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### Understanding [Embedded - Microprocessors](#)

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

### Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

#### Details

Product Status	Obsolete
Core Processor	PowerPC G4
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	1.4GHz
Co-Processors/DSP	Multimedia; SIMD
RAM Controllers	-
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	-
SATA	-
USB	-
Voltage - I/O	1.5V, 1.8V, 2.5V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	360-CLGA, FCCLGA
Supplier Device Package	360-FCCLGA (25x25)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/nxp-semiconductors/kmc7448vs1400nc">https://www.e-xfl.com/product-detail/nxp-semiconductors/kmc7448vs1400nc</a>

- Monitors all dispatched instructions and retires them in order
- Tracks unresolved branches and flushes instructions after a mispredicted branch
- Retires as many as three instructions per clock cycle
- Separate on-chip L1 instruction and data caches (Harvard architecture)
  - 32-Kbyte, eight-way set-associative instruction and data caches
  - Pseudo least-recently-used (PLRU) replacement algorithm
  - 32-byte (eight-word) L1 cache block
  - Physically indexed/physical tags
  - Cache write-back or write-through operation programmable on a per-page or per-block basis
  - Instruction cache can provide four instructions per clock cycle; data cache can provide four words per clock cycle
  - Caches can be disabled in software.
  - Caches can be locked in software.
  - MESI data cache coherency maintained in hardware
  - Separate copy of data cache tags for efficient snooping
  - Parity support on cache
  - No snooping of instruction cache except for **icbi** instruction
  - Data cache supports AltiVec LRU and transient instructions
  - Critical double- and/or quad-word forwarding is performed as needed. Critical quad-word forwarding is used for AltiVec loads and instruction fetches. Other accesses use critical double-word forwarding.
- Level 2 (L2) cache interface
  - On-chip, 1-Mbyte, eight-way set-associative unified instruction and data cache
  - Cache write-back or write-through operation programmable on a per-page or per-block basis
  - Parity support on cache tags
  - ECC or parity support on data
  - Error injection allows testing of error recovery software
- Separate memory management units (MMUs) for instructions and data
  - 52-bit virtual address, 32- or 36-bit physical address
  - Address translation for 4-Kbyte pages, variable-sized blocks, and 256-Mbyte segments
  - Memory programmable as write-back/write-through, caching-inhibited/caching-allowed, and memory coherency enforced/memory coherency not enforced on a page or block basis
  - Separate IBATs and DBATs (eight each) also defined as SPRs
  - Separate instruction and data translation lookaside buffers (TLBs)
    - Both TLBs are 128-entry, two-way set-associative and use an LRU replacement algorithm.
    - TLBs are hardware- or software-reloadable (that is, a page table search is performed in hardware or by system software on a TLB miss).

**Table 1. Microarchitecture Comparison (continued)**

Microarchitectural Specs	MPC7448	MPC7447A	MPC7447	MPC7445	MPC7441
<b>Execution Unit Timings (Latency-Throughput)</b>					
Aligned load (integer, float, vector)	3-1, 4-1, 3-1				
Misaligned load (integer, float, vector)	4-2, 5-2, 4-2				
L1 miss, L2 hit latency with ECC (data/instruction)	12/16	—			
L1 miss, L2 hit latency without ECC (data/instruction)	11/15	9/13			
SFX (add, sub, shift, rot, cmp, logicals)	1-1				
Integer multiply (32 × 8, 32 × 16, 32 × 32)	4-1, 4-1, 5-2				
Scalar float	5-1				
VSFX (vector simple)	1-1				
VCFX (vector complex)	4-1				
VFPU (vector float)	4-1				
VPER (vector permute)	2-1				
<b>MMUs</b>					
TLBs (instruction and data)	128-entry, 2-way				
Tablewalk mechanism	Hardware + software				
Instruction BATs/data BATs	8/8	8/8	8/8	8/8	4/4
<b>L1 I Cache/D Cache Features</b>					
Size	32K/32K				
Associativity	8-way				
Locking granularity	Way				
Parity on I cache	Word				
Parity on D cache	Byte				
Number of D cache misses (load/store)	5/2	5/1			
Data stream touch engines	4 streams				
<b>On-Chip Cache Features</b>					
Cache level	L2				
Size/associativity	1-Mbyte/ 8-way	512-Kbyte/8-way		256-Kbyte/8-way	
Access width	256 bits				
Number of 32-byte sectors/line	2	2			
Parity tag	Byte	Byte			
Parity data	Byte	Byte			
Data ECC	64-bit	—			
<b>Thermal Control</b>					
Dynamic frequency switching divide-by-two mode	Yes	Yes	No	No	No
Dynamic frequency switching divide-by-four mode	Yes	No	No	No	No
Thermal diode	Yes	Yes	No	No	No

Figure 3 provides the SYSCLK input timing diagram.

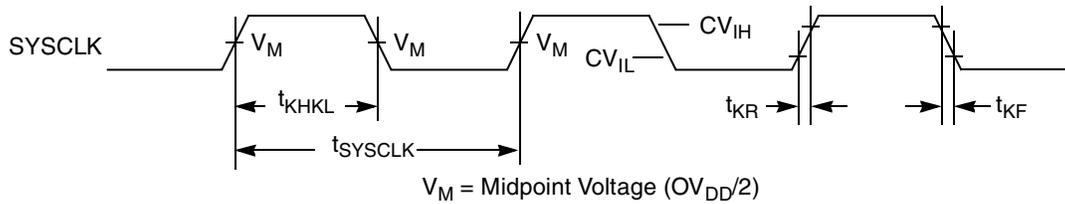


Figure 3. SYSCLK Input Timing Diagram

## 5.2.2 Processor Bus AC Specifications

Table 9 provides the processor bus AC timing specifications for the MPC7448 as defined in Figure 4 and Figure 5.

Table 9. Processor Bus AC Timing Specifications<sup>1</sup>

At recommended operating conditions. See Table 4.

Parameter	Symbol <sup>2</sup>	All Speed Grades		Unit	Notes
		Min	Max		
Input setup times: A[0:35], AP[0:4] D[0:63], DP[0:7] AACK, ARTRY, BG, CKSTP_IN, DBG, DTI[0:3], GBL, TT[0:4], QACK, TA, TBEN, TEA, TS, EXT_QUAL, PMON_IN, SHD[0:1] BMODE[0:1], BVSEL[0:1]	$t_{AVKH}$ $t_{DVKH}$ $t_{IVKH}$ $t_{MVKH}$	1.5 1.5 1.5 1.5	— — — —	ns	— — — 8
Input hold times: A[0:35], AP[0:4] D[0:63], DP[0:7] AACK, ARTRY, BG, CKSTP_IN, DBG, DTI[0:3], GBL, TT[0:4], QACK, TA, TBEN, TEA, TS, EXT_QUAL, PMON_IN, SHD[0:1] BMODE[0:1], BVSEL[0:1]	$t_{AXKH}$ $t_{DXKH}$ $t_{IXKH}$ $t_{MXKH}$	0 0 0 0	— — — —	ns	— — — 8
Output valid times: A[0:35], AP[0:4] D[0:63], DP[0:7] BR, CI, DRDY, GBL, HIT, PMON_OUT, QREQ, TBST, TSIZ[0:2], TT[0:4], WT TS ARTRY, SHD[0:1]	$t_{KHAV}$ $t_{KHVD}$ $t_{KHOV}$ $t_{KHTSV}$ $t_{KHARV}$	— — — — —	1.8 1.8 1.8 1.8 1.8	ns	
Output hold times: A[0:35], AP[0:4] D[0:63], DP[0:7] BR, CI, DRDY, GBL, HIT, PMON_OUT, QREQ, TBST, TSIZ[0:2], TT[0:4], WT TS ARTRY, SHD[0:1]	$t_{KHAX}$ $t_{KHDX}$ $t_{KHOX}$ $t_{KHTSX}$ $t_{KHARX}$	0.5 0.5 0.5 0.5 0.5	— — — — —	ns	
SYSCLK to output enable	$t_{KHOE}$	0.5	—	ns	5

### 5.2.3 IEEE Std. 1149.1 AC Timing Specifications

Table 10 provides the IEEE Std. 1149.1 (JTAG) AC timing specifications as defined in Figure 8 through Figure 11.

**Table 10. JTAG AC Timing Specifications (Independent of SYSCLK)<sup>1</sup>**

At recommended operating conditions. See Table 4.

Parameter	Symbol	Min	Max	Unit	Notes
TCK frequency of operation	$f_{TCLK}$	0	33.3	MHz	
TCK cycle time	$t_{TCLK}$	30	—	ns	
TCK clock pulse width measured at 1.4 V	$t_{HJL}$	15	—	ns	
TCK rise and fall times	$t_{JR}$ and $t_{JF}$	—	2	ns	
$\overline{TRST}$ assert time	$t_{TRST}$	25	—	ns	2
Input setup times: Boundary-scan data TMS, TDI	$t_{DVJH}$ $t_{IVJH}$	4 0	— —	ns	3
Input hold times: Boundary-scan data TMS, TDI	$t_{DXJH}$ $t_{IXJH}$	20 25	— —	ns	3
Valid times: Boundary-scan data TDO	$t_{JLDV}$ $t_{JLOV}$	4 4	20 25	ns	4
Output hold times: Boundary-scan data TDO	$t_{JLDX}$ $t_{JLOX}$	30 30	— —	ns	4
TCK to output high impedance: Boundary-scan data TDO	$t_{JLDZ}$ $t_{JLOZ}$	3 3	19 9	ns	4, 5

**Notes:**

- All outputs are measured from the midpoint voltage of the falling/rising edge of TCLK to the midpoint of the signal in question. The output timings are measured at the pins. All output timings assume a purely resistive 50- $\Omega$  load (see Figure 7). Time-of-flight delays must be added for trace lengths, vias, and connectors in the system.
- $\overline{TRST}$  is an asynchronous level sensitive signal. The time is for test purposes only.
- Non-JTAG signal input timing with respect to TCK.
- Non-JTAG signal output timing with respect to TCK.
- Guaranteed by design and characterization.

Figure 11 provides the test access port timing diagram.

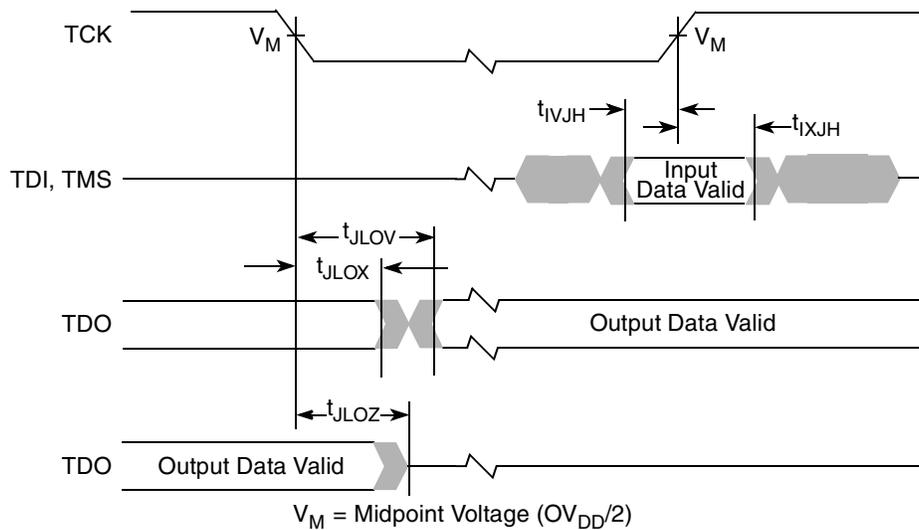


Figure 11. Test Access Port Timing Diagram

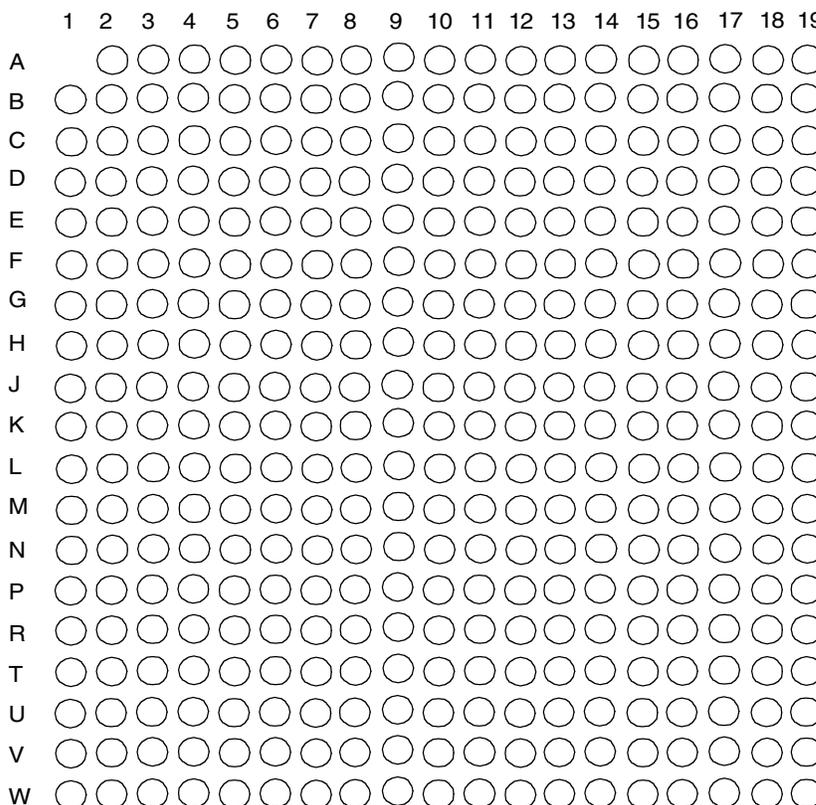
### 5.3 Voltage and Frequency Derating

Voltage and frequency derating is no longer supported for part numbers described by this document beginning with datecode 0613. (See [Section 11, “Part Numbering and Marking,”](#) for information on date code markings.) It is supported by some MPC7448 part numbers which target low-power applications; see [Section 11.2, “Part Numbers Not Fully Addressed by This Document”](#) and the referenced MPC7448 Hardware Specification Addenda for more information on these low-power devices. For those devices which previously supported this feature, information has been archived in the *Chip Errata for the MPC7448* (document order no. MPC7448CE).

# 6 Pin Assignments

Figure 12 (in Part A) shows the pinout of the MPC7448, 360 high coefficient of thermal expansion ceramic ball grid array (HCTE) package as viewed from the top surface. Part B shows the side profile of the HCTE package to indicate the direction of the top surface view.

**Part A**



Not to Scale

**Part B**

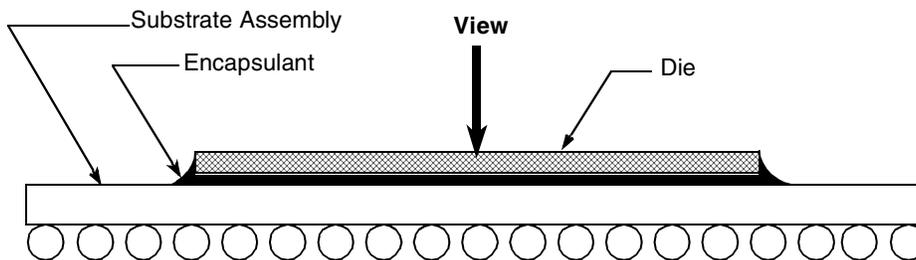


Figure 12. Pinout of the MPC7448, 360 HCTE Package as Viewed from the Top Surface

**Table 11. Pinout Listing for the MPC7448, 360 HCTE Package**

Signal Name	Pin Number	Active	I/O	Notes
A[0:35]	E11, H1, C11, G3, F10, L2, D11, D1, C10, G2, D12, L3, G4, T2, F4, V1, J4, R2, K5, W2, J2, K4, N4, J3, M5, P5, N3, T1, V2, U1, N5, W1, B12, C4, G10, B11	High	I/O	2
$\overline{\text{AACK}}$	R1	Low	Input	
AP[0:4]	C1, E3, H6, F5, G7	High	I/O	2
$\overline{\text{ARTRY}}$	N2	Low	I/O	3
AV <sub>DD</sub>	A8	—	Input	
$\overline{\text{BG}}$	M1	Low	Input	
$\overline{\text{BMODE0}}$	G9	Low	Input	4
$\overline{\text{BMODE1}}$	F8	Low	Input	5
$\overline{\text{BR}}$	D2	Low	Output	
BVSEL0	B7	High	Input	1, 6
BVSEL1	E10	High	Input	1, 20
$\overline{\text{CI}}$	J1	Low	Output	
$\overline{\text{CKSTP\_IN}}$	A3	Low	Input	
$\overline{\text{CKSTP\_OUT}}$	B1	Low	Output	
CLK_OUT	H2	High	Output	
D[0:63]	R15, W15, T14, V16, W16, T15, U15, P14, V13, W13, T13, P13, U14, W14, R12, T12, W12, V12, N11, N10, R11, U11, W11, T11, R10, N9, P10, U10, R9, W10, U9, V9, W5, U6, T5, U5, W7, R6, P7, V6, P17, R19, V18, R18, V19, T19, U19, W19, U18, W17, W18, T16, T18, T17, W3, V17, U4, U8, U7, R7, P6, R8, W8, T8	High	I/O	
$\overline{\text{DBG}}$	M2	Low	Input	
$\overline{\text{DFS2}}$	A12	Low	Input	20, 21
$\overline{\text{DFS4}}$	B6	Low	Input	12, 20, 21
DP[0:7]	T3, W4, T4, W9, M6, V3, N8, W6	High	I/O	
$\overline{\text{DRDY}}$	R3	Low	Output	7
DTI[0:3]	G1, K1, P1, N1	High	Input	8
EXT_QUAL	A11	High	Input	9
$\overline{\text{GBL}}$	E2	Low	I/O	
GND	B5, C3, D6, D13, E17, F3, G17, H4, H7, H9, H11, H13, J6, J8, J10, J12, K7, K3, K9, K11, K13, L6, L8, L10, L12, M4, M7, M9, M11, M13, N7, P3, P9, P12, R5, R14, R17, T7, T10, U3, U13, U17, V5, V8, V11, V15	—	—	
GND	A17, A19, B13, B16, B18, E12, E19, F13, F16, F18, G19, H18, J14, L14, M15, M17, M19, N14, N16, P15, P19	—	—	15
GND_SENSE	G12, N13	—	—	19
$\overline{\text{HIT}}$	B2	Low	Output	7
$\overline{\text{HRESET}}$	D8	Low	Input	
$\overline{\text{INT}}$	D4	Low	Input	
L1_TSTCLK	G8	High	Input	9
L2_TSTCLK	B3	High	Input	10

**Table 11. Pinout Listing for the MPC7448, 360 HCTE Package (continued)**

Signal Name	Pin Number	Active	I/O	Notes
$\overline{\text{LVRAM}}$	B10	—	—	12, 20, 22
NC (no connect)	A6, A14, A15, B14, B15, C14, C15, C16, C17, C18, C19, D14, D15, D16, D17, D18, D19, E14, E15, F14, F15, G14, G15, H15, H16, J15, J16, J17, J18, J19, K15, K16, K17, K18, K19, L15, L16, L17, L18, L19	—	—	11
$\overline{\text{LSSD\_MODE}}$	E8	Low	Input	6, 12
$\overline{\text{MCP}}$	C9	Low	Input	
$\text{OV}_{\text{DD}}$	B4, C2, C12, D5, F2, H3, J5, K2, L5, M3, N6, P2, P8, P11, R4, R13, R16, T6, T9, U2, U12, U16, V4, V7, V10, V14	—	—	
$\text{OVDD\_SENSE}$	E18, G18	—	—	16
$\text{PLL\_CFG}[0:4]$	B8, C8, C7, D7, A7	High	Input	
$\text{PLL\_CFG}[5]$	D10	High	Input	9, 20
$\overline{\text{PMON\_IN}}$	D9	Low	Input	13
$\overline{\text{PMON\_OUT}}$	A9	Low	Output	
$\overline{\text{QACK}}$	G5	Low	Input	
$\overline{\text{QREQ}}$	P4	Low	Output	
$\overline{\text{SHD}}[0:1]$	E4, H5	Low	I/O	3
$\overline{\text{SMI}}$	F9	Low	Input	
$\overline{\text{SRESET}}$	A2	Low	Input	
$\text{SYSCLK}$	A10	—	Input	
$\overline{\text{TA}}$	K6	Low	Input	
TBEN	E1	High	Input	
$\overline{\text{TBST}}$	F11	Low	Output	
TCK	C6	High	Input	
TDI	B9	High	Input	6
TDO	A4	High	Output	
$\overline{\text{TEA}}$	L1	Low	Input	
TEMP_ANODE	N18	—	—	17
TEMP_CATHODE	N19	—	—	17
TMS	F1	High	Input	6
$\overline{\text{TRST}}$	A5	Low	Input	6, 14
$\overline{\text{TS}}$	L4	Low	I/O	3
$\text{TSIZ}[0:2]$	G6, F7, E7	High	Output	
$\text{TT}[0:4]$	E5, E6, F6, E9, C5	High	I/O	
$\overline{\text{WT}}$	D3	Low	Output	
$\text{V}_{\text{DD}}$	H8, H10, H12, J7, J9, J11, J13, K8, K10, K12, K14, L7, L9, L11, L13, M8, M10, M12	—	—	
$\text{V}_{\text{DD}}$	A13, A16, A18, B17, B19, C13, E13, E16, F12, F17, F19, G11, G16, H14, H17, H19, M14, M16, M18, N15, N17, P16, P18	—	—	15

## 8.2 Mechanical Dimensions for the MPC7448, 360 HCTE BGA

Figure 13 provides the mechanical dimensions and bottom surface nomenclature for the MPC7448, 360 HCTE BGA package.

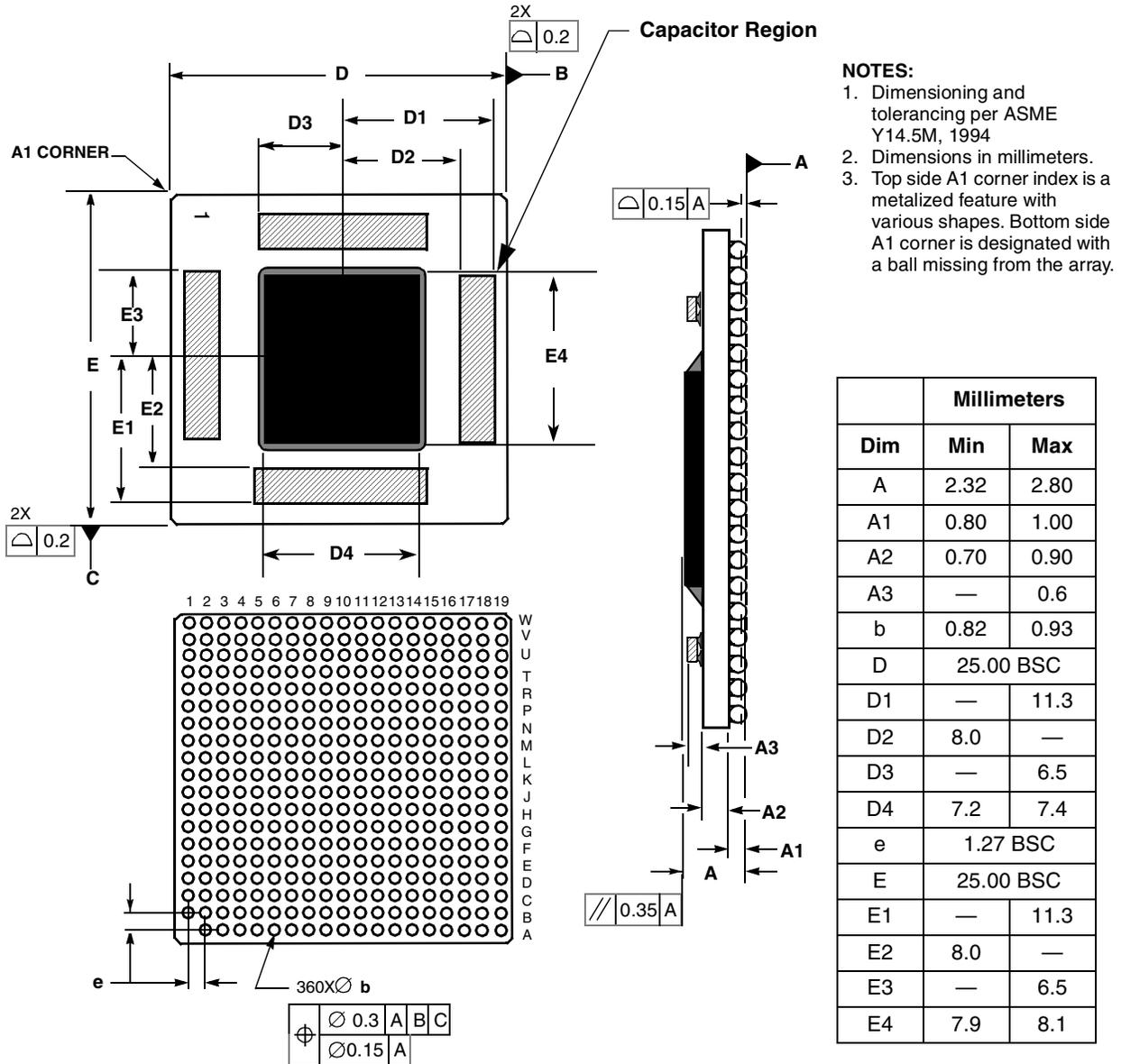


Figure 13. Mechanical Dimensions and Bottom Surface Nomenclature for the MPC7448, 360 HCTE BGA Package

## 8.4 Mechanical Dimensions for the MPC7448, 360 HCTE LGA

Figure 13 provides the mechanical dimensions and bottom surface nomenclature for the MPC7448, 360 HCTE LGA package.

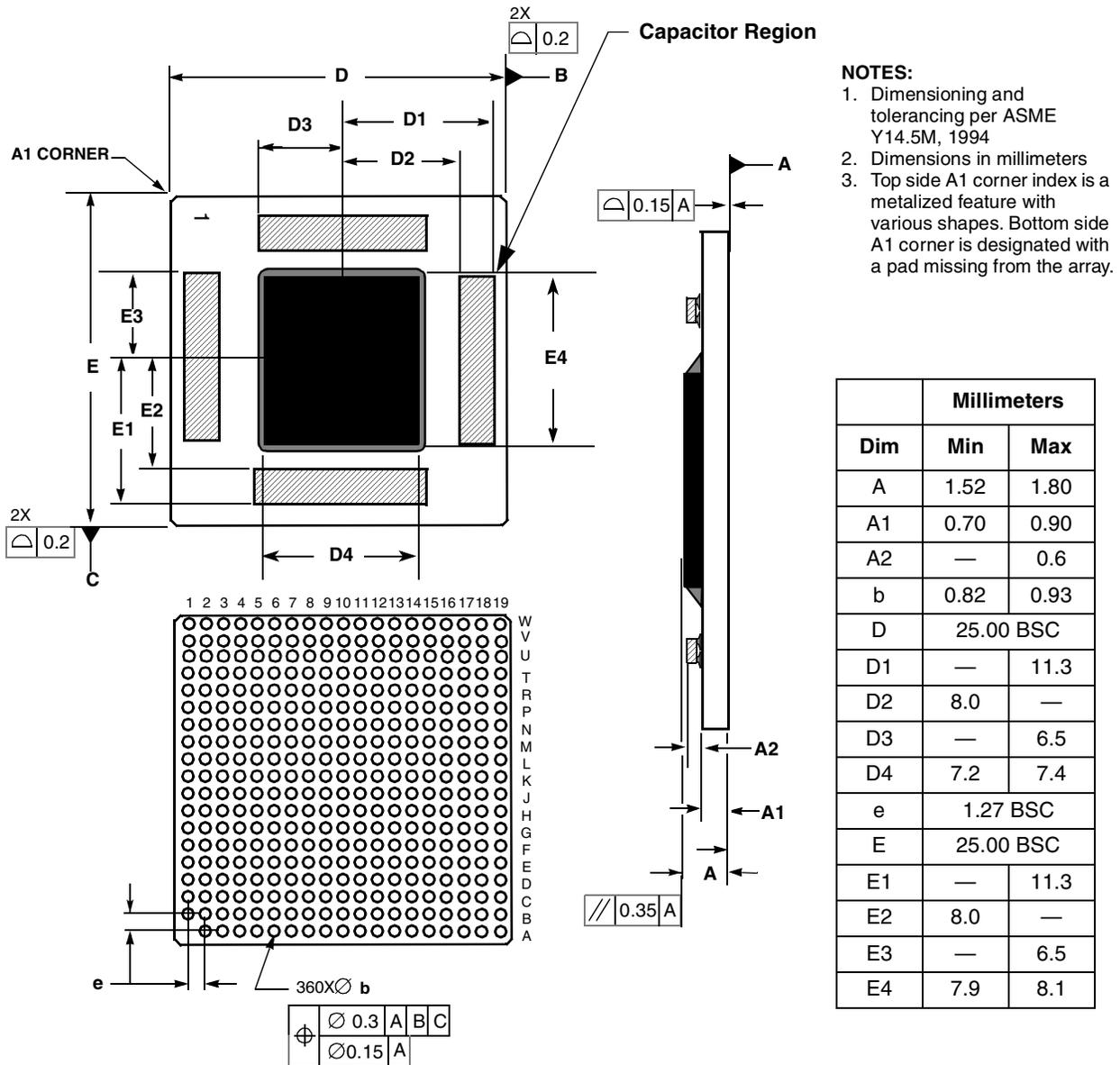


Figure 14. Mechanical Dimensions and Bottom Surface Nomenclature for the MPC7448, 360 HCTE LGA Package

## 8.6 Mechanical Dimensions for the MPC7448, 360 HCTE RoHS-Compliant BGA

Figure 13 provides the mechanical dimensions and bottom surface nomenclature for the MPC7448, 360 HCTE BGA package with RoHS-compliant lead-free spheres.

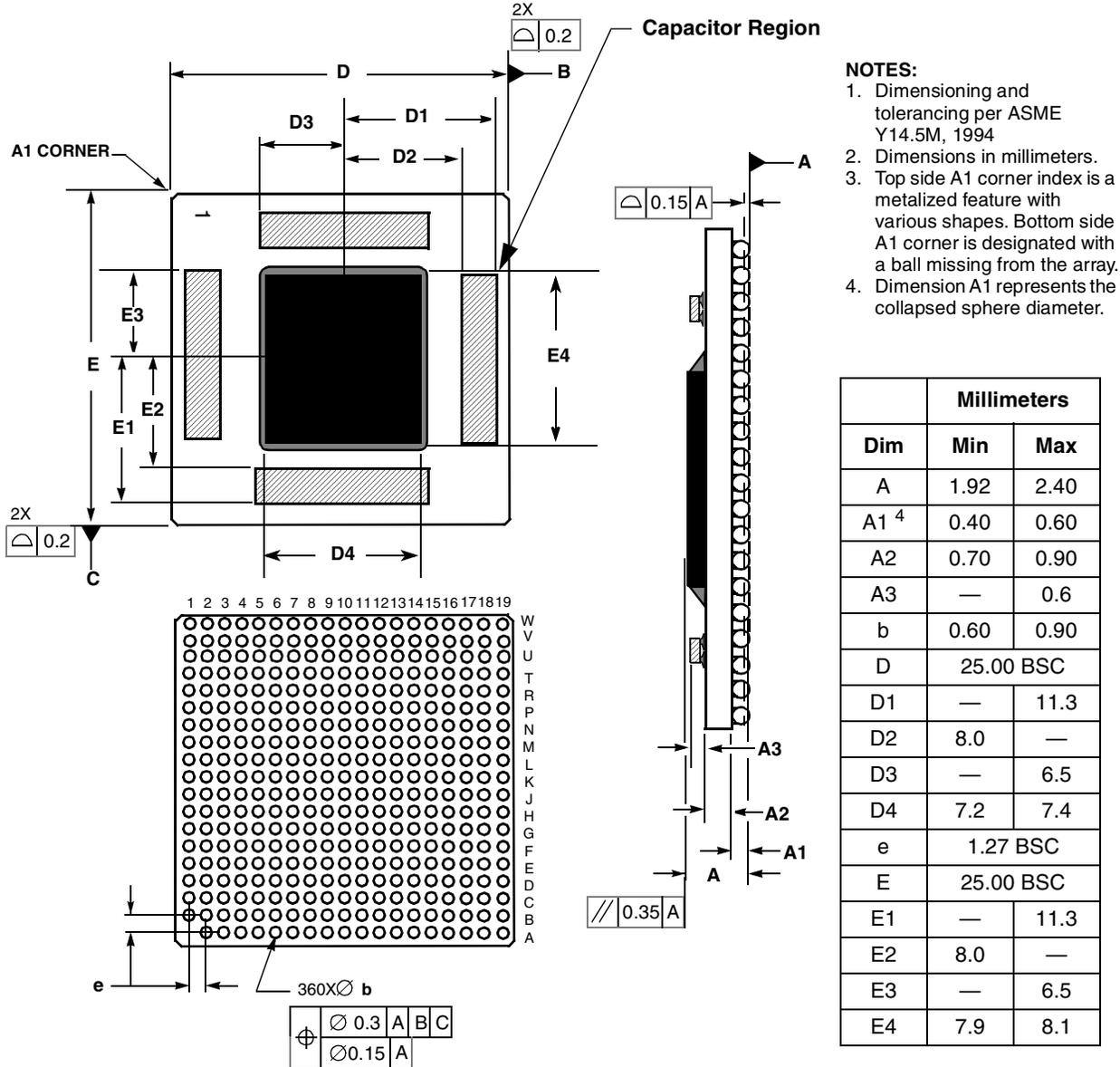
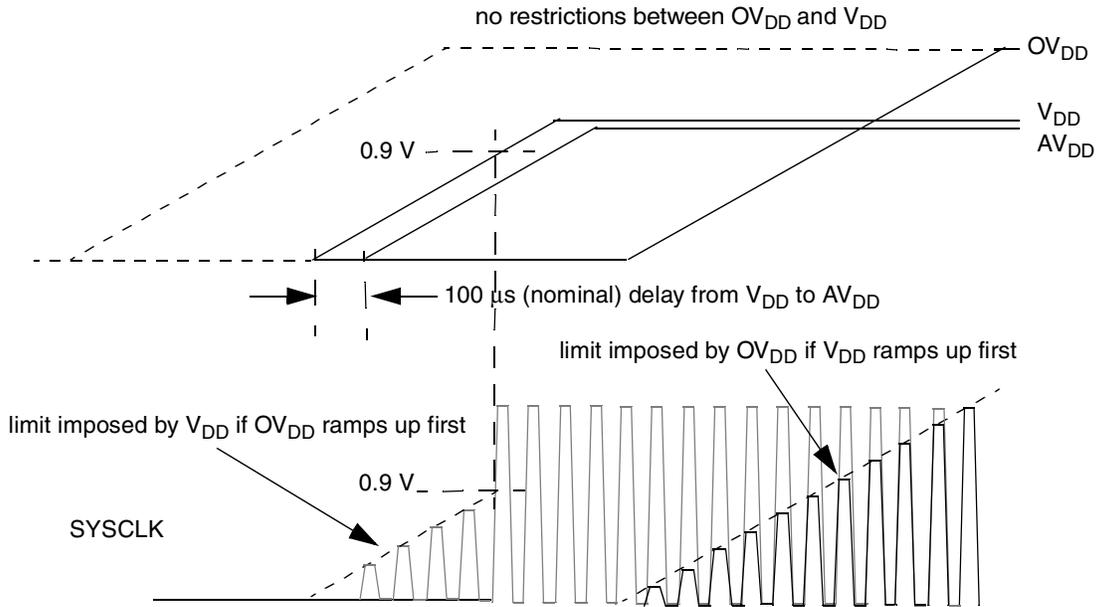


Figure 15. Mechanical Dimensions and Bottom Surface Nomenclature for the MPC7448, 360 HCTE RoHS-Compliant BGA Package

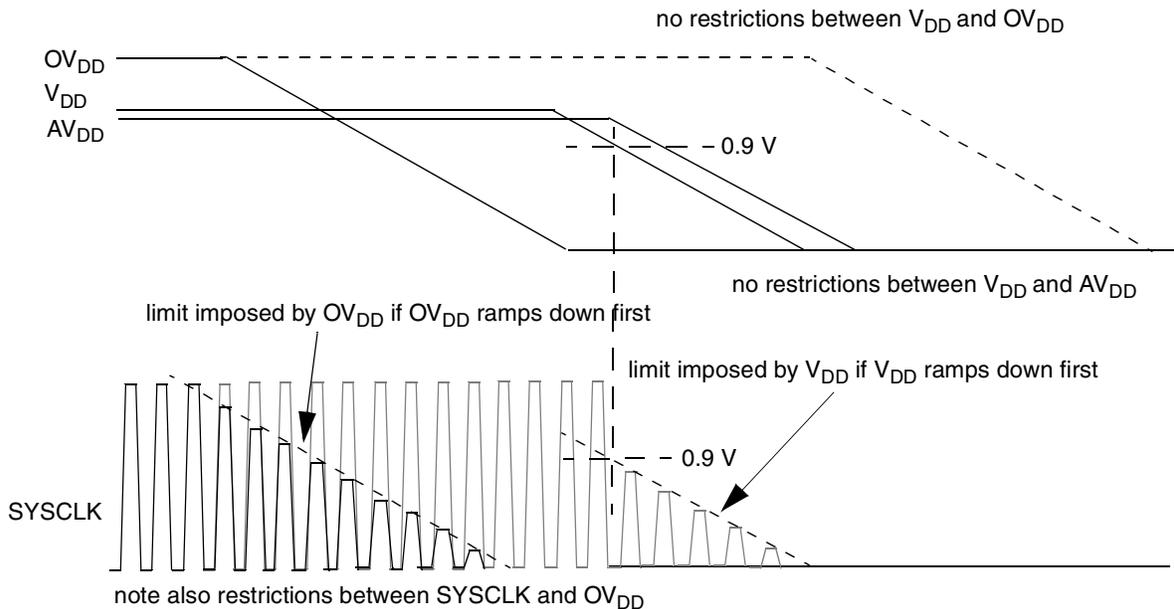
These requirements are shown graphically in [Figure 16](#).



**Figure 16. MPC7448 Power Up Sequencing Requirements**

Certain stipulations also apply to the manner in which the power rails of the MPC7448 power down, as follows:

- $OV_{DD}$  may ramp down any time before or after  $V_{DD}$ .
- The voltage at the SYSCLK input must not exceed  $V_{DD}$  once  $V_{DD}$  has ramped down below 0.9 V.
- The voltage at the SYSCLK input must not exceed  $OV_{DD}$  by more 20% during transients (see overshoot/undershoot specifications in [Figure 2](#)) or 0.3 V DC (see [Table 2](#)) at any time.



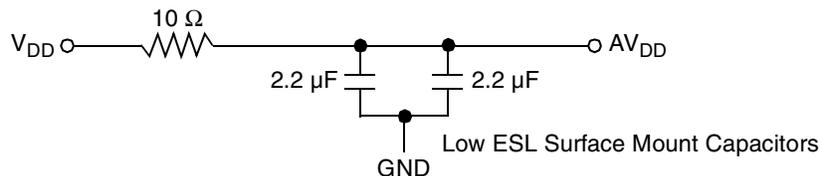
**Figure 17. MPC7448 Power Down Sequencing Requirements**

There is no requirement regarding  $AV_{DD}$  during power down, but it is recommended that  $AV_{DD}$  track  $V_{DD}$  within the RC time constant of the PLL filter circuit described in [Section 9.2.2, “PLL Power Supply Filtering”](#) (nominally 100  $\mu$ s).

## 9.2.2 PLL Power Supply Filtering

The  $AV_{DD}$  power signal is provided on the MPC7448 to provide power to the clock generation PLL. To ensure stability of the internal clock, the power supplied to the  $AV_{DD}$  input signal should be filtered of any noise in the 500-KHz to 10-MHz resonant frequency range of the PLL. The circuit shown in [Figure 18](#) using surface mount capacitors with minimum effective series inductance (ESL) is strongly recommended. In addition to filtering noise from the  $AV_{DD}$  input, it also provides the required delay between  $V_{DD}$  and  $AV_{DD}$  as described in [Section 9.2.1, “Power Supply Sequencing.”](#)

The circuit should be placed as close as possible to the  $AV_{DD}$  pin to minimize noise coupled from nearby circuits. It is often possible to route directly from the capacitors to the  $AV_{DD}$  pin, which is on the periphery of the device footprint.



**Figure 18. PLL Power Supply Filter Circuit**

## 9.2.3 Transient Specifications

To ensure the long-term reliability of the device, the MPC7448 requires that transients on the core power rail ( $V_{DD}$ ) be constrained. The recommended operating voltage specifications provided in [Table 4](#) are DC specifications. That is, the device may be operated continuously with  $V_{DD}$  within the specified range without adversely affecting the device's reliability. Excursions above the stated recommended operation range, including overshoot during power-up, can impact the long-term reliability of the device. Excursions are described by their amplitude and duration. Duration is defined as the time period during which the  $V_{DD}$  power plane, as measured at the VDD\_SENSE pins, will be within a specific voltage range, expressed as percentage of the total time the device will be powered up over the device lifetime. In practice, the period over which transients are measured can be any arbitrary period of time that accurately represents the expected range of processor and system activity. The voltage ranges and durations for normal operation and transients are described in [Table 14](#).

**Table 14. VDD Power Supply Transient Specifications**

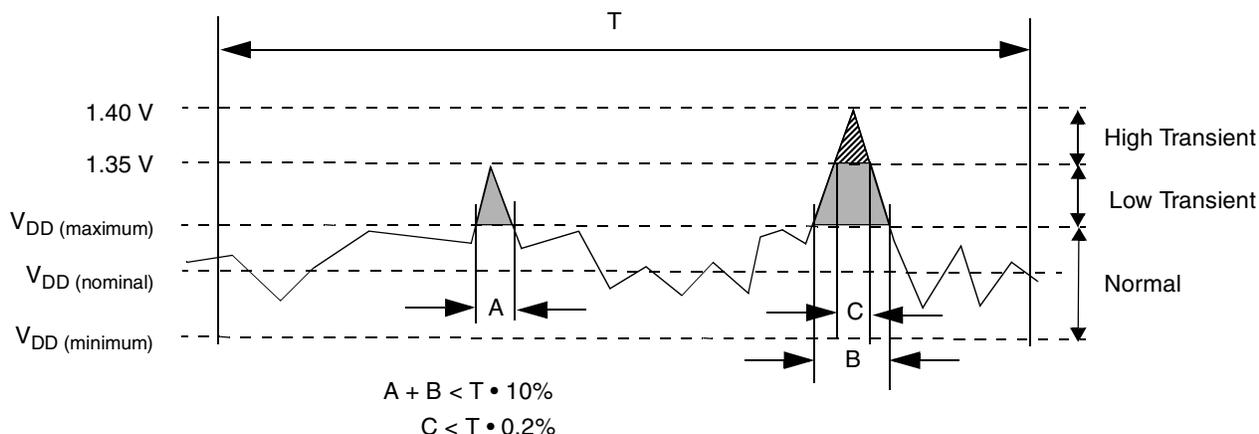
At recommended operating temperatures. See [Table 4](#).

Voltage Region	Voltage Range (V)		Permitted Duration <sup>1</sup>	Notes
	Min	Max		
Normal	$V_{DD}$ minimum	$V_{DD}$ maximum	100%	2
Low Transient	$V_{DD}$ maximum	1.35 V	10%	2, 3
High Transient	1.35 V	1.40 V	0.2%	4

**Notes:**

1. Permitted duration is defined as the percentage of the total time the device is powered on that the  $V_{DD}$  power supply voltage may exist within the specified voltage range.
2. See [Table 4](#) for nominal  $V_{DD}$  specifications.
3. To simplify measurement, excursions into the High Transient region are included in this duration.
4. Excursions above the absolute maximum rating of 1.4 V are not permitted; see [Table 2](#).

Note that, to simplify transient measurements, the duration of the excursion into the High Transient region is also included in the Low Transient duration, so that only the time the voltage is above each threshold must be considered. [Figure 19](#) shows an example of measuring voltage transients.



**Figure 19. Voltage Transient Example**

## 9.2.4 Decoupling Recommendations

Due to the MPC7448 dynamic power management feature, large address and data buses, and high operating frequencies, the MPC7448 can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. This noise must be prevented from reaching other components in the MPC7448 system, and the MPC7448 itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer use sufficient decoupling capacitors, typically one capacitor for every  $V_{DD}$  pin, and a similar amount for the  $OV_{DD}$  pins, placed as close as possible to the power pins of the MPC7448. It is also recommended that these decoupling capacitors receive their power from separate  $V_{DD}$ ,  $OV_{DD}$ , and GND power planes in the PCB, using short traces to minimize inductance.

These capacitors should have a value of 0.01 or 0.1  $\mu\text{F}$ . Only ceramic surface mount technology (SMT) capacitors should be used to minimize lead inductance. Orientations where connections are made along the length of the part, such as 0204, are preferable but not mandatory. Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993) and contrary to previous recommendations for decoupling Freescale microprocessors, multiple small capacitors of equal value are recommended over using multiple values of capacitance.

In addition, it is recommended that there be several bulk storage capacitors distributed around the PCB, feeding the  $V_{DD}$  and  $OV_{DD}$  planes, to enable quick recharging of the smaller chip capacitors. These bulk capacitors should have a low equivalent series resistance (ESR) rating to ensure the quick response time necessary. They should also be connected to the power and ground planes through two vias to minimize inductance. Suggested bulk capacitors are 100–330  $\mu\text{F}$  (AVX TPS tantalum or Sanyo OSCON).

## 9.3 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. Unless otherwise noted, unused active low inputs should be tied to  $OV_{DD}$  and unused active high inputs should be connected to GND. All NC (no connect) signals must remain unconnected.

Power and ground connections must be made to all external  $V_{DD}$ ,  $OV_{DD}$ , and GND pins in the MPC7448. For backward compatibility with the MPC7447, MPC7445, and MP7441, or for migrating a system originally designed for one of these devices to the MPC7448, the new power and ground signals (formerly NC, see [Table 11](#)) may be left unconnected if the core frequency is 1 GHz or less. Operation above 1 GHz requires that these additional power and ground signals be connected, and it is strongly recommended that all new designs include the additional connections. See also [Section 7, “Pinout Listings,”](#) for additional information.

The MPC7448 provides  $VDD\_SENSE$ ,  $OVDD\_SENSE$ , and  $GND\_SENSE$  pins. These pins connect directly to the power/ground planes in the device package and are intended to allow an external device to measure the voltage present on the  $V_{DD}$ ,  $OV_{DD}$  and GND planes in the device package. The most common use for these signals is as a feedback signal to a power supply regulator to allow it to compensate for board losses and supply the correct voltage at the device. (Note that all voltage parameters are specified at the pins of the device.) If not used for this purpose, it is recommended that these signals be connected to test points that can be used in the event that an accurate measurement of the voltage at the device is needed during system debug. Otherwise, these signals should be connected to the appropriate power/ground planes on the circuit board or left unconnected.

## 9.4 Output Buffer DC Impedance

The MPC7448 processor bus drivers are characterized over process, voltage, and temperature. To measure  $Z_0$ , an external resistor is connected from the chip pad to  $OV_{DD}$  or GND. The value of each resistor is varied until the pad voltage is  $OV_{DD}/2$ . Figure 20 shows the driver impedance measurement.

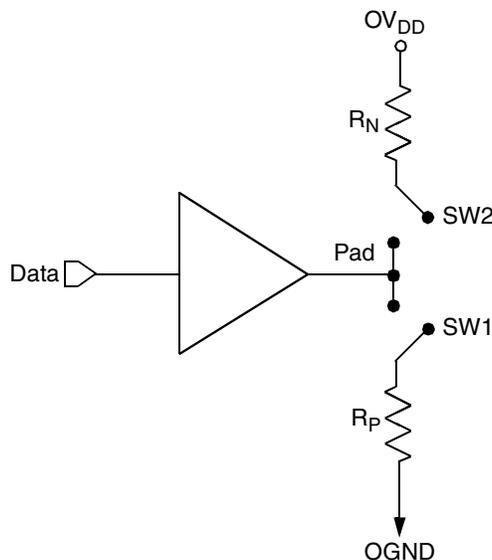


Figure 20. Driver Impedance Measurement

The output impedance is the average of two components—the resistances of the pull-up and pull-down devices. When data is held low, SW2 is closed (SW1 is open), and  $R_N$  is trimmed until the voltage at the pad equals  $OV_{DD}/2$ .  $R_N$  then becomes the resistance of the pull-down devices. When data is held high, SW1 is closed (SW2 is open), and  $R_P$  is trimmed until the voltage at the pad equals  $OV_{DD}/2$ .  $R_P$  then becomes the resistance of the pull-up devices.  $R_P$  and  $R_N$  are designed to be close to each other in value. Then,  $Z_0 = (R_P + R_N)/2$ .

Table 15 summarizes the signal impedance results. The impedance increases with junction temperature and is relatively unaffected by bus voltage.

Table 15. Impedance Characteristics

At recommended operating conditions. See Table 4

Impedance		Processor Bus	Unit
$Z_0$	Typical	33–42	$\Omega$
	Maximum	31–51	$\Omega$

## 9.5 Pull-Up/Pull-Down Resistor Requirements

The MPC7448 requires high-resistive (weak: 4.7-K $\Omega$ ) pull-up resistors on several control pins of the bus interface to maintain the control signals in the negated state after they have been actively negated and released by the MPC7448 or other bus masters. These pins are:  $\overline{TS}$ ,  $\overline{ARTRY}$ ,  $\overline{SHDO}$ , and  $\overline{SHDI}$ .

Some pins designated as being factory test pins must be pulled up to  $OV_{DD}$  or down to GND to ensure proper device operation. The pins that must be pulled up to  $OV_{DD}$  are  $\overline{LSSD\_MODE}$  and TEST[0:3]; the pins that must be pulled down to GND are L1\_TSTCLK and TEST[4]. The CKSTP\_IN signal should

to fully control the processor. If the target system has independent reset sources, such as voltage monitors, watchdog timers, power supply failures, or push-button switches, then the COP reset signals must be merged into these signals with logic.

The arrangement shown in [Figure 21](#) allows the COP port to independently assert  $\overline{\text{HRESET}}$  or  $\overline{\text{TRST}}$ , while ensuring that the target can drive  $\overline{\text{HRESET}}$  as well. If the JTAG interface and COP header will not be used,  $\overline{\text{TRST}}$  should be tied to  $\overline{\text{HRESET}}$  through a 0- $\Omega$  isolation resistor so that it is asserted when the system reset signal ( $\overline{\text{HRESET}}$ ) is asserted, ensuring that the JTAG scan chain is initialized during power-on. Although Freescale recommends that the COP header be designed into the system as shown in [Figure 21](#), if this is not possible, the isolation resistor will allow future access to  $\overline{\text{TRST}}$  in the case where a JTAG interface may need to be wired onto the system in debug situations.

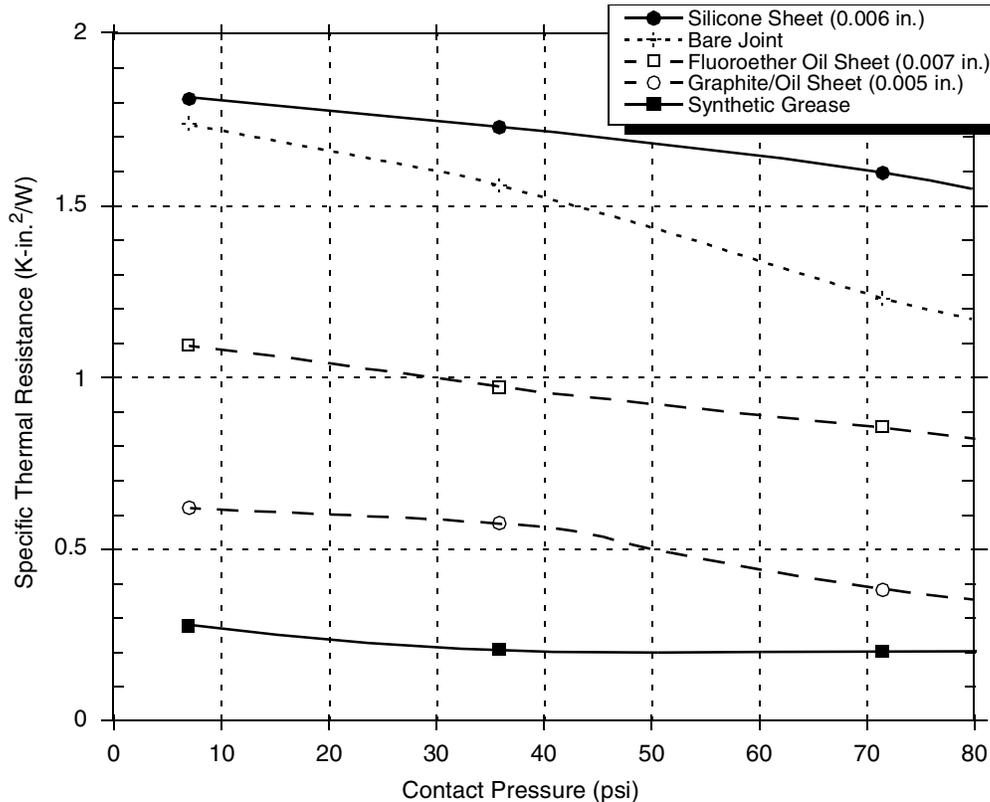
The COP header shown in [Figure 21](#) adds many benefits—breakpoints, watchpoints, register and memory examination/modification, and other standard debugger features are possible through this interface—and can be as inexpensive as an unpopulated footprint for a header to be added when needed.

The COP interface has a standard header for connection to the target system, based on the 0.025" square-post, 0.100" centered header assembly (often called a Berg header). The connector typically has pin 14 removed as a connector key.

There is no standardized way to number the COP header shown in [Figure 21](#); consequently, many different pin numbers have been observed from emulator vendors. Some are numbered top-to-bottom then left-to-right, while others use left-to-right then top-to-bottom, while still others number the pins counter clockwise from pin 1 (as with an IC). Regardless of the numbering, the signal placement recommended in [Figure 21](#) is common to all known emulators.

The  $\overline{\text{QACK}}$  signal shown in [Figure 21](#) is usually connected to the bridge chip or other system control logic in a system and is an input to the MPC7448 informing it that it can go into the quiescent state. Under normal operation this occurs during a low-power mode selection. In order for COP to work, the MPC7448 must see this signal asserted (pulled down). While shown on the COP header, not all emulator products drive this signal. If the product does not, a pull-down resistor can be populated to assert this signal. Additionally, some emulator products implement open-drain type outputs and can only drive  $\overline{\text{QACK}}$  asserted; for these tools, a pull-up resistor can be implemented to ensure this signal is negated when it is not being driven by the tool. Note that the pull-up and pull-down resistors on the  $\overline{\text{QACK}}$  signal are mutually exclusive and it is never necessary to populate both in a system. To preserve correct power-down operation,  $\overline{\text{QACK}}$  should be merged through logic so that it also can be driven by the bridge or system logic.

of any thermal interface material depends on many factors—thermal performance requirements, manufacturability, service temperature, dielectric properties, cost, and so on.



**Figure 25. Thermal Performance of Select Thermal Interface Material**

The board designer can choose between several types of thermal interfaces. Heat sink adhesive materials should be selected based on high conductivity and mechanical strength to meet equipment shock/vibration requirements. There are several commercially available thermal interfaces and adhesive materials provided by the following vendors:

The Bergquist Company 18930 West 78 <sup>th</sup> St. Chanhassen, MN 55317 Internet: <a href="http://www.bergquistcompany.com">www.bergquistcompany.com</a>	800-347-4572
Chomerics, Inc. 77 Dragon Ct. Woburn, MA 01801 Internet: <a href="http://www.chomerics.com">www.chomerics.com</a>	781-935-4850
Dow-Corning Corporation Corporate Center P.O. Box 994. Midland, MI 48686-0994 Internet: <a href="http://www.dowcorning.com">www.dowcorning.com</a>	800-248-2481

Due to the complexity and variety of system-level boundary conditions for today's microelectronic equipment, the combined effects of the heat transfer mechanisms (radiation, convection, and conduction) may vary widely. For these reasons, we recommend using conjugate heat transfer models for the board as well as system-level designs.

For system thermal modeling, the MPC7448 thermal model is shown in Figure 26. Four volumes represent this device. Two of the volumes, solder ball-air and substrate, are modeled using the package outline size of the package. The other two, die and bump-underfill, have the same size as the die. The silicon die should be modeled  $8.0 \times 7.3 \times 0.86 \text{ mm}^3$  with the heat source applied as a uniform source at the bottom of the volume. The bump and underfill layer is modeled as  $8.0 \times 7.3 \times 0.07 \text{ mm}^3$  collapsed in the z-direction with a thermal conductivity of  $5.0 \text{ W}/(\text{m} \cdot \text{K})$  in the z-direction. The substrate volume is  $25 \times 25 \times 1.14 \text{ mm}^3$  and has  $9.9 \text{ W}/(\text{m} \cdot \text{K})$  isotropic conductivity in the xy-plane and  $2.95 \text{ W}/(\text{m} \cdot \text{K})$  in the direction of the z-axis. The solder ball and air layer are modeled with the same horizontal dimensions as the substrate and is  $0.8 \text{ mm}$  thick. For the LGA package the solder and air layer is  $0.1 \text{ mm}$  thick, but the material properties are the same. It can also be modeled as a collapsed volume using orthotropic material properties:  $0.034 \text{ W}/(\text{m} \cdot \text{K})$  in the xy-plane direction and  $11.2 \text{ W}/(\text{m} \cdot \text{K})$  in the direction of the z-axis.

Conductivity	Value	Unit
<b>Die (<math>8.0 \times 7.3 \times 0.86 \text{ mm}^3</math>)</b>		
Silicon	Temperature-dependent	$\text{W}/(\text{m} \cdot \text{K})$
<b>Bump and Underfill (<math>8.0 \times 7.3 \times 0.07 \text{ mm}^3</math>)</b>		
$k_z$	5.0	$\text{W}/(\text{m} \cdot \text{K})$
<b>Substrate (<math>25 \times 25 \times 1.14 \text{ mm}^3</math>)</b>		
$k_x$	9.9	$\text{W}/(\text{m} \cdot \text{K})$
$k_y$	9.9	
$k_z$	2.95	
<b>Solder Ball and Air (<math>25 \times 25 \times 0.8 \text{ mm}^3</math>)</b>		
$k_x$	0.034	$\text{W}/(\text{m} \cdot \text{K})$
$k_y$	0.034	
$k_z$	11.2	

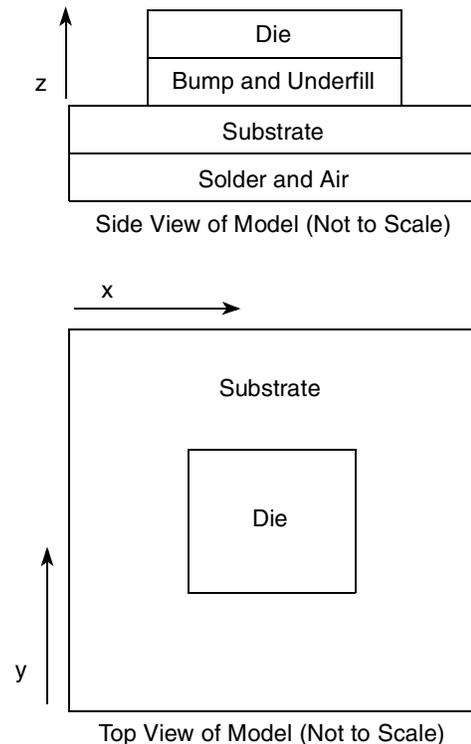


Figure 26. Recommended Thermal Model of MPC7448

## 9.7.4 Temperature Diode

The MPC7448 has a temperature diode on the microprocessor that can be used in conjunction with other system temperature monitoring devices (such as Analog Devices, ADT7461™). These devices use the negative temperature coefficient of a diode operated at a constant current to determine the temperature of the microprocessor and its environment. For proper operation, the monitoring device used should auto-calibrate the device by canceling out the  $V_{BE}$  variation of each MPC7448's internal diode.

The following are the specifications of the MPC7448 on-board temperature diode:

$$V_f > 0.40 \text{ V}$$

$$V_f < 0.90 \text{ V}$$

Operating range 2–300  $\mu\text{A}$

Diode leakage < 10 nA @ 125°C

Ideality factor over 5–150  $\mu\text{A}$  at 60°C:  $n = 1.0275 \pm 0.9\%$

Ideality factor is defined as the deviation from the ideal diode equation:

$$I_{fw} = I_s \left[ e^{\frac{qV_f}{nKT}} - 1 \right]$$

Another useful equation is:

$$V_H - V_L = n \frac{KT}{q} \left[ \ln \frac{I_H}{I_L} \right] - 1$$

Where:

$I_{fw}$  = Forward current

$I_s$  = Saturation current

$V_d$  = Voltage at diode

$V_f$  = Voltage forward biased

$V_H$  = Diode voltage while  $I_H$  is flowing

$V_L$  = Diode voltage while  $I_L$  is flowing

$I_H$  = Larger diode bias current

$I_L$  = Smaller diode bias current

$q$  = Charge of electron ( $1.6 \times 10^{-19}$  C)

$n$  = Ideality factor (normally 1.0)

$K$  = Boltzman's constant ( $1.38 \times 10^{-23}$  Joules/K)

$T$  = Temperature (Kelvins)

The ratio of  $I_H$  to  $I_L$  is usually selected to be 10:1. The previous equation simplifies to the following:

$$V_H - V_L = 1.986 \times 10^{-4} \times nT$$