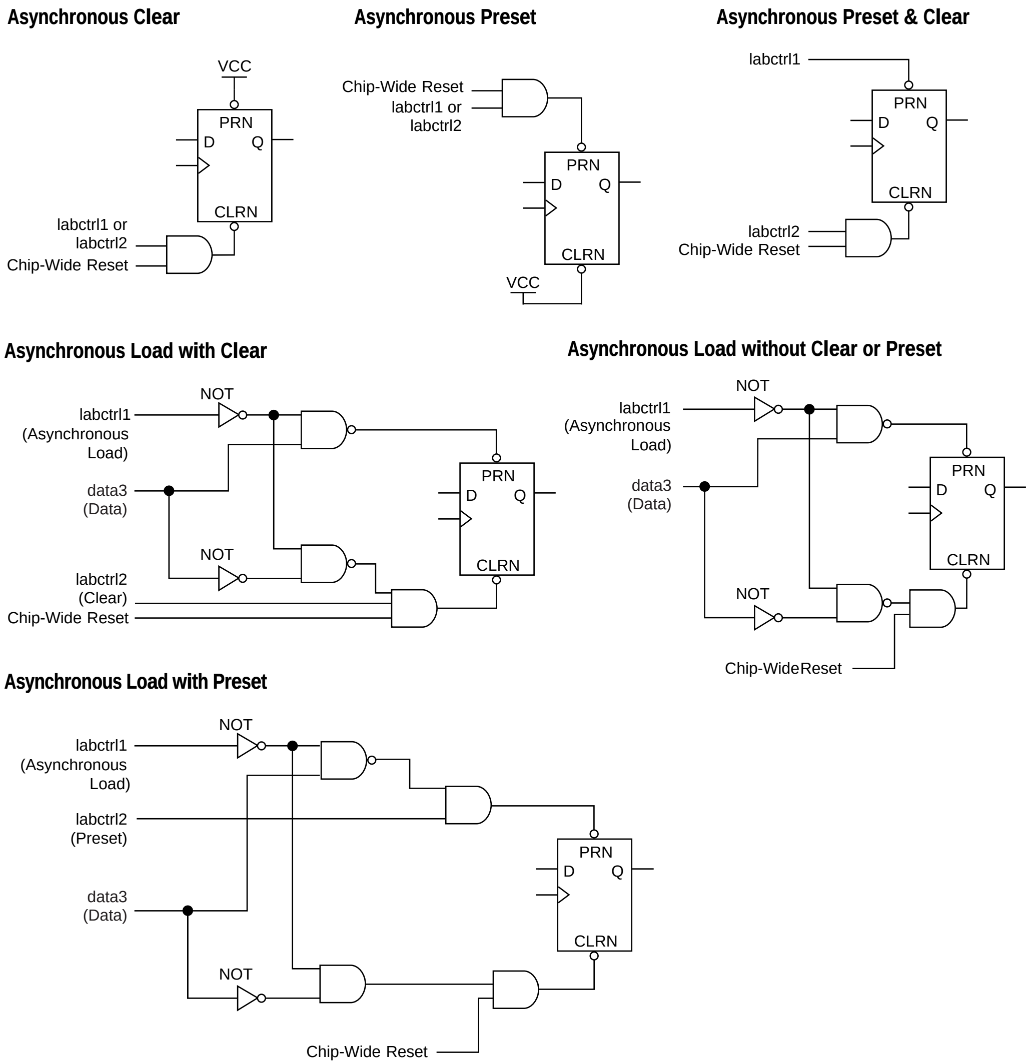
During compilation, the Altera Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

In addition to the six clear and preset modes, FLEX 10KE devices provide a chip-wide reset pin that can reset all registers in the device. Use of this feature is set during design entry. In any of the clear and preset modes, the chip-wide reset overrides all other signals. Registers with asynchronous presets may be preset when the chip-wide reset is asserted. Inversion can be used to implement the asynchronous preset. Figure 12 shows examples of how to setup the preset and clear inputs for the desired functionality.

Figure 12. FLEX 10KE LE Clear & Preset Modes


Asynchronous Clear

Asynchronous Preset

Asynchronous Preset & Clear

Asynchronous Load with Clear

Asynchronous Load without Clear or Preset

Asynchronous Load with Preset

Row-to-IOE Connections

When an IOE is used as an input signal, it can drive two separate row channels. The signal is accessible by all LEs within that row. When an IOE is used as an output, the signal is driven by a multiplexer that selects a signal from the row channels. Up to eight IOEs connect to each side of each row channel (see Figure 16).

Figure 16. FLEX 10KE Row-to-IOE Connections

The values for m and n are provided in Table 10.

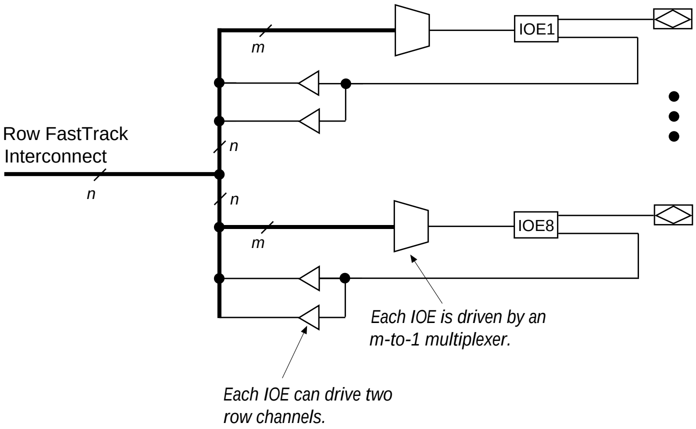


Table 10 lists the FLEX 10KE row-to-IOE interconnect resources.

Table 10. FLEX 10KE Row-to-IOE Interconnect Resources

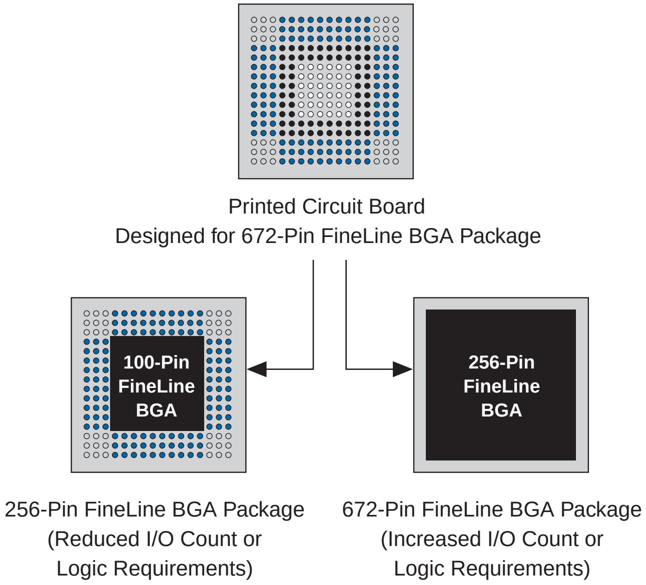
Device	Channels per Row *(n)*	Row Channels per Pin *(m)*
EPF10K30E	216	27
EPF10K50E EPF10K50S	216	27
EPF10K100E	312	39
EPF10K130E	312	39
EPF10K200E EPF10K200S	312	39

SameFrame Pin-Outs

FLEX 10KE devices support the SameFrame pin-out feature for FineLine BGA packages. The SameFrame pin-out feature is the arrangement of balls on FineLine BGA packages such that the lower-ball-count packages form a subset of the higher-ball-count packages. SameFrame pin-outs provide the flexibility to migrate not only from device to device within the same package, but also from one package to another. A given printed circuit board (PCB) layout can support multiple device density / package combinations. For example, a single board layout can support a range of devices from an EPF10K30E device in a 256-pin FineLine BGA package to an EPF10K200S device in a 672-pin FineLine BGA package.

The Altera software provides support to design PCBs with SameFrame pin-out devices. Devices can be defined for present and future use. The Altera software generates pin-outs describing how to lay out a board to take advantage of this migration (see Figure 18).

Figure 18. SameFrame Pin-Out Example



Tables 12 and 13 summarize the ClockLock and ClockBoost parameters for -1 and -2 speed-grade devices, respectively.

Table 12. ClockLock & ClockBoost Parameters for -1 Speed-Grade Devices

Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_R	Input rise time				5	ns
t_F	Input fall time				5	ns
t_{INDUTY}	Input duty cycle		40		60	%
f_{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)		25		180	MHz
f_{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)		16		90	MHz
f_{CLKDEV}	Input deviation from user specification in the MAX+PLUS II software *(1)*				25,000 *(2)*	PPM
$t_{INCLKSTB}$	Input clock stability (measured between adjacent clocks)				100	ps
t_{LOCK}	Time required for ClockLock or ClockBoost to acquire lock *(3)*				10	µs
t_{JITTER}	Jitter on ClockLock or ClockBoost-generated clock *(4)*	$t_{INCLKSTB} < 100$			250	ps
		$t_{INCLKSTB} < 50$			200 *(4)*	ps
$t_{OUTDUTY}$	Duty cycle for ClockLock or ClockBoost-generated clock		40	50	60	%

Table 22. FLEX 10KE 2.5-V Device DC Operating Conditions *Notes (6), (7)*

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{IH}	High-level input voltage		1.7, 0.5 × V_{CCIO} *(8)*		5.75	V
V_{IL}	Low-level input voltage		–0.5		0.8, 0.3 × V_{CCIO} *(8)*	V
V_{OH}	3.3-V high-level TTL output voltage	I_{OH} = –8 mA DC, V_{CCIO} = 3.00 V *(9)*	2.4			V
	3.3-V high-level CMOS output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 3.00 V *(9)*	V_{CCIO} – 0.2			V
	3.3-V high-level PCI output voltage	I_{OH} = –0.5 mA DC, V_{CCIO} = 3.00 to 3.60 V *(9)*	0.9 × V_{CCIO}			V
	2.5-V high-level output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 2.30 V *(9)*	2.1			V
		I_{OH} = –1 mA DC, V_{CCIO} = 2.30 V *(9)*	2.0			V
		I_{OH} = –2 mA DC, V_{CCIO} = 2.30 V *(9)*	1.7			V
V_{OL}	3.3-V low-level TTL output voltage	I_{OL} = 12 mA DC, V_{CCIO} = 3.00 V *(10)*			0.45	V
	3.3-V low-level CMOS output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 3.00 V *(10)*			0.2	V
	3.3-V low-level PCI output voltage	I_{OL} = 1.5 mA DC, V_{CCIO} = 3.00 to 3.60 V *(10)*			0.1 × V_{CCIO}	V
	2.5-V low-level output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 2.30 V *(10)*			0.2	V
		I_{OL} = 1 mA DC, V_{CCIO} = 2.30 V *(10)*			0.4	V
		I_{OL} = 2 mA DC, V_{CCIO} = 2.30 V *(10)*			0.7	V
I_I	Input pin leakage current	V_I = $V_{CCIOmax}$ to 0 V *(11)*	–10		10	μA
I_{OZ}	Tri-stated I/O pin leakage current	V_O = $V_{CCIOmax}$ to 0 V *(11)*	–10		10	μA
I_{CC0}	V_{CC} supply current (standby)	V_I = ground, no load, no toggling inputs		5		mA
		V_I = ground, no load, no toggling inputs *(12)*		10		mA
R_{CONF}	Value of I/O pin pull-up resistor before and during configuration	V_{CCIO} = 3.0 V *(13)*	20		50	k¾
		V_{CCIO} = 2.3 V *(13)*	30		80	k¾

Table 23. FLEX 10KE Device Capacitance Note (14)					
Symbol	**Parameter**	**Conditions**	**Min**	**Max**	**Unit**
C_{IN}	Input capacitance	V_{IN} = 0 V, f = 1.0 MHz		10	pF
C_{INCLK}	Input capacitance on dedicated clock pin	V_{IN} = 0 V, f = 1.0 MHz		12	pF
C_{OUT}	Output capacitance	V_{OUT} = 0 V, f = 1.0 MHz		10	pF

Notes to tables:

(1) See the *Operating Requirements for Altera Devices Data Sheet*.

(2) Minimum DC input voltage is –0.5 V. During transitions, the inputs may undershoot to –2.0 V for input currents less than 100 mA and periods shorter than 20 ns.

(3) Numbers in parentheses are for industrial-temperature-range devices.

(4) Maximum V_{CC} rise time is 100 ms, and V_{CC} must rise monotonically.

(5) All pins, including dedicated inputs, clock, I/O, and JTAG pins, may be driven before V_{CCINT} and V_{CCIO} are powered.

(6) Typical values are for T_A = 25° C, V_{CCINT} = 2.5 V, and V_{CCIO} = 2.5 V or 3.3 V.

(7) These values are specified under the FLEX 10KE Recommended Operating Conditions shown in Tables 20 and 21.

(8) The FLEX 10KE input buffers are compatible with 2.5-V, 3.3-V (LVTTL and LVCMOS), and 5.0-V TTL and CMOS signals. Additionally, the input buffers are 3.3-V PCI compliant when V_{CCIO} and V_{CCINT} meet the relationship shown in Figure 22.

(9) The I_{OH} parameter refers to high-level TTL, PCI, or CMOS output current.

(10) The I_{OL} parameter refers to low-level TTL, PCI, or CMOS output current. This parameter applies to open-drain pins as well as output pins.

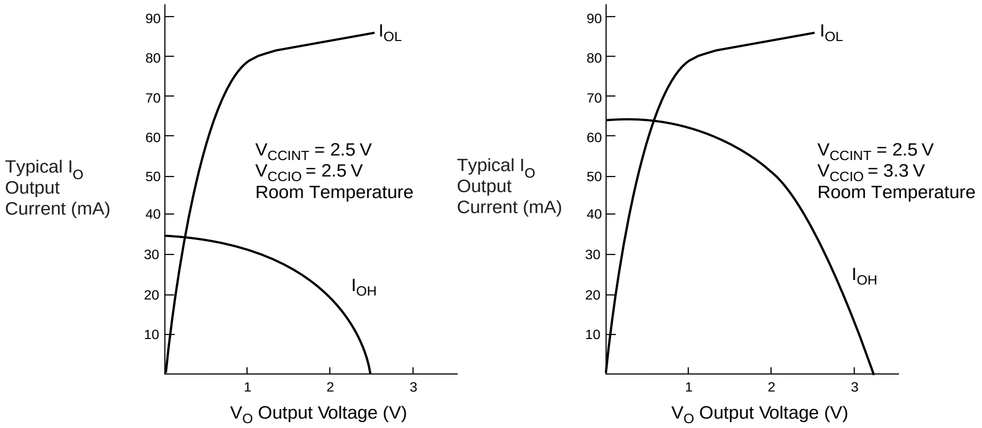
(11) This value is specified for normal device operation. The value may vary during power-up.

(12) This parameter applies to -1 speed-grade commercial-temperature devices and -2 speed-grade-industrial temperature devices.

(13) Pin pull-up resistance values will be lower if the pin is driven higher than V_{CCIO} by an external source.

(14) Capacitance is sample-tested only.

Figure 23. Output Drive Characteristics of FLEX 10KE Devices *Note (1)*



Note:

(1) These are transient (AC) currents.

Timing Model

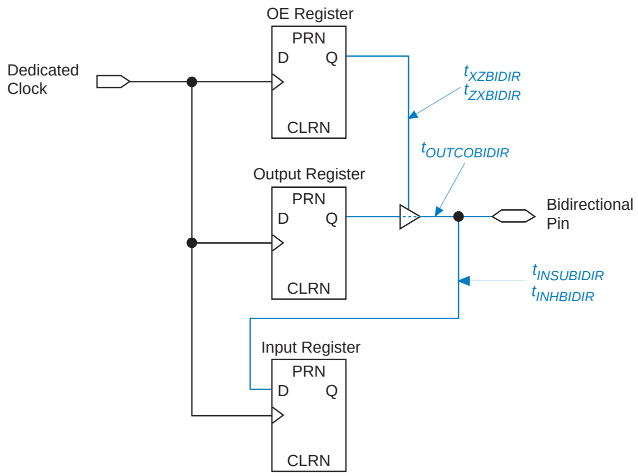
The continuous, high-performance FastTrack Interconnect routing resources ensure predictable performance and accurate simulation and timing analysis. This predictable performance contrasts with that of FPGAs, which use a segmented connection scheme and therefore have unpredictable performance.

Device performance can be estimated by following the signal path from a source, through the interconnect, to the destination. For example, the registered performance between two LEs on the same row can be calculated by adding the following parameters:

- LE register clock-to-output delay (t_{CO})
- Interconnect delay ($t_{SAMEROW}$)
- LE look-up table delay (t_{LUT})
- LE register setup time (t_{SU})

The routing delay depends on the placement of the source and destination LEs. A more complex registered path may involve multiple combinatorial LEs between the source and destination LEs.

Figure 28. Synchronous Bidirectional Pin External Timing Model



Tables 24 through 28 describe the FLEX 10KE device internal timing parameters. Tables 29 through 30 describe the FLEX 10KE external timing parameters and their symbols.

Table 24. LE Timing Microparameters (Part 1 of 2) Note (1)

Symbol	Parameter	Condition
t_{LUT}	LUT delay for data-in	
t_{CLUT}	LUT delay for carry-in	
t_{RLUT}	LUT delay for LE register feedback	
t_{PACKED}	Data-in to packed register delay	
t_{EN}	LE register enable delay	
t_{CICO}	Carry-in to carry-out delay	
t_{CGEN}	Data-in to carry-out delay	
t_{CGENR}	LE register feedback to carry-out delay	
t_{CASC}	Cascade-in to cascade-out delay	
t_C	LE register control signal delay	
t_{CO}	LE register clock-to-output delay	
t_{COMB}	Combinatorial delay	
t_{SU}	LE register setup time for data and enable signals before clock; LE register recovery time after asynchronous clear, preset, or load	
t_H	LE register hold time for data and enable signals after clock	
t_{PRE}	LE register preset delay	

Table 24. LE Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	Parameter	Condition
t_{CLR}	LE register clear delay	
t_{CH}	Minimum clock high time from clock pin	
t_{CL}	Minimum clock low time from clock pin	

Table 25. IOE Timing Microparameters *Note (1)*

Symbol	Parameter	Conditions
t_{IOD}	IOE data delay	
t_{IOC}	IOE register control signal delay	
t_{IOCO}	IOE register clock-to-output delay	
t_{IOCOMB}	IOE combinatorial delay	
t_{IOSU}	IOE register setup time for data and enable signals before clock; IOE register recovery time after asynchronous clear	
t_{IOH}	IOE register hold time for data and enable signals after clock	
t_{IOCLR}	IOE register clear time	
t_{OD1}	Output buffer and pad delay, slow slew rate = off, V_{CCIO} = 3.3 V	C1 = 35 pF *(2)*
t_{OD2}	Output buffer and pad delay, slow slew rate = off, V_{CCIO} = 2.5 V	C1 = 35 pF *(3)*
t_{OD3}	Output buffer and pad delay, slow slew rate = on	C1 = 35 pF *(4)*
t_{XZ}	IOE output buffer disable delay	
t_{ZX1}	IOE output buffer enable delay, slow slew rate = off, V_{CCIO} = 3.3 V	C1 = 35 pF *(2)*
t_{ZX2}	IOE output buffer enable delay, slow slew rate = off, V_{CCIO} = 2.5 V	C1 = 35 pF *(3)*
t_{ZX3}	IOE output buffer enable delay, slow slew rate = on	C1 = 35 pF *(4)*
t_{INREG}	IOE input pad and buffer to IOE register delay	
t_{IOFD}	IOE register feedback delay	
t_{INCOMB}	IOE input pad and buffer to FastTrack Interconnect delay	

Figures 29 and 30 show the asynchronous and synchronous timing waveforms, respectively, or the EAB macroparameters in Tables 26 and 27.

Figure 29. EAB Asynchronous Timing Waveforms

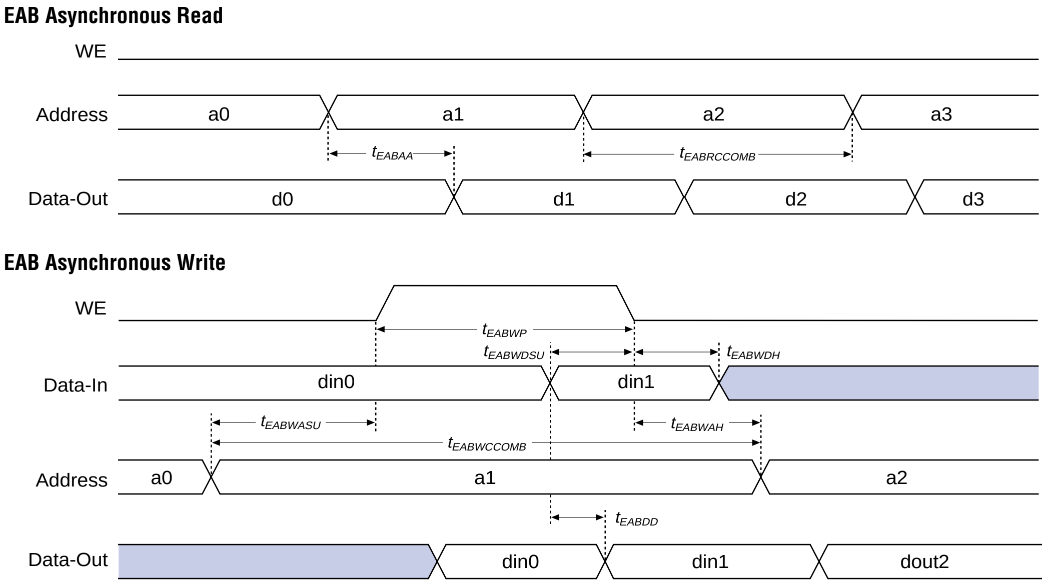


Table 40. EPF10K50E Device EAB Internal Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.7		2.0		2.7	ns
$t_{EABDATA1}$		0.6		0.7		0.9	ns
t_{EABWE1}		1.1		1.3		1.8	ns
t_{EABWE2}		0.4		0.4		0.6	ns
t_{EABRE1}		0.8		0.9		1.2	ns
t_{EABRE2}		0.4		0.4		0.6	ns
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		0.3		0.3		0.5	ns
$t_{EABBYPASS}$		0.5		0.6		0.8	ns
t_{EABSU}	0.9		1.0		1.4		ns
t_{EABH}	0.4		0.4		0.6		ns
t_{EABCLR}	0.3		0.3		0.5		ns
t_{AA}		3.2		3.8		5.1	ns
t_{WP}	2.5		2.9		3.9		ns
t_{RP}	0.9		1.1		1.5		ns
t_{WDSU}	0.9		1.0		1.4		ns
t_{WDH}	0.1		0.1		0.2		ns
t_{WASU}	1.7		2.0		2.7		ns
t_{WAH}	1.8		2.1		2.9		ns
t_{RASU}	3.1		3.7		5.0		ns
t_{RAH}	0.2		0.2		0.3		ns
t_{WO}		2.5		2.9		3.9	ns
t_{DD}		2.5		2.9		3.9	ns
t_{EABOUT}		0.5		0.6		0.8	ns
t_{EABCH}	1.5		2.0		2.5		ns
t_{EABCL}	2.5		2.9		3.9		ns

Table 41. EPF10K50E Device EAB Internal Timing Macroparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{EABAA}		6.4		7.6		10.2	ns
$t_{EABRCOMB}$	6.4		7.6		10.2		ns
$t_{EABRCREG}$	4.4		5.1		7.0		ns
t_{EABWP}	2.5		2.9		3.9		ns
$t_{EABWCOMB}$	6.0		7.0		9.5		ns
$t_{EABWCREG}$	6.8		7.8		10.6		ns
t_{EABDD}		5.7		6.7		9.0	ns
$t_{EABDATACO}$		0.8		0.9		1.3	ns
$t_{EABDATASU}$	1.5		1.7		2.3		ns
$t_{EABDATAH}$	0.0		0.0		0.0		ns
$t_{EABWESU}$	1.3		1.4		2.0		ns
t_{EABWEH}	0.0		0.0		0.0		ns
$t_{EABWDSU}$	1.5		1.7		2.3		ns
t_{EABWDH}	0.0		0.0		0.0		ns
$t_{EABWASU}$	3.0		3.6		4.8		ns
t_{EABWAH}	0.5		0.5		0.8		ns
t_{EABWO}		5.1		6.0		8.1	ns

Table 42. EPF10K50E Device Interconnect Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{DIN2IOE}$		3.5		4.3		5.6	ns
t_{DIN2LE}		2.1		2.5		3.4	ns
$t_{DIN2DATA}$		2.2		2.4		3.1	ns
$t_{DCLK2IOE}$		2.9		3.5		4.7	ns
$t_{DCLK2LE}$		2.1		2.5		3.4	ns
$t_{SAMELAB}$		0.1		0.1		0.2	ns
$t_{SAMEROW}$		1.1		1.1		1.5	ns
$t_{SAMECOLUMN}$		0.8		1.0		1.3	ns
$t_{DIFFROW}$		1.9		2.1		2.8	ns
$t_{TWOROWS}$		3.0		3.2		4.3	ns
$t_{LEPERIPH}$		3.1		3.3		3.7	ns
$t_{LABCARRY}$		0.1		0.1		0.2	ns
$t_{LABCASC}$		0.3		0.3		0.5	ns

Tables 52 through 58 show EPF10K130E device internal and external timing parameters.

Table 52. EPF10K130E Device LE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.6		0.9		1.3	ns
t_{CLUT}		0.6		0.8		1.0	ns
t_{RLUT}		0.7		0.9		0.2	ns
t_{PACKED}		0.3		0.5		0.6	ns
t_{EN}		0.2		0.3		0.4	ns
t_{CICO}		0.1		0.1		0.2	ns
t_{CGEN}		0.4		0.6		0.8	ns
t_{CGENR}		0.1		0.1		0.2	ns
t_{CASC}		0.6		0.9		1.2	ns
t_C		0.3		0.5		0.6	ns
t_{CO}		0.5		0.7		0.8	ns
t_{COMB}		0.3		0.5		0.6	ns
t_{SU}	0.5		0.7		0.8		ns
t_H	0.6		0.7		1.0		ns
t_{PRE}		0.9		1.2		1.6	ns
t_{CLR}		0.9		1.2		1.6	ns
t_{CH}	1.5		1.5		2.5		ns
t_{CL}	1.5		1.5		2.5		ns

Table 53. EPF10K130E Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.3		1.5		2.0	ns
t_{IOC}		0.0		0.0		0.0	ns
t_{IOCO}		0.6		0.8		1.0	ns
t_{IOCOMB}		0.6		0.8		1.0	ns
t_{IOSU}	1.0		1.2		1.6		ns
t_{IOH}	0.9		0.9		1.4		ns
t_{IOCLR}		0.6		0.8		1.0	ns
t_{OD1}		2.8		4.1		5.5	ns
t_{OD2}		2.8		4.1		5.5	ns

Table 58. EPF10K130E External Bidirectional Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{INSUBIDIR}$ (3)	2.2		2.4		3.2		ns
$t_{INHBIDIR}$ (3)	0.0		0.0		0.0		ns
$t_{INSUBIDIR}$ (4)	2.8		3.0		–		ns
$t_{INHBIDIR}$ (4)	0.0		0.0		–		ns
$t_{OUTCOBIDIR}$ (3)	2.0	5.0	2.0	7.0	2.0	9.2	ns
$t_{XZBIDIR}$ (3)		5.6		8.1		10.8	ns
$t_{ZXBIDIR}$ (3)		5.6		8.1		10.8	ns
$t_{OUTCOBIDIR}$ (4)	0.5	4.0	0.5	6.0	–	–	ns
$t_{XZBIDIR}$ (4)		4.6		7.1		–	ns
$t_{ZXBIDIR}$ (4)		4.6		7.1		–	ns

Notes to tables:

(1) All timing parameters are described in Tables 24 through 30 in this data sheet.
(2) These parameters are specified by characterization.
(3) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
(4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Tables 59 through 65 show EPF10K200E device internal and external timing parameters.

Table 59. EPF10K200E Device LE Timing Microparameters *(Part 1 of 2)* *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.7		0.8		1.2	ns
t_{CLUT}		0.4		0.5		0.6	ns
t_{RLUT}		0.6		0.7		0.9	ns
t_{PACKED}		0.3		0.5		0.7	ns
t_{EN}		0.4		0.5		0.6	ns
t_{CICO}		0.2		0.2		0.3	ns
t_{CGEN}		0.4		0.4		0.6	ns
t_{CGENR}		0.2		0.2		0.3	ns
t_{CASC}		0.7		0.8		1.2	ns
t_C		0.5		0.6		0.8	ns
t_{CO}		0.5		0.6		0.8	ns
t_{COMB}		0.4		0.6		0.8	ns
t_{SU}	0.4		0.6		0.7		ns

Power Consumption

The supply power (P) for FLEX 10KE devices can be calculated with the following equation:

$$P = P_{INT} + P_{IO} = (I_{CCSTANDBY} + I_{CCACTIVE}) \times V_{CC} + P_{IO}$$

The $I_{CCACTIVE}$ value depends on the switching frequency and the application logic. This value is calculated based on the amount of current that each LE typically consumes. The P_{IO} value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

Compared to the rest of the device, the embedded array consumes a negligible amount of power. Therefore, the embedded array can be ignored when calculating supply current.

The $I_{CCACTIVE}$ value can be calculated with the following equation:

$$I_{CCACTIVE} = K \times \mathbf{f_{MAX}} \times N \times \mathbf{tog_{LC}} \times \frac{\mu A}{MHz \times LE}$$

Where:

$\mathbf{f_{MAX}}$ = Maximum operating frequency in MHz
N = Total number of LEs used in the device
$\mathbf{tog_{LC}}$ = Average percent of LEs toggling at each clock (typically 12.5%)
K = Constant

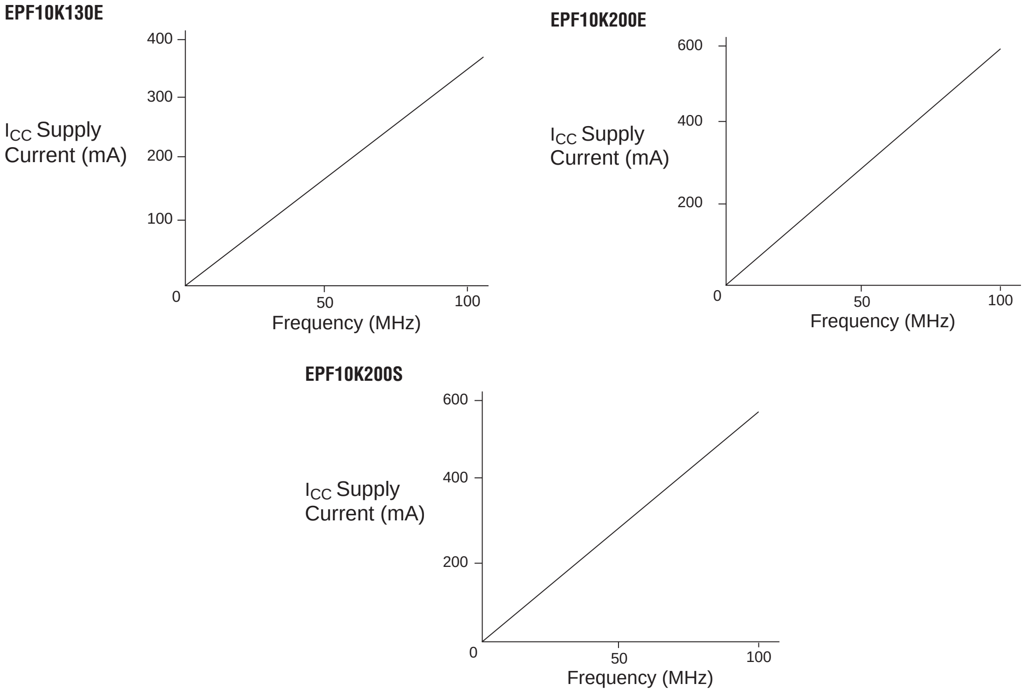
Table 80 provides the constant (K) values for FLEX 10KE devices.

Table 80. FLEX 10KE K Constant Values

Device	K Value
EPF10K30E	4.5
EPF10K50E	4.8
EPF10K50S	4.5
EPF10K100E	4.5
EPF10K130E	4.6
EPF10K200E	4.8
EPF10K200S	4.6

This calculation provides an I_{CC} estimate based on typical conditions with no output load. The actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

Figure 31. FLEX 10KE $I_{CCACTIVE}$ vs. Operating Frequency (Part 2 of 2)



Configuration & Operation

The FLEX 10KE architecture supports several configuration schemes. This section summarizes the device operating modes and available device configuration schemes.

Operating Modes

The FLEX 10KE architecture uses SRAM configuration elements that require configuration data to be loaded every time the circuit powers up. The process of physically loading the SRAM data into the device is called *configuration*. Before configuration, as V_{CC} rises, the device initiates a Power-On Reset (POR). This POR event clears the device and prepares it for configuration. The FLEX 10KE POR time does not exceed 50 µs.

When configuring with a configuration device, refer to the respective configuration device data sheet for POR timing information.