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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	360
Number of Logic Elements/Cells	2880
Total RAM Bits	40960
Number of I/O	189
Number of Gates	199000
Voltage - Supply	2.3V ~ 2.7V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	240-BFQFP
Supplier Device Package	240-PQFP (32x32)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k50eqc240-2aa

- Software design support and automatic place-and-route provided by Altera's development systems for Windows-based PCs and Sun SPARCstation, and HP 9000 Series 700/800
- Flexible package options
 - Available in a variety of packages with 144 to 672 pins, including the innovative FineLine BGA™ packages (see [Tables 3 and 4](#))
 - SameFrame™ pin-out compatibility between FLEX 10KA and FLEX 10KE devices across a range of device densities and pin counts
- Additional design entry and simulation support provided by EDIF 2.0.0 and 3.0.0 netlist files, library of parameterized modules (LPM), DesignWare components, Verilog HDL, VHDL, and other interfaces to popular EDA tools from manufacturers such as Cadence, Exemplar Logic, Mentor Graphics, OrCAD, Synopsys, Synplcity, VeriBest, and Viewlogic

Table 3. FLEX 10KE Package Options & I/O Pin Count *Notes (1), (2)*

Device	144-Pin TQFP	208-Pin PQFP	240-Pin PQFP RQFP	256-Pin FineLine BGA	356-Pin BGA	484-Pin FineLine BGA	599-Pin PGA	600-Pin BGA	672-Pin FineLine BGA
EPF10K30E	102	147		176		220			220 (3)
EPF10K50E	102	147	189	191		254			254 (3)
EPF10K50S	102	147	189	191	220	254			254 (3)
EPF10K100E		147	189	191	274	338			338 (3)
EPF10K130E			186		274	369		424	413
EPF10K200E							470	470	470
EPF10K200S			182		274	369	470	470	470

Notes:

- (1) FLEX 10KE device package types include thin quad flat pack (TQFP), plastic quad flat pack (PQFP), power quad flat pack (RQFP), pin-grid array (PGA), and ball-grid array (BGA) packages.
- (2) Devices in the same package are pin-compatible, although some devices have more I/O pins than others. When planning device migration, use the I/O pins that are common to all devices.
- (3) This option is supported with a 484-pin FineLine BGA package. By using SameFrame pin migration, all FineLine BGA packages are pin-compatible. For example, a board can be designed to support 256-pin, 484-pin, and 672-pin FineLine BGA packages. The Altera software automatically avoids conflicting pins when future migration is set.

Similar to the FLEX 10KE architecture, embedded gate arrays are the fastest-growing segment of the gate array market. As with standard gate arrays, embedded gate arrays implement general logic in a conventional “sea-of-gates” architecture. Additionally, embedded gate arrays have dedicated die areas for implementing large, specialized functions. By embedding functions in silicon, embedded gate arrays reduce die area and increase speed when compared to standard gate arrays. While embedded megafunctions typically cannot be customized, FLEX 10KE devices are programmable, providing the designer with full control over embedded megafunctions and general logic, while facilitating iterative design changes during debugging.

Each FLEX 10KE device contains an embedded array and a logic array. The embedded array is used to implement a variety of memory functions or complex logic functions, such as digital signal processing (DSP), wide data-path manipulation, microcontroller applications, and data-transformation functions. The logic array performs the same function as the sea-of-gates in the gate array and is used to implement general logic such as counters, adders, state machines, and multiplexers. The combination of embedded and logic arrays provides the high performance and high density of embedded gate arrays, enabling designers to implement an entire system on a single device.

FLEX 10KE devices are configured at system power-up with data stored in an Altera serial configuration device or provided by a system controller. Altera offers the EPC1, EPC2, and EPC16 configuration devices, which configure FLEX 10KE devices via a serial data stream. Configuration data can also be downloaded from system RAM or via the Altera BitBlaster™, ByteBlasterMV™, or MasterBlaster download cables. After a FLEX 10KE device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Because reconfiguration requires less than 85 ms, real-time changes can be made during system operation.

FLEX 10KE devices contain an interface that permits microprocessors to configure FLEX 10KE devices serially or in-parallel, and synchronously or asynchronously. The interface also enables microprocessors to treat a FLEX 10KE device as memory and configure it by writing to a virtual memory location, making it easy to reconfigure the device.



For more information on FLEX device configuration, see the following documents:

- *Configuration Devices for APEX & FLEX Devices Data Sheet*
- *BitBlaster Serial Download Cable Data Sheet*
- *ByteBlasterMV Parallel Port Download Cable Data Sheet*
- *MasterBlaster Download Cable Data Sheet*
- *Application Note 116 (Configuring APEX 20K, FLEX 10K, & FLEX 6000 Devices)*

FLEX 10KE devices are supported by the Altera development systems, which are integrated packages that offer schematic, text (including AHDL), and waveform design entry, compilation and logic synthesis, full simulation and worst-case timing analysis, and device configuration. The Altera software provides EDIF 2 0 0 and 3 0 0, LPM, VHDL, Verilog HDL, and other interfaces for additional design entry and simulation support from other industry-standard PC- and UNIX workstation-based EDA tools.

The Altera software works easily with common gate array EDA tools for synthesis and simulation. For example, the Altera software can generate Verilog HDL files for simulation with tools such as Cadence Verilog-XL. Additionally, the Altera software contains EDA libraries that use device-specific features such as carry chains, which are used for fast counter and arithmetic functions. For instance, the Synopsys Design Compiler library supplied with the Altera development system includes DesignWare functions that are optimized for the FLEX 10KE architecture.

The Altera development system runs on Windows-based PCs and Sun SPARCstation, and HP 9000 Series 700/800.



See the *MAX+PLUS II Programmable Logic Development System & Software Data Sheet* and the *Quartus Programmable Logic Development System & Software Data Sheet* for more information.

Functional Description

Each FLEX 10KE device contains an enhanced embedded array to implement memory and specialized logic functions, and a logic array to implement general logic.

The embedded array consists of a series of EABs. When implementing memory functions, each EAB provides 4,096 bits, which can be used to create RAM, ROM, dual-port RAM, or first-in first-out (FIFO) functions. When implementing logic, each EAB can contribute 100 to 600 gates towards complex logic functions, such as multipliers, microcontrollers, state machines, and DSP functions. EABs can be used independently, or multiple EABs can be combined to implement larger functions.

The logic array consists of logic array blocks (LABs). Each LAB contains eight LEs and a local interconnect. An LE consists of a four-input look-up table (LUT), a programmable flipflop, and dedicated signal paths for carry and cascade functions. The eight LEs can be used to create medium-sized blocks of logic—such as 8-bit counters, address decoders, or state machines—or combined across LABs to create larger logic blocks. Each LAB represents about 96 usable gates of logic.

Signal interconnections within FLEX 10KE devices (as well as to and from device pins) are provided by the FastTrack Interconnect routing structure, which is a series of fast, continuous row and column channels that run the entire length and width of the device.

Each I/O pin is fed by an I/O element (IOE) located at the end of each row and column of the FastTrack Interconnect routing structure. Each IOE contains a bidirectional I/O buffer and a flipflop that can be used as either an output or input register to feed input, output, or bidirectional signals. When used with a dedicated clock pin, these registers provide exceptional performance. As inputs, they provide setup times as low as 0.9 ns and hold times of 0 ns. As outputs, these registers provide clock-to-output times as low as 3.0 ns. IOEs provide a variety of features, such as JTAG BST support, slew-rate control, tri-state buffers, and open-drain outputs.

The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock, clear, and preset control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the output of the LE.

The LE has two outputs that drive the interconnect: one drives the local interconnect and the other drives either the row or column FastTrack Interconnect routing structure. The two outputs can be controlled independently. For example, the LUT can drive one output while the register drives the other output. This feature, called register packing, can improve LE utilization because the register and the LUT can be used for unrelated functions.

The FLEX 10KE architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. The carry chain supports high-speed counters and adders and the cascade chain implements wide-input functions with minimum delay. Carry and cascade chains connect all LEs in a LAB as well as all LABs in the same row. Intensive use of carry and cascade chains can reduce routing flexibility. Therefore, the use of these chains should be limited to speed-critical portions of a design.

Carry Chain

The carry chain provides a very fast (as low as 0.2 ns) carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the FLEX 10KE architecture to implement high-speed counters, adders, and comparators of arbitrary width efficiently. Carry chain logic can be created automatically by the Altera Compiler during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of carry chains.

Carry chains longer than eight LEs are automatically implemented by linking LABs together. For enhanced fitting, a long carry chain skips alternate LABs in a row. A carry chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB. For example, the last LE of the first LAB in a row carries to the first LE of the third LAB in the row. The carry chain does not cross the EAB at the middle of the row. For instance, in the EPF10K50E device, the carry chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB.

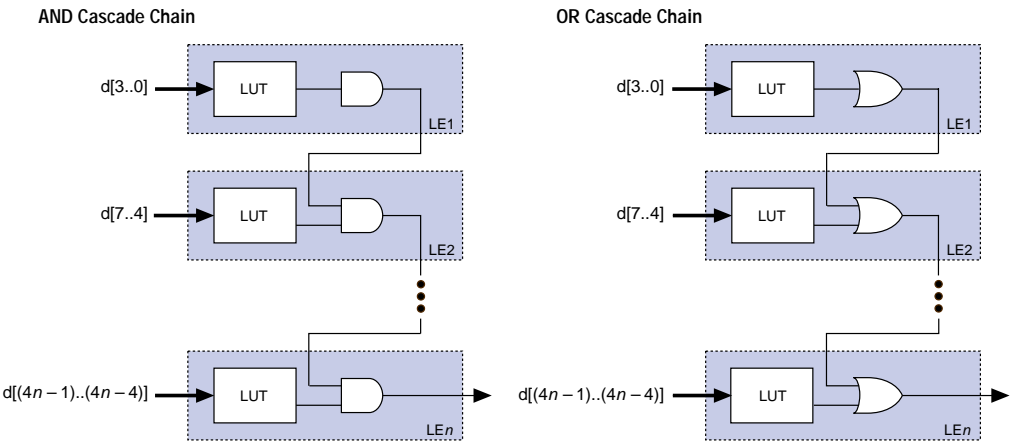
Cascade Chain

With the cascade chain, the FLEX 10KE architecture can implement functions that have a very wide fan-in. Adjacent LUTs can be used to compute portions of the function in parallel; the cascade chain serially connects the intermediate values. The cascade chain can use a logical AND or logical OR (via De Morgan's inversion) to connect the outputs of adjacent LEs. An a delay as low as 0.6 ns per LE, each additional LE provides four more inputs to the effective width of a function. Cascade chain logic can be created automatically by the Altera Compiler during design processing, or manually by the designer during design entry.

Cascade chains longer than eight bits are implemented automatically by linking several LABs together. For easier routing, a long cascade chain skips every other LAB in a row. A cascade chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB (e.g., the last LE of the first LAB in a row cascades to the first LE of the third LAB). The cascade chain does not cross the center of the row (e.g., in the EPF10K50E device, the cascade chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB). This break is due to the EAB's placement in the middle of the row.

Figure 10 shows how the cascade function can connect adjacent LEs to form functions with a wide fan-in. These examples show functions of $4n$ variables implemented with n LEs. The LE delay is 0.9 ns; the cascade chain delay is 0.6 ns. With the cascade chain, 2.7 ns are needed to decode a 16-bit address.

Figure 10. FLEX 10KE Cascade Chain Operation



LE Operating Modes

The FLEX 10KE LE can operate in the following four modes:

- Normal mode
- Arithmetic mode
- Up/down counter mode
- Clearable counter mode

Each of these modes uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. Three inputs to the LE provide clock, clear, and preset control for the register. The Altera software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions that use a specific LE operating mode for optimal performance.

The architecture provides a synchronous clock enable to the register in all four modes. The Altera software can set `DATA1` to enable the register synchronously, providing easy implementation of fully synchronous designs.

Normal Mode

The normal mode is suitable for general logic applications and wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a four-input LUT. The Altera Compiler automatically selects the carry-in or the `DATA3` signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. Either the register or the LUT can be used to drive both the local interconnect and the FastTrack Interconnect routing structure at the same time.

The LUT and the register in the LE can be used independently (register packing). To support register packing, the LE has two outputs; one drives the local interconnect, and the other drives the FastTrack Interconnect routing structure. The `DATA4` signal can drive the register directly, allowing the LUT to compute a function that is independent of the registered signal; a three-input function can be computed in the LUT, and a fourth independent signal can be registered. Alternatively, a four-input function can be generated, and one of the inputs to this function can be used to drive the register. The register in a packed LE can still use the clock enable, clear, and preset signals in the LE. In a packed LE, the register can drive the FastTrack Interconnect routing structure while the LUT drives the local interconnect, or vice versa.

Arithmetic Mode

The arithmetic mode offers 2 three-input LUTs that are ideal for implementing adders, accumulators, and comparators. One LUT computes a three-input function; the other generates a carry output. As shown in [Figure 11](#) on [page 22](#), the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, in an adder, this output is the sum of three signals: `a`, `b`, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

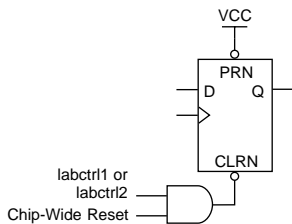
Up/Down Counter Mode

The up/down counter mode offers counter enable, clock enable, synchronous up/down control, and data loading options. These control signals are generated by the data inputs from the LAB local interconnect, the carry-in signal, and output feedback from the programmable register. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading. Data can also be loaded asynchronously with the clear and preset register control signals without using the LUT resources.

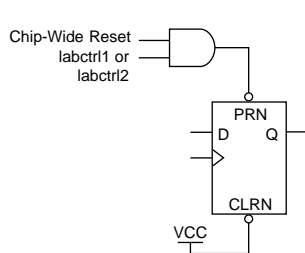
In addition to the six clear and preset modes, FLEX 10KE devices provide a chip-wide reset pin that can reset all registers in the device. Use of this feature is set during design entry. In any of the clear and preset modes, the chip-wide reset overrides all other signals. Registers with asynchronous presets may be preset when the chip-wide reset is asserted. Inversion can be used to implement the asynchronous preset. Figure 12 shows examples of how to setup the preset and clear inputs for the desired functionality.

Figure 12. FLEX 10KE LE Clear & Preset Modes

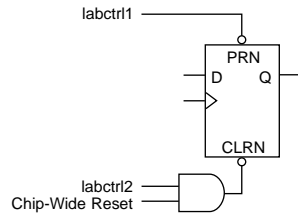
Asynchronous Clear



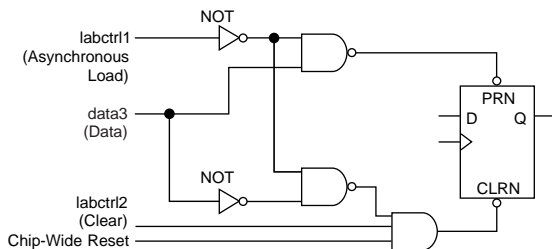
Asynchronous Preset



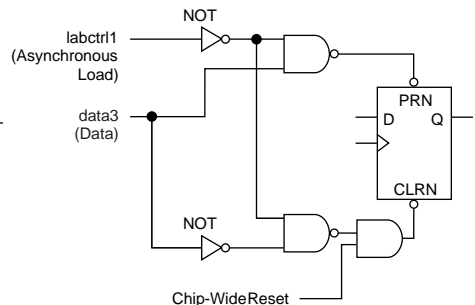
Asynchronous Preset & Clear



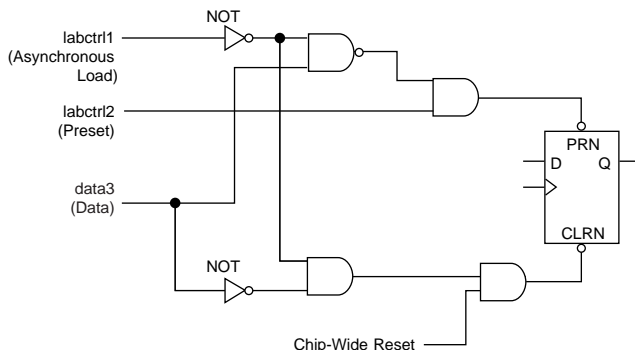
Asynchronous Load with Clear



Asynchronous Load without Clear or Preset



Asynchronous Load with Preset



Asynchronous Clear

The flipflop can be cleared by either LABCTRL1 or LABCTRL2. In this mode, the preset signal is tied to VCC to deactivate it.

Asynchronous Preset

An asynchronous preset is implemented as an asynchronous load, or with an asynchronous clear. If DATA3 is tied to VCC, asserting LABCTRL1 asynchronously loads a one into the register. Alternatively, the Altera software can provide preset control by using the clear and inverting the input and output of the register. Inversion control is available for the inputs to both LEs and IOEs. Therefore, if a register is preset by only one of the two LABCTRL signals, the DATA3 input is not needed and can be used for one of the LE operating modes.

Asynchronous Preset & Clear

When implementing asynchronous clear and preset, LABCTRL1 controls the preset and LABCTRL2 controls the clear. DATA3 is tied to VCC, so that asserting LABCTRL1 asynchronously loads a one into the register, effectively presetting the register. Asserting LABCTRL2 clears the register.

Asynchronous Load with Clear

When implementing an asynchronous load in conjunction with the clear, LABCTRL1 implements the asynchronous load of DATA3 by controlling the register preset and clear. LABCTRL2 implements the clear by controlling the register clear; LABCTRL2 does not have to feed the preset circuits.

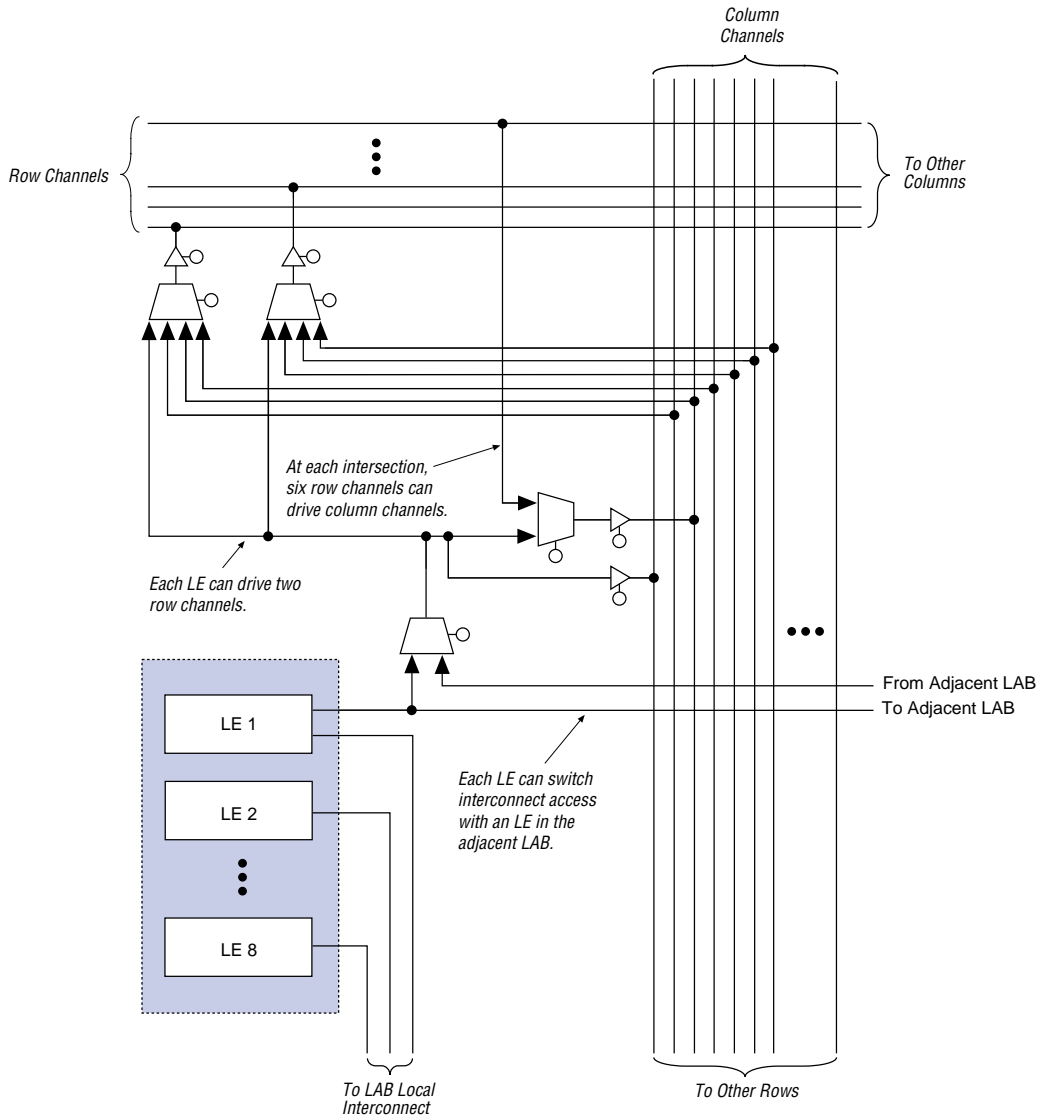
Asynchronous Load with Preset

When implementing an asynchronous load in conjunction with preset, the Altera software provides preset control by using the clear and inverting the input and output of the register. Asserting LABCTRL2 presets the register, while asserting LABCTRL1 loads the register. The Altera software inverts the signal that drives DATA3 to account for the inversion of the register's output.

Asynchronous Load without Preset or Clear

When implementing an asynchronous load without preset or clear, LABCTRL1 implements the asynchronous load of DATA3 by controlling the register preset and clear.

Figure 13. FLEX 10KE LAB Connections to Row & Column Interconnect

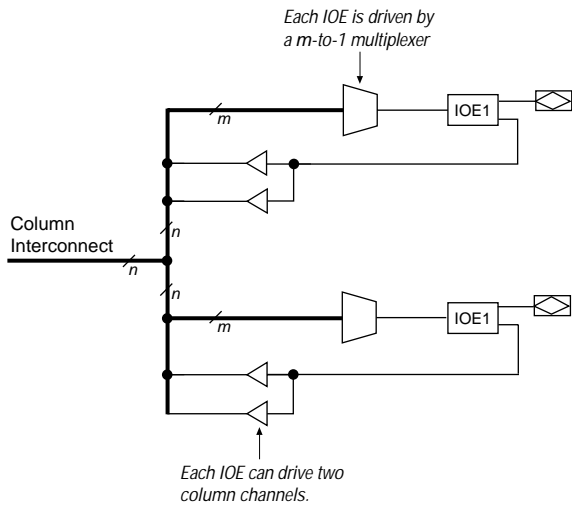


Column-to-IOE Connections

When an IOE is used as an input, it can drive up to two separate column channels. When an IOE is used as an output, the signal is driven by a multiplexer that selects a signal from the column channels. Two IOEs connect to each side of the column channels. Each IOE can be driven by column channels via a multiplexer. The set of column channels is different for each IOE (see [Figure 17](#)).

Figure 17. FLEX 10KE Column-to-IOE Connections

The values for *m* and *n* are provided in [Table 11](#).



[Table 11](#) lists the FLEX 10KE column-to-IOE interconnect resources.

Table 11. FLEX 10KE Column-to-IOE Interconnect Resources		
Device	Channels per Column (<i>n</i>)	Column Channels per Pin (<i>m</i>)
EPF10K30E	24	16
EPF10K50E EPF10K50S	24	16
EPF10K100E	24	16
EPF10K130E	32	24
EPF10K200E EPF10K200S	48	40

Table 17. 32-Bit IDCODE for FLEX 10KE Devices *Note (1)*

Device	IDCODE (32 Bits)			
	Version (4 Bits)	Part Number (16 Bits)	Manufacturer's Identity (11 Bits)	1 (1 Bit) (2)
EPF10K30E	0001	0001 0000 0011 0000	00001101110	1
EPF10K50E EPF10K50S	0001	0001 0000 0101 0000	00001101110	1
EPF10K100E	0010	0000 0001 0000 0000	00001101110	1
EPF10K130E	0001	0000 0001 0011 0000	00001101110	1
EPF10K200E EPF10K200S	0001	0000 0010 0000 0000	00001101110	1

Notes:

- (1) The most significant bit (MSB) is on the left.
 (2) The least significant bit (LSB) for all JTAG IDCODEs is 1.

FLEX 10KE devices include weak pull-up resistors on the JTAG pins.



For more information, see the following documents:

- *Application Note 39 (IEEE Std. 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices)*
- *BitBlaster Serial Download Cable Data Sheet*
- *ByteBlasterMV Parallel Port Download Cable Data Sheet*
- *Jam Programming & Test Language Specification*

Table 22. FLEX 10KE 2.5-V Device DC Operating Conditions

Notes (6), (7)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{IH}	High-level input voltage		1.7, $0.5 \times V_{CCIO}$ (8)		5.75	V
V_{IL}	Low-level input voltage		-0.5		0.8, $0.3 \times V_{CCIO}$ (8)	V
V_{OH}	3.3-V high-level TTL output voltage	$I_{OH} = -8$ mA DC, $V_{CCIO} = 3.00$ V (9)	2.4			V
	3.3-V high-level CMOS output voltage	$I_{OH} = -0.1$ mA DC, $V_{CCIO} = 3.00$ V (9)	$V_{CCIO} - 0.2$			V
	3.3-V high-level PCI output voltage	$I_{OH} = -0.5$ mA DC, $V_{CCIO} = 3.00$ to 3.60 V (9)	$0.9 \times V_{CCIO}$			V
	2.5-V high-level output voltage	$I_{OH} = -0.1$ mA DC, $V_{CCIO} = 2.30$ V (9)	2.1			V
		$I_{OH} = -1$ mA DC, $V_{CCIO} = 2.30$ V (9)	2.0			V
		$I_{OH} = -2$ mA DC, $V_{CCIO} = 2.30$ V (9)	1.7			V
V_{OL}	3.3-V low-level TTL output voltage	$I_{OL} = 12$ mA DC, $V_{CCIO} = 3.00$ V (10)			0.45	V
	3.3-V low-level CMOS output voltage	$I_{OL} = 0.1$ mA DC, $V_{CCIO} = 3.00$ V (10)			0.2	V
	3.3-V low-level PCI output voltage	$I_{OL} = 1.5$ mA DC, $V_{CCIO} = 3.00$ to 3.60 V (10)			$0.1 \times V_{CCIO}$	V
	2.5-V low-level output voltage	$I_{OL} = 0.1$ mA DC, $V_{CCIO} = 2.30$ V (10)			0.2	V
		$I_{OL} = 1$ mA DC, $V_{CCIO} = 2.30$ V (10)			0.4	V
		$I_{OL} = 2$ mA DC, $V_{CCIO} = 2.30$ V (10)			0.7	V
I_I	Input pin leakage current	$V_I = V_{CCIOmax}$ to 0 V (11)	-10		10	μ A
I_{OZ}	Tri-stated I/O pin leakage current	$V_O = V_{CCIOmax}$ to 0 V (11)	-10		10	μ A
I_{CC0}	V_{CC} supply current (standby)	$V_I =$ ground, no load, no toggling inputs		5		mA
		$V_I =$ ground, no load, no toggling inputs (12)		10		mA
R_{CONF}	Value of I/O pin pull-up resistor before and during configuration	$V_{CCIO} = 3.0$ V (13)	20		50	$k\frac{3}{4}$
		$V_{CCIO} = 2.3$ V (13)	30		80	$k\frac{3}{4}$

Table 37. EPF10K30E External Bidirectional Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{\text{INSUBIDIR}}$ (3)	2.8		3.9		5.2		ns
t_{INHBIDIR} (3)	0.0		0.0		0.0		ns
$t_{\text{INSUBIDIR}}$ (4)	3.8		4.9		—		ns
t_{INHBIDIR} (4)	0.0		0.0		—		ns
$t_{\text{OUTCOBIDIR}}$ (3)	2.0	4.9	2.0	5.9	2.0	7.6	ns
t_{XZBIDIR} (3)		6.1		7.5		9.7	ns
t_{ZXBIDIR} (3)		6.1		7.5		9.7	ns
$t_{\text{OUTCOBIDIR}}$ (4)	0.5	3.9	0.5	4.9	—	—	ns
t_{XZBIDIR} (4)		5.1		6.5		—	ns
t_{ZXBIDIR} (4)		5.1		6.5		—	ns

Notes to tables:

- (1) All timing parameters are described in Tables 24 through 30 in this data sheet.
- (2) These parameters are specified by characterization.
- (3) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
- (4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Tables 38 through 44 show EPF10K50E device internal and external timing parameters.

Table 38. EPF10K50E Device LE Timing Microparameters (Part 1 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.6		0.9		1.3	ns
t_{CLUT}		0.5		0.6		0.8	ns
t_{RLUT}		0.7		0.8		1.1	ns
t_{PACKED}		0.4		0.5		0.6	ns
t_{EN}		0.6		0.7		0.9	ns
t_{CICO}		0.2		0.2		0.3	ns
t_{CGEN}		0.5		0.5		0.8	ns
t_{CGENR}		0.2		0.2		0.3	ns
t_{CASC}		0.8		1.0		1.4	ns
t_{C}		0.5		0.6		0.8	ns
t_{CO}		0.7		0.7		0.9	ns
t_{COMB}		0.5		0.6		0.8	ns
t_{SU}	0.7		0.7		0.8		ns

Table 43. EPF10K50E External Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		8.5		10.0		13.5	ns
t_{INSU}	2.7		3.2		4.3		ns
t_{INH}	0.0		0.0		0.0		ns
t_{OUTCO}	2.0	4.5	2.0	5.2	2.0	7.3	ns
t_{PCISU}	3.0		4.2		-		ns
t_{PCIH}	0.0		0.0		-		ns
t_{PCICO}	2.0	6.0	2.0	7.7	-	-	ns

Table 44. EPF10K50E External Bidirectional Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{\text{INSUBIDIR}}$	2.7		3.2		4.3		ns
t_{INHBIDIR}	0.0		0.0		0.0		ns
$t_{\text{OUTCOBIDIR}}$	2.0	4.5	2.0	5.2	2.0	7.3	ns
t_{XZBIDIR}		6.8		7.8		10.1	ns
t_{ZXBIDIR}		6.8		7.8		10.1	ns

Notes to tables:

- (1) All timing parameters are described in Tables 24 through 30 in this data sheet.
 (2) These parameters are specified by characterization.

Tables 45 through 51 show EPF10K100E device internal and external timing parameters.

Table 45. EPF10K100E Device LE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.7		1.0		1.5	ns
t_{CLUT}		0.5		0.7		0.9	ns
t_{RLUT}		0.6		0.8		1.1	ns
t_{PACKED}		0.3		0.4		0.5	ns
t_{EN}		0.2		0.3		0.3	ns
t_{CICO}		0.1		0.1		0.2	ns
t_{CGEN}		0.4		0.5		0.7	ns

Table 45. EPF10K100E Device LE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{CGENR}		0.1		0.1		0.2	ns
t_{CASC}		0.6		0.9		1.2	ns
t_C		0.8		1.0		1.4	ns
t_{CO}		0.6		0.8		1.1	ns
t_{COMB}		0.4		0.5		0.7	ns
t_{SU}	0.4		0.6		0.7		ns
t_H	0.5		0.7		0.9		ns
t_{PRE}		0.8		1.0		1.4	ns
t_{CLR}		0.8		1.0		1.4	ns
t_{CH}	1.5		2.0		2.5		ns
t_{CL}	1.5		2.0		2.5		ns

Table 46. EPF10K100E Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.7		2.0		2.6	ns
t_{IOC}		0.0		0.0		0.0	ns
t_{IOCO}		1.4		1.6		2.1	ns
t_{IOCOMB}		0.5		0.7		0.9	ns
t_{IOSU}	0.8		1.0		1.3		ns
t_{IOH}	0.7		0.9		1.2		ns
t_{IOCLR}		0.5		0.7		0.9	ns
t_{OD1}		3.0		4.2		5.6	ns
t_{OD2}		3.0		4.2		5.6	ns
t_{OD3}		4.0		5.5		7.3	ns
t_{XZ}		3.5		4.6		6.1	ns
t_{ZX1}		3.5		4.6		6.1	ns
t_{ZX2}		3.5		4.6		6.1	ns
t_{ZX3}		4.5		5.9		7.8	ns
t_{INREG}		2.0		2.6		3.5	ns
t_{OFD}		0.5		0.8		1.2	ns
t_{INCOMB}		0.5		0.8		1.2	ns

Table 77. EPF10K200S Device Interconnect Timing Microparameters (Part 2 of 2) *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{LABCASC}$		0.5		1.0		1.4	ns

Table 78. EPF10K200S External Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{DRR}		9.0		12.0		16.0	ns
$t_{INSU}^{(2)}$	3.1		3.7		4.7		ns
$t_{INH}^{(2)}$	0.0		0.0		0.0		ns
$t_{OUTCO}^{(2)}$	2.0	3.7	2.0	4.4	2.0	6.3	ns
$t_{INSU}^{(3)}$	2.1		2.7		—		ns
$t_{INH}^{(3)}$	0.0		0.0		—		ns
$t_{OUTCO}^{(3)}$	0.5	2.7	0.5	3.4	—	—	ns
t_{PCISU}	3.0		4.2		—		ns
t_{PCIH}	0.0		0.0		—		ns
t_{PCICO}	2.0	6.0	2.0	8.9	—	—	ns

Table 79. EPF10K200S External Bidirectional Timing Parameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{INSUBIDIR}^{(2)}$	2.3		3.4		4.4		ns
$t_{INHBIDIR}^{(2)}$	0.0		0.0		0.0		ns
$t_{INSUBIDIR}^{(3)}$	3.3		4.4		—		ns
$t_{INHBIDIR}^{(3)}$	0.0		0.0		—		ns
$t_{OUTCOBIDIR}^{(2)}$	2.0	3.7	2.0	4.4	2.0	6.3	ns
$t_{XZBIDIR}^{(2)}$		6.9		7.6		9.2	ns
$t_{ZXBIDIR}^{(2)}$		5.9		6.6		—	ns
$t_{OUTCOBIDIR}^{(3)}$	0.5	2.7	0.5	3.4	—	—	ns
$t_{XZBIDIR}^{(3)}$		6.9		7.6		9.2	ns
$t_{ZXBIDIR}^{(3)}$		5.9		6.6		—	ns

Notes to tables:

- (1) All timing parameters are described in Tables 24 through 30 in this data sheet.
 (2) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
 (3) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Device Pin-Outs

See the Altera web site (<http://www.altera.com>) or the Altera Digital Library for pin-out information.

Revision History

The information contained in the *FLEX 10KE Embedded Programmable Logic Data Sheet* version 2.5 supersedes information published in previous versions.

Version 2.5

The following changes were made to the *FLEX 10KE Embedded Programmable Logic Data Sheet* version 2.5:

- *Note (1)* added to **Figure 23**.
- Text added to “**I/O Element**” section on **page 34**.
- Updated **Table 22**.

Version 2.4

The following changes were made to the *FLEX 10KE Embedded Programmable Logic Data Sheet* version 2.4: updated text on **page 34** and **page 63**.



101 Innovation Drive
San Jose, CA 95134
(408) 544-7000
<http://www.altera.com>
Applications Hotline:
(800) 800-EPLD
Literature Services:
lit_req@altera.com

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