



Welcome to [E-XFL.COM](#)

### What is "[Embedded - Microcontrollers](#)"?

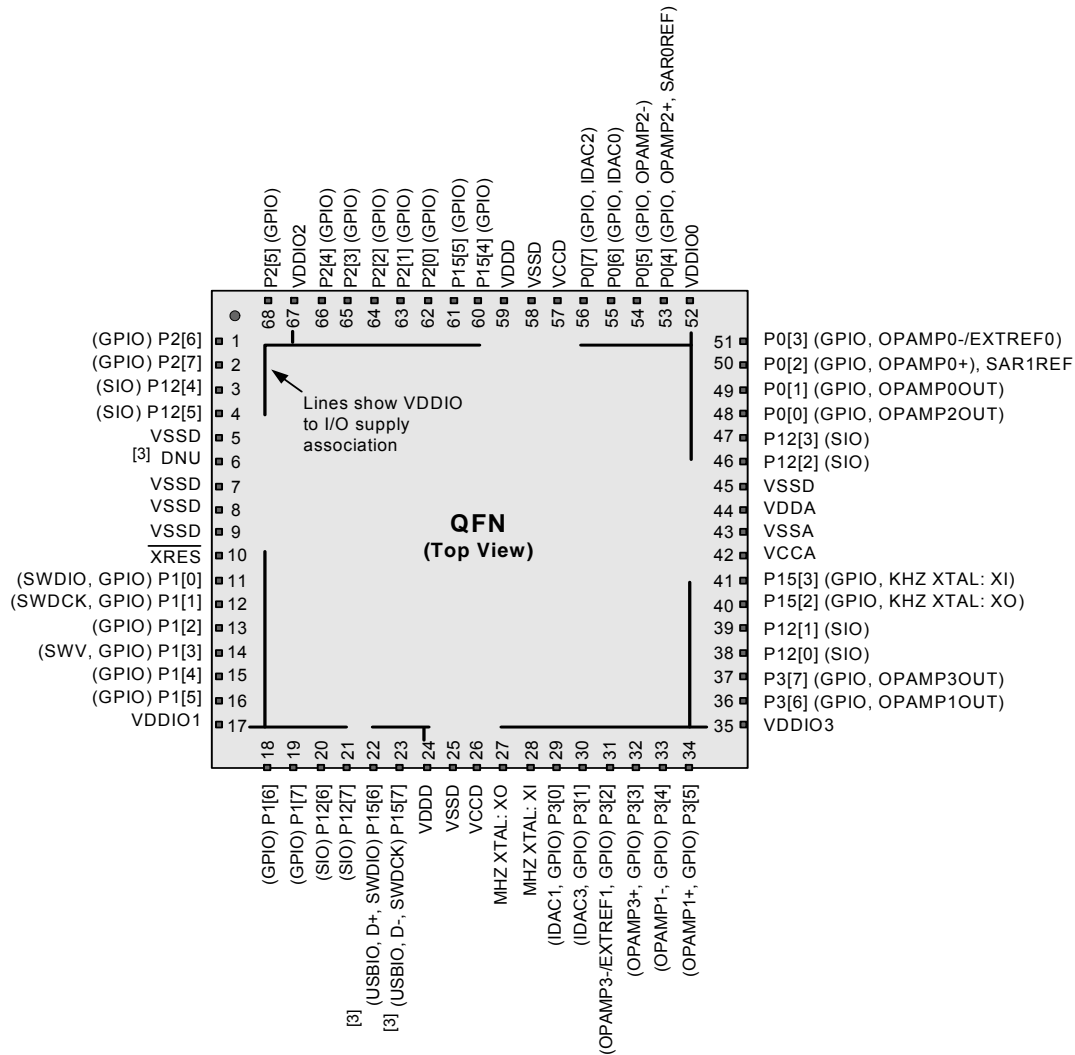
"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Obsolete
Core Processor	ARM® Cortex®-M3
Core Size	32-Bit Single-Core
Speed	67MHz
Connectivity	I <sup>2</sup> C, LINbus, SPI, UART/USART, USB
Peripherals	CapSense, DMA, LCD, POR, PWM, WDT
Number of I/O	60
Program Memory Size	256KB (256K x 8)
Program Memory Type	FLASH
EEPROM Size	2K x 8
RAM Size	64K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 2x12b; D/A 4x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/infineon-technologies/cy8c5468axi-018">https://www.e-xfl.com/product-detail/infineon-technologies/cy8c5468axi-018</a>

**Figure 2-1. 68-pin QFN Part Pinout<sup>[2]</sup>**



**Notes**

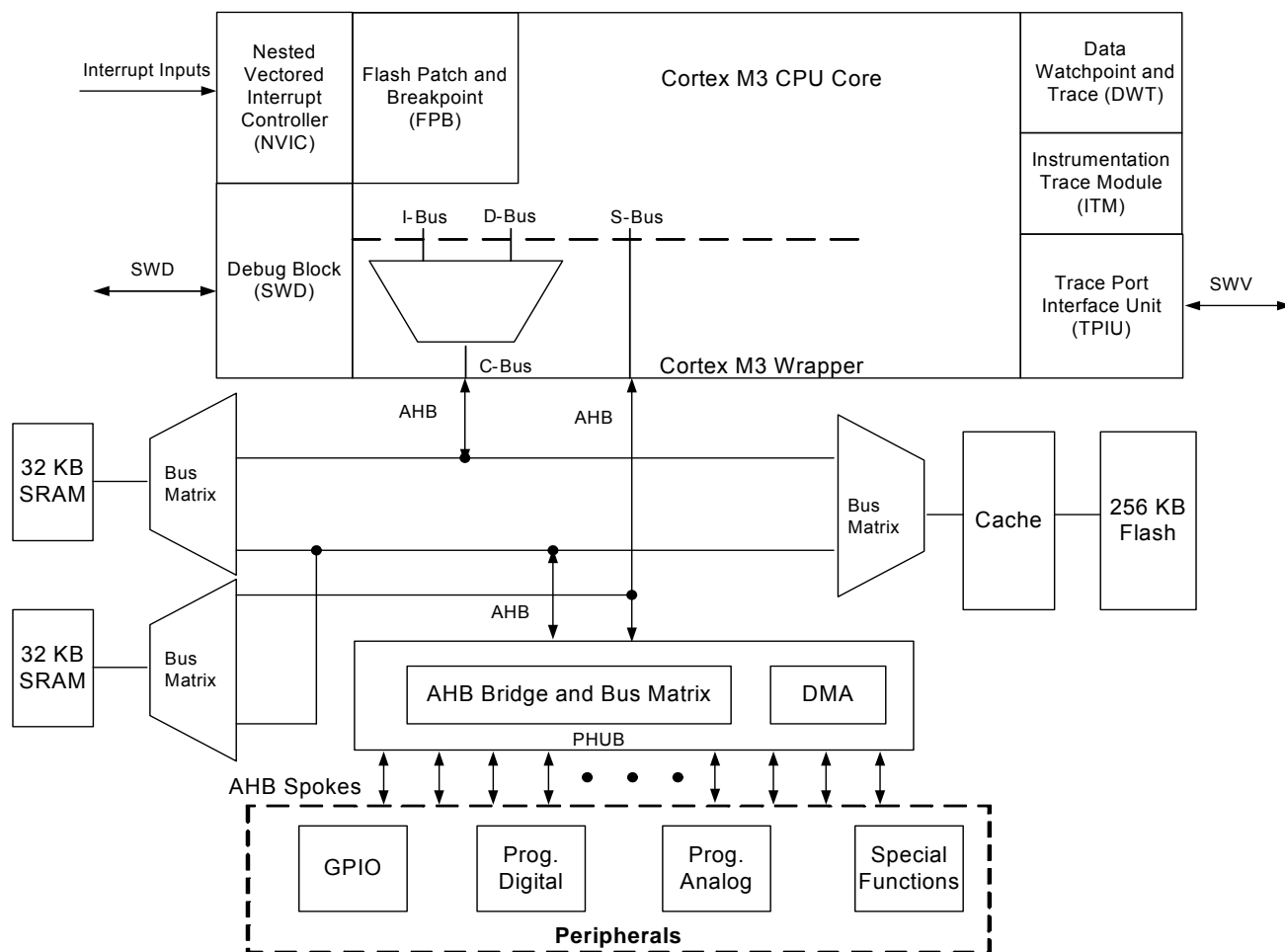
- The center pad on the QFN package should be connected to digital ground ( $V_{SSD}$ ) for best mechanical, thermal, and electrical performance. If not connected to ground, it should be electrically floated and not connected to any other signal.
- Pins labeled Do Not Use (DNU) must be left floating. USB pins on devices without USB are DNU.

## 4. CPU

### 4.1 ARM Cortex-M3 CPU

The CY8C54 family of devices has an ARM Cortex-M3 CPU core. The Cortex-M3 is a low power 32-bit three-stage pipelined Harvard architecture CPU that delivers 1.25 DMIPS/MHz. It is intended for deeply embedded applications that require fast interrupt handling features.

**Figure 4-1. ARM Cortex-M3 Block Diagram**



The Cortex-M3 CPU subsystem includes these features:

- ARM Cortex-M3 CPU
- Programmable Nested Vectored Interrupt Controller (NVIC), tightly integrated with the CPU core
- Full featured debug and trace module, tightly integrated with the CPU core
- Up to 256 KB of flash memory, 2 KB of EEPROM, and 64 KB of SRAM
- Cache controller with 128 bytes of memory
- Peripheral HUB (PHUB)
- DMA controller

#### 4.1.1 Cortex-M3 Features

The Cortex-M3 CPU features include:

- 4 GB address space. Predefined address regions for code, data, and peripherals. Multiple buses for efficient and simultaneous accesses of instructions, data, and peripherals.
- The Thumb®-2 instruction set, which offers ARM-level performance at Thumb-level code density. This includes 16-bit and 32-bit instructions. Advanced instructions include:
  - Bit-field control
  - Hardware multiply and divide
  - Saturation
  - If-Then
  - Wait for events and interrupts
  - Exclusive access and barrier
  - Special register access

The Cortex-M3 does not support ARM instructions.

There are two PHUB masters: the CPU and the DMA controller. Both masters may initiate transactions on the bus. The DMA channels can handle peripheral communication without CPU intervention. The arbiter in the central hub determines which DMA channel is the highest priority if there are multiple requests.

#### 4.3.1 PHUB Features

- CPU and DMA controller are both bus masters to the PHUB
- Eight Multi-layer AHB Bus parallel access paths (spokes) for peripheral access
- Simultaneous CPU and DMA access to peripherals located on different spokes
- Simultaneous DMA source and destination burst transactions on different spokes
- Supports 8, 16, 24, and 32-bit addressing and data

**Table 4-3. PHUB Spokes and Peripherals**

PHUB Spokes	Peripherals
0	SRAM
1	IOs, PICU
2	PHUB local configuration, Power manager, Clocks, IC, EEPROM, Flash programming interface
3	Analog interface and trim, Decimator
4	USB, I <sup>2</sup> C, Timers, Counters, and PWMs
5	DFB
6	UDBs group 1
7	UDBs group 2

#### 4.3.2 DMA Features

- 24 DMA channels
- Each channel has one or more Transaction Descriptors (TDs) to configure channel behavior. Up to 127 total TDs can be defined
- TDs can be dynamically updated
- Eight levels of priority per channel
- Any digitally routable signal, the CPU, or another DMA channel, can trigger a transaction
- Each channel can generate up to two interrupts per transfer
- Transactions can be stalled or canceled
- Supports transaction size of infinite or 1 to 64k bytes
- Large transactions may be broken into smaller bursts of 1 to 127 bytes
- TDs may be nested and/or chained for complex transactions

#### 4.3.3 Priority Levels

The CPU always has higher priority than the DMA controller when their accesses require the same bus resources. Due to the system architecture, the CPU can never starve the DMA. DMA channels of higher priority (lower priority number) may interrupt current DMA transfers. In the case of an interrupt, the current transfer is allowed to complete its current transaction. To ensure latency limits when multiple DMA accesses are requested simultaneously, a fairness algorithm guarantees an interleaved minimum percentage of bus bandwidth for priority levels 2 through 7. Priority levels 0 and 1 do not take part in the fairness algorithm and may use 100% of the bus bandwidth. If a tie occurs on two DMA requests of the same priority level, a simple round robin method is used to evenly share the allocated bandwidth. The round robin allocation can be disabled for each DMA channel, allowing it to always be at the head of the line. Priority levels 2 to 7 are guaranteed the minimum bus bandwidth shown in [Table 4-4](#) after the CPU and DMA priority levels 0 and 1 have satisfied their requirements.

**Table 4-4. Priority Levels**

Priority Level	% Bus Bandwidth
0	100.0
1	100.0
2	50.0
3	25.0
4	12.5
5	6.2
6	3.1
7	1.5

When the fairness algorithm is disabled, DMA access is granted based solely on the priority level; no bus bandwidth guarantees are made.

#### 4.3.4 Transaction Modes Supported

The flexible configuration of each DMA channel and the ability to chain multiple channels allow the creation of both simple and complex use cases. General use cases include, but are not limited to:

##### 4.3.4.1 Simple DMA

In a simple DMA case, a single TD transfers data between a source and sink (peripherals or memory location). The basic timing diagrams of DMA read and write cycles are shown in [Figure 4-2](#). For more description on other transfer modes, refer to the Technical Reference Manual.

**Table 4-6. Interrupt Vector Table**

Interrupt #	Cortex-M3 Exception #	Fixed Function	DMA	UDB
0	16	Low voltage detect (LVD)	phub_termout0[0]	udb_intr[0]
1	17	Cache	phub_termout0[1]	udb_intr[1]
2	18	Reserved	phub_termout0[2]	udb_intr[2]
3	19	Pwr Mgr	phub_termout0[3]	udb_intr[3]
4	20	PICU[0]	phub_termout0[4]	udb_intr[4]
5	21	PICU[1]	phub_termout0[5]	udb_intr[5]
6	22	PICU[2]	phub_termout0[6]	udb_intr[6]
7	23	PICU[3]	phub_termout0[7]	udb_intr[7]
8	24	PICU[4]	phub_termout0[8]	udb_intr[8]
9	25	PICU[5]	phub_termout0[9]	udb_intr[9]
10	26	PICU[6]	phub_termout0[10]	udb_intr[10]
11	27	PICU[12]	phub_termout0[11]	udb_intr[11]
12	28	PICU[15]	phub_termout0[12]	udb_intr[12]
13	29	Comparators Combined	phub_termout0[13]	udb_intr[13]
14	30	Switched Caps Combined	phub_termout0[14]	udb_intr[14]
15	31	I <sup>2</sup> C	phub_termout0[15]	udb_intr[15]
16	32	Reserved	phub_termout1[0]	udb_intr[16]
17	33	Reserved	phub_termout1[1]	udb_intr[17]
18	34	Reserved	phub_termout1[2]	udb_intr[18]
19	35	Reserved	phub_termout1[3]	udb_intr[19]
20	36	Reserved	phub_termout1[4]	udb_intr[20]
21	37	USB SOF Int	phub_termout1[5]	udb_intr[21]
22	38	USB Arb Int	phub_termout1[6]	udb_intr[22]
23	39	USB Bus Int	phub_termout1[7]	udb_intr[23]
24	40	USB Endpoint[0]	phub_termout1[8]	udb_intr[24]
25	41	USB Endpoint Data	phub_termout1[9]	udb_intr[25]
26	42	Reserved	phub_termout1[10]	udb_intr[26]
27	43	Reserved	phub_termout1[11]	udb_intr[27]
28	44	DFB Int	phub_termout1[12]	udb_intr[28]
29	45	Decimator Int	phub_termout1[13]	udb_intr[29]
30	46	phub_err_int	phub_termout1[14]	udb_intr[30]
31	47	eeeprom_fault_int	phub_termout1[15]	udb_intr[31]

## 5. Memory

### 5.1 Static RAM

CY8C54 Static RAM (SRAM) is used for temporary data storage. Code can be executed at full speed from the portion of SRAM that is located in the code space. This process is slower from SRAM above 0x20000000. The device provides up to 64 KB of SRAM. The CPU or the DMA controller can access all of SRAM. The SRAM can be accessed simultaneously by the Cortex-M3 CPU and the DMA controller if accessing different 32 KB blocks.

### 5.2 Flash Program Memory

Flash memory in PSoC devices provides nonvolatile storage for user firmware, user configuration data and bulk data storage. The main flash memory area contains up to 256 KB of user program space. Up to an additional 32 KB of flash space is available for storing device configuration data and bulk user data. User code may not be run out of this flash memory section. The flash output is 9 bytes wide with 8 bytes of data and one additional byte.

The flash programming interface performs flash erasing, programming and setting code protection levels. Flash In System Serial Programming (ISSP), typically used for production programming, is possible through the SWD interface. In-system programming, typically used for bootloaders, is also possible using serial interfaces such as I<sup>2</sup>C, USB, UART, and SPI, or any communications protocol.

### 5.3 Flash Security

All PSoC devices include a flexible flash protection model that prevents access and visibility to on-chip flash memory. This prevents duplication or reverse engineering of proprietary code. Flash memory is organized in blocks, where each block contains 256 bytes of program or data and 32 bytes of configuration or general-purpose data.

The device offers the ability to assign one of four protection levels to each row of flash. Table 5-1 lists the protection modes available. Flash protection levels can only be changed by performing a complete flash erase. The Full Protection and Field Upgrade settings disable external access (through a debugging tool such as PSoC Creator, for example). If your application requires code update through a boot loader, then use the Field Upgrade setting. Use the Unprotected setting only when no security is needed in your application. The PSoC device also offers an advanced security feature called Device Security which permanently disables all test, programming, and debug ports, protecting your application from external access (see the

“Device Security” section on page 52). For more information on how to take full advantage of the security features in PSoC, see the PSoC 5 TRM.

**Table 5-1. Flash Protection**

Protection Setting	Allowed	Not Allowed
Unprotected	External read and write + internal read and write	–
Factory Upgrade	External write + internal read and write	External read
Field Upgrade	Internal read and write	External read and write
Full Protection	Internal read	External read and write + internal write

### Disclaimer

Note the following details of the flash code protection features on Cypress devices.

Cypress products meet the specifications contained in their particular Cypress data sheets. Cypress believes that its family of products is one of the most secure families of its kind on the market today, regardless of how they are used. There may be methods, unknown to Cypress, that can breach the code protection features. Any of these methods, to our knowledge, would be dishonest and possibly illegal. Neither Cypress nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as “unbreakable.”

Cypress is willing to work with the customer who is concerned about the integrity of their code. Code protection is constantly evolving. We at Cypress are committed to continuously improving the code protection features of our products.

### 5.4 EEPROM

PSoC EEPROM memory is a byte addressable nonvolatile memory. The CY8C54 has 2 KB of EEPROM memory to store user data. Reads from EEPROM are random access at the byte level. Reads are done directly; writes are done by sending write commands to an EEPROM programming interface. CPU code execution can continue from flash during EEPROM writes. EEPROM is erasable and writeable at the row level. The EEPROM is divided into two sections, each containing 64 rows of 16 bytes each. The CPU cannot execute out of EEPROM.

## 6. System Integration

### 6.1 Clocking System

The clocking system generates, divides, and distributes clocks throughout the PSoC system. The IMO and PLL together can generate up to a 67 MHz clock, accurate to  $\pm 5\%$  over voltage and temperature. Additional internal and external clock sources allow each design to optimize accuracy, power, and cost. All of the system clock sources can be used to generate other clock frequencies in the 16-bit clock dividers and UDBs for anything the user wants, for example a UART baud rate generator.

Clock generation and distribution is automatically configured through the PSoC Creator IDE graphical interface. This is based on the complete system's requirements. It greatly speeds the design process. PSoC Creator allows designers to build clocking systems with minimal input. The designer can specify desired clock frequencies and accuracies, and the software locates or builds a clock that meets the required specifications. This is possible because of the programmability inherent PSoC.

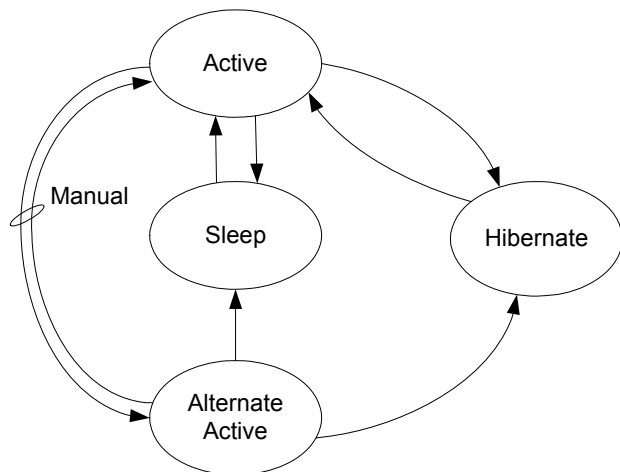
Key features of the clocking system include:

- Seven general purpose clock sources
  - 3 to 48 MHz IMO,  $\pm 5\%$  at 3 MHz
  - 4 to 25 MHz External Crystal Oscillator (MHzECO)
  - Clock doubler provides a doubled clock frequency output for the USB block, see [USB Clock Domain](#) on page 21
  - DSI signal from an external I/O pin or other logic
  - 24 to 67 MHz fractional Phase-Locked Loop (PLL) sourced from IMO, MHzECO, or DSI
  - 1 kHz, 33 kHz, 100 kHz ILO for Watch Dog Timer (WDT) and Sleep Timer
  - 32.768 kHz External Crystal Oscillator (kHzECO) for Real Time Clock (RTC)
- Independently sourced clock dividers in all clocks
- Eight 16-bit clock dividers for the digital system
- Four 16-bit clock dividers for the analog system
- Dedicated 16-bit divider for the CPU bus and CPU clock
- Automatic clock configuration in PSoC Creator

**Table 6-1. Oscillator Summary**

Source	Fmin	Tolerance at Fmin	Fmax	Tolerance at Fmax	Startup Time
IMO	3 MHz	$\pm 5\%$ over voltage and temperature	48 MHz	$\pm 10\%$	12 $\mu$ s max
MHzECO	4 MHz	Crystal dependent	25 MHz	Crystal dependent	5 ms typ, max is crystal dependent
DSI	0 MHz	Input dependent	66 MHz	Input dependent	Input dependent
PLL	24 MHz	Input dependent	67 MHz	Input dependent	250 $\mu$ s max
Doubler	48 MHz	Input dependent	48 MHz	Input dependent	1 $\mu$ s max
ILO	1 kHz	-50%, +100%	100 kHz	-55%, +100%	15 ms max in lowest power mode
kHzECO	32 kHz	Crystal dependent	32 kHz	Crystal dependent	500 ms typ, max is crystal dependent

**Figure 6-5. Power Mode Transitions**



### 6.2.1.1 Active Mode

Active mode is the primary operating mode of the device. When in active mode, the active configuration template bits control which available resources are enabled or disabled. When a resource is disabled, the digital clocks are gated, analog bias currents are disabled, and leakage currents are reduced as appropriate. User firmware can dynamically control subsystem power by setting and clearing bits in the active configuration template. The CPU can disable itself, in which case the CPU is automatically reenabled at the next wakeup event.

When a wakeup event occurs, the global mode is always returned to active, and the CPU is automatically enabled, regardless of its template settings. Active mode is the default global power mode upon boot.

### 6.2.1.2 Alternate Active Mode

Alternate Active mode is very similar to Active mode. In alternate active mode, fewer subsystems are enabled, to reduce power consumption. One possible configuration is to turn off the CPU and flash, and run peripherals at full speed.

### 6.2.1.3 Sleep Mode

Sleep mode powers down the CPU and other internal circuitry to reduce power consumption. However, supervisory services such as the central timewheel (CTW) remain available in this mode. The device can wake up using CTW or system reset. The wake up time from sleep mode is 125  $\mu$ s (typical).

### 6.2.1.4 Hibernate Mode

In hibernate mode nearly all of the internal functions are disabled. Internal voltages are reduced to the minimal level to keep vital systems alive. Configuration state is preserved in hibernate mode and SRAM memory is retained. GPIOs configured as digital outputs maintain their previous values and external GPIO pin interrupt settings are preserved. The device can only return from hibernate mode in response to an external reset (XRES).

### 6.2.1.5 Wakeup Events

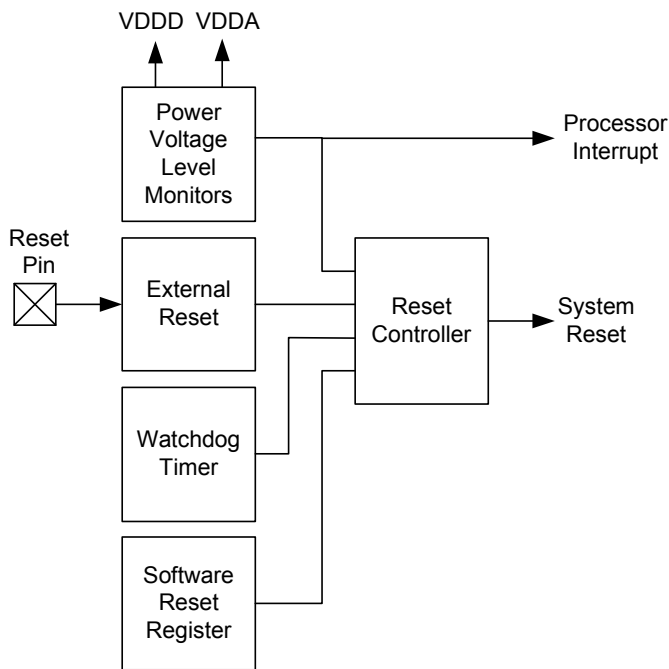
Wakeup events can come from the central timewheel or device reset. A wakeup event restores the system to active mode. The central timewheel allows the system to periodically wake up, poll peripherals, do voltage monitoring, or perform real-time functions. Reset event sources include the external reset pin (XRES).

## 6.3 Reset

CY8C54 has multiple internal and external reset sources available. The reset sources are:

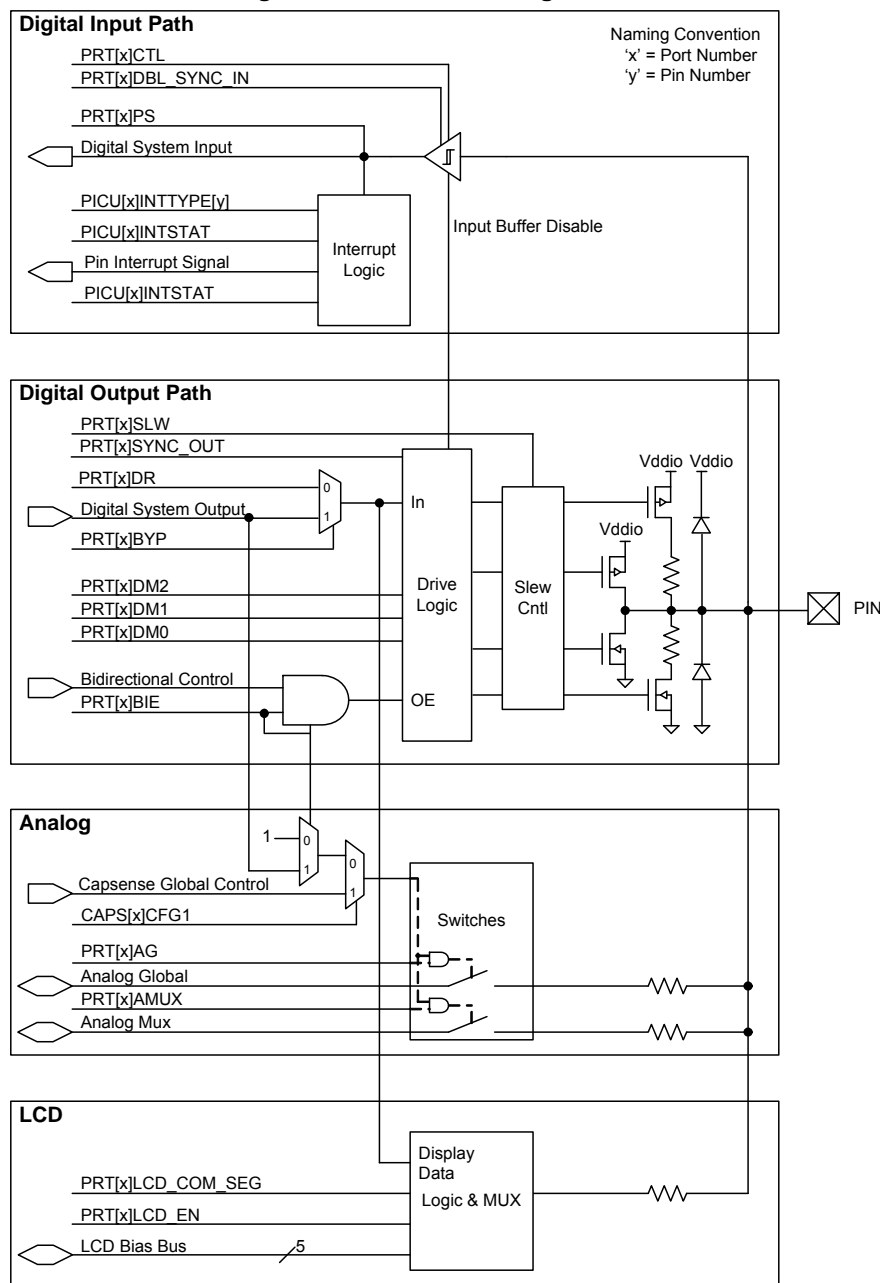
- **Power source monitoring** - The analog and digital power voltages,  $V_{DDA}$ ,  $V_{DDD}$ ,  $V_{CCA}$ , and  $V_{CCD}$  are monitored in several different modes during power up and active mode. The monitors are programmable to generate an interrupt to the processor under certain conditions.
- **External** - The device can be reset from an external source by pulling the reset pin (XRES) low. The XRES pin includes an internal pull up to  $V_{ddio1}$ .  $V_{DDD}$ ,  $V_{DDA}$ , and  $V_{ddio1}$  must all have voltage applied before the part comes out of reset.
- **Watchdog timer** - A watchdog timer monitors the execution of instructions by the processor. If the watchdog timer is not reset by firmware within a certain period of time, the watchdog timer generates a reset. The watchdog timer can be used only when the part remains in active mode.
- **Software** - The device can be reset under program control.

**Figure 6-6. Resets**



- Continuous 100  $\mu$ A clamp current capability
- Standard drive strength down to 2.7 V
- Additional features only provided on SIO pins:
  - Higher drive strength than GPIO
  - Hot swap capability (5 V tolerance at any operating  $V_{DD}$ )
  - Programmable and regulated high input and output drive levels down to 1.2 V
  - No analog input or LCD capability
  - Over voltage tolerance up to 5.5 V
- SIO can act as a general purpose analog comparator
- USBIO features:
  - Full speed USB 2.0 I/O
  - Highest drive strength for general purpose use
  - Input, output, or both for CPU and DMA
  - Input, output, or both for digital peripherals
  - Digital output (CMOS) drive mode
  - Each pin can be an interrupt source configured as rising edge, falling edge, or both edges

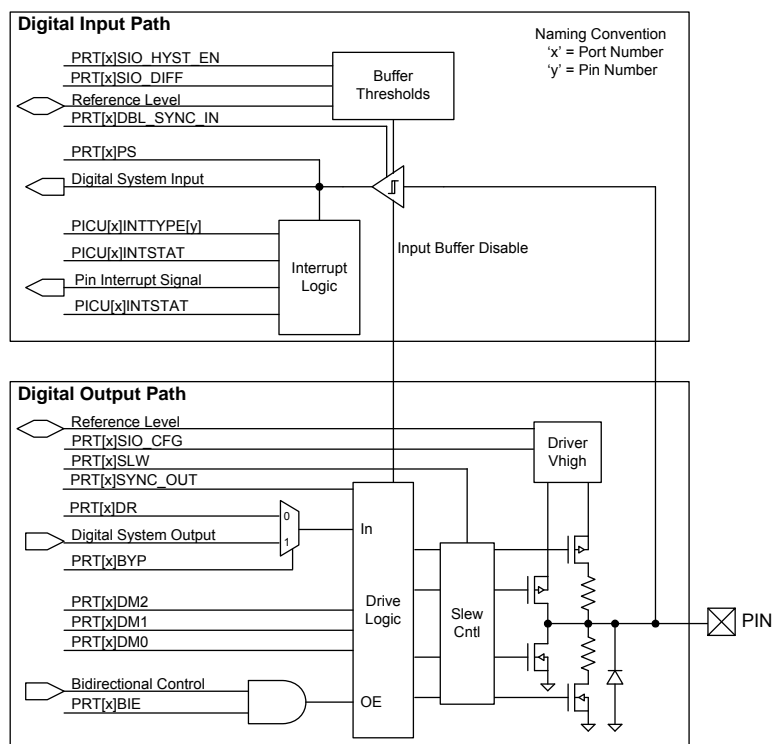
**Figure 6-7. GPIO Block Diagram**



**Note**

11. GPIOs with opamp outputs are not recommended for use with CapSense.

**Figure 6-8. SIO Input/Output Block Diagram**



**Figure 6-9. USBIO Block Diagram**

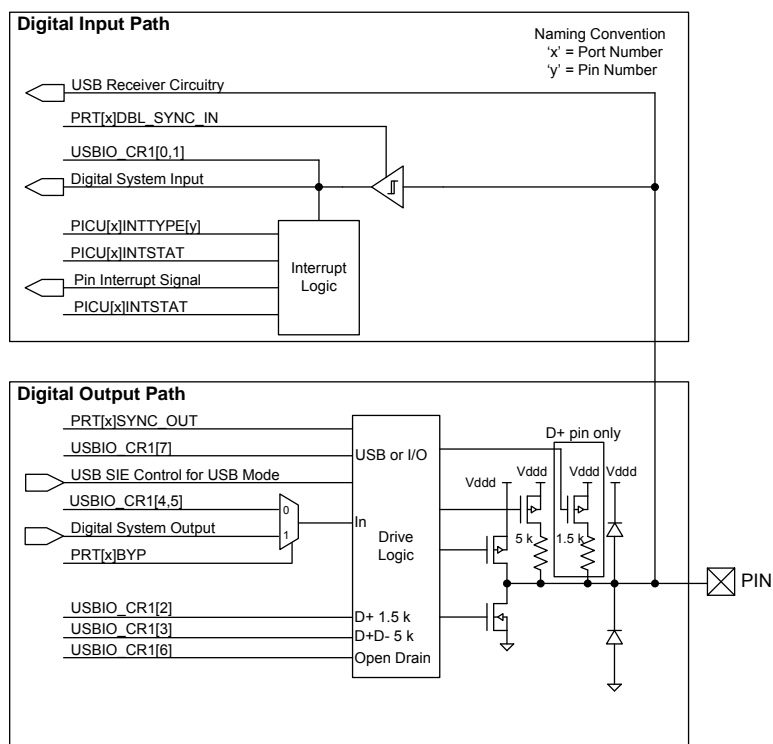
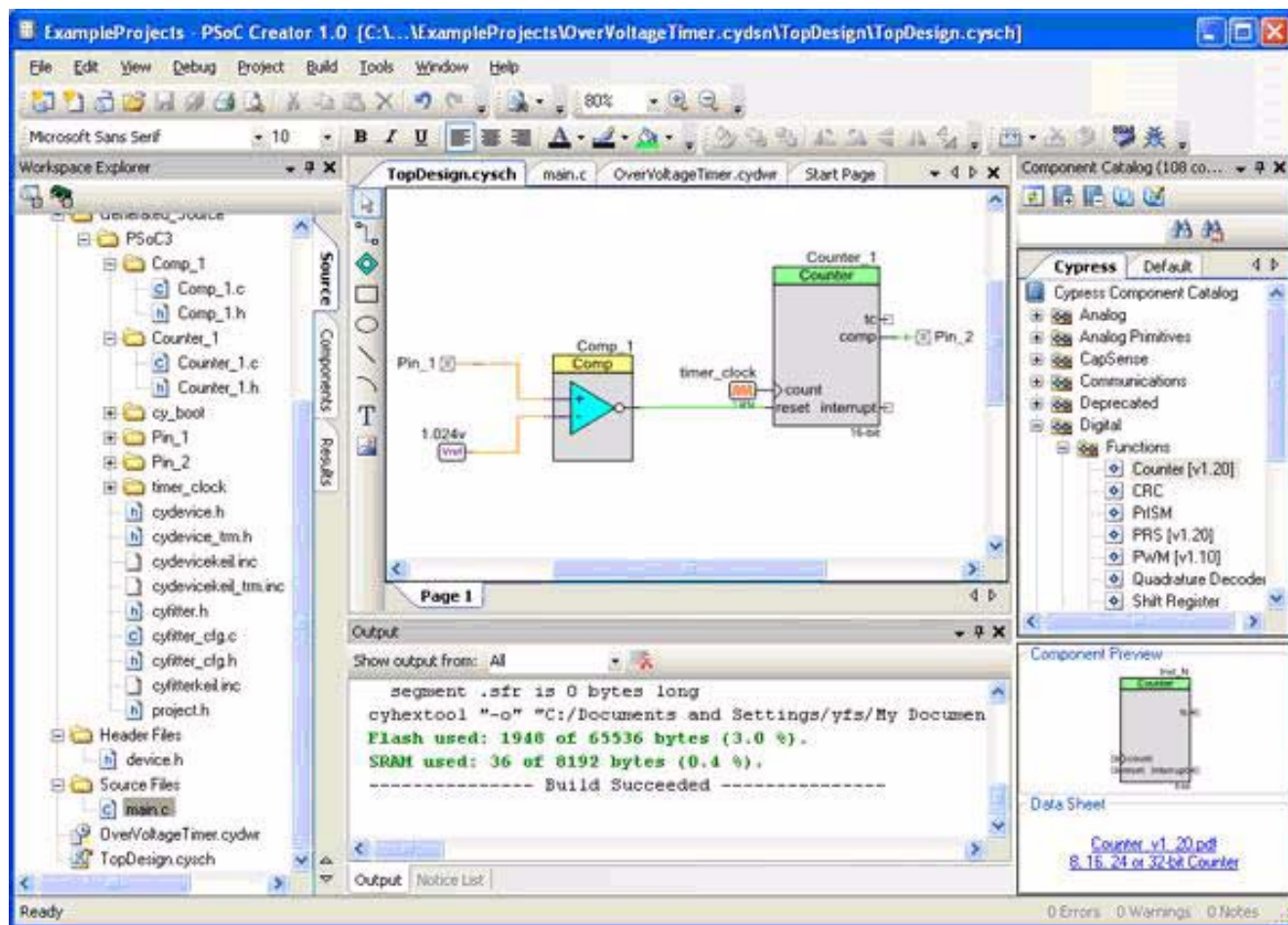


Figure 7-2. PSoc Creator Framework



Analog local buses (abus) are routing resources located within the analog subsystem and are used to route signals between different analog blocks. There are eight abus routes in PSoC 5, four in the left half (abusl [0:3]) and four in the right half (abusr [0:3]) as shown in Figure 8-2. Using the abus saves the analog globals and analog mux buses from being used for interconnecting the analog blocks.

Multiplexers and switches exist on the various buses to direct signals into and out of the analog blocks. A multiplexer can have only one connection on at a time, whereas a switch can have multiple connections on simultaneously. In Figure 8-2, multiplexers are indicated by grayed ovals and switches are indicated by transparent ovals.

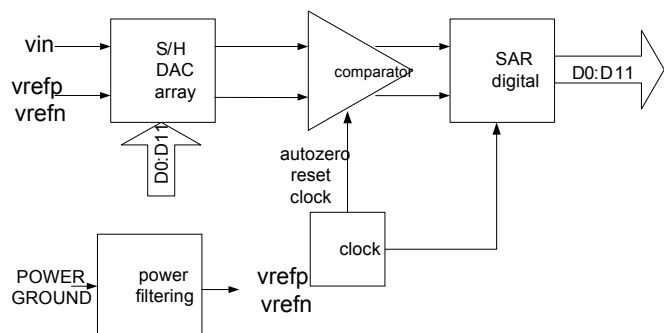
## 8.2 Successive Approximation ADCs

The CY8C54 family of devices has two Successive Approximation (SAR) ADCs. These ADCs are 12-bit at up to 700 ksp/s, with single-ended or differential inputs, making them useful for a wide variety of sampling and control applications.

### 8.2.1 Functional Description

In a SAR ADC an analog input signal is sampled and compared with the output of a DAC. A binary search algorithm is applied to the DAC and used to determine the output bits in succession from MSB to LSB. A block diagram of one SAR ADC is shown in Figure 8-3.

**Figure 8-3. SAR ADC Block Diagram**



The input is connected to the analog globals and muxes. The maximum clock rate is 14 MHz.

### 8.2.2 Conversion Signals

Writing a start bit or assertion of a Start of Frame (SOF) signal is used to start a conversion. SOF can be used in applications

where the sampling period is longer than the conversion time, or when the ADC needs to be synchronized to other hardware. This signal is optional and does not need to be connected if the SAR ADC is running in a continuous mode. A digital clock or UDB output can be used to drive this input. When the SAR is first powered up or awakened from any of the sleeping modes, there is a power up wait time of 10  $\mu$ s before it is ready to start the first conversion.

When the conversion is complete, a status bit is set and the output signal End of Frame (EOF) asserts and remains asserted until the value is read by either the DMA controller or the CPU. The EOF signal may be used to trigger an interrupt or a DMA request.

### 8.2.3 Operational Modes

A ONE\_SHOT control bit is used to set the SAR ADC conversion mode to either continuous or one conversion per SOF signal. DMA transfer of continuous samples, without CPU intervention, is supported.

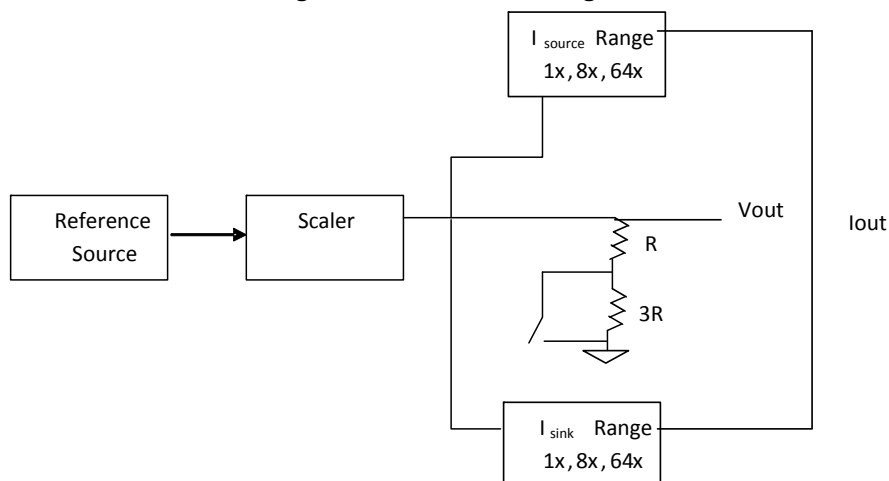
## 8.3 Comparators

The CY8C54 family of devices contains four comparators. Comparators have these features:

- Input offset factory trimmed to less than 15 mV
- Rail-to-rail common mode input range ( $V_{SSA}$  to  $V_{CCA}$ )
- Speed and power can be traded off by using one of three modes: fast, slow, or ultra low power
- Comparator outputs can be routed to look up tables to perform simple logic functions and then can also be routed to digital blocks
- The positive input of the comparators may be optionally passed through a low pass filter. Two filters are provided
- Comparator inputs can be connections to GPIO, DAC outputs and SC block outputs

### 8.3.1 Input and Output Interface

The positive and negative inputs to the comparators come from the analog global buses, the analog mux line, the analog local bus and precision reference through multiplexers. The output from each comparator could be routed to any of the two input LUTs. The output of that LUT is routed to the UDB Digital System Interface.

**Figure 8-10. DAC Block Diagram**


### 8.9.1 Current DAC

The current DAC (IDAC) can be configured for the ranges 0 to 31.875  $\mu\text{A}$ , 0 to 255  $\mu\text{A}$ , and 0 to 2.04 mA. The IDAC can be configured to source or sink current.

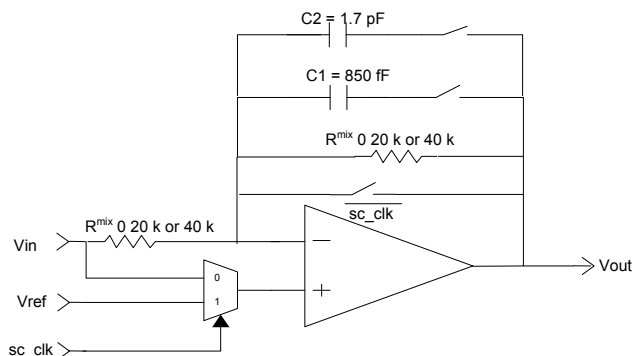
### 8.9.2 Voltage DAC

For the voltage DAC (VDAC), the current DAC output is routed through resistors. The two ranges available for the VDAC are 0 to 1.02 V and 0 to 4.08 V. In voltage mode any load connected to the output of a DAC should be purely capacitive (the output of the VDAC is not buffered).

## 8.10 Up/Down Mixer

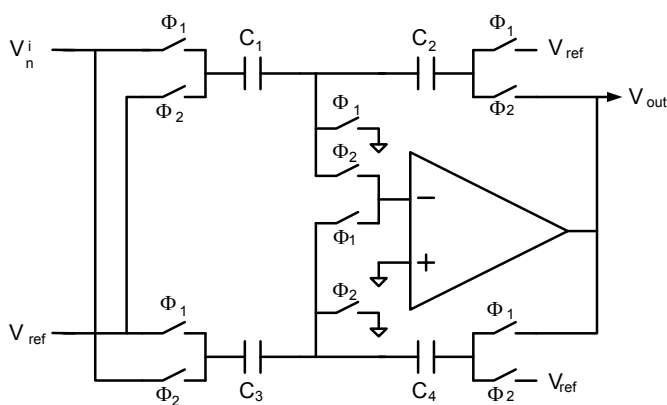
In continuous time mode, the SC/CT block components are used to build an up or down mixer. Any mixing application contains an input signal frequency and a local oscillator frequency. The polarity of the clock, Fclk, switches the amplifier between inverting or noninverting gain. The output is the product of the input and the switching function from the local oscillator, with frequency components at the local oscillator plus and minus the signal frequency ( $F_{\text{clk}} + F_{\text{in}}$  and  $F_{\text{clk}} - F_{\text{in}}$ ) and reduced-level frequency components at odd integer multiples of the local oscillator frequency. The local oscillator frequency is provided by the selected clock source for the mixer.

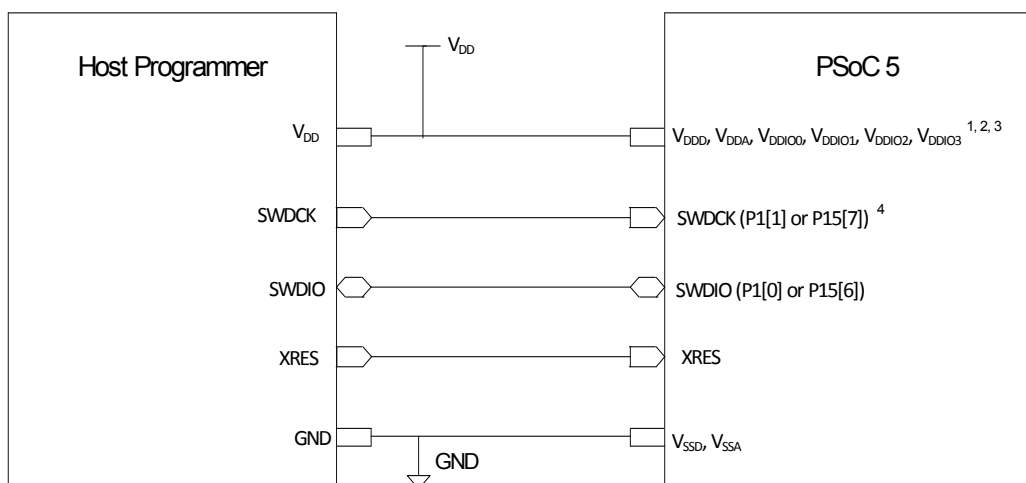
Continuous time up and down mixing works for applications with input signals and local oscillator frequencies up to 1 MHz.

**Figure 8-11. Mixer Configuration**


## 8.11 Sample and Hold

The main application for a sample and hold, is to hold a value stable while an ADC is performing a conversion. Some applications require multiple signals to be sampled simultaneously, such as for power calculations (V and I).

**Figure 8-12. Sample and Hold Topology ( $\Phi_1$  and  $\Phi_2$  are opposite phases of a clock)**


**Figure 9-1. SWD Interface Connections between PSoC 5 and Programmer**


<sup>1</sup> The voltage levels of the Host Programmer and the PSoC 5 voltage domains involved in programming should be the same. XRES pin is powered by V<sub>DDIO1</sub>. The USB SWD pins are powered by V<sub>DD</sub>. So for programming using the USB SWD pins with XRES pin, the V<sub>DD</sub>, V<sub>DDIO1</sub> of PSoC 5 should be at the same voltage level as Host V<sub>DD</sub>. Rest of PSoC 5 voltage domains (V<sub>DDA</sub>, V<sub>DDIO0</sub>, V<sub>DDIO2</sub>, V<sub>DDIO3</sub>) need not be at the same voltage level as host Programmer. The Port 1 SWD pins are powered by V<sub>DDIO1</sub>. So V<sub>DDIO1</sub> of PSoC 5 should be at same voltage level as host V<sub>DD</sub> for Port 1 SWD programming. Rest of PSoC 5 voltage domains (V<sub>DD</sub>, V<sub>DDA</sub>, V<sub>DDIO0</sub>, V<sub>DDIO2</sub>, V<sub>DDIO3</sub>) need not be at the same voltage level as host Programmer.

<sup>2</sup> V<sub>dda</sub> must be greater than or equal to all other power supplies (V<sub>ddd</sub>, V<sub>ddio</sub>'s) in PSoC 5.

<sup>3</sup> For Power cycle mode Programming, XRES pin is not required. But the Host programmer must have the capability to toggle power (V<sub>ddd</sub>, V<sub>dda</sub>, All V<sub>ddio</sub>'s) to PSoC 5. This may typically require external interface circuitry to toggle power which will depend on the programming setup. The power supplies can be brought up in any sequence, however, once stable, V<sub>DDA</sub> must be greater than or equal to all other supplies.

<sup>4</sup> When USB SWD pins are used for Programming, the P1[1] SWDCK pin must be externally connected to Ground using external pull-down resistor (around 100 K resistor). This is required for P15[7] SWDCK signal to be seen by PSoC 5's internal logic.

## 11.2 Device Level Specifications

Specifications are valid for  $-40\text{ }^{\circ}\text{C} \leq T_A \leq 85\text{ }^{\circ}\text{C}$  and  $T_J \leq 100\text{ }^{\circ}\text{C}$ , except where noted. Specifications are valid for 2.7 V to 5.5 V, except where noted.

### 11.2.1 Device Level Specifications

**Table 11-2. DC Specifications**

Parameter	Description	Conditions				Min	Typ	Max	Units
V <sub>DDA</sub>	Analog supply voltage and input to analog core regulator	Analog core regulator enabled				2.7	–	5.5	V
V <sub>DDD</sub>	Digital supply voltage relative to V <sub>SSD</sub>	Digital core regulator enabled				2.7	–	V <sub>DDA</sub> <sup>[17]</sup>	V
V <sub>DDIO</sub> <sup>[18]</sup>	I/O supply voltage relative to V <sub>SSIO</sub>					1.71	–	V <sub>DDA</sub> <sup>[17]</sup>	V
		Device Configuration	V <sub>DDX</sub>	F <sub>CPU</sub>	Temp				
I <sub>DD</sub> <sup>[19]</sup>	Active Mode	Only IMO and CPU clock enabled. CPU executing simple loop from cache	2.7 V to 5.5 V	6 MHz	–40 °C	–	2.2	3	mA
					25 °C	–	2.4	3.5	
					85 °C	–	2.8	3.5	
		IMO enabled, bus clock and CPU clock enabled. CPU executing complex program from flash	2.7 V to 5.5 V	3 MHz	–40 °C	–	3.4	4	
					25 °C	–	3.6	4.5	
					85 °C	–	4.2	5	
			6 MHz	–40 °C	–	5.6	6		
				25 °C	–	6	7		
				85 °C	–	6.6	7.5		
			12 MHz	–40 °C	–	10	11		
				25 °C	–	11	12		
				85 °C	–	12	13		
		24 MHz	–40 °C	–	17	19			
			25 °C	–	18	20			
			85 °C	–	20	22			
48 MHz	–40 °C	–	31	34					
	25 °C	–	33	35					
	85 °C	–	36	39					
63 MHz	–40 °C	–	36	39					
	25 °C	–	37	40					
	85 °C	–	40	44					

#### Notes

17.  $V_{DDD}$  and  $V_{DDA}$  must be brought up in synchronization with each other, that is, at the same rates and levels.  $V_{DDA}$  must be greater than or equal to all other supplies.
18. The  $V_{DDIO}$  supply voltage must be greater than the maximum analog voltage on the associated GPIO pins. Maximum analog voltage on GPIO pin  $\leq V_{DDIO} \leq V_{DDA}$ .
19. The current consumption of additional peripherals that are implemented only in programmed logic blocks can be found in their respective data sheets, available in PSoC Creator, the integrated design environment. To estimate total current, find CPU current at frequency of interest and add peripheral currents for your particular system from the device data sheet and component data sheets.

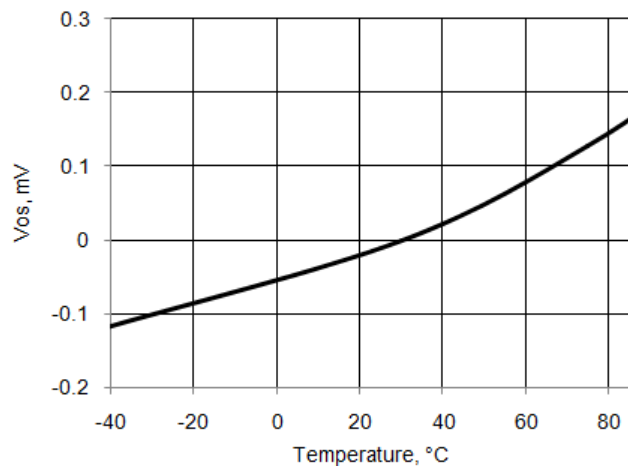
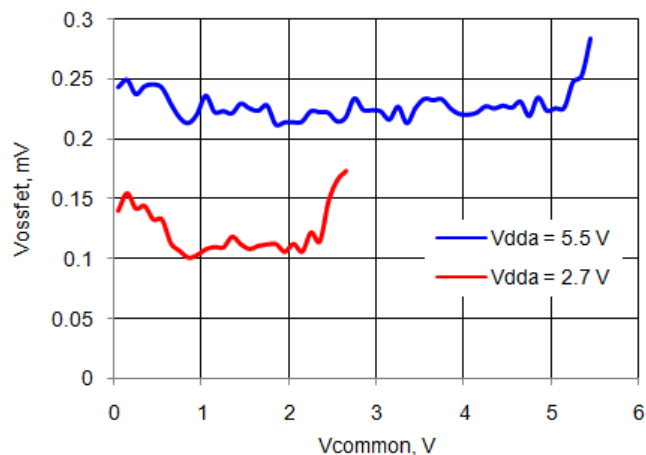
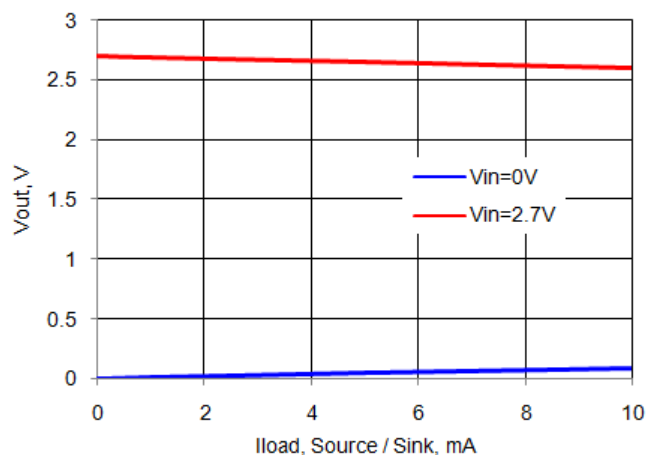
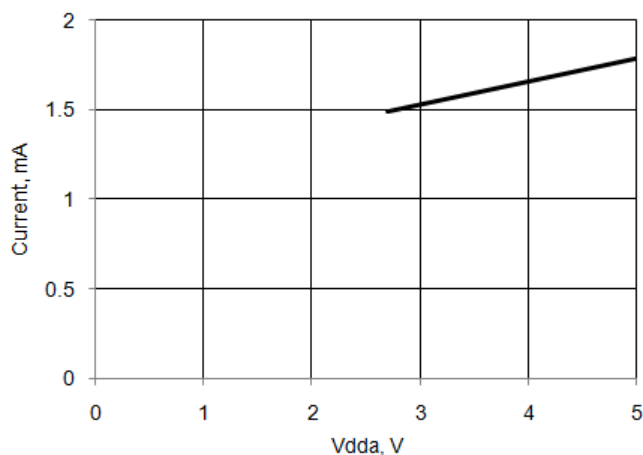
## 11.5 Analog Peripherals

Specifications are valid for  $-40\text{ }^{\circ}\text{C} \leq T_A \leq 85\text{ }^{\circ}\text{C}$  and  $T_J \leq 100\text{ }^{\circ}\text{C}$ , except where noted. Specifications are valid for 2.7 V to 5.5 V, except where noted.

### 11.5.1 Opamp

**Table 11-15. Opamp DC Specifications**

Parameter	Description	Conditions	Min	Typ	Max	Units
$V_I$	Input voltage range		$V_{SSA}$	–	$V_{DDA}$	V
$V_{OS}$	Input offset voltage	Operating temperature $> 70\text{ }^{\circ}\text{C}$	–	–	3	mV
		Operating temperature $-40\text{ }^{\circ}\text{C}$ to $70\text{ }^{\circ}\text{C}$	–	–	2	mV
$TCV_{OS}$	Input offset voltage drift with temperature		–	–	$\pm 30$	$\mu\text{V} / ^{\circ}\text{C}$
$Ge1$	Gain error, unity gain buffer mode	$R_{load} = 1\text{ k}\Omega$	–	–	$\pm 0.1$	%
$C_{IN}$	Input capacitance	Routing from pin	–	–	18	pF
$V_O$	Output voltage range	1 mA, source or sink	$V_{SSA} + 0.05$	–	$V_{DDA} - 0.05$	V
$I_{OUT}$	Output current, source or sink	$V_{SSA} + 500\text{ mV} \leq V_{out} \leq V_{DDA} - 500\text{ mV}$	10	–	–	mA
$I_{DD}$	Quiescent current	$V_{SSA} + 50\text{ mV} < V_{IN} < V_{DDA} - 50\text{ mV}$	–	1	2.5	mA
CMRR	Common mode rejection ratio		80	–	–	dB
PSRR	Power supply rejection ratio		75	–	–	dB

**Figure 11-17. Opamp Voffset vs Temperature,  $V_{DDA} = 5\text{ V}$** 

**Figure 11-19. Opamp Voffset vs Vcommon and  $V_{DDA}$ , 25 °C**

**Figure 11-18. Opamp Output Voltage vs Load Current, 25 °C,  $V_{DDA} = 2.7\text{ V}$** 

**Figure 11-20. Opamp Operating Current vs  $V_{DDA}$** 

**Table 11-16. Opamp AC Specifications**

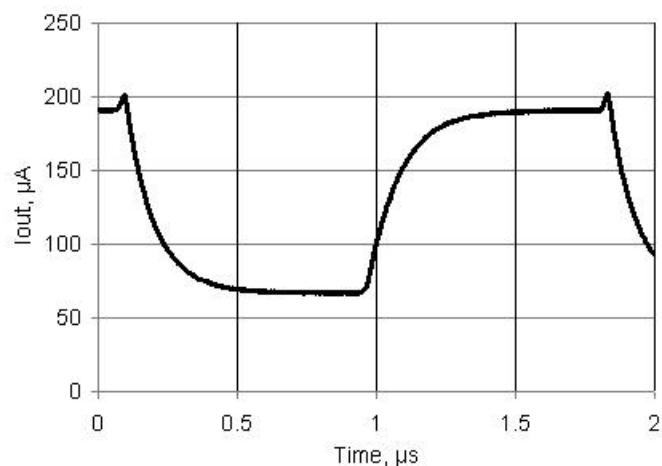
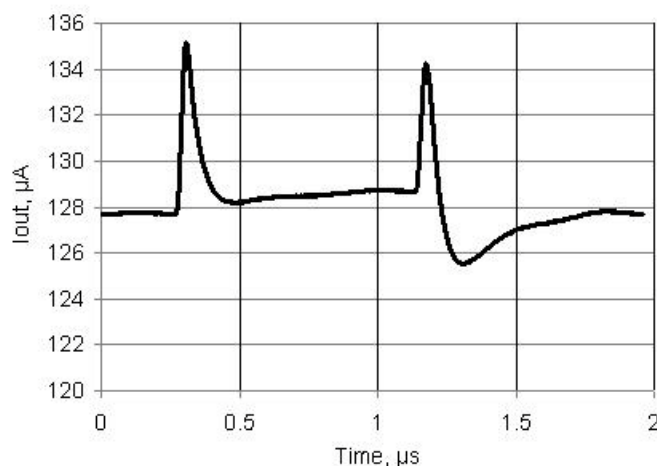
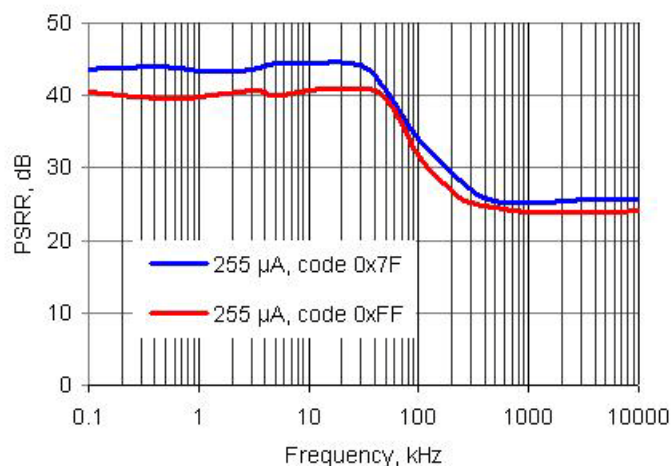
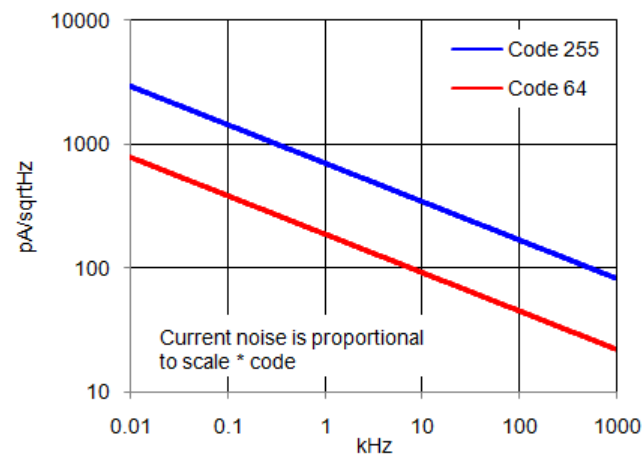
Parameter	Description	Conditions	Min	Typ	Max	Units
GBW	Gain-bandwidth product	200 pF load	3	—	—	MHz
SR	Slew rate, 20% - 80%	200 pF load	3	—	—	V/μs
$e_n$	Input noise density	$V_{dda} = 5\text{ V}$ , at 100 kHz	—	45	—	nV/sqrtHz

**Note**

35. Based on device characterization (Not production tested).

**Table 11-24. IDAC AC Specifications**

Parameter	Description	Conditions	Min	Typ	Max	Units
$F_{DAC}$	Update rate		–	–	5.5	Msp/s
$T_{SETTLE}$	Settling time to 0.5 LSB	Range = 31.875 $\mu$ A or 255 $\mu$ A, full scale transition, fast mode, 600 $\Omega$ 15-pF load	–	–	180	ns
	Current noise	Range = 255 $\mu$ A, source mode, fast mode, $V_{DDA} = 5$ V, 10 kHz	–	340	–	pA/sqrtHz

**Figure 11-40. IDAC Step Response, Codes 0x40 - 0xC0, 255  $\mu$ A Mode, Source Mode, Fast Mode,  $V_{DDA} = 5$  V**

**Figure 11-42. IDAC Glitch Response, Codes 0x7F - 0x80, 255  $\mu$ A Mode, Source Mode, Fast Mode,  $V_{DDA} = 5$  V**

**Figure 11-41. IDAC PSRR vs Frequency**

**Figure 11-43. IDAC Current Noise, 255  $\mu$ A Mode, Source Mode, Fast Mode,  $V_{DDA} = 5$  V**


## 13. Packaging

Table 13-1. Package Characteristics

Parameter	Description	Conditions	Min	Typ	Max	Units
T <sub>A</sub>	Operating ambient temperature		-40	25	85	°C
T <sub>J</sub>	Operating junction temperature		-40	–	100	°C
T <sub>ja</sub>	Package θJA (68-pin QFN)		–	15	–	°C/Watt
T <sub>ja</sub>	Package θJA (100-pin TQFP)		–	34	–	°C/Watt
T <sub>jc</sub>	Package θJC (68-pin QFN)		–	13	–	°C/Watt
T <sub>jc</sub>	Package θJC (100-pin TQFP)		–	10	–	°C/Watt

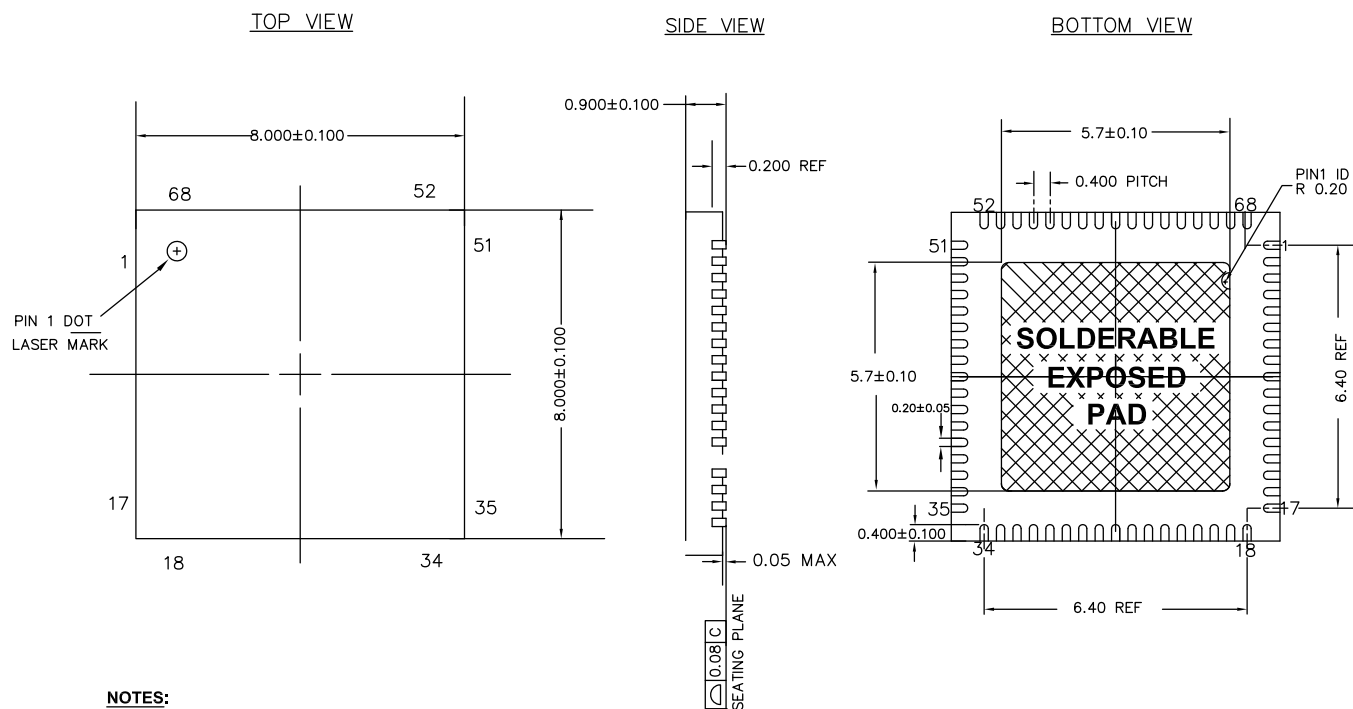
Table 13-2. Solder Reflow Peak Temperature

Package	Maximum Peak Temperature	Maximum Time at Peak Temperature
68-pin QFN	260 °C	30 seconds
100-pin TQFP	260 °C	30 seconds

Table 13-3. Package Moisture Sensitivity Level (MSL), IPC/JEDEC J-STD-2

Package	MSL
68-pin QFN	MSL 3
100-pin TQFP	MSL 3

Figure 13-1. 68-pin QFN 8 × 8 with 0.4 mm Pitch Package Outline (Sawn Version)



**NOTES:**

1. HATCH AREA IS SOLDERABLE EXPOSED METAL.
2. REFERENCE JEDEC#: MO-220
3. PACKAGE WEIGHT: 0.17g
4. ALL DIMENSIONS ARE IN MILLIMETERS

001-09618 \*D

## 14. Acronyms

**Table 14-1. Acronyms Used in this Document**

Acronym	Description
abus	analog local bus
ADC	analog-to-digital converter
AG	analog global
AHB	AMBA (advanced microcontroller bus architecture) high-performance bus, an ARM data transfer bus
ALU	arithmetic logic unit
AMUXBUS	analog multiplexer bus
API	application programming interface
APSR	application program status register
ARM®	advanced RISC machine, a CPU architecture
ATM	automatic thump mode
BW	bandwidth
CMRR	common-mode rejection ratio
CPU	central processing unit
CRC	cyclic redundancy check, an error-checking protocol
DAC	digital-to-analog converter, see also IDAC, VDAC
DFB	digital filter block
DIO	digital input/output, GPIO with only digital capabilities, no analog. See GPIO.
DMA	direct memory access, see also TD
DNL	differential nonlinearity, see also INL
DNU	do not use
DR	port write data registers
DSI	digital system interconnect
DWT	data watchpoint and trace
ECO	external crystal oscillator
EEPROM	electrically erasable programmable read-only memory
EMI	electromagnetic interference
EOC	end of conversion
EOF	end of frame
EPSR	execution program status register
ESD	electrostatic discharge
FIR	finite impulse response, see also IIR
FPB	flash patch and breakpoint
FS	full-speed
GPIO	general-purpose input/output, applies to a PSoC pin

**Table 14-1. Acronyms Used in this Document** (continued)

Acronym	Description
HVI	high-voltage interrupt, see also LVI, LVD
IC	integrated circuit
IDAC	current DAC, see also DAC, VDAC
IDE	integrated development environment
I <sup>2</sup> C, or IIC	Inter-Integrated Circuit, a communications protocol
IIR	infinite impulse response, see also FIR
ILO	internal low-speed oscillator, see also IMO
IMO	internal main oscillator, see also ILO
INL	integral nonlinearity, see also DNL
I/O	input/output, see also GPIO, DIO, SIO, USBIO
IPOR	initial power-on reset
IPSR	interrupt program status register
IRQ	interrupt request
ITM	instrumentation trace macrocell
LCD	liquid crystal display
LIN	Local Interconnect Network, a communications protocol.
LR	link register
LUT	lookup table
LVD	low-voltage detect, see also LVI
LVI	low-voltage interrupt, see also HVI
LVTTL	low-voltage transistor-transistor logic
MAC	multiply-accumulate
MCU	microcontroller unit
MISO	master-in slave-out
NC	no connect
NMI	nonmaskable interrupt
NRZ	non-return-to-zero
NVIC	nested vectored interrupt controller
NVL	nonvolatile latch, see also WOL
opamp	operational amplifier
PAL	programmable array logic, see also PLD
PC	program counter
PCB	printed circuit board
PGA	programmable gain amplifier
PHUB	peripheral hub
PHY	physical layer
PICU	port interrupt control unit
PLA	programmable logic array

## 16. Document Conventions

### 16.1 Units of Measure

Table 16-1. Units of Measure

Symbol	Unit of Measure
°C	degrees Celsius
dB	decibels
fF	femtofarads
Hz	hertz
KB	1024 bytes
kbps	kilobits per second
Khr	kilohours
kHz	kilohertz
kΩ	kilohms
ksps	kilosamples per second
LSB	least significant bit
Mbps	megabits per second
MHz	megahertz
MΩ	megaohms
Msps	megasamples per second
μA	microamperes
μF	microfarads
μH	microhenrys
μs	microseconds
μV	microvolts
μW	microwatts
mA	milliamperes
ms	milliseconds
mV	millivolts
nA	nanoamperes
ns	nanoseconds
nV	nanovolts
Ω	ohms
pF	picofarads
ppm	parts per million
ps	picoseconds
s	seconds
sps	samples per second
sqrtHz	square root of hertz
V	volts