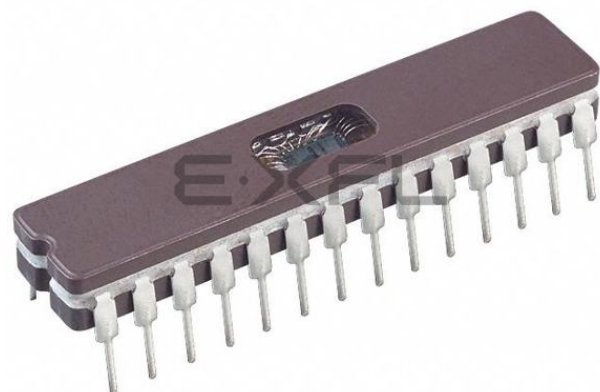


Welcome to [E-XFL.COM](https://www.e-xfl.com)

What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"



Details

Product Status	Obsolete
Core Processor	PIC
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LVD, POR, PWM, WDT
Number of I/O	22
Program Memory Size	16KB (8K x 16)
Program Memory Type	EPROM, UV
EEPROM Size	-
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	4.2V ~ 5.5V
Data Converters	A/D 5x10b
Oscillator Type	External
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Through Hole
Package / Case	28-CDIP (0.300", 7.62mm) Window
Supplier Device Package	28-CerDip
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18c242-jw

PIC18CXX2

TABLE 3-3: INITIALIZATION CONDITIONS FOR ALL REGISTERS

Register	Applicable Devices				Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt
TOSU	242	442	252	452	---0 0000	---0 0000	---0 uuuu ⁽³⁾
TOSH	242	442	252	452	0000 0000	0000 0000	uuuu uuuu ⁽³⁾
TOSL	242	442	252	452	0000 0000	0000 0000	uuuu uuuu ⁽³⁾
STKPTR	242	442	252	452	00-0 0000	00-0 0000	uu-u uuuu ⁽³⁾
PCLATU	242	442	252	452	---0 0000	---0 0000	---u uuuu
PCLATH	242	442	252	452	0000 0000	0000 0000	uuuu uuuu
PCL	242	442	252	452	0000 0000	0000 0000	PC + 2 ⁽²⁾
TBLPTRU	242	442	252	452	--00 0000	--00 0000	--uu uuuu
TBLPTRH	242	442	252	452	0000 0000	0000 0000	uuuu uuuu
TBLPTRL	242	442	252	452	0000 0000	0000 0000	uuuu uuuu
TABLAT	242	442	252	452	0000 0000	0000 0000	uuuu uuuu
PRODH	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
PRODL	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
INTCON	242	442	252	452	0000 000x	0000 000u	uuuu uuuu ⁽¹⁾
INTCON2	242	442	252	452	1111 -1-1	1111 -1-1	uuuu -u-u ⁽¹⁾
INTCON3	242	442	252	452	11-0 0-00	11-0 0-00	uu-u u-uu ⁽¹⁾
INDF0	242	442	252	452	N/A	N/A	N/A
POSTINC0	242	442	252	452	N/A	N/A	N/A
POSTDEC0	242	442	252	452	N/A	N/A	N/A
PREINC0	242	442	252	452	N/A	N/A	N/A
PLUSW0	242	442	252	452	N/A	N/A	N/A
FSR0H	242	442	252	452	---- 0000	---- 0000	---- uuuu
FSR0L	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
WREG	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
INDF1	242	442	252	452	N/A	N/A	N/A
POSTINC1	242	442	252	452	N/A	N/A	N/A
POSTDEC1	242	442	252	452	N/A	N/A	N/A
PREINC1	242	442	252	452	N/A	N/A	N/A
PLUSW1	242	442	252	452	N/A	N/A	N/A

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition

Note 1: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

- 2:** When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).
- 3:** When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.
- 4:** See Table 3-2 for RESET value for specific condition.
- 5:** Bit 6 of PORTA, LATA, and TRISA are enabled in ECIO and RCIO oscillator modes only. In all other oscillator modes, they are disabled and read '0'.
- 6:** The long write enable is only reset on a POR or MCLR Reset.
- 7:** Bit 6 of PORTA, LATA and TRISA are not available on all devices. When unimplemented, they are read as '0'.

TABLE 3-3: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Register	Applicable Devices				Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt
TRISE	242	442	252	452	0000 -111	0000 -111	uuuu -uuu
TRISD	242	442	252	452	1111 1111	1111 1111	uuuu uuuu
TRISC	242	442	252	452	1111 1111	1111 1111	uuuu uuuu
TRISB	242	442	252	452	1111 1111	1111 1111	uuuu uuuu
TRISA ^(5, 7)	242	442	252	452	-111 1111 ⁽⁵⁾	-111 1111 ⁽⁵⁾	-uuu uuuu ⁽⁵⁾
LATE	242	442	252	452	---- -xxx	---- -uuu	---- -uuu
LATD	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
LATC	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
LATB	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
LATA ^(5, 7)	242	442	252	452	-xxx xxxx ⁽⁵⁾	-uuu uuuu ⁽⁵⁾	-uuu uuuu ⁽⁵⁾
PORTE	242	442	252	452	---- -000	---- -000	---- -uuu
PORTD	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTC	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTB	242	442	252	452	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTA ^(5, 7)	242	442	252	452	-x0x 0000 ⁽⁵⁾	-u0u 0000 ⁽⁵⁾	-uuu uuuu ⁽⁵⁾

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition

Note 1: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

4: See Table 3-2 for RESET value for specific condition.

5: Bit 6 of PORTA, LATA, and TRISA are enabled in ECIO and RCIO oscillator modes only. In all other oscillator modes, they are disabled and read '0'.

6: The long write enable is only reset on a POR or MCLR Reset.

7: Bit 6 of PORTA, LATA and TRISA are not available on all devices. When unimplemented, they are read as '0'.

4.0 MEMORY ORGANIZATION

There are two memory blocks in Enhanced MCU devices. These memory blocks are:

- Program Memory
- Data Memory

Program and data memory use separate buses so that concurrent access can occur.

4.1 Program Memory Organization

A 21-bit program counter is capable of addressing the 2-Mbyte program memory space. Accessing a location between the physically implemented memory and the 2-Mbyte address will cause a read of all '0's (a NOP instruction).

PIC18C252 and PIC18C452 have 32 Kbytes of EPROM, while PIC18C242 and PIC18C442 have 16 Kbytes of EPROM. This means that PIC18CX52 devices can store up to 16K of single word instructions, and PIC18CX42 devices can store up to 8K of single word instructions.

The RESET vector address is at 0000h and the interrupt vector addresses are at 0008h and 0018h.

Figure 4-1 shows the Program Memory Map for PIC18C242/442 devices and Figure 4-2 shows the Program Memory Map for PIC18C252/452 devices.

4.2 Return Address Stack

The return address stack allows any combination of up to 31 program calls and interrupts to occur. The PC (Program Counter) is pushed onto the stack when a `CALL` or `RCALL` instruction is executed, or an interrupt is acknowledged. The PC value is pulled off the stack on a `RETURN`, `RETLW` or a `RETFIE` instruction. `PCLATU` and `PCLATH` are not affected by any of the call or return instructions.

The stack operates as a 31-word by 21-bit RAM and a 5-bit stack pointer, with the stack pointer initialized to 00000b after all RESETS. There is no RAM associated with stack pointer 00000b. This is only a RESET value. During a `CALL` type instruction causing a push onto the stack, the stack pointer is first incremented and the RAM location pointed to by the stack pointer is written with the contents of the PC. During a `RETURN` type instruction causing a pop from the stack, the contents of the RAM location pointed to by the `STKPTR` is transferred to the PC and then the stack pointer is decremented.

The stack space is not part of either program or data space. The stack pointer is readable and writable, and the address on the top of the stack is readable and writable through SFR registers. Data can also be pushed to, or popped from, the stack, using the top-of-stack SFRs. Status bits indicate if the stack pointer is at, or beyond the 31 levels provided.

4.2.1 TOP-OF-STACK ACCESS

The top of the stack is readable and writable. Three register locations, `TOSU`, `TOSH` and `TOSL` hold the contents of the stack location pointed to by the `STKPTR` register. This allows users to implement a software stack, if necessary. After a `CALL`, `RCALL` or interrupt, the software can read the pushed value by reading the `TOSU`, `TOSH` and `TOSL` registers. These values can be placed on a user defined software stack. At return time, the software can replace the `TOSU`, `TOSH` and `TOSL` and do a return.

The user must disable the global interrupt enable bits during this time to prevent inadvertent stack operations.

4.2.2 RETURN STACK POINTER (STKPTR)

The `STKPTR` register contains the stack pointer value, the `STKFUL` (stack full) status bit, and the `STKUNF` (stack underflow) status bits. Register 4-1 shows the `STKPTR` register. The value of the stack pointer can be 0 through 31. The stack pointer increments when values are pushed onto the stack and decrements when values are popped off the stack. At RESET, the stack pointer value will be 0. The user may read and write the stack pointer value. This feature can be used by a Real Time Operating System for return stack maintenance.

After the PC is pushed onto the stack 31 times (without popping any values off the stack), the `STKFUL` bit is set. The `STKFUL` bit can only be cleared in software or by a POR.

The action that takes place when the stack becomes full, depends on the state of the `STVREN` (Stack Overflow Reset Enable) configuration bit. Refer to Section 18.0 for a description of the device configuration bits. If `STVREN` is set (default), the 31st push will push the $(PC + 2)$ value onto the stack, set the `STKFUL` bit, and reset the device. The `STKFUL` bit will remain set and the stack pointer will be set to 0.

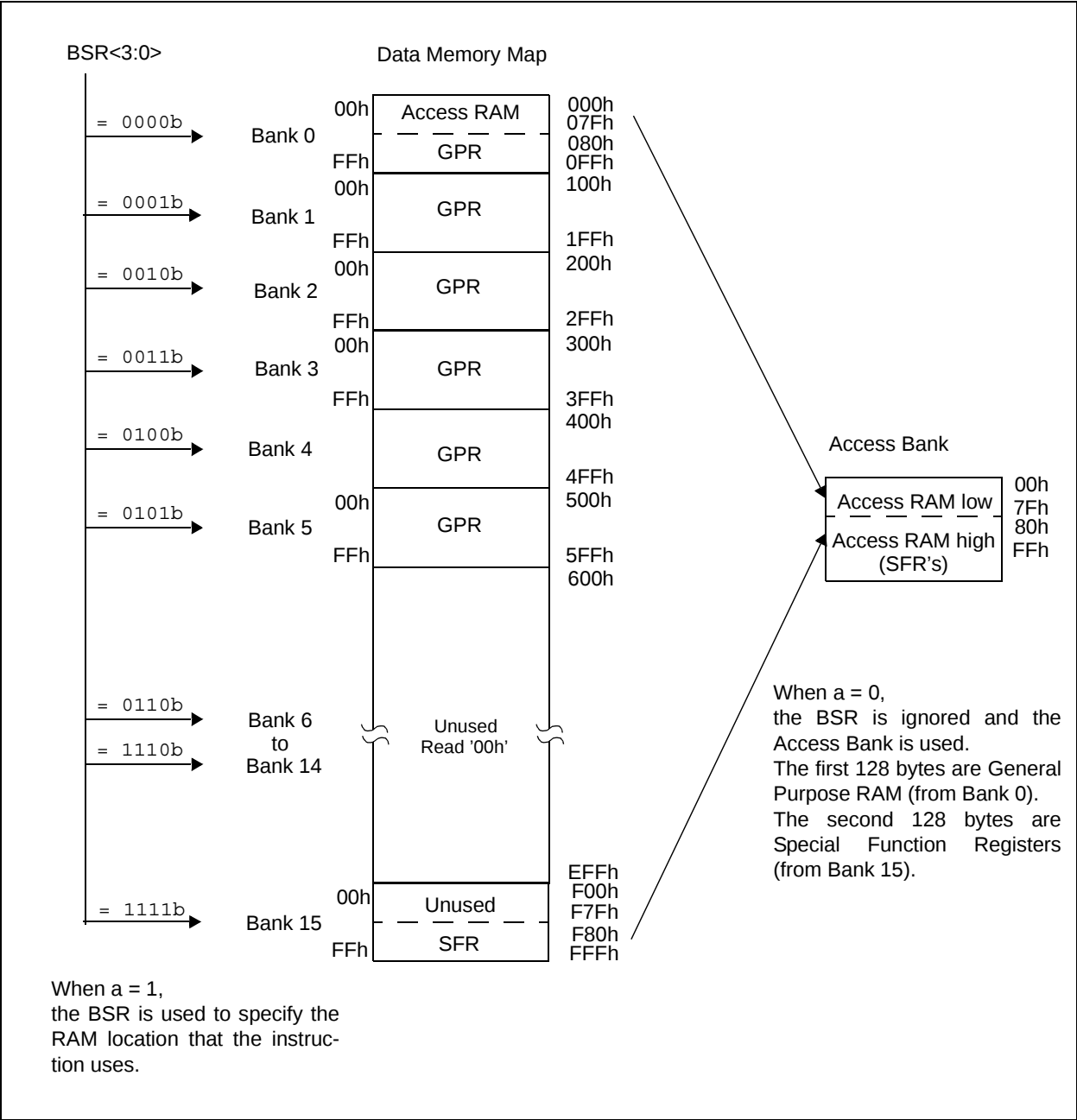
If `STVREN` is cleared, the `STKFUL` bit will be set on the 31st push and the stack pointer will increment to 31. Any additional pushes will not overwrite the 31st push and `STKPTR` will remain at 31.

When the stack has been popped enough times to unload the stack, the next pop will return a value of zero to the PC and sets the `STKUNF` bit, while the stack pointer remains at 0. The `STKUNF` bit will remain set until cleared in software or a POR occurs.

Note: Returning a value of zero to the PC on an underflow, has the effect of vectoring the program to the RESET vector, where the stack conditions can be verified and appropriate actions can be taken.

PIC18CXX2

FIGURE 4-7: DATA MEMORY MAP FOR PIC18C252/452



REGISTER 7-7: PERIPHERAL INTERRUPT ENABLE REGISTER 2 (PIE2)

U-0	U-0	U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	—	—	BCLIE	LVDIE	TMR3IE	CCP2IE
bit 7				bit 0			

- bit 7-4 **Unimplemented:** Read as '0'
- bit 3 **BCLIE:** Bus Collision Interrupt Enable bit
 1 = Enabled
 0 = Disabled
- bit 2 **LVDIE:** Low Voltage Detect Interrupt Enable bit
 1 = Enabled
 0 = Disabled
- bit 1 **TMR3IE:** TMR3 Overflow Interrupt Enable bit
 1 = Enables the TMR3 overflow interrupt
 0 = Disables the TMR3 overflow interrupt
- bit 0 **CCP2IE:** CCP2 Interrupt Enable bit
 1 = Enables the CCP2 interrupt
 0 = Disables the CCP2 interrupt

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown

PIC18CXX2

14.3.4 MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK. The master determines when the slave (Processor 2, Figure 14-2) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPBUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPBUF register as if a normal received byte (interrupts and status bits appropriately set). This could be useful in receiver applications as a "Line Activity Monitor" mode.

The clock polarity is selected by appropriately programming the CKP bit (SSPCON1<4>). This then, would give waveforms for SPI communication as shown in

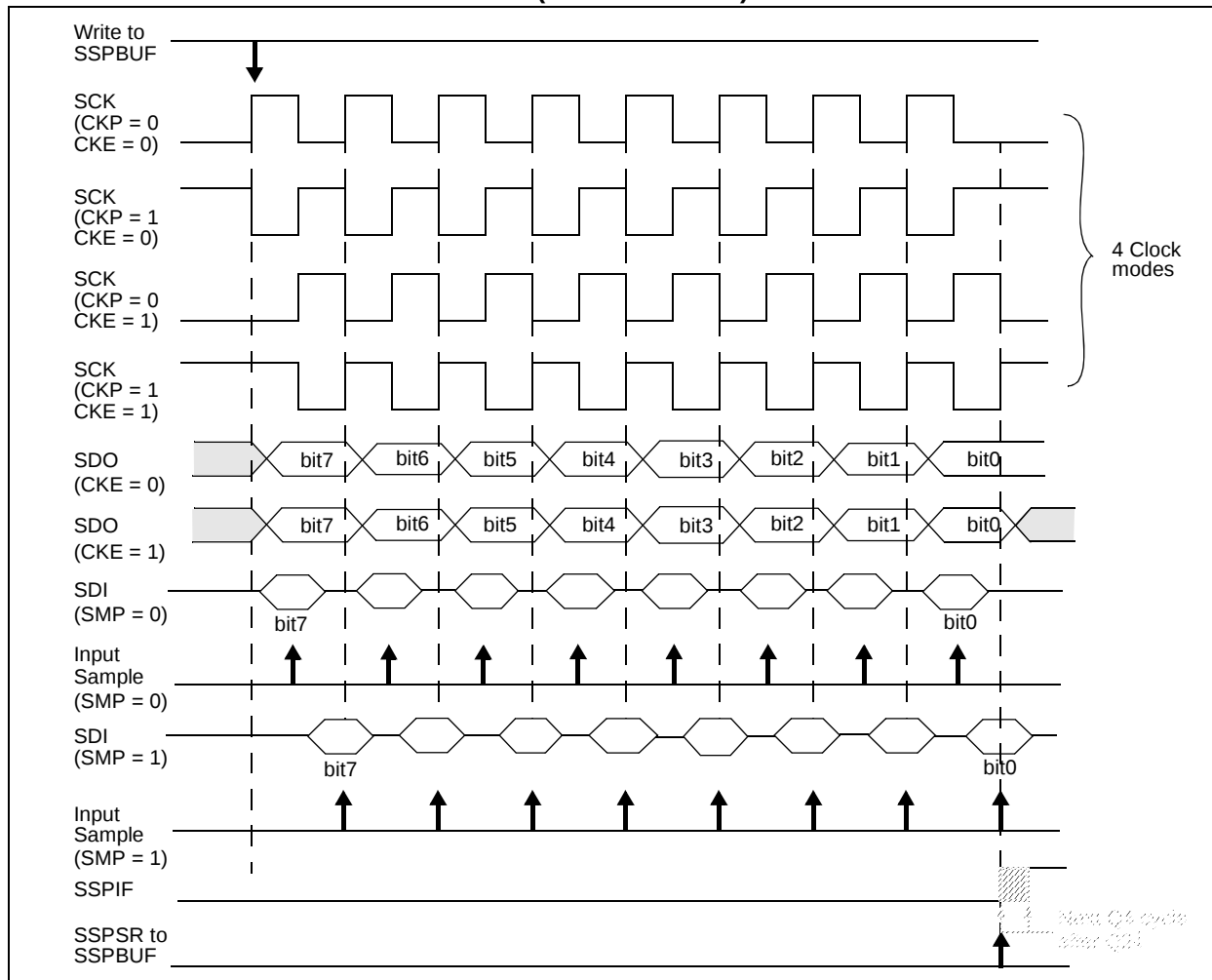
Figure 14-3, Figure 14-5, and Figure 14-6, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- $F_{osc}/4$ (or T_{CY})
- $F_{osc}/16$ (or $4 \cdot T_{CY}$)
- $F_{osc}/64$ (or $16 \cdot T_{CY}$)
- $\text{Timer2 output}/2$

This allows a maximum data rate (at 40 MHz) of 10.00 Mbps.

Figure 14-3 shows the waveforms for Master mode. When the CKE bit is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPBUF is loaded with the received data is shown.

FIGURE 14-3: SPI MODE WAVEFORM (MASTER MODE)



PIC18CXX2

FIGURE 15-2: ASYNCHRONOUS TRANSMISSION

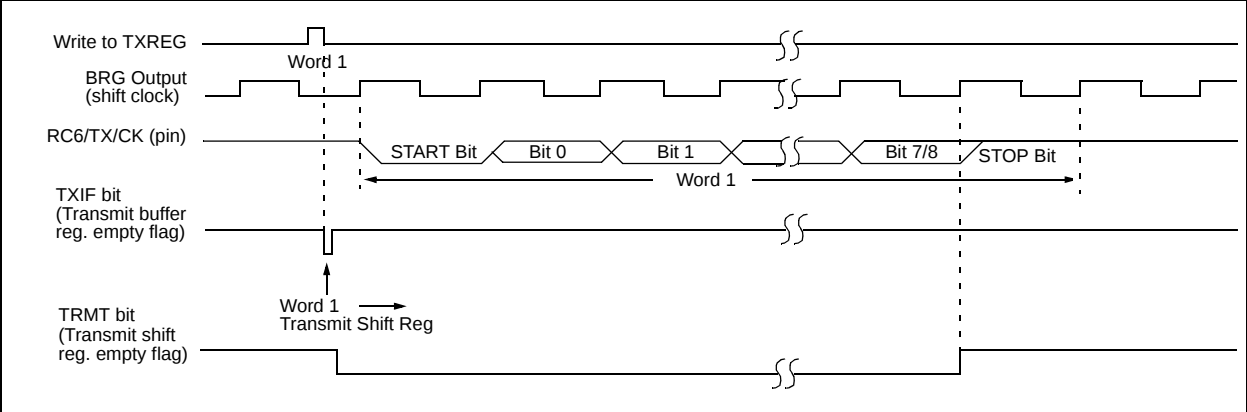


FIGURE 15-3: ASYNCHRONOUS TRANSMISSION (BACK TO BACK)

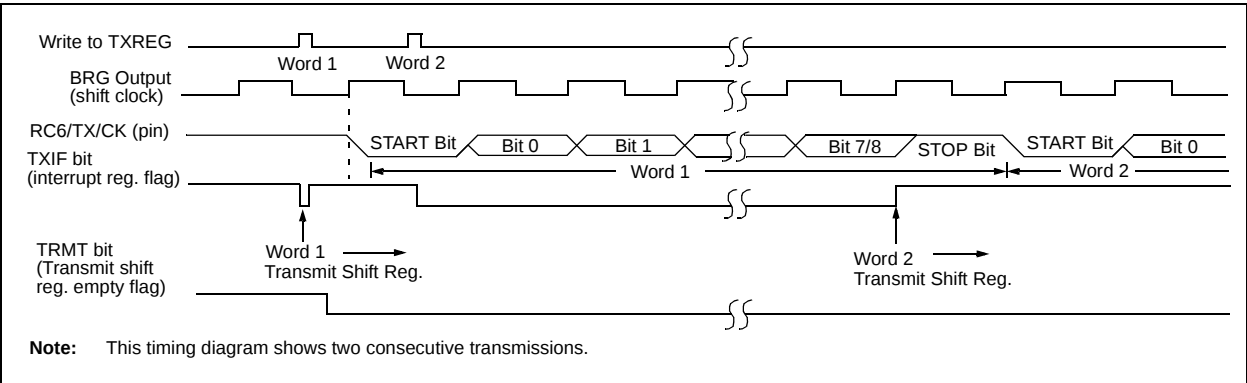


TABLE 15-6: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other RESETS
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	0000 000x	0000 000u
PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
IPR1	PSPIP ⁽¹⁾	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 -00x	0000 -00x
TXREG	USART Transmit Register								0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010
SPBRG	Baud Rate Generator Register								0000 0000	0000 0000

Legend: x = unknown, - = unimplemented locations read as '0'.

Shaded cells are not used for Asynchronous Transmission.

Note 1: The PSPIF, PSPIE and PSPIP bits are reserved on the PIC18C2X2 devices. Always maintain these bits clear.

PIC18CXX2

FIGURE 15-5: ASYNCHRONOUS RECEPTION

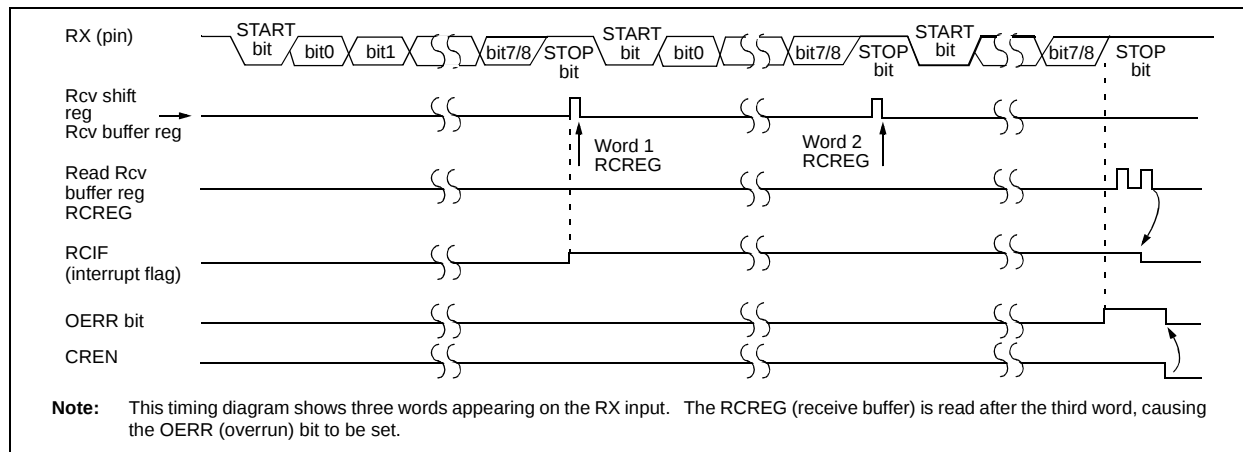


TABLE 15-7: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other RESETS
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	0000 000x	0000 000u
PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
IPR1	PSPIP ⁽¹⁾	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 -00x	0000 -00x
RCREG	USART Receive Register								0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010
SPBRG	Baud Rate Generator Register								0000 0000	0000 0000

Legend: x = unknown, - = unimplemented locations read as '0'.

Shaded cells are not used for Asynchronous Reception.

Note 1: The PSPIF, PSPIE and PSPIP bits are reserved on the PIC18C2X2 devices. Always maintain these bits clear.

16.0 COMPATIBLE 10-BIT ANALOG-TO-DIGITAL CONVERTER (A/D) MODULE

The analog-to-digital (A/D) converter module has five inputs for the PIC18C2x2 devices and eight for the PIC18C4x2 devices. This module has the ADCON0 and ADCON1 register definitions that are compatible with the mid-range A/D module.

The A/D allows conversion of an analog input signal to a corresponding 10-bit digital number.

The A/D module has four registers. These registers are:

- A/D Result High Register (ADRESH)
- A/D Result Low Register (ADRESL)
- A/D Control Register 0 (ADCON0)
- A/D Control Register 1 (ADCON1)

The ADCON0 register, shown in Register 16-1, controls the operation of the A/D module. The ADCON1 register, shown in Register 16-2, configures the functions of the port pins.

REGISTER 16-1: ADCON0 REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0
ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/DONE	—	ADON
bit 7							bit 0

bit 7-6 **ADCS1:ADCS0:** A/D Conversion Clock Select bits (ADCON0 bits in **bold**)

ADCON1 <ADCS2>	ADCON0 <ADCS1:ADCS0>	Clock Conversion
0	00	Fosc/2
0	01	Fosc/8
0	10	Fosc/32
0	11	FRC (clock derived from the internal A/D RC oscillator)
1	00	Fosc/4
1	01	Fosc/16
1	10	Fosc/64
1	11	FRC (clock derived from the internal A/D RC oscillator)

bit 5-3 **CHS2:CHS0:** Analog Channel Select bits

000 = channel 0 (AN0)
 001 = channel 1 (AN1)
 010 = channel 2 (AN2)
 011 = channel 3 (AN3)
 100 = channel 4 (AN4)
 101 = channel 5 (AN5)
 110 = channel 6 (AN6)
 111 = channel 7 (AN7)

Note: The PIC18C2X2 devices do not implement the full 8 A/D channels; the unimplemented selections are reserved. Do not select any unimplemented channel.

bit 2 **GO/DONE:** A/D Conversion Status bit

When ADON = 1:

1 = A/D conversion in progress (setting this bit starts the A/D conversion which is automatically cleared by hardware when the A/D conversion is complete)
 0 = A/D conversion not in progress

bit 1 **Unimplemented:** Read as '0'

bit 0 **ADON:** A/D On bit

1 = A/D converter module is powered up
 0 = A/D converter module is shut-off and consumes no operating current

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
- n = Value at POR reset	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

PIC18CXX2

The value that is in the ADRESH/ADRESL registers is not modified for a Power-on Reset. The ADRESH/ADRESL registers will contain unknown data after a Power-on Reset.

After the A/D module has been configured as desired, the selected channel must be acquired before the conversion is started. The analog input channels must have their corresponding TRIS bits selected as an input. To determine acquisition time, see Section 16.1. After this acquisition time has elapsed, the A/D conversion can be started. The following steps should be followed for doing an A/D conversion:

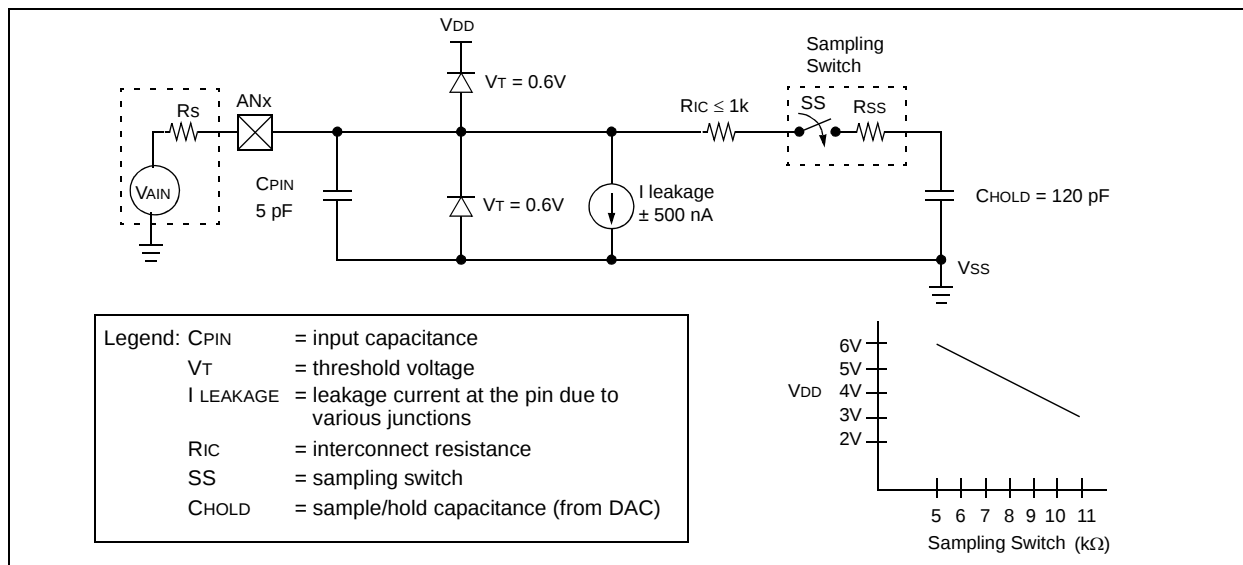
1. Configure the A/D module:
 - Configure analog pins, voltage reference and digital I/O (ADCON1)
 - Select A/D input channel (ADCON0)
 - Select A/D conversion clock (ADCON0)
 - Turn on A/D module (ADCON0)
2. Configure A/D interrupt (if desired):
 - Clear ADIF bit
 - Set ADIE bit
 - Set GIE bit
3. Wait the required acquisition time.
4. Start conversion:
 - Set $\overline{\text{GO/DONE}}$ bit (ADCON0)
5. Wait for A/D conversion to complete, by either:
 - Polling for the $\overline{\text{GO/DONE}}$ bit to be cleared
 OR
 - Waiting for the A/D interrupt
6. Read A/D Result registers (ADRESH/ADRESL); clear bit ADIF if required.
7. For next conversion, go to step 1 or step 2, as required. The A/D conversion time per bit is defined as T_{AD} . A minimum wait of $2T_{AD}$ is required before next acquisition starts.

16.1 A/D Acquisition Requirements

For the A/D converter to meet its specified accuracy, the charge holding capacitor (C_{HOLD}) must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 16-2. The source impedance (R_s) and the internal sampling switch (R_{SS}) impedance directly affect the time required to charge the capacitor C_{HOLD} . The sampling switch (R_{SS}) impedance varies over the device voltage (V_{DD}). The source impedance affects the offset voltage at the analog input (due to pin leakage current). **The maximum recommended impedance for analog sources is 2.5 k Ω .** After the analog input channel is selected (changed), this acquisition must be done before the conversion can be started.

Note: When the conversion is started, the holding capacitor is disconnected from the input pin.

FIGURE 16-2: ANALOG INPUT MODEL



To calculate the minimum acquisition time, Equation 16-1 may be used. This equation assumes that 1/2 LSB error is used (1024 steps for the A/D). The 1/2 LSB error is the maximum error allowed for the A/D to meet its specified resolution.

EQUATION 16-1: ACQUISITION TIME

$$\begin{aligned} T_{ACQ} &= \text{Amplifier Settling Time} + \text{Holding Capacitor Charging Time} + \text{Temperature Coefficient} \\ &= T_{AMP} + T_C + T_{COFF} \end{aligned}$$

EQUATION 16-2: A/D MINIMUM CHARGING TIME

$$\begin{aligned} V_{HOLD} &= (V_{REF} - (V_{REF}/2048)) \cdot (1 - e^{(-T_C/CHOLD(RIC + R_{SS} + R_S))}) \\ \text{or} \\ T_C &= -(120 \text{ pF})(1 \text{ k}\Omega + R_{SS} + R_S) \ln(1/2047) \end{aligned}$$

Example 16-1 shows the calculation of the minimum required acquisition time T_{ACQ} . This calculation is based on the following application system assumptions:

- $CHOLD = 120 \text{ pF}$
- $R_S = 2.5 \text{ k}\Omega$
- Conversion Error $\leq 1/2 \text{ LSB}$
- $V_{DD} = 5V \rightarrow R_{SS} = 7 \text{ k}\Omega$
- Temperature = 50°C (system max.)
- $V_{HOLD} = 0V$ @ time = 0

EXAMPLE 16-1: CALCULATING THE MINIMUM REQUIRED ACQUISITION TIME

$$\begin{aligned} T_{ACQ} &= T_{AMP} + T_C + T_{COFF} \\ \text{Temperature coefficient is only required for temperatures } > 25^\circ\text{C}. \\ T_{ACQ} &= 2 \mu\text{s} + T_C + [(Temp - 25^\circ\text{C})(0.05 \mu\text{s}/^\circ\text{C})] \\ T_C &= -CHOLD (RIC + R_{SS} + R_S) \ln(1/2047) \\ &= -120 \text{ pF} (1 \text{ k}\Omega + 7 \text{ k}\Omega + 2.5 \text{ k}\Omega) \ln(0.0004885) \\ &= -120 \text{ pF} (10.5 \text{ k}\Omega) \ln(0.0004885) \\ &= -1.26 \mu\text{s} (-7.6241) \\ &= 9.61 \mu\text{s} \\ T_{ACQ} &= 2 \mu\text{s} + 9.61 \mu\text{s} + [(50^\circ\text{C} - 25^\circ\text{C})(0.05 \mu\text{s}/^\circ\text{C})] \\ &= 11.61 \mu\text{s} + 1.25 \mu\text{s} \\ &= 12.86 \mu\text{s} \end{aligned}$$

PIC18CXX2

TABLE 16-3: SUMMARY OF A/D REGISTERS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other RESETS
INTCON	GIE/ GIEH	PEIE/ GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	0000 000x	0000 000u
PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
IPR1	PSPIP ⁽¹⁾	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	0000 0000	0000 0000
PIR2	—	—	—	—	BCLIF	LVDIF	TMR3IF	CCP2IF	---- 0000	---- 0000
PIE2	—	—	—	—	BCLIE	LVDIE	TMR3IE	CCP2IE	---- 0000	---- 0000
IPR2	—	—	—	—	BCLIP	LVDIP	TMR3IP	CCP2IP	---- 0000	---- 0000
ADRESH	A/D Result Register								xxxx xxxx	uuuu uuuu
ADRESL	A/D Result Register								xxxx xxxx	uuuu uuuu
ADCON0	ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/ DONE	—	ADON	0000 00-0	0000 00-0
ADCON1	ADFM	ADCS2	—	—	PCFG3	PCFG2	PCFG1	PCFG0	---- -000	---- -000
PORTA	—	RA6	RA5	RA4	RA3	RA2	RA1	RA0	--0x 0000	--0u 0000
TRISA	—	PORTA Data Direction Register							--11 1111	--11 1111
PORTE	—	—	—	—	—	RE2	RE1	RE0	---- -000	---- -000
LATE	—	—	—	—	—	LATE2	LATE1	LATE0	---- -xxx	---- -uuu
TRISE	IBF	OBF	IBOV	PSPMODE	—	PORTE Data Direction bits			0000 -111	0000 -111

Legend: x = unknown, u = unchanged, — = unimplemented, read as '0'. Shaded cells are not used for A/D conversion.

Note 1: The PSPIF, PSPIE and PSPIP bits are reserved on the PIC18CXX2 devices. Always maintain these bits clear.

17.0 LOW VOLTAGE DETECT

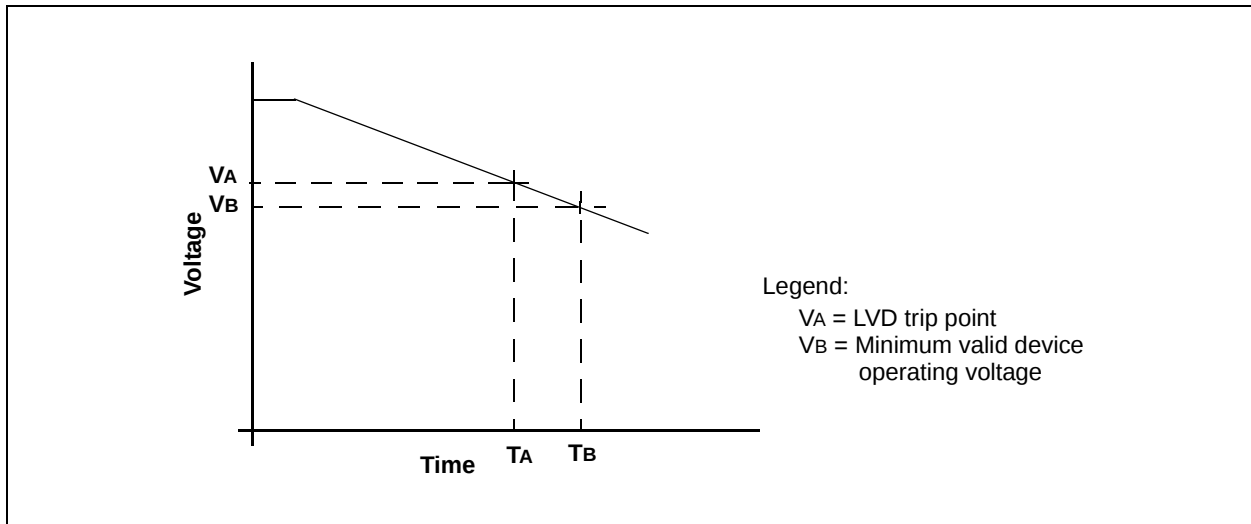
In many applications, the ability to determine if the device voltage (V_{DD}) is below a specified voltage level is a desirable feature. A window of operation for the application can be created, where the application software can do “housekeeping tasks” before the device voltage exits the valid operating range. This can be done using the Low Voltage Detect module.

This module is a software programmable circuitry, where a device voltage trip point can be specified. When the voltage of the device becomes lower than the specified point, an interrupt flag is set. If the interrupt is enabled, the program execution will branch to the interrupt vector address and the software can then respond to that interrupt source.

The Low Voltage Detect circuitry is completely under software control. This allows the circuitry to be “turned off” by the software, which minimizes the current consumption for the device.

Figure 17-1 shows a possible application voltage curve (typically for batteries). Over time, the device voltage decreases. When the device voltage equals voltage V_A , the LVD logic generates an interrupt. This occurs at time T_A . The application software then has the time, until the device voltage is no longer in valid operating range, to shut-down the system. Voltage point V_B is the minimum valid operating voltage specification. This occurs at time T_B . The difference $T_B - T_A$ is the total time for shut-down.

FIGURE 17-1: TYPICAL LOW VOLTAGE DETECT APPLICATION



The block diagram for the LVD module is shown in Figure 17-2. A comparator uses an internally generated reference voltage as the set point. When the selected tap output of the device voltage crosses the set point (is lower than), the LVDIF bit is set.

Each node in the resistor divider represents a “trip point” voltage. The “trip point” voltage is the minimum supply voltage level at which the device can operate before the LVD module asserts an interrupt. When the

supply voltage is equal to the trip point, the voltage tapped off of the resistor array is equal to the 1.2V internal reference voltage generated by the voltage reference module. The comparator then generates an interrupt signal setting the LVDIF bit. This voltage is software programmable to any one of 16 values (see Figure 17-2). The trip point is selected by programming the LVDL3:LVDL0 bits ($LVDCON<3:0>$).

PIC18CXX2

REGISTER 18-5: CONFIGURATION REGISTER 3 HIGH (CONFIG3H: BYTE ADDRESS 300005h)

U-0	U-0	U-0	U-0	U-0	U-0	U-0	R/P-1
—	—	—	—	—	—	—	CCP2MX
bit 7							bit 0

bit 7-1 **Unimplemented:** Read as '0'

bit 0 **CCP2MX:** CCP2 Mux bit

1 = CCP2 input/output is multiplexed with RC1
0 = CCP2 input/output is multiplexed with RB3

Legend:

R = Readable bit P = Programmable bit U = Unimplemented bit, read as '0'
- n = Value when device is unprogrammed u = Unchanged from programmed state

REGISTER 18-6: CONFIGURATION REGISTER 4 LOW (CONFIG4L: BYTE ADDRESS 300006h)

U-0	U-0	U-0	U-0	U-0	U-0	R/P-1	R/P-1
—	—	—	—	—	—	Reserved	STVREN
bit 7							bit 0

bit 7-2 **Unimplemented:** Read as '0'

bit 1 **Reserved:** Maintain this bit set

bit 0 **STVREN:** Stack Full/Underflow Reset Enable bit

1 = Stack Full/Underflow will cause RESET
0 = Stack Full/Underflow will not cause RESET

Legend:

R = Readable bit P = Programmable bit U = Unimplemented bit, read as '0'
- n = Value when device is unprogrammed u = Unchanged from programmed state

PIC18CXX2

GOTO Unconditional Branch

Syntax: `[label] GOTO k`

Operands: $0 \leq k \leq 1048575$

Operation: $k \rightarrow PC<20:1>$

Status Affected: None

Encoding:

1st word ($k<7:0>$)

1110

1111

k_7kkk

$kkkk_0$

2nd word ($k<19:8>$)

1111

$k_{19}kkk$

$kkkk$

$kkkk_8$

Description: GOTO allows an unconditional branch anywhere within entire 2 Mbyte memory range. The 20-bit value 'k' is loaded into PC<20:1>. GOTO is always a two-cycle instruction.

Words: 2

Cycles: 2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'<7:0>.	No operation	Read literal 'k'<19:8>, Write to PC
No operation	No operation	No operation	No operation

Example: GOTO THERE

After Instruction

PC = Address (THERE)

INCF Increment f

Syntax: `[label] INCF f[,d[,a]]`

Operands: $0 \leq f \leq 255$

$d \in [0,1]$

$a \in [0,1]$

Operation: $(f) + 1 \rightarrow \text{dest}$

Status Affected: C,DC,N,OV,Z

Encoding:

0010

10da

ffff

ffff

Description: The contents of register 'f' are incremented. If 'd' is 0, the result is placed in WREG. If 'd' is 1, the result is placed back in register 'f' (default). If 'a' is 0, the Access Bank will be selected, overriding the BSR value. If 'a' = 1, then the bank will be selected as per the BSR value (default).

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: INCF CNT, 1, 0

Before Instruction

CNT = 0xFF
Z = 0
C = ?
DC = ?

After Instruction

CNT = 0x00
Z = 1
C = 1
DC = 1

FIGURE 22-13: I_{PD} vs. V_{DD} (SLEEP MODE, ALL PERIPHERALS DISABLED)

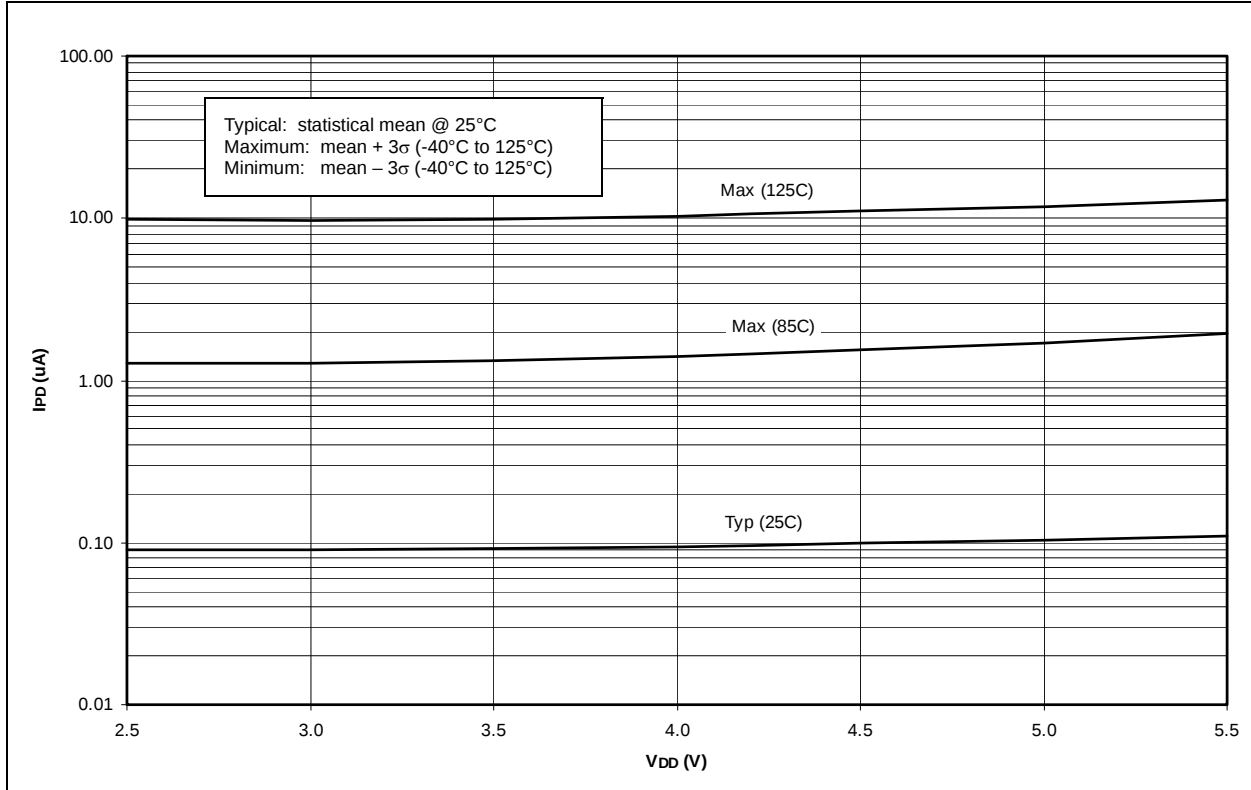
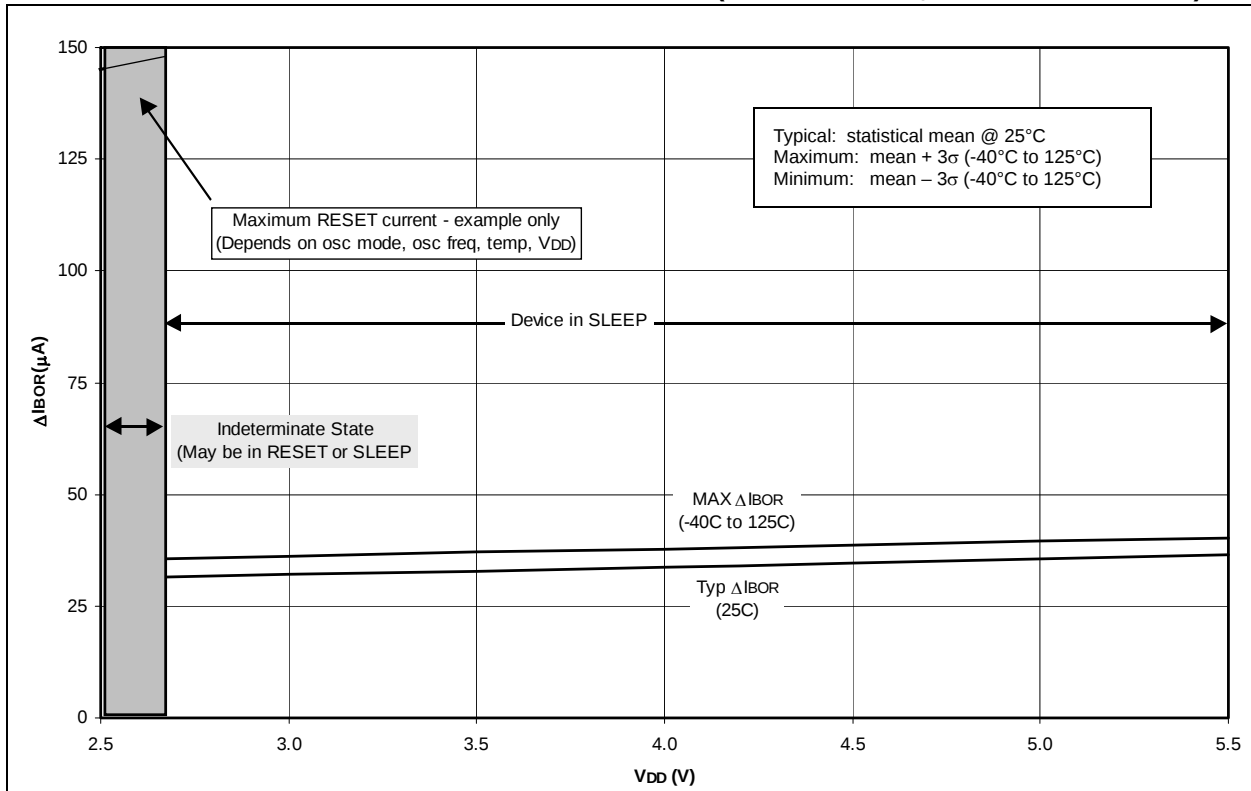


FIGURE 22-14: ΔI_{BOR} vs. V_{DD} OVER TEMPERATURE (BOR ENABLED, $V_{BOR} = 2.50V - 2.66V$)



PIC18CXX2

FIGURE 22-23: TYPICAL AND MAXIMUM Vol vs. IoL (VDD = 3V, -40°C TO +125°C)

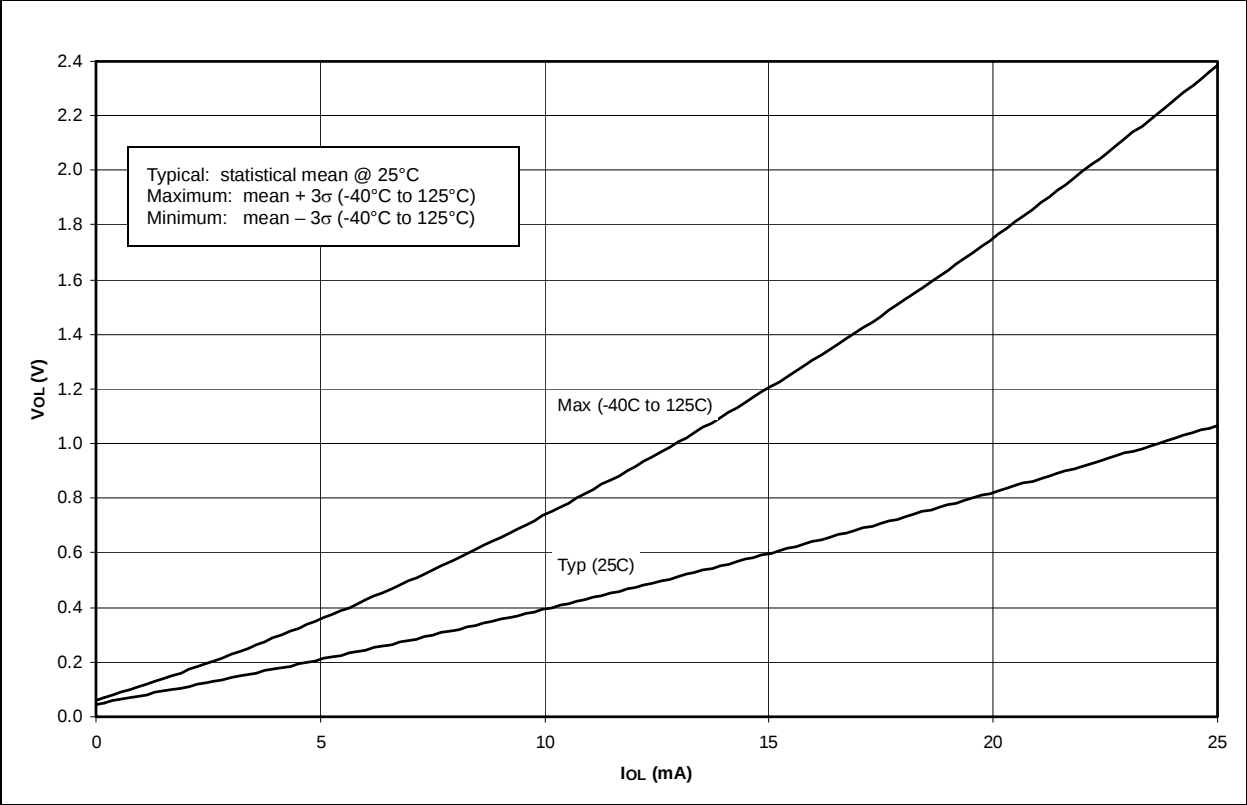
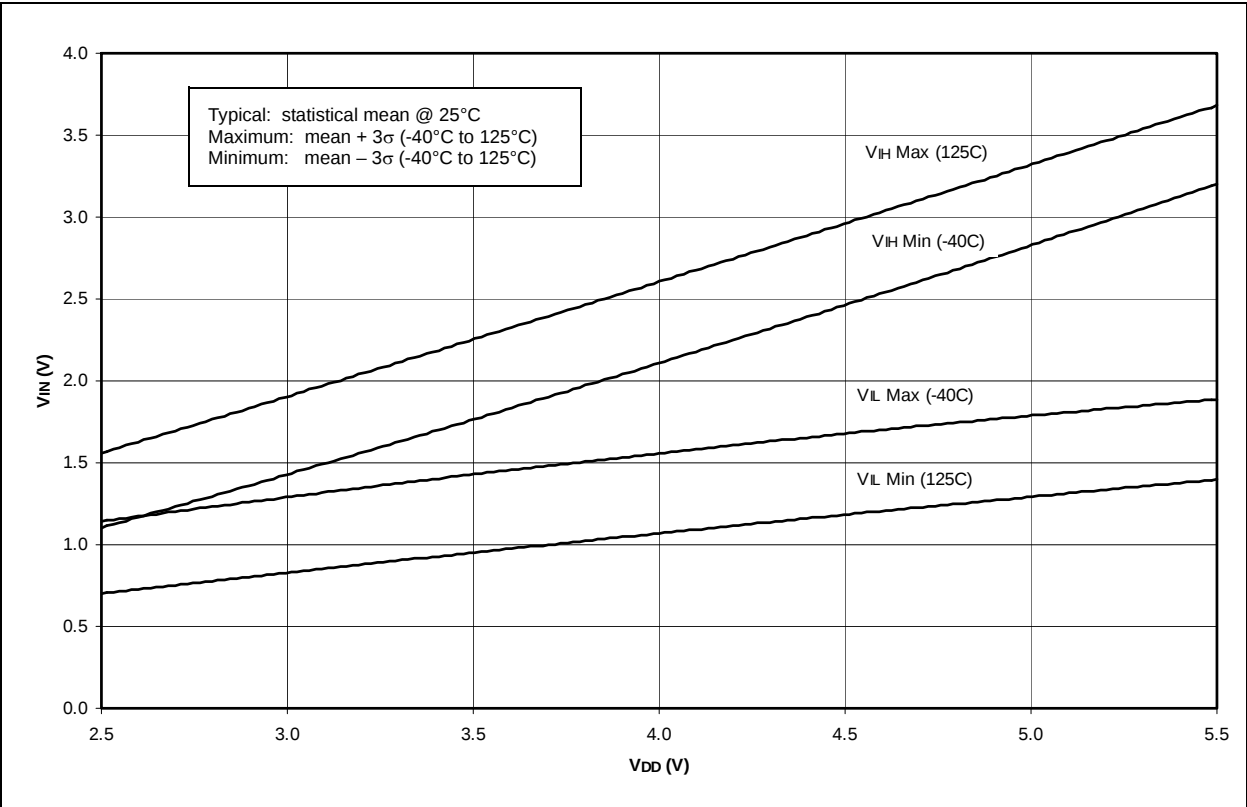
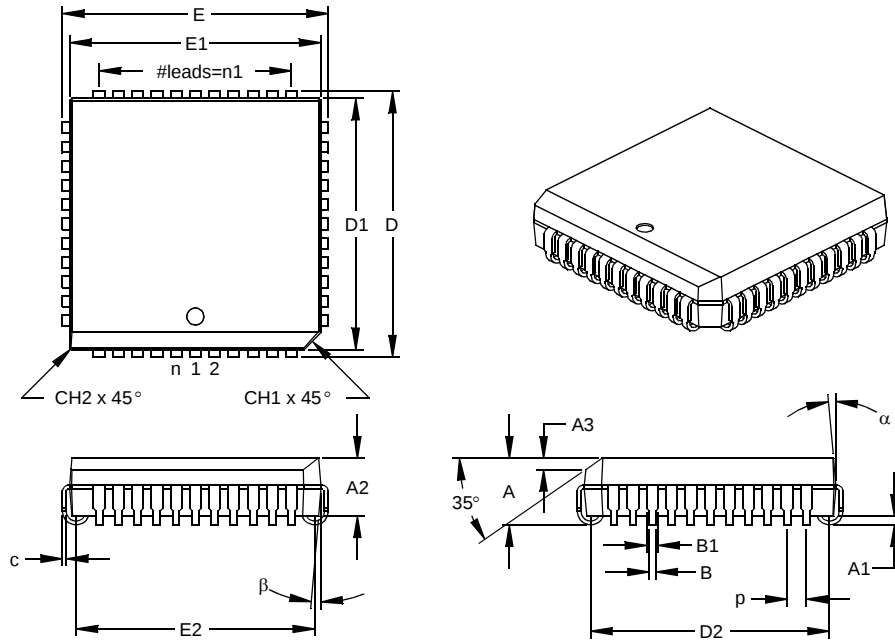


FIGURE 22-24: MINIMUM AND MAXIMUM VIN vs. VDD (ST INPUT, -40°C TO +125°C)



44-Lead Plastic Leaded Chip Carrier (L) – Square (PLCC)

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Units		INCHES*			MILLIMETERS		
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		44			44	
Pitch	p		.050			1.27	
Pins per Side	n1		11			11	
Overall Height	A	.165	.173	.180	4.19	4.39	4.57
Molded Package Thickness	A2	.145	.153	.160	3.68	3.87	4.06
Standoff §	A1	.020	.028	.035	0.51	0.71	0.89
Side 1 Chamfer Height	A3	.024	.029	.034	0.61	0.74	0.86
Corner Chamfer 1	CH1	.040	.045	.050	1.02	1.14	1.27
Corner Chamfer (others)	CH2	.000	.005	.010	0.00	0.13	0.25
Overall Width	E	.685	.690	.695	17.40	17.53	17.65
Overall Length	D	.685	.690	.695	17.40	17.53	17.65
Molded Package Width	E1	.650	.653	.656	16.51	16.59	16.66
Molded Package Length	D1	.650	.653	.656	16.51	16.59	16.66
Footprint Width	E2	.590	.620	.630	14.99	15.75	16.00
Footprint Length	D2	.590	.620	.630	14.99	15.75	16.00
Lead Thickness	c	.008	.011	.013	0.20	0.27	0.33
Upper Lead Width	B1	.026	.029	.032	0.66	0.74	0.81
Lower Lead Width	B	.013	.020	.021	0.33	0.51	0.53
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10

* Controlling Parameter

§ Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side.

JEDEC Equivalent: MO-047

Drawing No. C04-048

PIC18CXX2

S

SCI. See USART.	
SCK	121
SDI	121
SDO	121
Serial Clock, SCK	121
Serial Communication Interface. See USART	
Serial Data In, SDI	121
Serial Data Out, SDO	121
Serial Peripheral Interface. See SPI	
SETF	221
Slave Select Synchronization	125
Slave Select, $\overline{SS}$	121
SLEEP	179, 185, 222
Software Simulator (MPLAB SIM)	230
Special Event Trigger. See Compare	
Special Features of the CPU	179
Configuration Registers	180–182
Special Function Registers	42
Map	45
SPI	
Master Mode	124
Serial Clock	121
Serial Data In	121
Serial Data Out	121
Slave Select	121
SPI Clock	124
SPI Mode	121
SPI Master/Slave Connection	123
SPI Module	
Master/Slave Connection	123
Slave Mode	125
Slave Select Synchronization	125
Slave Synch Timing	125
Slave Timing with CKE = 0	126
Slave Timing with CKE = 1	126
$\overline{SS}$	121
SSP	115
Block Diagram (SPI Mode)	121
I ² C Mode. See I ² C.	
SPI Mode	121
Associated Registers	127
Block Diagram	121
SPI Mode. See SPI.	
SSPBUF	124
SSPCON1	118
SSPCON2 Register	120
SSPSR	124
SSPSTAT	116
TMR2 Output for Clock Shift	101, 102
SSP Module	
SPI Master Mode	124
SPI Master/Slave Connection	123
SPI Slave Mode	125
SSPCON1 Register	118
SSPOV	139
SSPSTAT Register	116
R/W Bit	129
STATUS Register	52
STKPTR Register	38
SUBFWB	222
SUBLW	223
SUBWF	223
SUBWFB	224
SWAPF	224
Synchronous Serial Port. See SSP.	

T

TABLAT Register	57
Table Pointer Operations (Table)	57
Table Read Operation, Diagram	55
Table Write Operation, Diagram	55
TBLPTR Register	57
TBLRD	225
TBLWT	226
Timer0	93
Clock Source Edge Select (T0SE Bit)	95
Clock Source Select (T0CS Bit)	95
Overflow Interrupt	95
Prescaler. See Prescaler, Timer0	
T0CON Register	93
Timing Diagram	249
Timer1	97
Block Diagram	98
Block Diagram (16-bit R/W Mode)	98
Oscillator	97, 99, 103, 105
Overflow Interrupt	97, 99, 103, 105
Prescaler.	98
Special Event Trigger (CCP)	99, 105, 110
T1CON Register	97
Timing Diagram	249
TMR1H Register	97, 103
TMR1L Register	97, 103
Timer2	101
Associated Registers	102
Block Diagram	102
Postscaler. See Postscaler, Timer2.	
PR2 Register	101, 112
Prescaler. See Prescaler, Timer2.	
SSP Clock Shift	101, 102
T2CON Register	101
TMR2 Register	101
TMR2 to PR2 Match Interrupt	101, 102, 112
Timer3	103
Associated Registers	105
Block Diagram	104
Block Diagram (16-bit R/W Mode)	104
T3CON Register	103
Timing Diagrams	
Acknowledge Sequence Timing	142
Baud Rate Generator with Clock Arbitration	136
BRG Reset Due to SDA Collision	146
Bus Collision	
START Condition Timing	145
Bus Collision During a Repeated START	
Condition (Case 1)	147
Bus Collision During a Repeated START	
Condition (Case2)	147
Bus Collision During a START Condition	
(SCL = 0)	146
Bus Collision During a STOP Condition	148
Bus Collision for Transmit and Acknowledge	144
I ² C Bus Data	259
I ² C Master Mode First START Bit Timing	137
I ² C Master Mode Reception Timing	141
I ² C Master Mode Transmission Timing	140
Master Mode Transmit Clock Arbitration	143
Repeat START Condition	138
Slave Synchronization	125
SPI Mode Timing (Master Mode) SPI Mode	
Master Mode Timing Diagram	124
SPI Mode Timing (Slave Mode with CKE = 0)	126
SPI Mode Timing (Slave Mode with CKE = 1)	126