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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	360
Number of Logic Elements/Cells	2880
Total RAM Bits	40960
Number of I/O	102
Number of Gates	199000
Voltage - Supply	2.3V ~ 2.7V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf10k50etc144-2n

Device	144-Pin TQFP	208-Pin PQFP	240-Pin PQFP RQFP	256-Pin FineLine BGA	356-Pin BGA	484-Pin FineLine BGA	599-Pin PGA	600-Pin BGA	672-Pin FineLine BGA
Pitch (mm)	0.50	0.50	0.50	1.0	1.27	1.0	–	1.27	1.0
Area (mm²)	484	936	1,197	289	1,225	529	3,904	2,025	729
Length × width (mm × mm)	22 × 22	30.6 × 30.6	34.6 × 34.6	17 × 17	35 × 35	23 × 23	62.5 × 62.5	45 × 45	27 × 27

Table 4. FLEX 10KE Package Sizes

General Description

Altera FLEX 10KE devices are enhanced versions of FLEX 10K devices. Based on reconfigurable CMOS SRAM elements, the FLEX architecture incorporates all features necessary to implement common gate array megafunctions. With up to 200,000 typical gates, FLEX 10KE devices provide the density, speed, and features to integrate entire systems, including multiple 32-bit buses, into a single device.

The ability to reconfigure FLEX 10KE devices enables 100% testing prior to shipment and allows the designer to focus on simulation and design verification. FLEX 10KE reconfigurability eliminates inventory management for gate array designs and generation of test vectors for fault coverage.

Table 5 shows FLEX 10KE performance for some common designs. All performance values were obtained with Synopsys DesignWare or LPM functions. Special design techniques are not required to implement the applications; the designer simply infers or instantiates a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Embedded Array Block

The EAB is a flexible block of RAM, with registers on the input and output ports, that is used to implement common gate array megafunctions. Because it is large and flexible, the EAB is suitable for functions such as multipliers, vector scalars, and error correction circuits. These functions can be combined in applications such as digital filters and microcontrollers.

Logic functions are implemented by programming the EAB with a read-only pattern during configuration, thereby creating a large LUT. With LUTs, combinatorial functions are implemented by looking up the results, rather than by computing them. This implementation of combinatorial functions can be faster than using algorithms implemented in general logic, a performance advantage that is further enhanced by the fast access times of EABs. The large capacity of EABs enables designers to implement complex functions in one logic level without the routing delays associated with linked LEs or field-programmable gate array (FPGA) RAM blocks. For example, a single EAB can implement any function with 8 inputs and 16 outputs. Parameterized functions such as LPM functions can take advantage of the EAB automatically.

The FLEX 10KE EAB provides advantages over FPGAs, which implement on-board RAM as arrays of small, distributed RAM blocks. These small FPGA RAM blocks must be connected together to make RAM blocks of manageable size. The RAM blocks are connected together using multiplexers implemented with more logic blocks. These extra multiplexers cause extra delay, which slows down the RAM block. FPGA RAM blocks are also prone to routing problems because small blocks of RAM must be connected together to make larger blocks. In contrast, EABs can be used to implement large, dedicated blocks of RAM that eliminate these timing and routing concerns.

The FLEX 10KE enhanced EAB adds dual-port capability to the existing EAB structure. The dual-port structure is ideal for FIFO buffers with one or two clocks. The FLEX 10KE EAB can also support up to 16-bit-wide RAM blocks and is backward-compatible with any design containing FLEX 10K EABs. The FLEX 10KE EAB can act in dual-port or single-port mode. When in dual-port mode, separate clocks may be used for EAB read and write sections, which allows the EAB to be written and read at different rates. It also has separate synchronous clock enable signals for the EAB read and write sections, which allow independent control of these sections.

The EAB can also be used for bidirectional, dual-port memory applications where two ports read or write simultaneously. To implement this type of dual-port memory, two EABs are used to support two simultaneous read or writes.

Alternatively, one clock and clock enable can be used to control the input registers of the EAB, while a different clock and clock enable control the output registers (see Figure 2).

Figure 2. FLEX 10KE Device in Dual-Port RAM Mode   *Notes (1)*
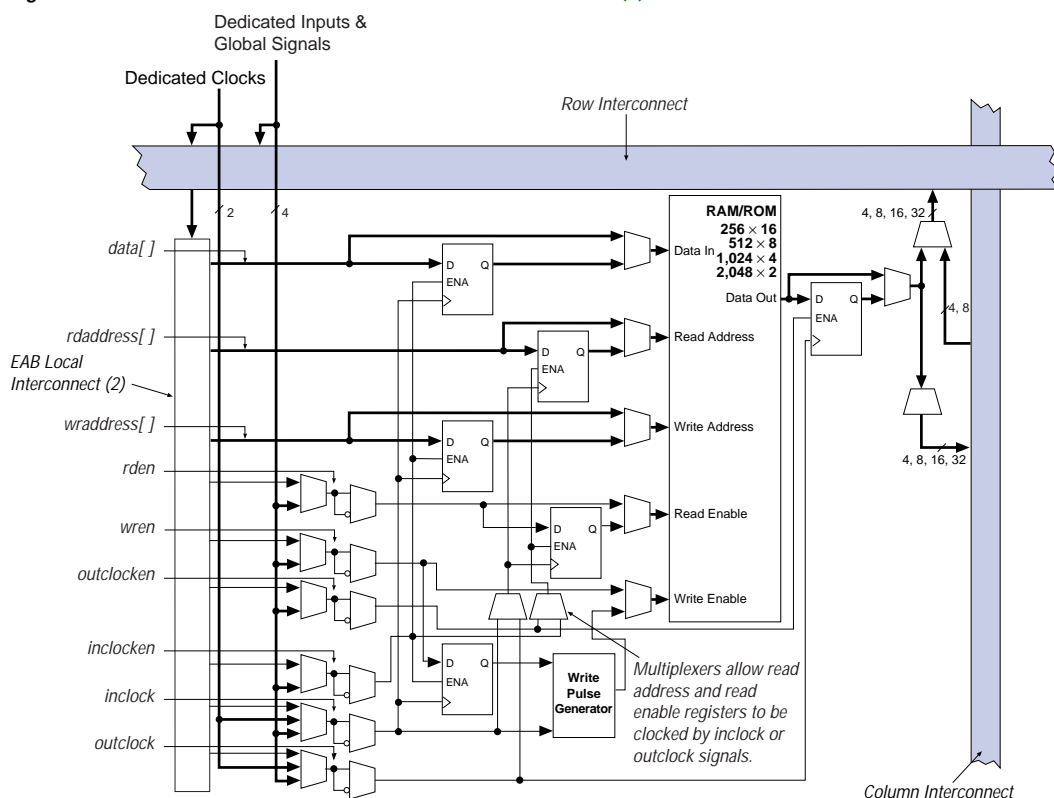
Notes:

(1)   All registers can be asynchronously cleared by EAB local interconnect signals, global signals, or the chip-wide reset.
(2)   EPF10K30E and EPF10K50E devices have 88 EAB local interconnect channels; EPF10K100E, EPF10K130E, and EPF10K200E devices have 104 EAB local interconnect channels.

EABs provide flexible options for driving and controlling clock signals. Different clocks and clock enables can be used for reading and writing to the EAB. Registers can be independently inserted on the data input, EAB output, write address, write enable signals, read address, and read enable signals. The global signals and the EAB local interconnect can drive write enable, read enable, and clock enable signals. The global signals, dedicated clock pins, and EAB local interconnect can drive the EAB clock signals. Because the LEs drive the EAB local interconnect, the LEs can control write enable, read enable, clear, clock, and clock enable signals.

An EAB is fed by a row interconnect and can drive out to row and column interconnects. Each EAB output can drive up to two row channels and up to two column channels; the unused row channel can be driven by other LEs. This feature increases the routing resources available for EAB outputs (see Figures 2 and 4). The column interconnect, which is adjacent to the EAB, has twice as many channels as other columns in the device.

Logic Array Block

An LAB consists of eight LEs, their associated carry and cascade chains, LAB control signals, and the LAB local interconnect. The LAB provides the coarse-grained structure to the FLEX 10KE architecture, facilitating efficient routing with optimum device utilization and high performance (see Figure 7).

Figure 7. FLEX 10KE LAB
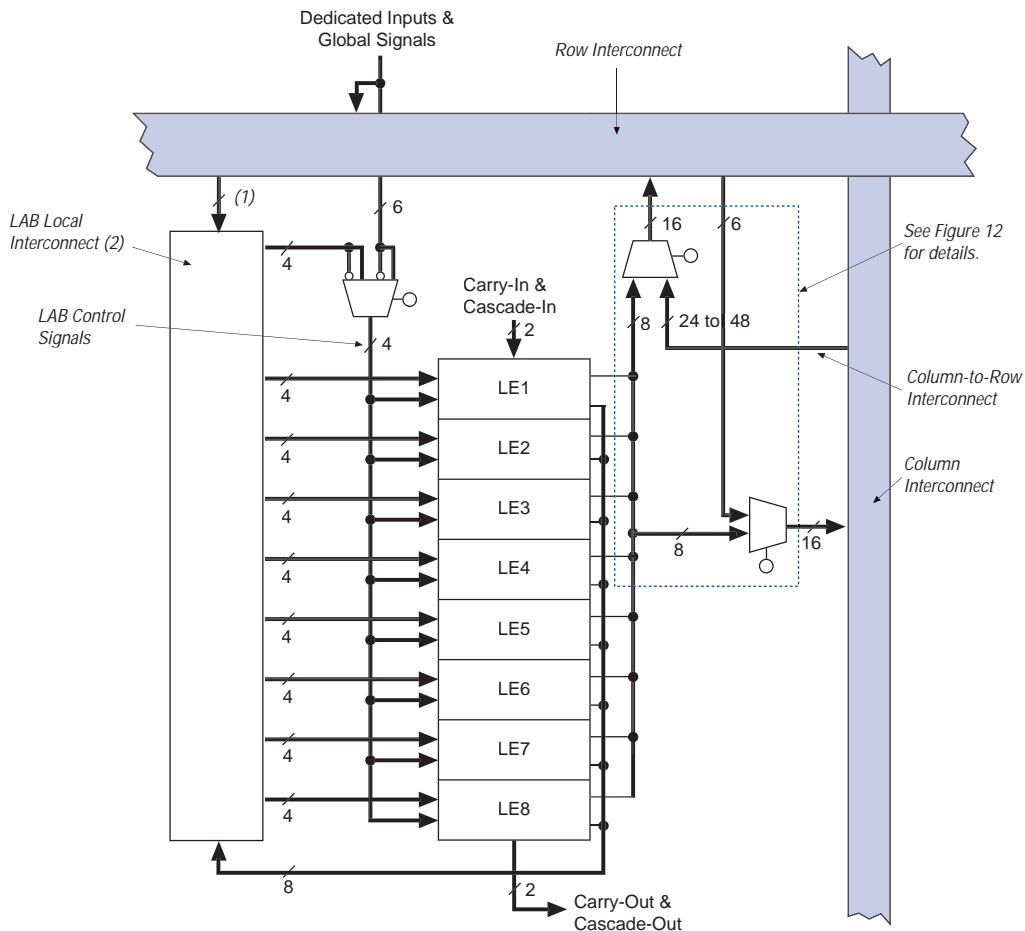
Notes:
(1)   EPF10K30E, EPF10K50E, and EPF10K50S devices have 22 inputs to the LAB local interconnect channel from the
        row; EPF10K100E, EPF10K130E, EPF10K200E, and EPF10K200S devices have 26.
(2)   EPF10K30E, EPF10K50E, and EPF10K50S devices have 30 LAB local interconnect channels; EPF10K100E,
        EPF10K130E, EPF10K200E, and EPF10K200S devices have 34.

Each LAB provides four control signals with programmable inversion that can be used in all eight LEs. Two of these signals can be used as clocks, the other two can be used for clear/preset control. The LAB clocks can be driven by the dedicated clock input pins, global signals, I/O signals, or internal signals via the LAB local interconnect. The LAB preset and clear control signals can be driven by the global signals, I/O signals, or internal signals via the LAB local interconnect. The global control signals are typically used for global clock, clear, or preset signals because they provide asynchronous control with very low skew across the device. If logic is required on a control signal, it can be generated in one or more LE in any LAB and driven into the local interconnect of the target LAB. In addition, the global control signals can be generated from LE outputs.

Logic Element

The LE, the smallest unit of logic in the FLEX 10KE architecture, has a compact size that provides efficient logic utilization. Each LE contains a four-input LUT, which is a function generator that can quickly compute any function of four variables. In addition, each LE contains a programmable flipflop with a synchronous clock enable, a carry chain, and a cascade chain. Each LE drives both the local and the FastTrack Interconnect routing structure (see Figure 8).

Figure 8. FLEX 10KE Logic Element
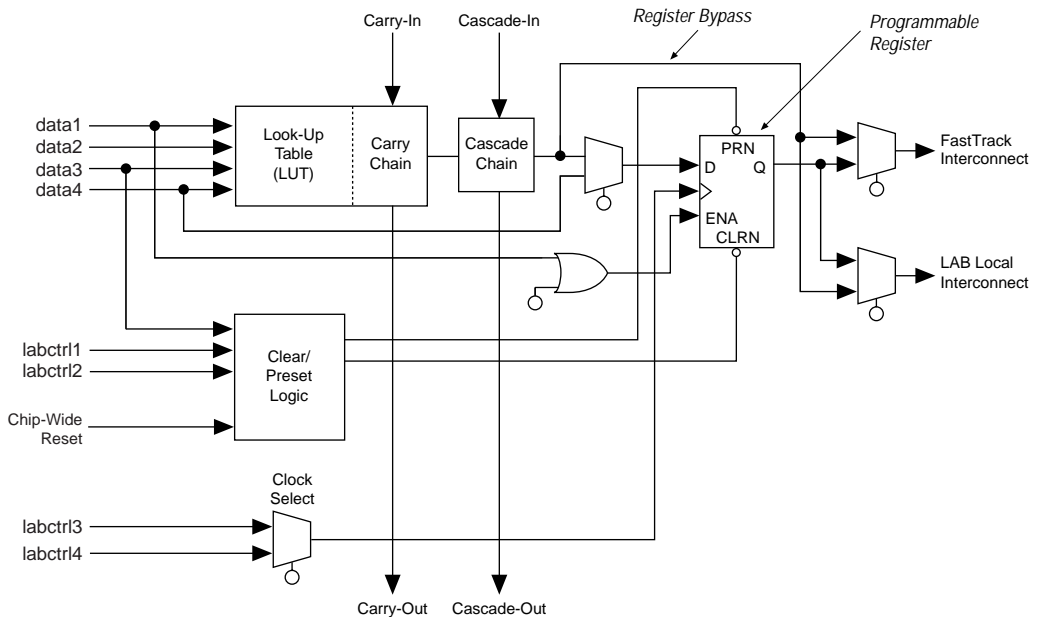
Normal Mode

The normal mode is suitable for general logic applications and wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a four-input LUT. The Altera Compiler automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. Either the register or the LUT can be used to drive both the local interconnect and the FastTrack Interconnect routing structure at the same time.

The LUT and the register in the LE can be used independently (register packing). To support register packing, the LE has two outputs; one drives the local interconnect, and the other drives the FastTrack Interconnect routing structure. The DATA4 signal can drive the register directly, allowing the LUT to compute a function that is independent of the registered signal; a three-input function can be computed in the LUT, and a fourth independent signal can be registered. Alternatively, a four-input function can be generated, and one of the inputs to this function can be used to drive the register. The register in a packed LE can still use the clock enable, clear, and preset signals in the LE. In a packed LE, the register can drive the FastTrack Interconnect routing structure while the LUT drives the local interconnect, or vice versa.

Arithmetic Mode

The arithmetic mode offers 2 three-input LUTs that are ideal for implementing adders, accumulators, and comparators. One LUT computes a three-input function; the other generates a carry output. As shown in Figure 11 on page 22, the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, in an adder, this output is the sum of three signals: a, b, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

Up/Down Counter Mode

The up/down counter mode offers counter enable, clock enable, synchronous up/down control, and data loading options. These control signals are generated by the data inputs from the LAB local interconnect, the carry-in signal, and output feedback from the programmable register. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading. Data can also be loaded asynchronously with the clear and preset register control signals without using the LUT resources.

Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Use 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is AND ed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-states without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the Altera Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

FastTrack Interconnect Routing Structure

In the FLEX 10KE architecture, connections between LEs, EABs, and device I/O pins are provided by the FastTrack Interconnect routing structure, which is a series of continuous horizontal and vertical routing channels that traverses the device. This global routing structure provides predictable performance, even in complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.

The FastTrack Interconnect routing structure consists of row and column interconnect channels that span the entire device. Each row of LABs is served by a dedicated row interconnect. The row interconnect can drive I/O pins and feed other LABs in the row. The column interconnect routes signals between rows and can drive I/O pins.

Row channels drive into the LAB or EAB local interconnect. The row signal is buffered at every LAB or EAB to reduce the effect of fan-out on delay. A row channel can be driven by an LE or by one of three column channels. These four signals feed dual 4-to-1 multiplexers that connect to two specific row channels. These multiplexers, which are connected to each LE, allow column channels to drive row channels even when all eight LEs in a LAB drive the row interconnect.

Each column of LABs or EABs is served by a dedicated column interconnect. The column interconnect that serves the EABs has twice as many channels as other column interconnects. The column interconnect can then drive I/O pins or another row's interconnect to route the signals to other LABs or EABs in the device. A signal from the column interconnect, which can be either the output of a LE or an input from an I/O pin, must be routed to the row interconnect before it can enter a LAB or EAB. Each row channel that is driven by an IOE or EAB can drive one specific column channel.

Access to row and column channels can be switched between LEs in adjacent pairs of LABs. For example, a LE in one LAB can drive the row and column channels normally driven by a particular LE in the adjacent LAB in the same row, and vice versa. This flexibility enables routing resources to be used more efficiently (see Figure 13).

Figure 14. FLEX 10KE Interconnect Resources
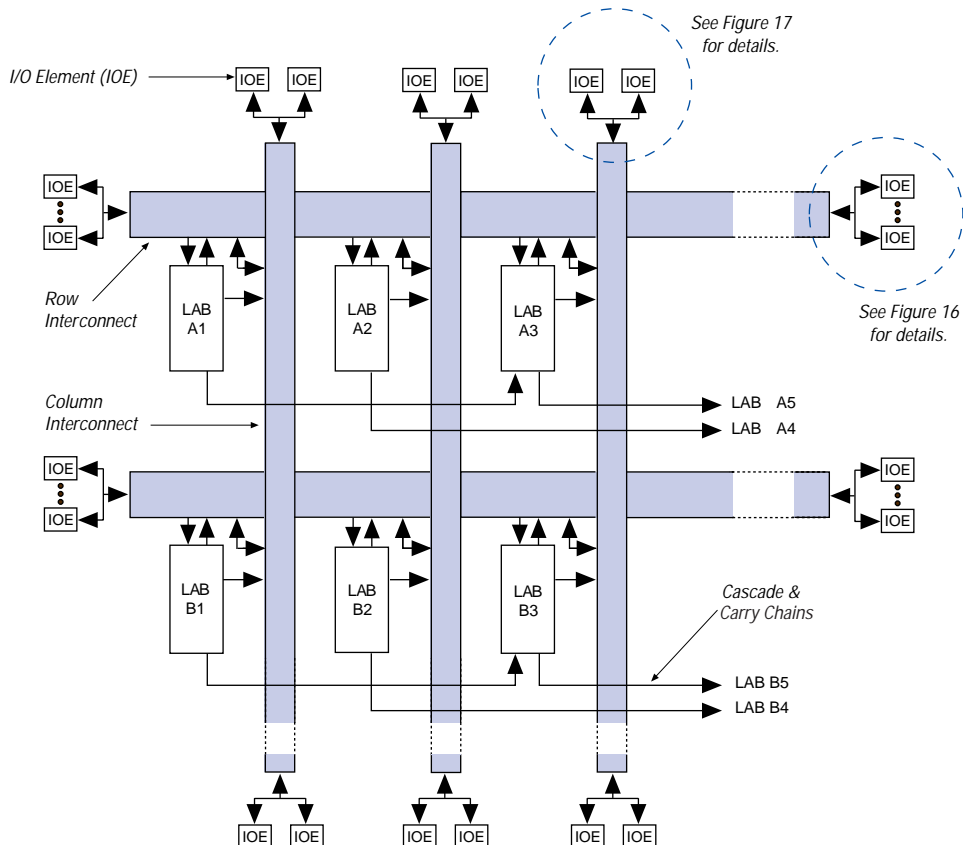
I/O Element

An IOE contains a bidirectional I/O buffer and a register that can be used either as an input register for external data that requires a fast setup time, or as an output register for data that requires fast clock-to-output performance. In some cases, using an LE register for an input register will result in a faster setup time than using an IOE register. IOEs can be used as input, output, or bidirectional pins. For bidirectional registered I/O implementation, the output register should be in the IOE, and the data input and output enable registers should be LE registers placed adjacent to the bidirectional pin. The Altera Compiler uses the programmable inversion option to invert signals from the row and column interconnect automatically where appropriate. Figure 15 shows the bidirectional I/O registers.

The VCCINT pins must always be connected to a 2.5-V power supply. With a 2.5-V V_{CCINT} level, input voltages are compatible with 2.5-V, 3.3-V, and 5.0-V inputs. The VCCIO pins can be connected to either a 2.5-V or 3.3-V power supply, depending on the output requirements. When the VCCIO pins are connected to a 2.5-V power supply, the output levels are compatible with 2.5-V systems. When the VCCIO pins are connected to a 3.3-V power supply, the output high is at 3.3 V and is therefore compatible with 3.3-V or 5.0-V systems. Devices operating with V_{CCIO} levels higher than 3.0 V achieve a faster timing delay of t_{OD2} instead of t_{OD1}.

Table 14 summarizes FLEX 10KE MultiVolt I/O support.

Table 14. FLEX 10KE MultiVolt I/O Support

V_{CCIO} (V)	Input Signal (V)			Output Signal (V)		
	2.5	3.3	5.0	2.5	3.3	5.0
2.5	✓	✓ (1)	✓ (1)	✓		
3.3	✓	✓	✓ (1)	✓ (2)	✓	✓

Notes:
(1)   The PCI clamping diode must be disabled to drive an input with voltages higher than V_{CCIO}.
(2)   When V_{CCIO} = 3.3 V, a FLEX 10KE device can drive a 2.5-V device that has 3.3-V tolerant inputs.

Open-drain output pins on FLEX 10KE devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a V_{IH} of 3.5 V. When the open-drain pin is active, it will drive low. When the pin is inactive, the trace will be pulled up to 5.0 V by the resistor. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The I_{OL} current specification should be considered when selecting a pull-up resistor.

Power Sequencing & Hot-Socketing

Because FLEX 10KE devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The V_{CCIO} and V_{CCINT} power planes can be powered in any order.

Signals can be driven into FLEX 10KE devices before and during power up without damaging the device. Additionally, FLEX 10KE devices do not drive out during power up. Once operating conditions are reached, FLEX 10KE devices operate as specified by the user.

IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

All FLEX 10KE devices provide JTAG BST circuitry that complies with the IEEE Std. 1149.1-1990 specification. FLEX 10KE devices can also be configured using the JTAG pins through the BitBlaster or ByteBlasterMV download cable, or via hardware that uses the Jam™ STAPL programming and test language. JTAG boundary-scan testing can be performed before or after configuration, but not during configuration. FLEX 10KE devices support the JTAG instructions shown in Table 15.

Table 15. FLEX 10KE JTAG Instructions

JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins.
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation.
USERCODE	Selects the user electronic signature (USERCODE) register and places it between the TDI and TDO pins, allowing the USERCODE to be serially shifted out of TDO.
IDCODE	Selects the IDCODE register and places it between TDI and TDO, allowing the IDCODE to be serially shifted out of TDO.
ICR Instructions	These instructions are used when configuring a FLEX 10KE device via JTAG ports with a BitBlaster or ByteBlasterMV download cable, or using a Jam File (**.jam**) or Jam Byte-Code File (**.jbc**) via an embedded processor.

The instruction register length of FLEX 10KE devices is 10 bits. The USERCODE register length in FLEX 10KE devices is 32 bits; 7 bits are determined by the user, and 25 bits are pre-determined. Tables 16 and 17 show the boundary-scan register length and device IDCODE information for FLEX 10KE devices.

Table 16. FLEX 10KE Boundary-Scan Register Length

Device	Boundary-Scan Register Length
EPF10K30E	690
EPF10K50E EPF10K50S	798
EPF10K100E	1,050
EPF10K130E	1,308
EPF10K200E EPF10K200S	1,446

Figure 20 shows the timing requirements for the JTAG signals.
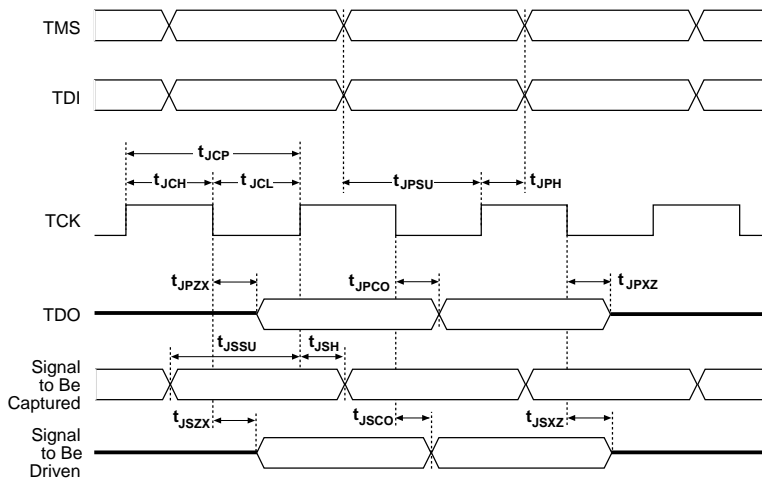
Figure 20. FLEX 10KE JTAG Waveforms

Table 18 shows the timing parameters and values for FLEX 10KE devices.

	Table 18. FLEX 10KE JTAG Timing Parameters & Values			
Symbol	Parameter	Min	Max	Unit
t$_{JCP}$	TCK clock period	100		ns
t$_{JCH}$	TCK clock high time	50		ns
t$_{JCL}$	TCK clock low time	50		ns
t$_{JPSU}$	JTAG port setup time	20		ns
t$_{JPH}$	JTAG port hold time	45		ns
t$_{JPCO}$	JTAG port clock to output		25	ns
t$_{JPZX}$	JTAG port high impedance to valid output		25	ns
t$_{JPXZ}$	JTAG port valid output to high impedance		25	ns
t$_{JSSU}$	Capture register setup time	20		ns
t$_{JSH}$	Capture register hold time	45		ns
t$_{JSCO}$	Update register clock to output		35	ns
t$_{JSZX}$	Update register high impedance to valid output		35	ns
t$_{JSXZ}$	Update register valid output to high impedance		35	ns

Figure 23. Output Drive Characteristics of FLEX 10KE Devices *Note (1)*
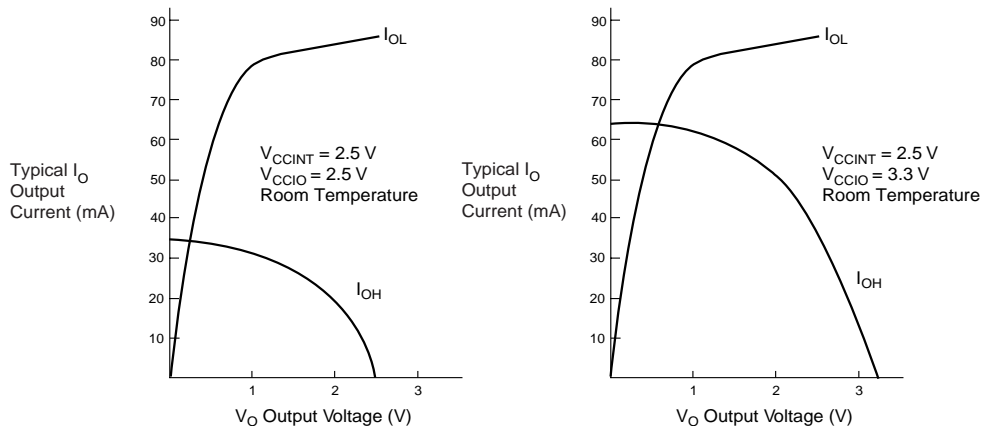
Note:
(1) These are transient (AC) currents.

Timing Model

The continuous, high-performance FastTrack Interconnect routing resources ensure predictable performance and accurate simulation and timing analysis. This predictable performance contrasts with that of FPGAs, which use a segmented connection scheme and therefore have unpredictable performance.

Device performance can be estimated by following the signal path from a source, through the interconnect, to the destination. For example, the registered performance between two LEs on the same row can be calculated by adding the following parameters:

- LE register clock-to-output delay (t_{CO})
- Interconnect delay ($t_{SAMEROW}$)
- LE look-up table delay (t_{LUT})
- LE register setup time (t_{SU})

The routing delay depends on the placement of the source and destination LEs. A more complex registered path may involve multiple combinatorial LEs between the source and destination LEs.
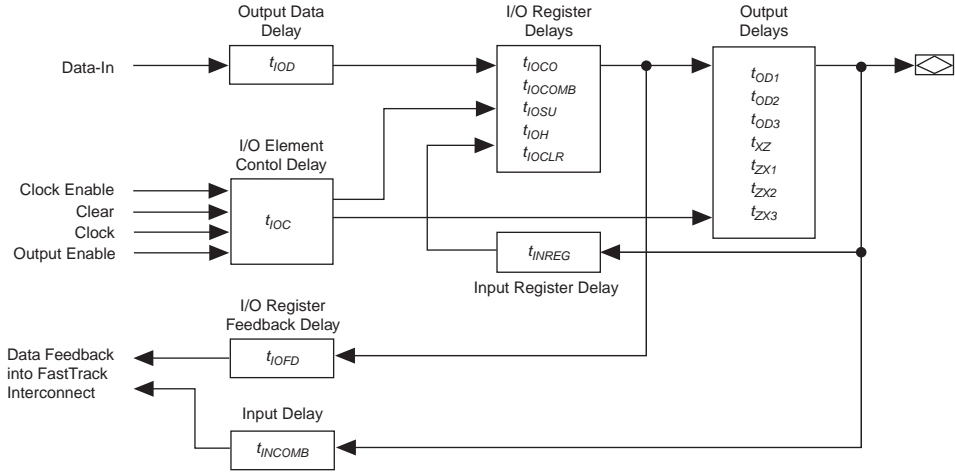
Figure 26. FLEX 10KE Device IOE Timing Model

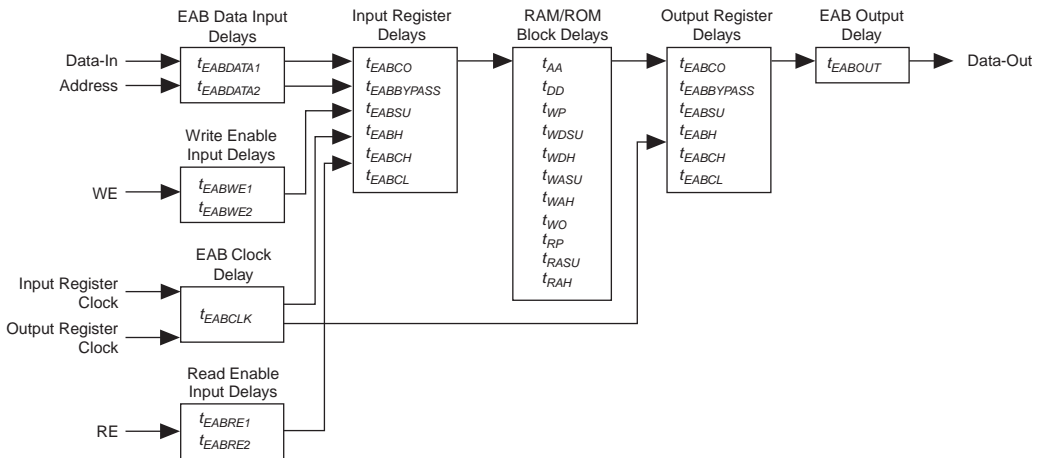Figure 27. FLEX 10KE Device EAB Timing Model

Table 24. LE Timing Microparameters (Part 2 of 2)   Note (1)

Symbol	Parameter	Condition
t_{CLR}	LE register clear delay	
t_{CH}	Minimum clock high time from clock pin	
t_{CL}	Minimum clock low time from clock pin	

Table 25. IOE Timing Microparameters   Note (1)

Symbol	Parameter	Conditions
t_{IOD}	IOE data delay	
t_{IOC}	IOE register control signal delay	
t_{IOCO}	IOE register clock-to-output delay	
t_{IOCOMB}	IOE combinatorial delay	
t_{IOSU}	IOE register setup time for data and enable signals before clock; IOE register recovery time after asynchronous clear	
t_{IOH}	IOE register hold time for data and enable signals after clock	
t_{IOCLR}	IOE register clear time	
t_{OD1}	Output buffer and pad delay, slow slew rate = off, V_{CCIO} = 3.3 V	C1 = 35 pF (2)
t_{OD2}	Output buffer and pad delay, slow slew rate = off, V_{CCIO} = 2.5 V	C1 = 35 pF (3)
t_{OD3}	Output buffer and pad delay, slow slew rate = on	C1 = 35 pF (4)
t_{XZ}	IOE output buffer disable delay	
t_{ZX1}	IOE output buffer enable delay, slow slew rate = off, V_{CCIO} = 3.3 V	C1 = 35 pF (2)
t_{ZX2}	IOE output buffer enable delay, slow slew rate = off, V_{CCIO} = 2.5 V	C1 = 35 pF (3)
t_{ZX3}	IOE output buffer enable delay, slow slew rate = on	C1 = 35 pF (4)
t_{INREG}	IOE input pad and buffer to IOE register delay	
t_{IOFD}	IOE register feedback delay	
t_{INCOMB}	IOE input pad and buffer to FastTrack Interconnect delay	

Table 53. EPF10K130E Device IOE Timing Microparameters *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{OD3}		4.0		5.6		7.5	ns
t_{XZ}		2.8		4.1		5.5	ns
t_{ZX1}		2.8		4.1		5.5	ns
t_{ZX2}		2.8		4.1		5.5	ns
t_{ZX3}		4.0		5.6		7.5	ns
t_{INREG}		2.5		3.0		4.1	ns
t_{IOFD}		0.4		0.5		0.6	ns
t_{INCOMB}		0.4		0.5		0.6	ns

Table 54. EPF10K130E Device EAB Internal Microparameters *(Part 1 of 2)* *Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{EABDATA1}$		1.5		2.0		2.6	ns
$t_{EABDATA2}$		0.0		0.0		0.0	ns
t_{EABWE1}		1.5		2.0		2.6	ns
t_{EABWE2}		0.3		0.4		0.5	ns
t_{EABRE1}		0.3		0.4		0.5	ns
t_{EABRE2}		0.0		0.0		0.0	ns
t_{EABCLK}		0.0		0.0		0.0	ns
t_{EABCO}		0.3		0.4		0.5	ns
$t_{EABBYPASS}$		0.1		0.1		0.2	ns
t_{EABSU}	0.8		1.0		1.4		ns
t_{EABH}	0.1		0.2		0.2		ns
t_{EABCLR}	0.3		0.4		0.5		ns
t_{AA}		4.0		5.0		6.6	ns
t_{WP}	2.7		3.5		4.7		ns
t_{RP}	1.0		1.3		1.7		ns
t_{WDSU}	1.0		1.3		1.7		ns
t_{WDH}	0.2		0.2		0.3		ns
t_{WASU}	1.6		2.1		2.8		ns
t_{WAH}	1.6		2.1		2.8		ns
t_{RASU}	3.0		3.9		5.2		ns
t_{RAH}	0.1		0.1		0.2		ns
t_{WO}		1.5		2.0		2.6	ns

Altera Corporation

Table 58. EPF10K130E External Bidirectional Timing Parameters *Notes (1), (2)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t$_{INSUBIDIR}$ (3)	2.2		2.4		3.2		ns
t$_{INHBIDIR}$ (3)	0.0		0.0		0.0		ns
t$_{INSUBIDIR}$ (4)	2.8		3.0		–		ns
t$_{INHBIDIR}$ (4)	0.0		0.0		–		ns
t$_{OUTCOBIDIR}$ (3)	2.0	5.0	2.0	7.0	2.0	9.2	ns
t$_{XZBIDIR}$ (3)		5.6		8.1		10.8	ns
t$_{ZXBIDIR}$ (3)		5.6		8.1		10.8	ns
t$_{OUTCOBIDIR}$ (4)	0.5	4.0	0.5	6.0	–	–	ns
t$_{XZBIDIR}$ (4)		4.6		7.1		–	ns
t$_{ZXBIDIR}$ (4)		4.6		7.1		–	ns

Notes to tables:

(1) All timing parameters are described in Tables 24 through 30 in this data sheet.
(2) These parameters are specified by characterization.
(3) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
(4) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Tables 59 through 65 show EPF10K200E device internal and external timing parameters.

Table 59. EPF10K200E Device LE Timing Microparameters *(Part 1 of 2) Note (1)*

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t$_{LUT}$		0.7		0.8		1.2	ns
t$_{CLUT}$		0.4		0.5		0.6	ns
t$_{RLUT}$		0.6		0.7		0.9	ns
t$_{PACKED}$		0.3		0.5		0.7	ns
t$_{EN}$		0.4		0.5		0.6	ns
t$_{CICO}$		0.2		0.2		0.3	ns
t$_{CGEN}$		0.4		0.4		0.6	ns
t$_{CGENR}$		0.2		0.2		0.3	ns
t$_{CASC}$		0.7		0.8		1.2	ns
t$_{C}$		0.5		0.6		0.8	ns
t$_{CO}$		0.5		0.6		0.8	ns
t$_{COMB}$		0.4		0.6		0.8	ns
t$_{SU}$	0.4		0.6		0.7		ns

Table 71. EPF10K50S External Timing Parameters   Note (1)

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t$_{DRR}$		8.0		9.5		12.5	ns
t$_{INSU}$ (2)	2.4		2.9		3.9		ns
t$_{INH}$ (2)	0.0		0.0		0.0		ns
t$_{OUTCO}$ (2)	2.0	4.3	2.0	5.2	2.0	7.3	ns
t$_{INSU}$ (3)	2.4		2.9				ns
t$_{INH}$ (3)	0.0		0.0				ns
t$_{OUTCO}$ (3)	0.5	3.3	0.5	4.1			ns
t$_{PCISU}$	2.4		2.9		–		ns
t$_{PCIH}$	0.0		0.0		–		ns
t$_{PCICO}$	2.0	6.0	2.0	7.7	–	–	ns

Table 72. EPF10K50S External Bidirectional Timing Parameters   Note (1)

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t$_{INSUBIDIR}$ (2)	2.7		3.2		4.3		ns
t$_{INHBIDIR}$ (2)	0.0		0.0		0.0		ns
t$_{INHBIDIR}$ (3)	0.0		0.0		–		ns
t$_{INSUBIDIR}$ (3)	3.7		4.2		–		ns
t$_{OUTCOBIDIR}$ (2)	2.0	4.5	2.0	5.2	2.0	7.3	ns
t$_{XZBIDIR}$ (2)		6.8		7.8		10.1	ns
t$_{ZXBIDIR}$ (2)		6.8		7.8		10.1	ns
t$_{OUTCOBIDIR}$ (3)	0.5	3.5	0.5	4.2	–	–	ns
t$_{XZBIDIR}$ (3)		6.8		8.4		–	ns
t$_{ZXBIDIR}$ (3)		6.8		8.4		–	ns

Notes to tables:

(1)   All timing parameters are described in Tables 24 through 30.

(2)   This parameter is measured without use of the ClockLock or ClockBoost circuits.

(3)   This parameter is measured with use of the ClockLock or ClockBoost circuits

Figure 31. FLEX 10KE I_CCACTIVE vs. Operating Frequency (Part 2 of 2)

EPF10K130E

EPF10K200E

EPF10K200S
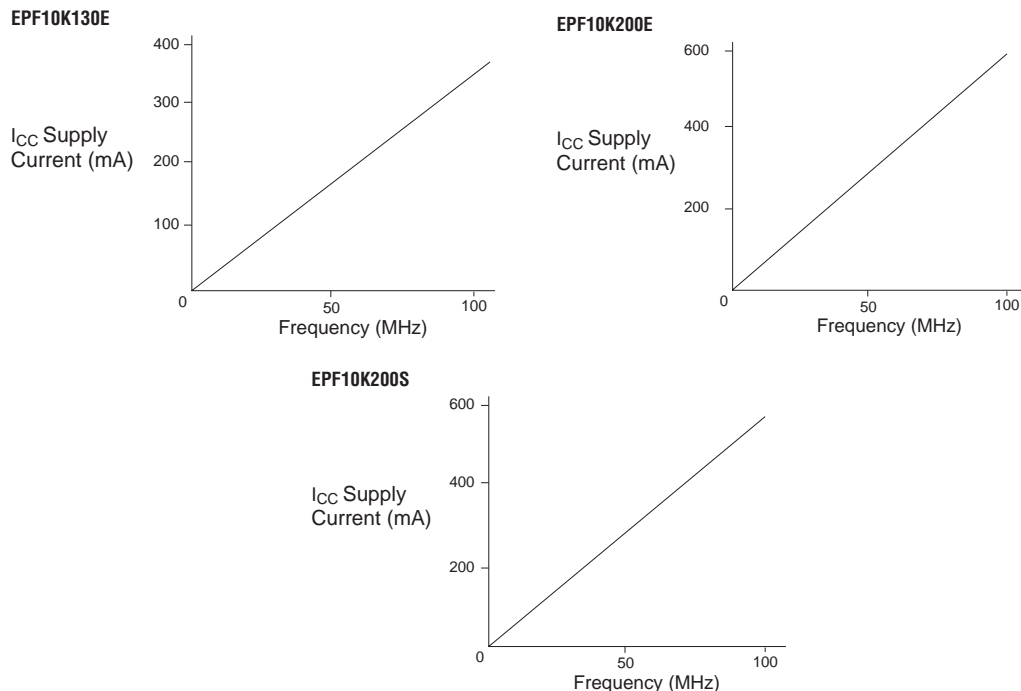
Configuration & Operation

The FLEX 10KE architecture supports several configuration schemes. This section summarizes the device operating modes and available device configuration schemes.

Operating Modes

The FLEX 10KE architecture uses SRAM configuration elements that require configuration data to be loaded every time the circuit powers up. The process of physically loading the SRAM data into the device is called *configuration*. Before configuration, as V_{CC} rises, the device initiates a Power-On Reset (POR). This POR event clears the device and prepares it for configuration. The FLEX 10KE POR time does not exceed 50 µs.

When configuring with a configuration device, refer to the respective configuration device data sheet for POR timing information.