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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I <sup>2</sup> C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	11
Program Memory Size	7KB (4K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	16-UQFN Exposed Pad
Supplier Device Package	16-UQFN (4x4)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1824-i-jq">https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1824-i-jq</a>

## 1.0 DEVICE OVERVIEW

The PIC16(L)F1824/8 are described within this data sheet. They are available in 14/20 pin packages. Figure 1-1 shows a block diagram of the PIC16(L)F1824/8 devices. Tables 1-2 and 1-3 show the pinout descriptions.

Reference Table 1-1 for peripherals available per device.

**TABLE 1-1: DEVICE PERIPHERAL SUMMARY**

Peripheral		PIC16(L)F1824	PIC16(L)F1828
ADC		•	•
Capacitive Sensing Module (CSM)		•	•
Data EEPROM		•	•
Digital-to-Analog Converter (DAC)		•	•
Digital Signal Modulator (DSM)		•	•
EUSART		•	•
Fixed Voltage Reference (FVR)		•	•
SR Latch		•	•
Capture/Compare/PWM Modules			
	ECCP1	•	•
	ECCP2	•	•
	CCP3	•	•
	CCP4	•	•
Comparators			
	C1	•	•
	C2	•	•
Master Synchronous Serial Ports			
	MSSP	•	•
Timers			
	Timer0	•	•
	Timer1	•	•
	Timer2	•	•
	Timer4	•	•
	Timer6	•	•

# PIC16(L)F1824/8

## 3.2.1.1 STATUS Register

The STATUS register, shown in Register 3-1, contains:

- the arithmetic status of the ALU
- the Reset status

The STATUS register can be the destination for any instruction, like any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the  $\overline{TO}$  and  $\overline{PD}$  bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, `CLRF STATUS` will clear the upper three bits and set the Z bit. This leaves the STATUS register as '000u u1uu' (where u = unchanged).

It is recommended, therefore, that only `BCF`, `BSF`, `SWAPF` and `MOVWF` instructions are used to alter the STATUS register, because these instructions do not affect any Status bits. For other instructions not affecting any Status bits (Refer to **Section 29.0 "Instruction Set Summary"**).

**Note 1:** The  $\overline{C}$  and  $\overline{DC}$  bits operate as Borrow and Digit Borrow out bits, respectively, in subtraction.

**REGISTER 3-1: STATUS: STATUS REGISTER**

U-0	U-0	U-0	R-1/q	R-1/q	R/W-0/u	R/W-0/u	R/W-0/u
—	—	—	$\overline{TO}$	$\overline{PD}$	Z	DC <sup>(1)</sup>	C <sup>(1)</sup>
bit 7			bit 0				

**Legend:**

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7-5	<b><math>\overline{TO}</math>:</b> Unimplemented: Read as '0'
bit 4	<b><math>\overline{TO}</math>:</b> Time-out bit 1 = After power-up, <code>CLRWDT</code> instruction or <code>SLEEP</code> instruction 0 = A WDT time-out occurred
bit 3	<b><math>\overline{PD}</math>:</b> Power-down bit 1 = After power-up or by the <code>CLRWDT</code> instruction 0 = By execution of the <code>SLEEP</code> instruction
bit 2	<b>Z:</b> Zero bit 1 = The result of an arithmetic or logic operation is zero 0 = The result of an arithmetic or logic operation is not zero
bit 1	<b>DC:</b> Digit Carry/Digit Borrow bit ( <code>ADDWF</code> , <code>ADDLW</code> , <code>SUBLW</code> , <code>SUBWF</code> instructions) <sup>(1)</sup> 1 = A carry-out from the 4th low-order bit of the result occurred 0 = No carry-out from the 4th low-order bit of the result
bit 0	<b>C:</b> Carry/Borrow bit ( <code>ADDWF</code> , <code>ADDLW</code> , <code>SUBLW</code> , <code>SUBWF</code> instructions) <sup>(1)</sup> 1 = A carry-out from the Most Significant bit of the result occurred 0 = No carry-out from the Most Significant bit of the result occurred

**Note 1:** For  $\overline{Borrow}$ , the polarity is reversed. A subtraction is executed by adding the two's complement of the second operand. For rotate (`RRF`, `RLF`) instructions, this bit is loaded with either the high-order or low-order bit of the source register.

## 7.11 Power Control (PCON) Register

The Power Control (PCON) register contains flag bits to differentiate between a:

- Power-on Reset ( $\overline{\text{POR}}$ )
- Brown-out Reset ( $\overline{\text{BOR}}$ )
- Reset Instruction Reset ( $\overline{\text{RI}}$ )
- Stack Overflow Reset (STKOVF)
- Stack Underflow Reset (STKUNF)
- MCLR Reset ( $\overline{\text{RMCLR}}$ )

The PCON register bits are shown in Register 7-2.

**REGISTER 7-2: PCON: POWER CONTROL REGISTER**

R/W/HS-0/q	R/W/HS-0/q	U-0	U-0	R/W/HC-1/q	R/W/HC-1/q	R/W/HC-q/u	R/W/HC-q/u
STKOVF	STKUNF	—	—	$\overline{\text{RMCLR}}$	$\overline{\text{RI}}$	$\overline{\text{POR}}$	$\overline{\text{BOR}}$
bit 7							bit 0

**Legend:**

HC = Bit is cleared by hardware

HS = Bit is set by hardware

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-m/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

q = Value depends on condition

bit 7	<b>STKOVF:</b> Stack Overflow Flag bit 1 = A Stack Overflow occurred 0 = A Stack Overflow has not occurred or set to '0' by firmware
bit 6	<b>STKUNF:</b> Stack Underflow Flag bit 1 = A Stack Underflow occurred 0 = A Stack Underflow has not occurred or set to '0' by firmware
bit 5-4	<b>Unimplemented:</b> Read as '0'
bit 3	<b><math>\overline{\text{RMCLR}}</math>:</b> MCLR Reset Flag bit 1 = A $\overline{\text{MCLR}}$ Reset has not occurred or set to '1' by firmware 0 = A $\overline{\text{MCLR}}$ Reset has occurred (set to '0' in hardware when a $\overline{\text{MCLR}}$ Reset occurs)
bit 2	<b><math>\overline{\text{RI}}</math>:</b> RESET Instruction Flag bit 1 = A RESET instruction has not been executed or set to '1' by firmware 0 = A RESET instruction has been executed (set to '0' in hardware upon executing a RESET instruction)
bit 1	<b><math>\overline{\text{POR}}</math>:</b> Power-on Reset Status bit 1 = No Power-on Reset occurred 0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs)
bit 0	<b><math>\overline{\text{BOR}}</math>:</b> Brown-out Reset Status bit 1 = No Brown-out Reset occurred 0 = A Brown-out Reset occurred (must be set in software after a Power-on Reset or Brown-out Reset occurs)

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## 11.6 Write Verify

Depending on the application, good programming practice may dictate that the value written to the data EEPROM or program memory should be verified (see Example 11-6) to the desired value to be written. Example 11-6 shows how to verify a write to EEPROM.

### EXAMPLE 11-6: EEPROM WRITE VERIFY

```
BANKSEL EEDATL      ;
MOVF    EEDATL, W    ;EEDATL not changed
                        ;from previous write
BSF      EECON1, RD   ;YES, Read the
                        ;value written
XORWF    EEDATL, W    ;
BTFSS    STATUS, Z    ;Is data the same
GOTO     WRITE_ERR    ;No, handle error
:         ;Yes, continue
```

## 13.0 INTERRUPT-ON-CHANGE

The PORTA pins can be configured to operate as Interrupt-on-Change (IOC) pins. On the PIC16(L)F1828 devices, the PORTB pins can also be configured to operate as IOC pins. An interrupt can be generated by detecting a signal that has either a rising edge or a falling edge. Any individual port pin, or combination of port pins, can be configured to generate an interrupt. The interrupt-on-change module has the following features:

- Interrupt-on-change enable (Master Switch)
- Individual pin configuration
- Rising and falling edge detection
- Individual pin interrupt flags

Figure 13-1 is a block diagram of the IOC module.

### 13.1 Enabling the Module

To allow individual port pins to generate an interrupt, the IOCIE bit of the INTCON register must be set. If the IOCIE bit is disabled, the edge detection on the pin will still occur, but an interrupt will not be generated.

### 13.2 Individual Pin Configuration

For each port pin, a rising edge detector and a falling edge detector are present. To enable a pin to detect a rising edge, the associated bit of the IOCxP register is set. To enable a pin to detect a falling edge, the associated bit of the IOCxN register is set.

A pin can be configured to detect rising and falling edges simultaneously by setting both associated bits of the IOCxP and IOCxN registers, respectively.

## 13.3 Interrupt Flags

The IOCAFx and IOCBFx bits located in the IOCAF and IOCBF registers, respectively, are status flags that correspond to the interrupt-on-change pins of the associated port. If an expected edge is detected on an appropriately enabled pin, then the status flag for that pin will be set, and an interrupt will be generated if the IOCIE bit is set. The IOCIF bit of the INTCON register reflects the status of all IOCAFx and IOCBFx bits.

### 13.4 Clearing Interrupt Flags

The individual status flags, (IOCAFx and IOCBFx bits), can be cleared by resetting them to zero. If another edge is detected during this clearing operation, the associated status flag will be set at the end of the sequence, regardless of the value actually being written.

In order to ensure that no detected edge is lost while clearing flags, only AND operations masking out known changed bits should be performed. The following sequence is an example of what should be performed.

#### EXAMPLE 13-1: CLEARING INTERRUPT FLAGS (PORTA EXAMPLE)

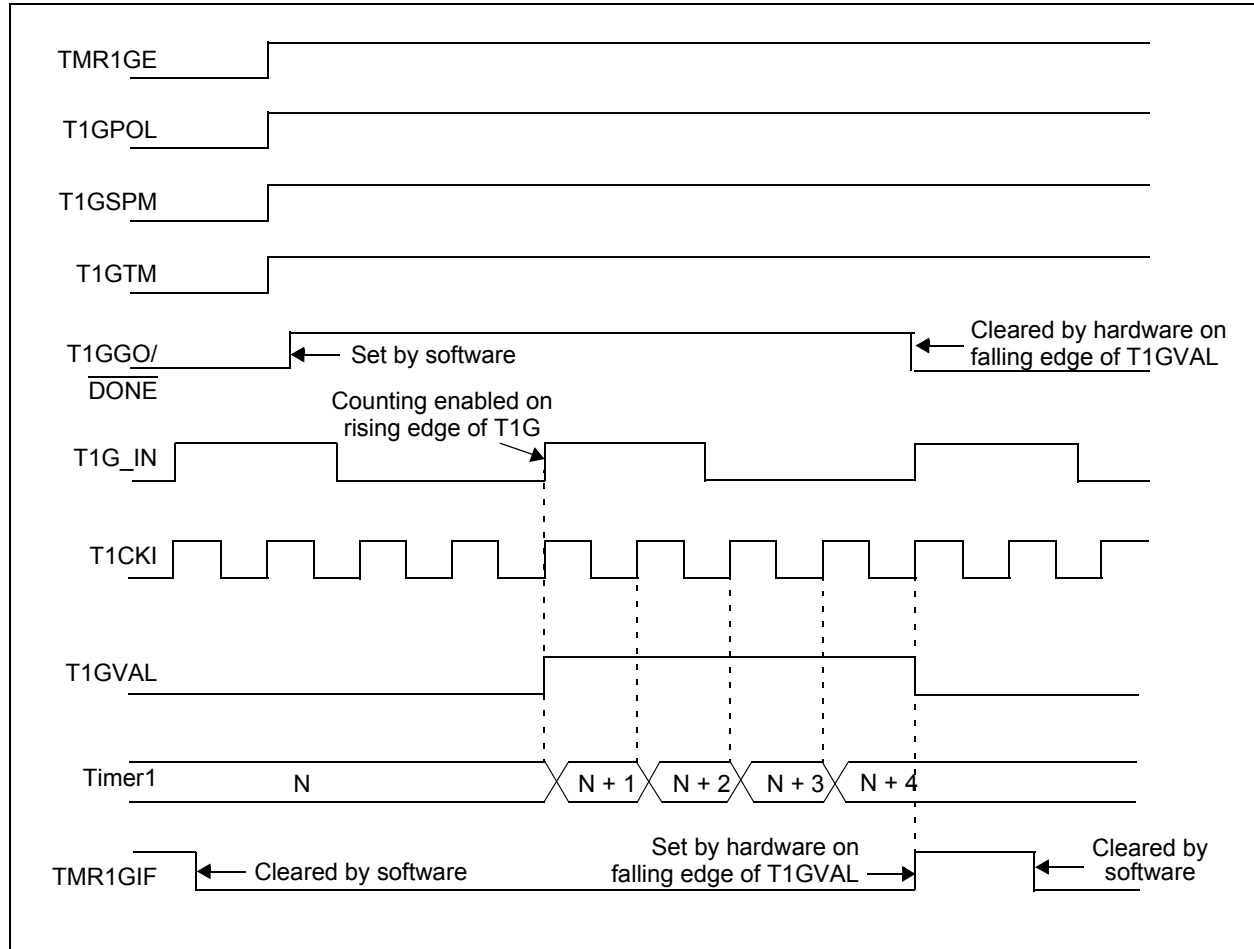
```
MOVLW    0xff
XORWF    IOCAF, W
ANDWF    IOCAF, F
```

### 13.5 Operation in Sleep

The interrupt-on-change interrupt sequence will wake the device from Sleep mode, if the IOCIE bit is set.

If an edge is detected while in Sleep mode, the IOCxF register will be updated prior to the first instruction executed out of Sleep.

**FIGURE 21-6: TIMER1 GATE SINGLE-PULSE AND TOGGLE COMBINED MODE**



## 22.0 TIMER2/4/6 MODULES

There are up to three identical Timer2-type modules available. To maintain pre-existing naming conventions, the Timers are called Timer2, Timer4 and Timer6 (also Timer2/4/6).

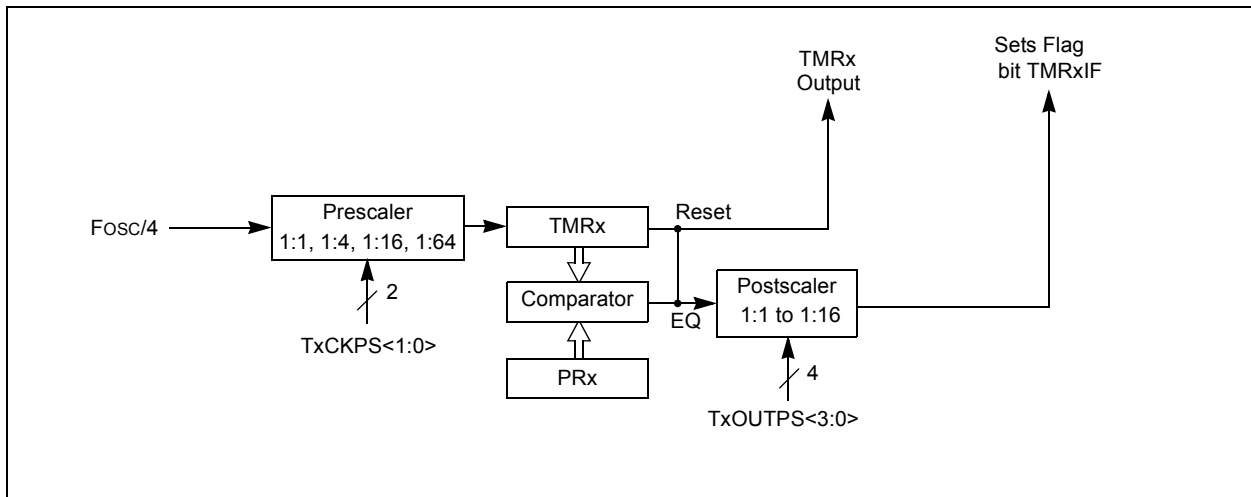
**Note:** The 'x' variable used in this section is used to designate Timer2, Timer4, or Timer6. For example, TxCON references T2CON, T4CON, or T6CON. PRx references PR2, PR4, or PR6.

The Timer2/4/6 modules incorporate the following features:

- 8-bit Timer and Period registers (TMRx and PRx, respectively)
- Readable and writable (both registers)
- Software programmable prescaler (1:1, 1:4, 1:16, and 1:64)
- Software programmable postscaler (1:1 to 1:16)
- Interrupt on TMRx match with PRx, respectively
- Optional use as the shift clock for the MSSPx modules (Timer2 only)

See Figure 22-1 for a block diagram of Timer2/4/6.

**FIGURE 22-1: TIMER2/4/6 BLOCK DIAGRAM**





## 25.2 SPI Mode Overview

The Serial Peripheral Interface (SPI) bus is a synchronous serial data communication bus that operates in Full Duplex mode. Devices communicate in a master/slave environment where the master device initiates the communication. A slave device is controlled through a chip select known as Slave Select.

The SPI bus specifies four signal connections:

- Serial Clock (SCK)
- Serial Data Out (SDO)
- Serial Data In (SDI)
- Slave Select ( $\overline{SS}$ )

Figure 25-1 shows the block diagram of the MSSP1 module when operating in SPI Mode.

The SPI bus operates with a single master device and one or more slave devices. When multiple slave devices are used, an independent Slave Select connection is required from the master device to each slave device.

Figure 25-4 shows a typical connection between a master device and multiple slave devices.

The master selects only one slave at a time. Most slave devices have tri-state outputs so their output signal appears disconnected from the bus when they are not selected.

Transmissions involve two shift registers, eight bits in size, one in the master and one in the slave. With either the master or the slave device, data is always shifted out one bit at a time, with the Most Significant bit (MSb) shifted out first. At the same time, a new Least Significant bit (LSb) is shifted into the same register.

Figure 25-5 shows a typical connection between two processors configured as master and slave devices.

Data is shifted out of both shift registers on the programmed clock edge and latched on the opposite edge of the clock.

The master device transmits information out on its SDO output pin which is connected to, and received by, the slave's SDI input pin. The slave device transmits information out on its SDO output pin, which is connected to, and received by, the master's SDI input pin.

To begin communication, the master device first sends out the clock signal. Both the master and the slave devices should be configured for the same clock polarity.

The master device starts a transmission by sending out the MSb from its shift register. The slave device reads this bit from that same line and saves it into the LSb position of its shift register.

During each SPI clock cycle, a full duplex data transmission occurs. This means that while the master device is sending out the MSb from its shift register (on its SDO pin) and the slave device is reading this bit and

saving it as the LSb of its shift register, that the slave device is also sending out the MSb from its shift register (on its SDO pin) and the master device is reading this bit and saving it as the LSb of its shift register.

After eight bits have been shifted out, the master and slave have exchanged register values.

If there is more data to exchange, the shift registers are loaded with new data and the process repeats itself.

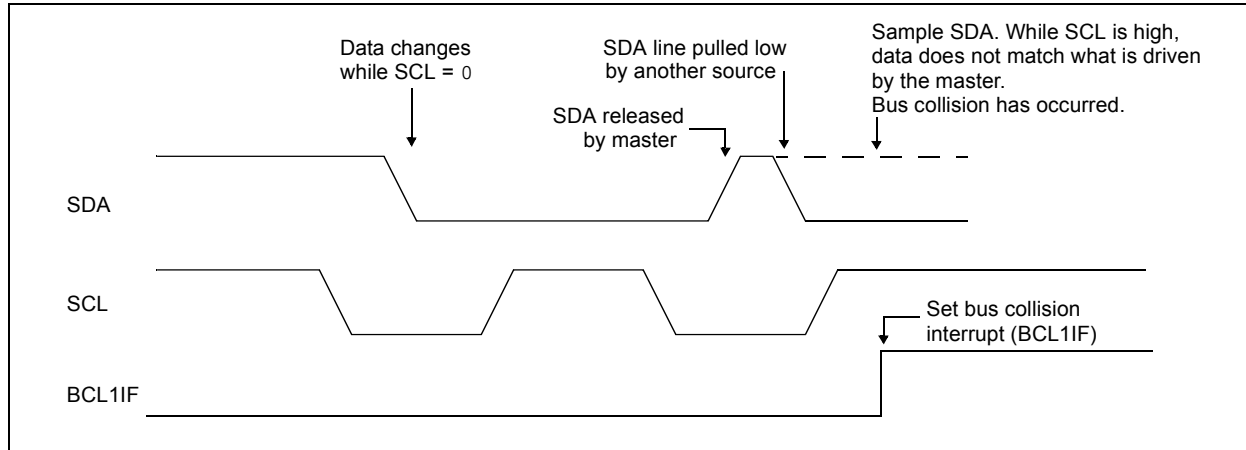
Whether the data is meaningful or not (dummy data), depends on the application software. This leads to three scenarios for data transmission:

- Master sends useful data and slave sends dummy data.
- Master sends useful data and slave sends useful data.
- Master sends dummy data and slave sends useful data.

Transmissions may involve any number of clock cycles. When there is no more data to be transmitted, the master stops sending the clock signal and it deselects the slave.

Every slave device connected to the bus that has not been selected through its slave select line must disregard the clock and transmission signals and must not transmit out any data of its own.

**FIGURE 25-32: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE**



## 25.6.13.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

- SDA or SCL are sampled low at the beginning of the Start condition (Figure 25-33).
- SCL is sampled low before SDA is asserted low (Figure 25-34).

During a Start condition, both the SDA and the SCL pins are monitored.

If the SDA pin is already low, or the SCL pin is already low, then all of the following occur:

- the Start condition is aborted,
- the BCL1IF flag is set and
- the MSSP1 module is reset to its Idle state (Figure 25-33).

The Start condition begins with the SDA and SCL pins deasserted. When the SDA pin is sampled high, the Baud Rate Generator is loaded and counts down. If the SCL pin is sampled low while SDA is high, a bus collision occurs because it is assumed that another master is attempting to drive a data '1' during the Start condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 25-35). If, however, a '1' is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to zero; if the SCL pin is sampled as '0' during this time, a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

**Note:** The reason that bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.

## REGISTER 25-2: SSP1CON1: SSP1 CONTROL REGISTER 1

R/C/HS-0/0	R/C/HS-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
WCOL	SSPOV	SSPEN	CKP	SSPM<3:0>			
bit 7							bit 0

### Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	HS = Bit is set by hardware C = User cleared

bit 7	<b>WCOL:</b> Write Collision Detect bit <b>Master mode:</b> 1 = A write to the SSP1BUF register was attempted while the I <sup>2</sup> C™ conditions were not valid for a transmission to be started 0 = No collision <b>Slave mode:</b> 1 = The SSP1BUF register is written while it is still transmitting the previous word (must be cleared in software) 0 = No collision
bit 6	<b>SSPOV:</b> Receive Overflow Indicator bit <sup>(1)</sup> <b>In SPI mode:</b> 1 = A new byte is received while the SSP1BUF register is still holding the previous data. In case of overflow, the data in SSP1SR is lost. Overflow can only occur in Slave mode. In Slave mode, the user must read the SSP1BUF, even if only transmitting data, to avoid setting overflow. In Master mode, the overflow bit is not set since each new reception (and transmission) is initiated by writing to the SSP1BUF register (must be cleared in software). 0 = No overflow <b>In I<sup>2</sup>C mode:</b> 1 = A byte is received while the SSP1BUF register is still holding the previous byte. SSPOV is a "don't care" in Transmit mode (must be cleared in software). 0 = No overflow
bit 5	<b>SSPEN:</b> Synchronous Serial Port Enable bit In both modes, when enabled, these pins must be properly configured as input or output <b>In SPI mode:</b> 1 = Enables serial port and configures SCK, SDO, SDI and $\overline{SS}$ as the source of the serial port pins <sup>(2)</sup> 0 = Disables serial port and configures these pins as I/O port pins <b>In I<sup>2</sup>C mode:</b> 1 = Enables the serial port and configures the SDA and SCL pins as the source of the serial port pins <sup>(3)</sup> 0 = Disables serial port and configures these pins as I/O port pins
bit 4	<b>CKP:</b> Clock Polarity Select bit <b>In SPI mode:</b> 1 = Idle state for clock is a high level 0 = Idle state for clock is a low level <b>In I<sup>2</sup>C Slave mode:</b> SCL release control 1 = Enable clock 0 = Holds clock low (clock stretch). (Used to ensure data setup time.) <b>In I<sup>2</sup>C Master mode:</b> Unused in this mode
bit 3-0	<b>SSPM&lt;3:0&gt;:</b> Synchronous Serial Port Mode Select bits 0000 = SPI Master mode, clock = Fosc/4 0001 = SPI Master mode, clock = Fosc/16 0010 = SPI Master mode, clock = Fosc/64 0011 = SPI Master mode, clock = TMR2 output/2 0100 = SPI Slave mode, clock = SCK pin, $\overline{SS}$ pin control enabled 0101 = I <sup>2</sup> C Slave mode, clock = SCK pin, $\overline{SS}$ pin control disabled, $\overline{SS}$ can be used as I/O pin 0110 = I <sup>2</sup> C Slave mode, 7-bit address 0111 = I <sup>2</sup> C Slave mode, 10-bit address 1000 = I <sup>2</sup> C Master mode, clock = Fosc/(4 * (SSP1ADD+1)) <sup>(4)</sup> 1001 = Reserved 1010 = SPI Master mode, clock = Fosc/(4 * (SSP1ADD+1)) <sup>(5)</sup> 1011 = I <sup>2</sup> C firmware controlled Master mode (Slave idle) 1100 = Reserved 1101 = Reserved 1110 = I <sup>2</sup> C Slave mode, 7-bit address with Start and Stop bit interrupts enabled 1111 = I <sup>2</sup> C Slave mode, 10-bit address with Start and Stop bit interrupts enabled

- Note**
- 1: In Master mode, the overflow bit is not set since each new reception (and transmission) is initiated by writing to the SSP1BUF register.
  - 2: When enabled, these pins must be properly configured as input or output.
  - 3: When enabled, the SDA and SCL pins must be configured as inputs.
  - 4: SSP1ADD values of 0, 1 or 2 are not supported for I<sup>2</sup>C™ mode.
  - 5: SSPxADD value of 0 is not supported. Use SSPxM = 0000 instead.

BCF	Bit Clear f
Syntax:	[ <i>label</i> ] BCF f,b
Operands:	$0 \leq f \leq 127$ $0 \leq b \leq 7$
Operation:	$0 \rightarrow (f \ll b)$
Status Affected:	None
Description:	Bit 'b' in register 'f' is cleared.

BTFSC	Bit Test f, Skip if Clear
Syntax:	[ <i>label</i> ] BTFSC f,b
Operands:	$0 \leq f \leq 127$ $0 \leq b \leq 7$
Operation:	skip if $(f \ll b) = 0$
Status Affected:	None
Description:	If bit 'b' in register 'f' is '1', the next instruction is executed. If bit 'b', in register 'f', is '0', the next instruction is discarded, and a NOP is executed instead, making this a 2-cycle instruction.

BRA	Relative Branch
Syntax:	[ <i>label</i> ] BRA label [ <i>label</i> ] BRA \$+k
Operands:	$-256 \leq \text{label} - \text{PC} + 1 \leq 255$ $-256 \leq k \leq 255$
Operation:	$(\text{PC}) + 1 + k \rightarrow \text{PC}$
Status Affected:	None
Description:	Add the signed 9-bit literal 'k' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $\text{PC} + 1 + k$ . This instruction is a 2-cycle instruction. This branch has a limited range.

BTFSS	Bit Test f, Skip if Set
Syntax:	[ <i>label</i> ] BTFSS f,b
Operands:	$0 \leq f \leq 127$ $0 \leq b < 7$
Operation:	skip if $(f \ll b) = 1$
Status Affected:	None
Description:	If bit 'b' in register 'f' is '0', the next instruction is executed. If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.

BRW	Relative Branch with W
Syntax:	[ <i>label</i> ] BRW
Operands:	None
Operation:	$(\text{PC}) + (W) \rightarrow \text{PC}$
Status Affected:	None
Description:	Add the contents of W (unsigned) to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $\text{PC} + 1 + (W)$ . This instruction is a 2-cycle instruction.

BSF	Bit Set f
Syntax:	[ <i>label</i> ] BSF f,b
Operands:	$0 \leq f \leq 127$ $0 \leq b \leq 7$
Operation:	$1 \rightarrow (f \ll b)$
Status Affected:	None
Description:	Bit 'b' in register 'f' is set.

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## CALL Call Subroutine

Syntax: [ *label* ] CALL k

Operands:  $0 \leq k \leq 2047$

Operation: (PC)+1 → TOS,  
k → PC<10:0>,  
(PCLATH<4:3>) → PC<12:11>

Status Affected: None

Description: Call Subroutine. First, return address (PC + 1) is pushed onto the stack. The 11-bit immediate address is loaded into PC bits <10:0>. The upper bits of the PC are loaded from PCLATH. CALL is a 2-cycle instruction.

## CALLW Subroutine Call With W

Syntax: [ *label* ] CALLW

Operands: None

Operation: (PC) + 1 → TOS,  
(W) → PC<7:0>,  
(PCLATH<6:0>) → PC<14:8>

Status Affected: None

Description: Subroutine call with W. First, the return address (PC + 1) is pushed onto the return stack. Then, the contents of W is loaded into PC<7:0>, and the contents of PCLATH into PC<14:8>. CALLW is a 2-cycle instruction.

## CLRF Clear f

Syntax: [ *label* ] CLRF f

Operands:  $0 \leq f \leq 127$

Operation: 00h → (f)  
1 → Z

Status Affected: Z

Description: The contents of register 'f' are cleared and the Z bit is set.

## CLRW Clear W

Syntax: [ *label* ] CLRW

Operands: None

Operation: 00h → (W)  
1 → Z

Status Affected: Z

Description: W register is cleared. Zero bit (Z) is set.

## CLRWDTClear Watchdog Timer

Syntax: [ *label* ] CLRWDTClear Watchdog Timer

Operands: None

Operation: 00h → WDT  
0 → WDT prescaler,  
1 →  $\overline{\text{TO}}$   
1 →  $\overline{\text{PD}}$

Status Affected:  $\overline{\text{TO}}$ ,  $\overline{\text{PD}}$

Description: CLRWDTClear Watchdog Timer. It also resets the prescaler of the WDT. Status bits  $\overline{\text{TO}}$  and  $\overline{\text{PD}}$  are set.

## COMF Complement f

Syntax: [ *label* ] COMF f,d

Operands:  $0 \leq f \leq 127$   
d ∈ [0,1]

Operation: ( $\bar{f}$ ) → (destination)

Status Affected: Z

Description: The contents of register 'f' are complemented. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

## DECF Decrement f

Syntax: [ *label* ] DECF f,d

Operands:  $0 \leq f \leq 127$   
d ∈ [0,1]

Operation: (f) - 1 → (destination)

Status Affected: Z

Description: Decrement register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

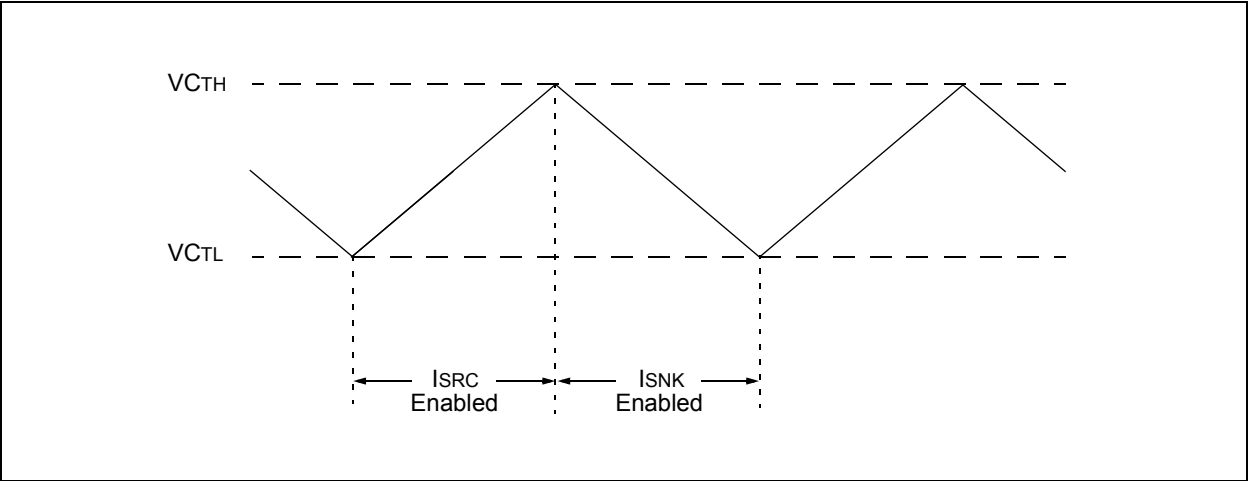
# PIC16(L)F1824/8

TABLE 30-17: CAP SENSE OSCILLATOR SPECIFICATIONS

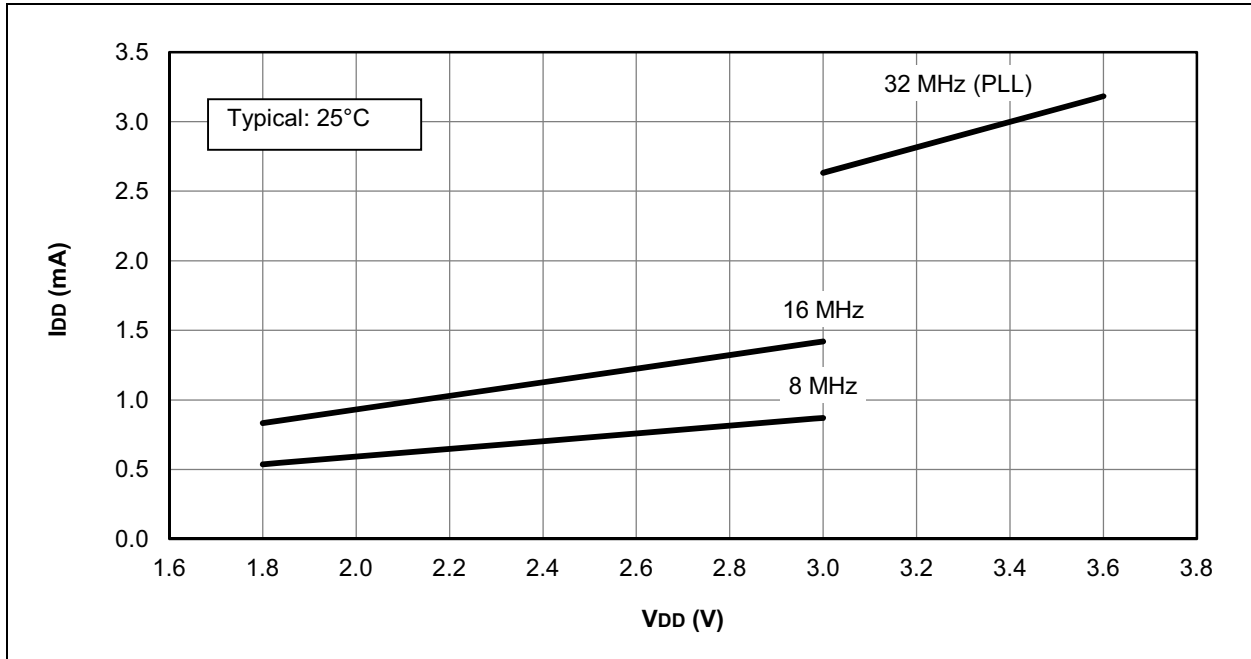
Param. No.	Symbol	Characteristic		Min.	Typ†	Max.	Units	Conditions
CS01	ISRC	Current Source	High	—	-8	—	μA	
			Medium	—	-1.5	—	μA	
			Low	—	-0.3	—	μA	
CS02	ISNK	Current Sink	High	—	7.5	—	μA	
			Medium	—	1.5	—	μA	
			Low	—	0.25	—	μA	
CS03	VCTH	Cap Threshold		—	0.8	—	V	
CS04	VCTL	Cap Threshold		—	0.4	—	V	
CS05	VCHYST	Cap Hysteresis (VCTH - VCTL)	High	—	525	—	mV	
			Medium	—	375	—	mV	
			Low	—	300	—	mV	

† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

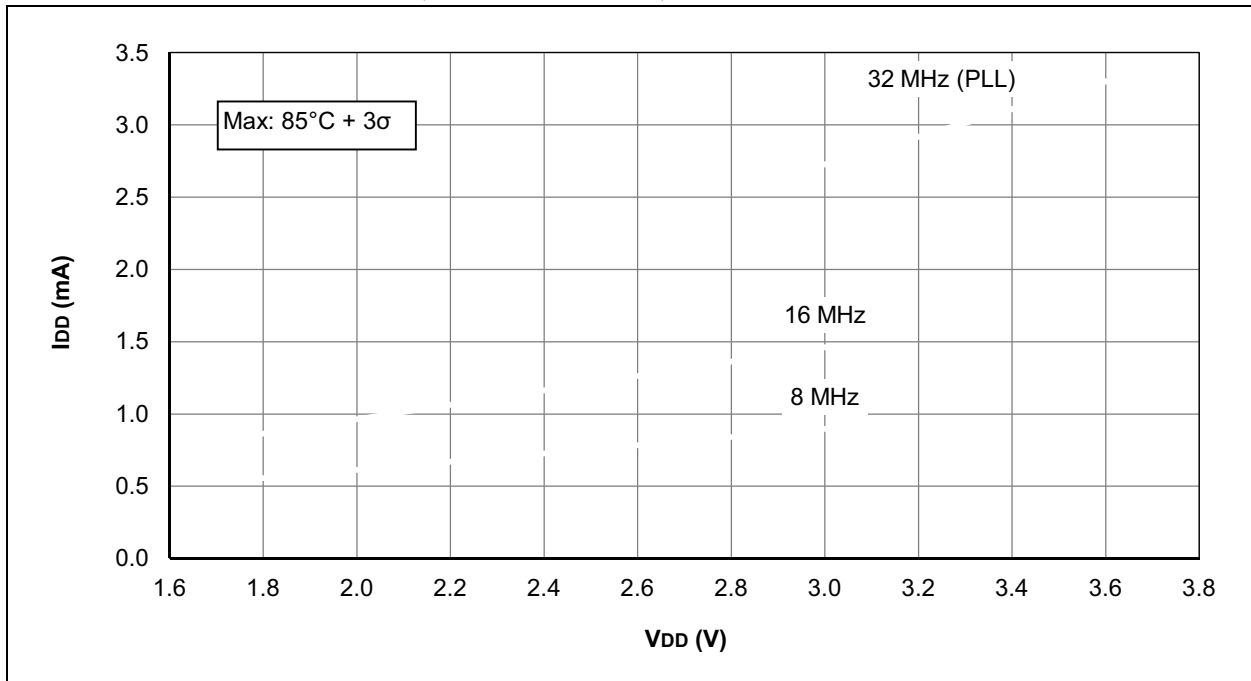
FIGURE 30-22: CAP SENSE OSCILLATOR



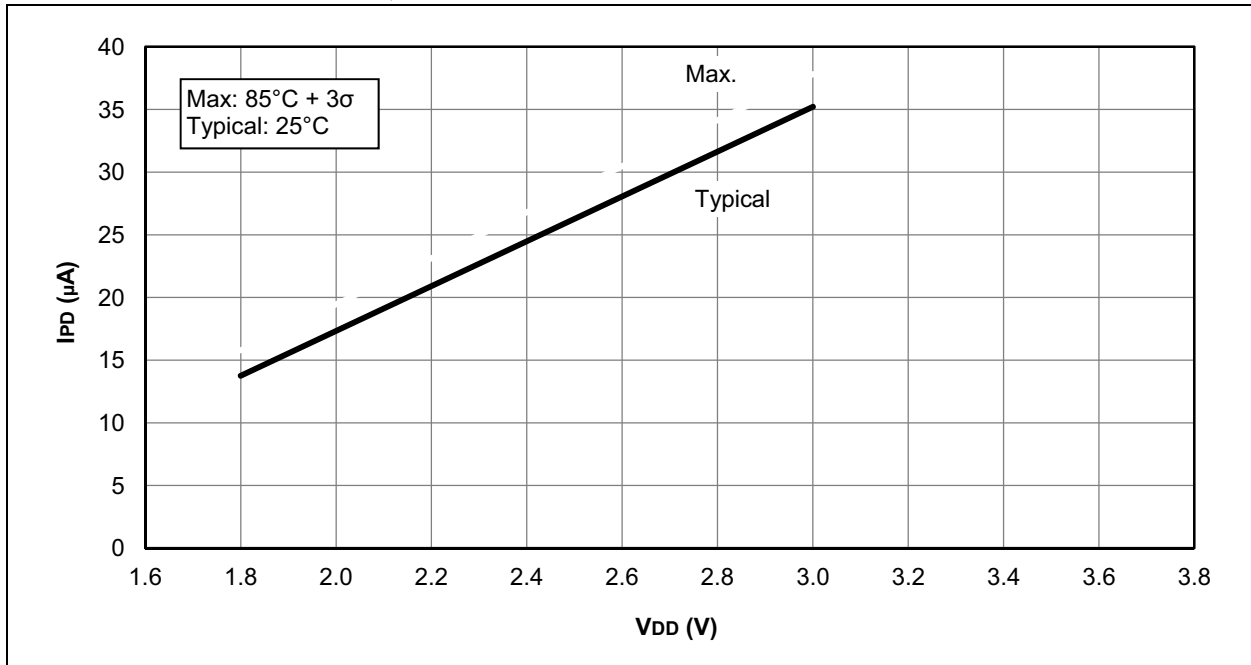
**FIGURE 31-15: I<sub>DD</sub> TYPICAL, HFINTOSC MODE, PIC16LF1824/8 ONLY**



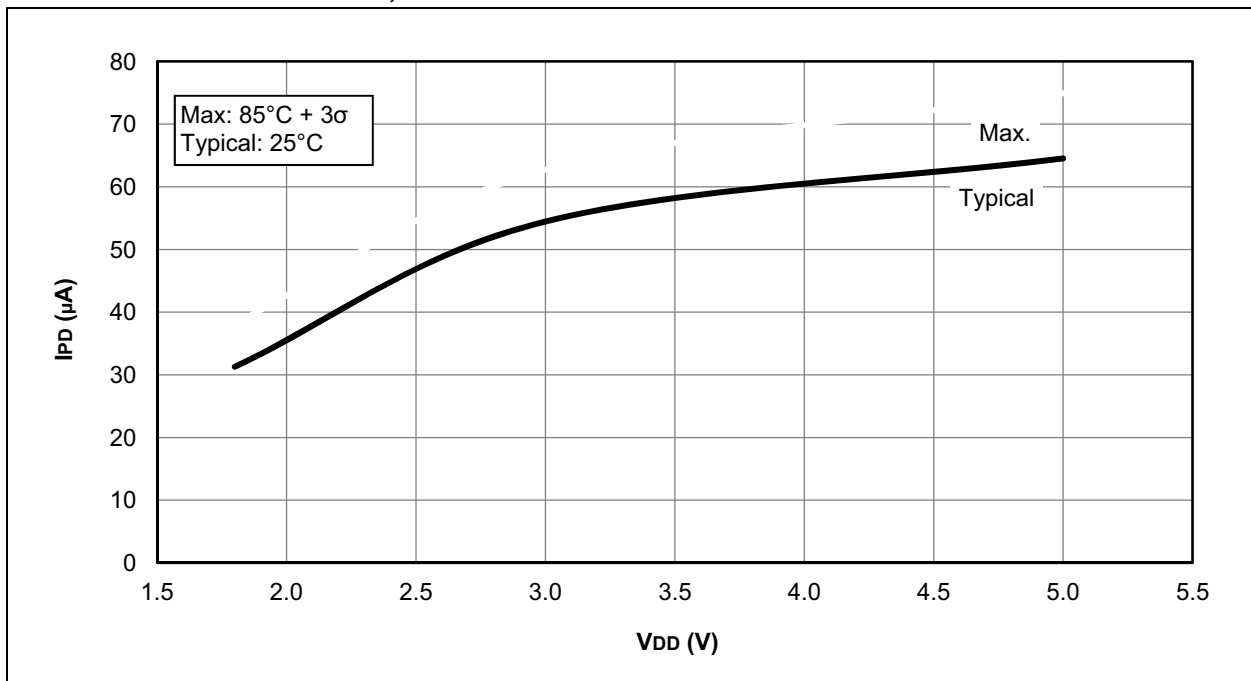
**FIGURE 31-16: I<sub>DD</sub> MAXIMUM, HFINTOSC MODE, PIC16LF1824/8 ONLY**



**FIGURE 31-35: I<sub>PD</sub>, CAPACITIVE SENSING (CPS) MODULE, HIGH-CURRENT RANGE, CPSRM = 0, PIC16LF1824/8 ONLY**

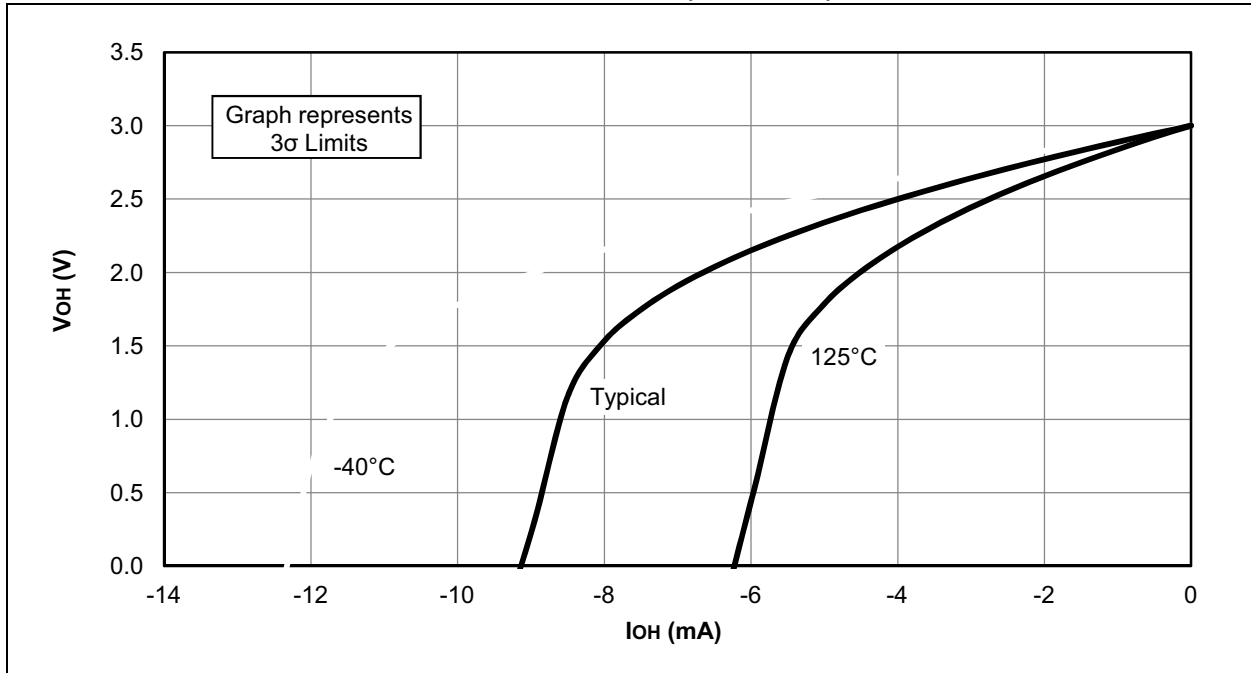


**FIGURE 31-36: I<sub>PD</sub>, CAPACITIVE SENSING (CPS) MODULE, HIGH-CURRENT RANGE, CPSRM = 0, PIC16F1824/8 ONLY**

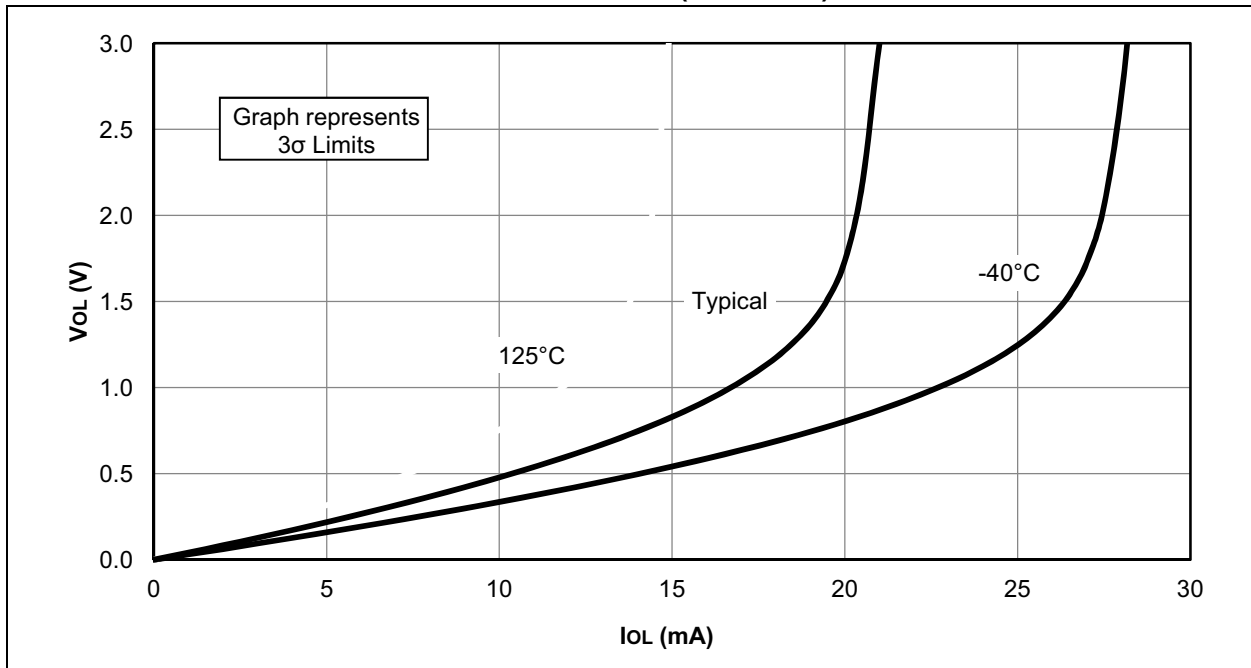




**FIGURE 31-43:  $V_{OH}$  vs.  $I_{OH}$  OVER TEMPERATURE ( $V_{DD} = 3.0V$ )**



**FIGURE 31-44:  $V_{OL}$  vs.  $I_{OL}$  OVER TEMPERATURE ( $V_{DD} = 3.0V$ )**



## 32.6 MPLAB X SIM Software Simulator

The MPLAB X SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB X SIM Software Simulator fully supports symbolic debugging using the MPLAB XC Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

## 32.7 MPLAB REAL ICE In-Circuit Emulator System

The MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs all 8, 16 and 32-bit MCU, and DSC devices with the easy-to-use, powerful graphical user interface of the MPLAB X IDE.

The emulator is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with in-circuit debugger systems (RJ-11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradable through future firmware downloads in MPLAB X IDE. MPLAB REAL ICE offers significant advantages over competitive emulators including full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, logic probes, a ruggedized probe interface and long (up to three meters) interconnection cables.

## 32.8 MPLAB ICD 3 In-Circuit Debugger System

The MPLAB ICD 3 In-Circuit Debugger System is Microchip's most cost-effective, high-speed hardware debugger/programmer for Microchip Flash DSC and MCU devices. It debugs and programs PIC Flash microcontrollers and dsPIC DSCs with the powerful, yet easy-to-use graphical user interface of the MPLAB IDE.

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

## 32.9 PICkit 3 In-Circuit Debugger/Programmer

The MPLAB PICkit 3 allows debugging and programming of PIC and dsPIC Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB IDE. The MPLAB PICkit 3 is connected to the design engineer's PC using a full-speed USB interface and can be connected to the target via a Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the Reset line to implement in-circuit debugging and In-Circuit Serial Programming™ (ICSP™).

## 32.10 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages, and a modular, detachable socket assembly to support various package types. The ICSP cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices, and incorporates an MMC card for file storage and data applications.

## 32.11 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM™ and dsPICDEM™ demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ® security ICs, CAN, IrDA®, PowerSmart battery management, SEEVAL® evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page ([www.microchip.com](http://www.microchip.com)) for the complete list of demonstration, development and evaluation kits.

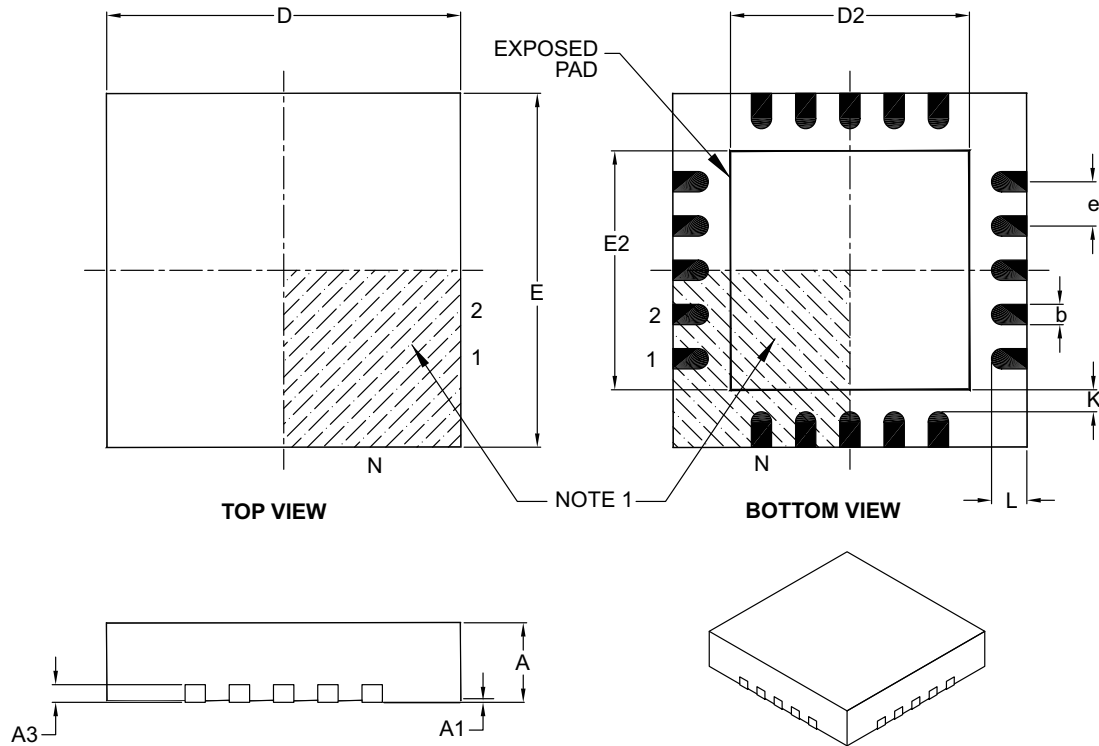
## 32.12 Third-Party Development Tools

Microchip also offers a great collection of tools from third-party vendors. These tools are carefully selected to offer good value and unique functionality.

- Device Programmers and Gang Programmers from companies, such as SoftLog and CCS
- Software Tools from companies, such as Gimpel and Trace Systems
- Protocol Analyzers from companies, such as Saleae and Total Phase
- Demonstration Boards from companies, such as MikroElektronika, Digilent® and Olimex
- Embedded Ethernet Solutions from companies, such as EZ Web Lynx, WIZnet and IPLogika®

## 20-Lead Plastic Quad Flat, No Lead Package (ML) – 4x4x0.9 mm Body [QFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N	20		
Pitch	e	0.50 BSC		
Overall Height	A	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3	0.20 REF		
Overall Width	E	4.00 BSC		
Exposed Pad Width	E2	2.60	2.70	2.80
Overall Length	D	4.00 BSC		
Exposed Pad Length	D2	2.60	2.70	2.80
Contact Width	b	0.18	0.25	0.30
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	K	0.20	—	—

### Notes:

- Pin 1 visual index feature may vary, but must be located within the hatched area.
- Package is saw singulated.
- Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
  - REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-126B

