STMicroelectronics - STM32F429VGT6TR Datasheet



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Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

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Product Status	Active
Core Processor	ARM® Cortex®-M4
Core Size	32-Bit Single-Core
Speed	180MHz
Connectivity	CANbus, EBI/EMI, Ethernet, I ² C, IrDA, LINbus, SPI, UART/USART, USB OTG
Peripherals	Brown-out Detect/Reset, DMA, I ² S, LCD, POR, PWM, WDT
Number of I/O	82
Program Memory Size	1MB (1M × 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	256K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 16x12b; D/A 2x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP
Supplier Device Package	100-LQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/stm32f429vgt6tr

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Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

communicate at speeds of up to 11.25 Mbit/s. The other available interfaces communicate at up to 5.62 bit/s.

USART1, USART2, USART3 and USART6 also provide hardware management of the CTS and RTS signals, Smart Card mode (ISO 7816 compliant) and SPI-like communication capability. All interfaces can be served by the DMA controller.

USART name	Standard features	Modem (RTS/CTS)	LIN	SPI master	irDA	Smartcard (ISO 7816)	Max. baud rate in Mbit/s (oversampling by 16)	Max. baud rate in Mbit/s (oversampling by 8)	APB mapping
USART1	х	х	x	х	х	х	5.62	11.25	APB2 (max. 90 MHz)
USART2	х	Х	x	х	х	Х	2.81	5.62	APB1 (max. 45 MHz)
USART3	х	Х	x	х	х	Х	2.81	5.62	APB1 (max. 45 MHz)
UART4	х	-	x	-	х	-	2.81	5.62	APB1 (max. 45 MHz)
UART5	х	-	х	-	х	-	2.81	5.62	APB1 (max. 45 MHz)
USART6	х	х	х	х	х	х	5.62	11.25	APB2 (max. 90 MHz)
UART7	х	-	х	-	х	-	2.81	5.62	APB1 (max. 45 MHz)
UART8	х	-	x	_	х	-	2.81	5.62	APB1 (max. 45 MHz)

1. X = feature supported.

3.25 Serial peripheral interface (SPI)

The devices feature up to six SPIs in slave and master modes in full-duplex and simplex communication modes. SPI1, SPI4, SPI5, and SPI6 can communicate at up to 45 Mbits/s, SPI2 and SPI3 can communicate at up to 22.5 Mbit/s. The 3-bit prescaler gives 8 master mode frequencies and the frame is configurable to 8 bits or 16 bits. The hardware CRC generation/verification supports basic SD Card/MMC modes. All SPIs can be served by the DMA controller.

The SPI interface can be configured to operate in TI mode for communications in master mode and slave mode.



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3.35 Digital camera interface (DCMI)

The devices embed a camera interface that can connect with camera modules and CMOS sensors through an 8-bit to 14-bit parallel interface, to receive video data. The camera interface can sustain a data transfer rate up to 54 Mbyte/s at 54 MHz. It features:

- Programmable polarity for the input pixel clock and synchronization signals
- Parallel data communication can be 8-, 10-, 12- or 14-bit
- Supports 8-bit progressive video monochrome or raw bayer format, YCbCr 4:2:2 progressive video, RGB 565 progressive video or compressed data (like JPEG)
- Supports continuous mode or snapshot (a single frame) mode
- Capability to automatically crop the image

3.36 Random number generator (RNG)

All devices embed an RNG that delivers 32-bit random numbers generated by an integrated analog circuit.

3.37 General-purpose input/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain, with or without pull-up or pull-down), as input (floating, with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. All GPIOs are high-current-capable and have speed selection to better manage internal noise, power consumption and electromagnetic emission.

The I/O configuration can be locked if needed by following a specific sequence in order to avoid spurious writing to the I/Os registers.

Fast I/O handling allowing maximum I/O toggling up to 90 MHz.

3.38 Analog-to-digital converters (ADCs)

Three 12-bit analog-to-digital converters are embedded and each ADC shares up to 16 external channels, performing conversions in the single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

Additional logic functions embedded in the ADC interface allow:

- Simultaneous sample and hold
- Interleaved sample and hold

The ADC can be served by the DMA controller. An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

To synchronize A/D conversion and timers, the ADCs could be triggered by any of TIM1, TIM2, TIM3, TIM4, TIM5, or TIM8 timer.



3.39 Temperature sensor

The temperature sensor has to generate a voltage that varies linearly with temperature. The conversion range is between 1.7 V and 3.6 V. The temperature sensor is internally connected to the same input channel as V_{BAT}, ADC1_IN18, which is used to convert the sensor output voltage into a digital value. When the temperature sensor and V_{BAT} conversion are enabled at the same time, only V_{BAT} conversion is performed.

As the offset of the temperature sensor varies from chip to chip due to process variation, the internal temperature sensor is mainly suitable for applications that detect temperature changes instead of absolute temperatures. If an accurate temperature reading is needed, then an external temperature sensor part should be used.

3.40 Digital-to-analog converter (DAC)

The two 12-bit buffered DAC channels can be used to convert two digital signals into two analog voltage signal outputs.

This dual digital Interface supports the following features:

- two DAC converters: one for each output channel
- 8-bit or 10-bit monotonic output
- left or right data alignment in 12-bit mode
- synchronized update capability
- noise-wave generation
- triangular-wave generation
- dual DAC channel independent or simultaneous conversions
- DMA capability for each channel
- external triggers for conversion
- input voltage reference V_{REF+}

Eight DAC trigger inputs are used in the device. The DAC channels are triggered through the timer update outputs that are also connected to different DMA streams.

3.41 Serial wire JTAG debug port (SWJ-DP)

The ARM SWJ-DP interface is embedded, and is a combined JTAG and serial wire debug port that enables either a serial wire debug or a JTAG probe to be connected to the target.

Debug is performed using 2 pins only instead of 5 required by the JTAG (JTAG pins could be re-use as GPIO with alternate function): the JTAG TMS and TCK pins are shared with SWDIO and SWCLK, respectively, and a specific sequence on the TMS pin is used to switch between JTAG-DP and SW-DP.

3.42 Embedded Trace Macrocell™

The ARM Embedded Trace Macrocell provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the STM32F42x through a small number of ETM pins to an external hardware trace port analyzer (TPA) device. The TPA is connected to a host computer using USB, Ethernet, or any other high-speed channel. Real-time instruction and data flow activity can be recorded and then formatted for display on the host computer that runs the debugger software. TPA hardware is commercially available from common development tool vendors.

The Embedded Trace Macrocell operates with third party debugger software tools.



			Pin nı	ımber	-								
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
82	115	C9	C12	143	C4	165	C12	PD1	I/O	FT	-	CAN1_TX, FMC_D3, EVENTOUT	-
83	116	В9	D12	144	A3	166	D12	PD2	I/O	FT	-	TIM3_ETR, UART5_RX, SDIO_CMD, DCMI_D11, EVENTOUT	-
84	117	A9	D11	145	B4	167	C11	PD3	I/O	FT	-	SPI2_SCK/I2S2_CK, USART2_CTS, FMC_CLK, DCMI_D5, LCD_G7, EVENTOUT	-
85	118	D8	D10	146	B5	168	D11	PD4	I/O	FT	-	USART2_RTS, FMC_NOE, EVENTOUT	-
86	119	C8	C11	147	A4	169	C10	PD5	I/O	FT	-	USART2_TX, FMC_NWE, EVENTOUT	-
-	120	-	D8	148	-	170	F8	V _{SS}	S		-	-	-
-	121	D6	C8	149	C5	171	E9	V _{DD}	S		-	-	-
87	122	B8	B11	150	F4	172	B11	PD6	I/O	FT	-	SPI3_MOSI/I2S3_SD, SAI1_SD_A, USART2_RX, FMC_NWAIT, DCMI_D10, LCD_B2, EVENTOUT	-
88	123	A8	A11	151	A5	173	A11	PD7	I/O	FT	-	USART2_CK, FMC_NE1/FMC_NCE2, EVENTOUT	-
-	-	-	-	-	-	174	B10	PJ12	I/O	FT	-	LCD_B0, EVENTOUT	-
_	-	-	-	-	-	175	B9	PJ13	I/O	FT	-	LCD_B1, EVENTOUT	-
-	-	-	-	-	-	176	C9	PJ14	I/O	FT	-	LCD_B2, EVENTOUT	-
_	-	-	-	-	-	177	D10	PJ15	I/O	FT	-	LCD_B3, EVENTOUT	-
-	124	NC (2)	C10	152	E5	178	D9	PG9	I/O	FT	-	USART6_RX, FMC_NE2/FMC_NCE3, DCMI_VSYNC ⁽⁸⁾ , EVENTOUT	-

 Table 10. STM32F427xx and STM32F429xx pin and ball definitions (continued)

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Bus	Boundary address	Peripheral
	0xE00F FFFF - 0xFFFF FFFF	Reserved
Cortex-M4	0xE000 0000 - 0xE00F FFFF	Cortex-M4 internal peripherals
	0xD000 0000 - 0xDFFF FFFF	FMC bank 6
	0xC000 0000 - 0xCFFF FFFF	FMC bank 5
	0xA000 1000 - 0xBFFF FFFF	Reserved
	0xA000 0000- 0xA000 0FFF	FMC control register
ALIDS	0x9000 0000 - 0x9FFF FFFF	FMC bank 4
	0x8000 0000 - 0x8FFF FFFF	FMC bank 3
	0x7000 0000 - 0x7FFF FFFF	FMC bank 2
	0x6000 0000 - 0x6FFF FFFF	FMC bank 1
	0x5006 0C00- 0x5FFF FFFF	Reserved
	0x5006 0800 - 0X5006 0BFF	RNG
	0x5005 0400 - X5006 07FF	Reserved
AHB2	0x5005 0000 - 0X5005 03FF	DCMI
	0x5004 0000- 0x5004 FFFF	Reserved
	0x5000 0000 - 0X5003 FFFF	USB OTG FS

 Table 13. STM32F427xx and STM32F429xx register boundary addresses



Bus	Boundary address	Peripheral
	0x4008 0000- 0x4FFF FFFF	Reserved
	0x4004 0000 - 0x4007 FFFF	USB OTG HS
	0x4002 BC00- 0x4003 FFFF	Reserved
	0x4002 B000 - 0x4002 BBFF	DMA2D
	0x4002 9400 - 0x4002 AFFF	Reserved
	0x4002 9000 - 0x4002 93FF	
	0x4002 8C00 - 0x4002 8FFF	
	0x4002 8800 - 0x4002 8BFF	ETHERNET MAC
	0x4002 8400 - 0x4002 87FF	
	0x4002 8000 - 0x4002 83FF	
	0x4002 6800 - 0x4002 7FFF	Reserved
	0x4002 6400 - 0x4002 67FF	DMA2
	0x4002 6000 - 0x4002 63FF	DMA1
	0X4002 5000 - 0X4002 5FFF	Reserved
	0x4002 4000 - 0x4002 4FFF	BKPSRAM
AHB1	0x4008 0000- 0x4FFF FFFF Reserved 0x4004 0000 - 0x4007 FFFF USB OTG HS 0x4002 BC00- 0x4003 FFFF Reserved 0x4002 9000 - 0x4002 BBFF DMA2D 0x4002 9000 - 0x4002 93FF Reserved 0x4002 8000 - 0x4002 8FFF Reserved 0x4002 8000 - 0x4002 8FFF ETHERNET MAC 0x4002 8000 - 0x4002 8FFF ETHERNET MAC 0x4002 8000 - 0x4002 8FFF DMA2 0x4002 6800 - 0x4002 8FFF DMA2 0x4002 6400 - 0x4002 8FFF DMA2 0x4002 6000 - 0x4002 67FF DMA2 0x4002 6000 - 0x4002 67FF DMA2 0x4002 6000 - 0x4002 67FF DMA1 0x4002 6000 - 0x4002 67FF Reserved 0x4002 6000 - 0x4002 7FFF Reserved 0x4002 6000 - 0x4002 7FFF Reserved 0x4002 3000 - 0x4002 3FFF Flash interface register 0x4002 3000 - 0x4002 3FFF Reserved 0x4002 2000 - 0x4002 3FFF Reserved 0x4002 2000 -	Flash interface register
	0x4002 3800 - 0x4002 3BFF	RCC
	0X4002 3400 - 0X4002 37FF	Reserved
	0x4002 3000 - 0x4002 33FF	CRC
	0x4002 2C00 - 0x4002 2FFF	Reserved
	0x4002 2800 - 0x4002 2BFF	GPIOK
	0x4002 2400 - 0x4002 27FF	GPIOJ
	0x4002 2000 - 0x4002 23FF	GPIOI
	0x4002 1C00 - 0x4002 1FFF	GPIOH
	0x4002 1800 - 0x4002 1BFF	GPIOG
	0x4002 1400 - 0x4002 17FF	GPIOF
	0x4002 1000 - 0x4002 13FF	GPIOE
	0X4002 0C00 - 0x4002 0FFF	GPIOD
	0x4002 0800 - 0x4002 0BFF	GPIOC
	0x4002 0400 - 0x4002 07FF	GPIOB
	0x4002 0000 - 0x4002 03FF	GPIOA

Table 13. STM32F427xx and STM32F429xx register boundary addresses (continued)



Table 24. Typical and maximum current consumption in Run mode, code with data processing
running from Flash memory (ART accelerator enabled except prefetch) or RAM ⁽¹⁾

Symbol	Parameter	Conditions	f _{HCLK} (MHz)	Тур	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C 141 ⁽⁵⁾ 133 ⁽⁵⁾ 115 112 85 66 45 46 43 39 36 34 33 87 ⁽⁵⁾ 83 ⁽⁵⁾ 73 72 56 41 28 26 24 35 34 33 33 33	Unit	
			180	98	104 ⁽⁵⁾	123	141 ⁽⁵⁾		
			168	89	98 ⁽⁵⁾	116	133 ⁽⁵⁾		
			150	75	84	100	115		
			144	72	81	96	112		
			120	54	58	72	85		
		All	90	43	45	56	66		
		Peripherals	60	29	30	38	45		
	enabled ⁽³⁾⁽⁴⁾ 30 16 20	34	46						
			25	13	16	30	43		
	Supply current in RUN mode		16	11	13	27	39	mΔ	
			8	5	9	23	36		
			4	4	8	21	34		
			2	2	7	20	33		
'DD			180	44	47 ⁽⁵⁾	69	87 ⁽⁵⁾		
			168	41	45 ⁽⁵⁾	66	83 ⁽⁵⁾		
			150	36	39	57	73		
			144	33	37	56	72		
			120	25	29	43	56		
		All	90	20	21	32	41		
		Peripherals	60	14	15	22	28		
		disabled(0)	30	8	8	12	26		
			25	7	7	10	24		
			16	7	9	22	35		
			8	3	7	21	34		
			4	3	6	20	33		
				2	2	6	20	33	

1. Code and data processing running from SRAM1 using boot pins.

2. Guaranteed by characterization.

3. When analog peripheral blocks such as ADCs, DACs, HSE, LSE, HSI, or LSI are ON, an additional power consumption should be considered.

4. When the ADC is ON (ADON bit set in the ADC_CR2 register), add an additional power consumption of 1.6 mA per ADC for the analog part.

5. Guaranteed by test in production.



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Figure 25. Typical V_{BAT} current consumption (LSE and RTC ON/backup RAM OFF)

Figure 26. Typical V_{BAT} current consumption (LSE and RTC ON/backup RAM ON)





I/O system current consumption

The current consumption of the I/O system has two components: static and dynamic.

I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in *Table 56: I/O static characteristics*.

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

Caution: Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

I/O dynamic current consumption

In addition to the internal peripheral current consumption (see *Table 35: Peripheral current consumption*), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the MCU supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DD} \times f_{SW} \times C$$

where

 I_{SW} is the current sunk by a switching I/O to charge/discharge the capacitive load

V_{DD} is the MCU supply voltage

f_{SW} is the I/O switching frequency

C is the total capacitance seen by the I/O pin: C = C_{INT} + C_{EXT}

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.



6.3.9 External clock source characteristics

High-speed external user clock generated from an external source

In bypass mode the HSE oscillator is switched off and the input pin is a standard I/O. The external clock signal has to respect the *Table 56: I/O static characteristics*. However, the recommended clock input waveform is shown in *Figure 27*.

The characteristics given in *Table 37* result from tests performed using an high-speed external clock source, and under ambient temperature and supply voltage conditions summarized in *Table 17*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{HSE_ext}	External user clock source frequency ⁽¹⁾		1	-	50	MHz
V _{HSEH}	OSC_IN input pin high level voltage		$0.7V_{DD}$	-	V_{DD}	V
V _{HSEL}	OSC_IN input pin low level voltage		V_{SS}	-	$0.3V_{\text{DD}}$	v
t _{w(HSE)} t _{w(HSE)}	OSC_IN high or low time ⁽¹⁾		5	-	-	ns
t _{r(HSE)} t _{f(HSE)}	OSC_IN rise or fall time ⁽¹⁾		-	-	10	15
C _{in(HSE)}	OSC_IN input capacitance ⁽¹⁾		-	5	-	pF
DuCy _(HSE)	Duty cycle		45	-	55	%
١ _L	OSC_IN Input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$	-	-	±1	μA

Table 37. High-speed external user clock characteristics

1. Guaranteed by design.



Symbol	Parameter	Conditions	S	Min	Тур	Мах	Unit
			RMS	-	25	-	
Jitter ⁽³⁾	Cycle-to-cycle jitter	System clock	peak to peak	-	±150	-	
		120 MHz	RMS	-	15	-	
	Period Jitter		peak to peak	-	<u>+200</u>	-	ps
	Main clock output (MCO) for RMII Ethernet	Cycle to cycle at 5 on 1000 samples	-	32	-		
	Main clock output (MCO) for MII Ethernet	Cycle to cycle at 25 MHz on 1000 samples		-	40	-	
	Bit Time CAN jitter	Cycle to cycle at 2 on 1000 samples	1 MHz	-	330	-	
I _{DD(PLL)} ⁽⁴⁾	PLL power consumption on VDD	VCO freq = 100 M VCO freq = 432 M	1Hz 1Hz	0.15 0.45	-	0.40 0.75	mA
I _{DDA(PLL)} ⁽⁴⁾	PLL power consumption on VDDA	Cycle to cycle at 50 MHz on 1000 samples Cycle to cycle at 25 MHz on 1000 samples Cycle to cycle at 1 MHz on 1000 samples VCO freq = 100 MHz VCO freq = 432 MHz VCO freq = 432 MHz		0.30 0.55	-	0.40 0.85	mA

Table 43. Main PLL characteristics (continued)

1. Take care of using the appropriate division factor M to obtain the specified PLL input clock values. The M factor is shared between PLL and PLLI2S.

2. Guaranteed by design.

3. The use of 2 PLLs in parallel could degraded the Jitter up to +30%.

4. Guaranteed by characterization results.

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
f _{PLLI2S_IN}	PLLI2S input clock ⁽¹⁾			0.95 ⁽²⁾	1	2.10	MHz
f _{PLLI2S_OUT}	PLLI2S multiplier output clock			-	-	216	MHz
f _{VCO_OUT}	PLLI2S VCO output			100	-	432	MHz
	PLLI2S lock time	VCO freq = 100 MHz	<u>z</u>	75	-	200	110
LOCK		VCO freq = 432 MHz	2	100	-	300	μο
LOCK		Cycle to cycle at	RMS	-	90	-	
	Master I2S clock jitter	12.288 MHz on 48KHz period, N=432, R=5	peak to peak	-	±280	-	ps
Jitter ⁽³⁾	DIParameterConditionsMinTypMinPLLI2S input clock $0.95^{(2)}$ 12.JTPLLI2S multiplier output clock $ -$ 2.PLLI2S word $0.95^{(2)}$ 100 $-$ 4.PLLI2S VCO output 100 $-$ 4.PLLI2S lock timeVCO freq = 100 MHz75 $-$ 2.VCO freq = 432 MHz 100 $-$ 3.VCO freq = 432 MHz 100 $ 3.6$ VCO freq = 432 MHz 0.90 $ -$ Master I2S clock jitter 2.88 MHz on 48 KHz period, $N = 432$, R = 5 on 1000 samples $ -$ WS I2S clock jitter $Cycle to cycle at 48$ KHz on 1000 samples $ 400$	-	ps				
	WS I2S clock jitter	Cycle to cycle at 48 I on 1000 samples	-	400	-	ps	

Table 44. PLLI2S (audio PLL) characteristics



SymbolParameterN _{END} Endurance		Parameter Conditions Value Min ⁽¹⁾		Unit
		$T_A = -40$ to +85 °C (6 suffix versions) $T_A = -40$ to +105 °C (7 suffix versions)	10	kcycles
		1 kcycle ⁽²⁾ at T _A = 85 °C	30	
t _{RET}	Data retention	1 kcycle ⁽²⁾ at T _A = 105 °C	10	Years
		10 kcycles ⁽²⁾ at T _A = 55 °C	20	

 Table 50. Flash memory endurance and data retention

1. Guaranteed by characterization results.

2. Cycling performed over the whole temperature range.

6.3.14 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A burst of fast transient voltage (positive and negative) is applied to V_{DD} and V_{SS} through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 51*. They are based on the EMS levels and classes defined in application note AN1709.

Symbol	Parameter	Conditions	Level/ Class
V _{FESD}	Voltage limits to be applied on any I/O pin to induce a functional disturbance	V _{DD} = 3.3 V, LQFP176, T _A = +25 °C, f _{HCLK} = 168 MHz, conforms to IEC 61000-4-2	2B
V _{EFTB}	Fast transient voltage burst limits to be applied through 100 pF on V_{DD} and V_{SS} pins to induce a functional disturbance	$V_{DD} = 3.3 \text{ V}, \text{LQFP176}, \text{T}_{\text{A}} =+25 \text{ °C}, \text{f}_{\text{HCLK}} = 168 \text{ MHz}, \text{ conforms to} \text{IEC 61000-4-2}$	4A

Table 51. EMS characteristics

When the application is exposed to a noisy environment, it is recommended to avoid pin exposition to disturbances. The pins showing a middle range robustness are: PA0, PA1, PA2, PH2, PH3, PH4, PH5, PA3, PA4, PA5, PA6, PA7, PC4, and PC5.

As a consequence, it is recommended to add a serial resistor (1 k Ω) located as close as possible to the MCU to the pins exposed to noise (connected to tracks longer than 50 mm on PCB).



I²S interface characteristics

Unless otherwise specified, the parameters given in *Table 63* for the I²S interface are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in *Table 17*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5V_{DD}

Refer to Section 6.3.17: I/O port characteristics for more details on the input/output alternate function characteristics (CK, SD, WS).

Symbol	Parameter	Conditions	Min	Max	Unit
f _{MCK}	I2S Main clock output	-	256x8K	256xFs ⁽²⁾	MHz
6	129 clock froguency	Master data: 32 bits	-		N411-
^I CK	123 Clock frequency	Slave data: 32 bits	-	64xFs	IVIHZ
D _{CK}	I2S clock frequency duty cycle	Slave receiver	30	70	%
t _{v(WS)}	WS valid time	Master mode	0	6	
t _{h(WS)}	WS hold time	Master mode	0	-	
t _{su(WS)}	WS setup time	Slave mode	1	-	
t _{h(WS)}	WS hold time	Slave mode	0	-	
t _{su(SD_MR)}	Data input sotup timo	Master receiver	7.5	-	
t _{su(SD_SR)}		Slave receiver	2	-	
t _{h(SD_MR)}	Data input hold time	Master receiver	0	-	ns
t _{h(SD_SR)}	Data input noid time	Slave receiver	0	-	
t _{v(SD_ST)}		Slave transmitter (after enable edge)	-	27	
t _{h(SD_ST)}	Data output valid time				
t _{v(SD_MT)}		Master transmitter (after enable edge)	-	20	
t _{h(SD_MT)}	Data output hold time	Master transmitter (after enable edge)	2.5	-	

Table 63.	I ² S d	vnamic characteristics ⁽¹⁾
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1. Guaranteed by characterization results.

2. The maximum value of 256xFs is 45 MHz (APB1 maximum frequency).

Note: Refer to t

Refer to the I2S section of RM0090 reference manual for more details on the sampling frequency (F_S).

 f_{MCK} , f_{CK} , and D_{CK} values reflect only the digital peripheral behavior. The values of these parameters might be slightly impacted by the source clock precision. D_{CK} depends mainly on the value of ODD bit. The digital contribution leads to a minimum value of (I2SDIV/(2*I2SDIV+ODD) and a maximum value of (I2SDIV+ODD)/(2*I2SDIV+ODD). F_S maximum value is supported for each mode/condition.





Figure 55. Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms

1. Mode 2/B, C and D only. In Mode 1, FMC_NADV is not used.

Table 86. Asynchronous non-multiplexed SRAM/PSRAM/NOR -
road timings $^{(1)(2)}$
read tinings

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FMC_NE low time	2T _{HCLK} – 0.5	2 T _{HCLK} +0.5	ns
t _{v(NOE_NE)}	FMC_NEx low to FMC_NOE low	0	1	ns
t _{w(NOE)}	FMC_NOE low time	2T _{HCLK}	2T _{HCLK} + 0.5	ns
t _{h(NE_NOE)}	FMC_NOE high to FMC_NE high hold time	0	-	ns
t _{v(A_NE)}	FMC_NEx low to FMC_A valid	-	2	ns
t _{h(A_NOE)}	Address hold time after FMC_NOE high	0	-	ns
t _{v(BL_NE)}	FMC_NEx low to FMC_BL valid	-	2	ns
t _{h(BL_NOE)}	FMC_BL hold time after FMC_NOE high	0	-	ns
t _{su(Data_NE)}	Data to FMC_NEx high setup time	T _{HCLK} + 2.5	-	ns
t _{su(Data_NOE)}	Data to FMC_NOEx high setup time	T _{HCLK} +2	-	ns



Symbol	Parameter	Min	Мах	Unit			
t _{h(Data_NOE)}	Data hold time after FMC_NOE high	0	-	ns			
t _{h(Data_NE)}	Data hold time after FMC_NEx high	0	-	ns			
t _{v(NADV_NE)}	FMC_NEx low to FMC_NADV low	-	0	ns			
t _{w(NADV)}	FMC_NADV low time	-	T _{HCLK} +1	ns			

Table 86. Asynchronous non-multiplexed SRAM/PSRAM/NOR - read timings $^{(1)(2)}$ (continued)

1. C_L = 30 pF.

2. Guaranteed by characterization results.

Table 87. Asynchronous non-multiplexed SRAM/PSRAM/NOR read -
NWAIT timings ⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	w(NE) FMC_NE low time		7T _{HCLK} +1	
t _{w(NOE)}	t _{w(NOE)} FMC_NWE low time		5T _{HCLK} +2	ns
t _{su(NWAIT_NE)} FMC_NWAIT valid before FMC_NEx high		5T _{HCLK} +1.5	-	
^t h(NE_NWAIT)	FMC_NEx hold time after FMC_NWAIT invalid	4T _{HCLK} +1	-	

1. C_L = 30 pF.

2. Guaranteed by characterization results.





Figure 56. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms

1. Mode 2/B, C and D only. In Mode 1, FMC_NADV is not used.

	(4)(0)
Table 88. Asynchronous non-multi	plexed SRAM/PSRAM/NOR write timings ⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FMC_NE low time	3T _{HCLK}	3T _{HCLK} +1	ns
t _{v(NWE_NE)}	FMC_NEx low to FMC_NWE low	T _{HCLK} – 0.5	T _{HCLK} + 0.5	ns
t _{w(NWE)}	FMC_NWE low time	T _{HCLK}	T _{HCLK} + 0.5	ns
t _{h(NE_NWE)}	FMC_NWE high to FMC_NE high hold time	T _{HCLK} +1.5	-	ns
t _{v(A_NE)}	FMC_NEx low to FMC_A valid	-	0	ns
t _{h(A_NWE)}	Address hold time after FMC_NWE high	T _{HCLK} +0.5	-	ns
t _{v(BL_NE)}	FMC_NEx low to FMC_BL valid	-	1.5	ns
t _{h(BL_NWE)}	FMC_BL hold time after FMC_NWE high	T _{HCLK} +0.5	-	ns
t _{v(Data_NE)}	Data to FMC_NEx low to Data valid	-	T _{HCLK} + 2	ns
t _{h(Data_NWE)}	Data hold time after FMC_NWE high	T _{HCLK} +0.5	-	ns
t _{v(NADV_NE)}	FMC_NEx low to FMC_NADV low	-	0.5	ns
t _{w(NADV)}	FMC_NADV low time	-	T _{HCLK} + 0.5	ns

1. C_L = 30 pF.

2. Guaranteed by characterization results.





Figure 74. SDRAM write access waveforms



Device marking for LQFP208

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

Other optional marking or inset/upset marks, which depends assembly location, are not indicated below.





 Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.



Table 116. UFBGA169 - 169-ball 7 x 7 mm 0.50 mm pitch, ultra fine pitch ball grid array package mechanical data (continued)

Symbol		millimeters	illimeters inch		inches ⁽¹⁾	nches ⁽¹⁾	
Symbol	Min	Тур	Мах	Min	Тур	Мах	
F	0.450	0.500	0.550	0.0177	0.0197	0.0217	
ddd	-	-	0.100	-	-	0.0039	
eee	-	-	0.150	-	-	0.0059	
fff	-	-	0.050	-	-	0.0020	

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 96. UFBGA169 - 169-ball, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array recommended footprint



Table 117. UFBGA169 recommended PCB design rules (0.5 mm pitch BGA)

Dimension	Recommended values
Pitch	0.5
Dpad	0.27 mm
Dsm	0.35 mm typ. (depends on the soldermask registration tolerance)
Solder paste	0.27 mm aperture diameter.

Note:Non-solder mask defined (NSMD) pads are recommended.4 to 6 mils solder paste screen printing process.

