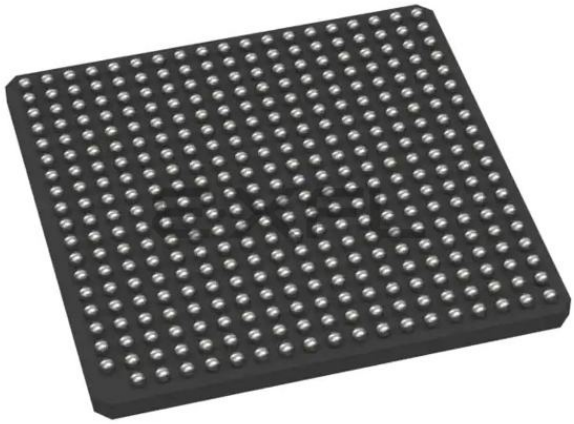

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC G4
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	1.7GHz
Co-Processors/DSP	Multimedia; SIMD
RAM Controllers	-
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	-
SATA	-
USB	-
Voltage - I/O	1.5V, 1.8V, 2.5V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	360-CBGA, FCCBGA
Supplier Device Package	360-FCCBGA (25x25)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmc7448vu1700lc

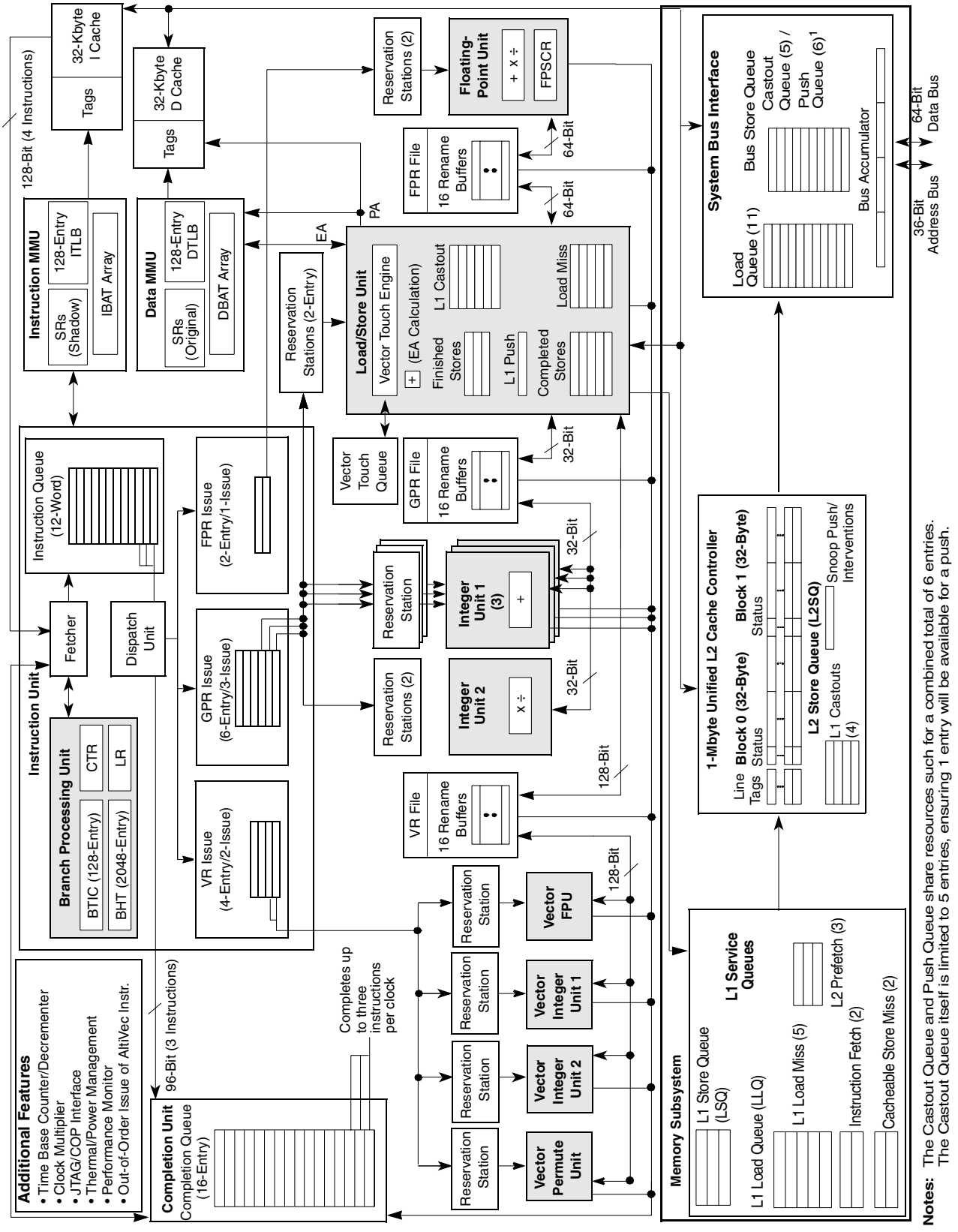


Figure 1. MPC7448 Block Diagram

- Reliability and serviceability
  - Parity checking on system bus
  - Parity checking on the L1 caches and L2 data tags
  - ECC or parity checking on L2 data

3 Comparison with the MPC7447A, MPC7447, MPC7445, and MPC7441

Table 1 compares the key features of the MPC7448 with the key features of the earlier MPC7447A, MPC7447, MPC7445, and MPC7441. All are based on the MPC7450 RISC microprocessor and are architecturally very similar. The MPC7448 is identical to the MPC7447A, but the MPC7448 supports 1 Mbyte of L2 cache with ECC and the use of dynamic frequency switching (DFS) with more bus-to-core ratios.

Table 1. Microarchitecture Comparison

Microarchitectural Specs	MPC7448	MPC7447A	MPC7447	MPC7445	MPC7441
Basic Pipeline Functions					
Logic inversions per cycle	18				
Pipeline stages up to execute	5				
Total pipeline stages (minimum)	7				
Pipeline maximum instruction throughput	3 + branch				
Pipeline Resources					
Instruction buffer size	12				
Completion buffer size	16				
Renames (integer, float, vector)	16, 16, 16				
Maximum Execution Throughput					
SFX	3				
Vector	2 (any 2 of 4 units)				
Scalar floating-point	1				
Out-of-Order Window Size in Execution Queues					
SFX integer units	1 entry × 3 queues				
Vector units	In order, 4 queues				
Scalar floating-point unit	In order				
Branch Processing Resources					
Prediction structures	BTIC, BHT, link stack				
BTIC size, associativity	128-entry, 4-way				
BHT size	2K-entry				
Link stack depth	8				
Unresolved branches supported	3				
Branch taken penalty (BTIC hit)	1				
Minimum misprediction penalty	6				

Table 1. Microarchitecture Comparison (continued)

Microarchitectural Specs	MPC7448	MPC7447A	MPC7447	MPC7445	MPC7441
Execution Unit Timings (Latency-Throughput)					
Aligned load (integer, float, vector)	3-1, 4-1, 3-1				
Misaligned load (integer, float, vector)	4-2, 5-2, 4-2				
L1 miss, L2 hit latency with ECC (data/instruction)	12/16	—			
L1 miss, L2 hit latency without ECC (data/instruction)	11/15	9/13			
SFX (add, sub, shift, rot, cmp, logicals)	1-1				
Integer multiply (32 × 8, 32 × 16, 32 × 32)	4-1, 4-1, 5-2				
Scalar float	5-1				
VSFX (vector simple)	1-1				
VCFX (vector complex)	4-1				
VFPU (vector float)	4-1				
VPER (vector permute)	2-1				
MMUs					
TLBs (instruction and data)	128-entry, 2-way				
Tablewalk mechanism	Hardware + software				
Instruction BATs/data BATs	8/8	8/8	8/8	8/8	4/4
L1 I Cache/D Cache Features					
Size	32K/32K				
Associativity	8-way				
Locking granularity	Way				
Parity on I cache	Word				
Parity on D cache	Byte				
Number of D cache misses (load/store)	5/2	5/1			
Data stream touch engines	4 streams				
On-Chip Cache Features					
Cache level	L2				
Size/associativity	1-Mbyte/ 8-way	512-Kbyte/8-way		256-Kbyte/8-way	
Access width	256 bits				
Number of 32-byte sectors/line	2	2			
Parity tag	Byte	Byte			
Parity data	Byte	Byte			
Data ECC	64-bit	—			
Thermal Control					
Dynamic frequency switching divide-by-two mode	Yes	Yes	No	No	No
Dynamic frequency switching divide-by-four mode	Yes	No	No	No	No
Thermal diode	Yes	Yes	No	No	No

Figure 2 shows the undershoot and overshoot voltage on the MPC7448.

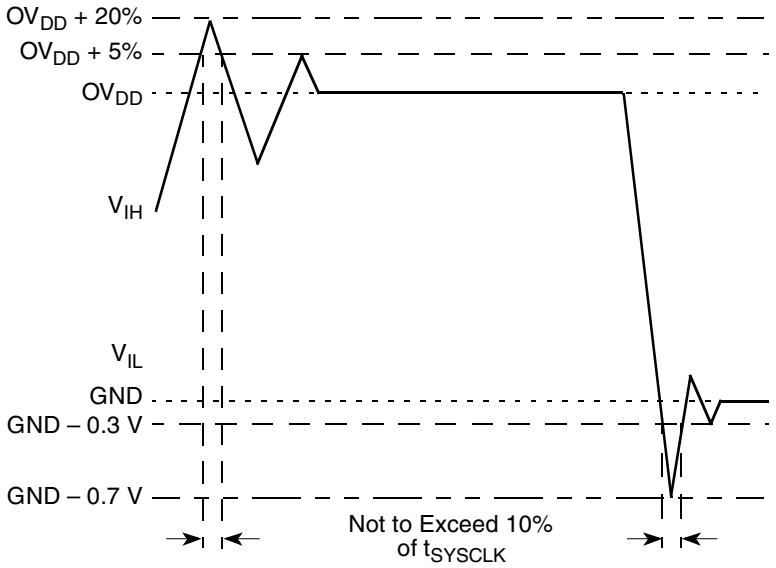


Figure 2. Overshoot/Undershoot Voltage

The MPC7448 provides several I/O voltages to support both compatibility with existing systems and migration to future systems. The MPC7448 core voltage must always be provided at the nominal voltage (see Table 4). The input voltage threshold for each bus is selected by sampling the state of the voltage select pins at the negation of the signal $\overline{\text{HRESET}}$. The output voltage will swing from GND to the maximum voltage applied to the OV_{DD} power pins. Table 3 provides the input threshold voltage settings. Because these settings may change in future products, it is recommended that BVSEL[0:1] be configured using resistor options, jumpers, or some other flexible means, with the capability to reconfigure the termination of this signal in the future, if necessary.

Table 3. Input Threshold Voltage Setting

BVSEL0	BVSEL1	I/O Voltage Mode[1]	Notes
0	0	1.8 V	2, 3
0	1	2.5 V	2, 4
1	0	1.5 V	2
1	1	2.5 V	4

Notes:

1. **Caution:** The I/O voltage mode selected must agree with the OV_{DD} voltages supplied. See Table 4.
2. If used, pull-down resistors should be less than 250 Ω.
3. The pin configuration used to select 1.8V mode on the MPC7448 is not compatible with the pin configuration used to select 1.8V mode on the MPC7447A and earlier devices.
4. The pin configuration used to select 2.5V mode on the MPC7448 is fully compatible with the pin configuration used to select 2.5V mode on the MPC7447A and earlier devices.

Table 4 provides the recommended operating conditions for the MPC7448 part numbers described by this document; see Section 11.1, “Part Numbers Fully Addressed by This Document,” for more information. See Section 9.2, “Power Supply Design and Sequencing” for power sequencing requirements.

Table 4. Recommended Operating Conditions[1]

Characteristic		Symbol	Recommended Value								Unit	Notes
			1000 MHz		1420 MHz		1600 MHz		1700 MHz			
			Min	Max	Min	Max	Min	Max	Min	Max		
Core supply voltage		V_{DD}	1.15 V ± 50 mV		1.2 V ± 50 mV		1.25 V ± 50 mV		1.3 V +20/ – 50 mV		V	3, 4, 5
PLL supply voltage		AV_{DD}	1.15 V ± 50 mV		1.2 V ± 50 mV		1.25 V ± 50 mV		1.3 V +20/ – 50 mV		V	2, 3, 4
Processor bus supply voltage	I/O Voltage Mode = 1.5 V	OV_{DD}	1.5 V ± 5%		1.5 V ± 5%		1.5 V ± 5%		1.5 V ± 5%		V	4
	I/O Voltage Mode = 1.8 V		1.8 V ± 5%		1.8 V ± 5%		1.8 V ± 5%		1.8 V ± 5%			4
	I/O Voltage Mode = 2.5 V		2.5 V ± 5%		2.5 V ± 5%		2.5 V ± 5%		2.5 V ± 5%			4
Input voltage	Processor bus	V_{in}	GND	OV_{DD}	GND	OV_{DD}	GND	OV_{DD}	GND	OV_{DD}	V	
	JTAG signals	V_{in}	GND	OV_{DD}	GND	OV_{DD}	GND	OV_{DD}	GND	OV_{DD}		
Die-junction temperature		T_j	0	105	0	105	0	105	0	105	•C	6

Notes:

1. These are the recommended and tested operating conditions.
2. This voltage is the input to the filter discussed in Section 9.2.2, “PLL Power Supply Filtering,” and not necessarily the voltage at the AV_{DD} pin, which may be reduced from V_{DD} by the filter.
3. Some early devices supported voltage and frequency derating whereby VDD (and AVDD) could be reduced to reduce power consumption. This feature has been superseded and is no longer supported. See Section 5.3, “Voltage and Frequency Derating,” for more information.
4. **Caution**: Power sequencing requirements must be met; see Section 9.2, “Power Supply Design and Sequencing”.
5. **Caution**: See Section 9.2.3, “Transient Specifications” for information regarding transients on this power supply.
6. For information on extended temperature devices, see Section 11.2, “Part Numbers Not Fully Addressed by This Document.”

5.2 AC Electrical Characteristics

This section provides the AC electrical characteristics for the MPC7448. After fabrication, functional parts are sorted by maximum processor core frequency as shown in Section 5.2.1, “Clock AC Specifications,” and tested for conformance to the AC specifications for that frequency. The processor core frequency, determined by the bus (SYSCLK) frequency and the settings of the PLL_CFG[0:5] signals, can be dynamically modified using dynamic frequency switching (DFS). Parts are sold by maximum processor core frequency; see Section 11, “Part Numbering and Marking,” for information on ordering parts. DFS is described in Section 9.7.5, “Dynamic Frequency Switching (DFS).”

5.2.1 Clock AC Specifications

Table 8 provides the clock AC timing specifications as defined in Figure 3 and represents the tested operating frequencies of the devices. The maximum system bus frequency, f_{SYSCLK}, given in Table 8, is considered a practical maximum in a typical single-processor system. This does not exclude multi-processor systems, but these typically require considerably more design effort to achieve the maximum rated bus frequency. The actual maximum SYSCLK frequency for any application of the MPC7448 will be a function of the AC timings of the microprocessor(s), the AC timings for the system controller, bus loading, circuit board topology, trace lengths, and so forth, and may be less than the value given in Table 8.

Table 9. Processor Bus AC Timing Specifications[1] (continued)

At recommended operating conditions. See Table 4.

Parameter	Symbol [2]	All Speed Grades		Unit	Notes
		Min	Max		
SYSCLK to output high impedance (all except $\overline{TS}$, $\overline{ARTRY}$, $\overline{SHD0}$, $\overline{SHD1}$)	t_{KHOZ}	—	1.8	ns	5
SYSCLK to $\overline{TS}$ high impedance after precharge	t_{KHTSPZ}	—	1	t_{SYSCLK}	3, 4, 5
Maximum delay to $\overline{ARTRY}$/$\overline{SHD0}$/$\overline{SHD1}$ precharge	t_{KHARP}	—	1	t_{SYSCLK}	3, 5, 6, 7
SYSCLK to $\overline{ARTRY}$/$\overline{SHD0}$/$\overline{SHD1}$ high impedance after precharge	t_{KHARPZ}	—	2	t_{SYSCLK}	3, 5, 6, 7

Notes:

1. All input specifications are measured from the midpoint of the signal in question to the midpoint of the rising edge of the input SYSCLK. All output specifications are measured from the midpoint of the rising edge of SYSCLK to the midpoint of the signal in question. All output timings assume a purely resistive 50-Ω load (see Figure 4). Input and output timings are measured at the pin; time-of-flight delays must be added for trace lengths, vias, and connectors in the system.
2. The symbology used for timing specifications herein follows the pattern of $t_{(signal)(state)(reference)(state)}$ for inputs and $t_{(reference)(state)(signal)(state)}$ for outputs. For example, t_{IVKH} symbolizes the time input signals (I) reach the valid state (V) relative to the SYSCLK reference (K) going to the high (H) state or input setup time. And t_{KHOV} symbolizes the time from SYSCLK(K) going high (H) until outputs (O) are valid (V) or output valid time. Input hold time can be read as the time that the input signal (I) went invalid (X) with respect to the rising clock edge (KH) (note the position of the reference and its state for inputs) and output hold time can be read as the time from the rising edge (KH) until the output went invalid (OX).
3. t_{sysclk} is the period of the external clock (SYSCLK) in ns. The numbers given in the table must be multiplied by the period of SYSCLK to compute the actual time duration (in ns) of the parameter in question.
4. According to the bus protocol, $\overline{TS}$ is driven only by the currently active bus master. It is asserted low and precharged high before returning to high impedance, as shown in Figure 6. The nominal precharge width for $\overline{TS}$ is t_{SYSCLK}, that is, one clock period. Since no master can assert $\overline{TS}$ on the following clock edge, there is no concern regarding contention with the precharge. Output valid and output hold timing is tested for the signal asserted. Output valid time is tested for precharge.The high-impedance behavior is guaranteed by design.
5. Guaranteed by design and not tested
6. According to the bus protocol, $\overline{ARTRY}$ can be driven by multiple bus masters through the clock period immediately following $\overline{AACK}$. Bus contention is not an issue because any master asserting $\overline{ARTRY}$ will be driving it low. Any master asserting it low in the first clock following $\overline{AACK}$ will then go to high impedance for a fraction of a cycle, then negated for up to an entire cycle (crossing a bus cycle boundary) before being three-stated again. The nominal precharge width for $\overline{ARTRY}$ is 1.0 t_{SYSCLK}; that is, it should be high impedance as shown in Figure 6 before the first opportunity for another master to assert $\overline{ARTRY}$. Output valid and output hold timing is tested for the signal asserted.The high-impedance behavior is guaranteed by design.
7. According to the MPX bus protocol, $\overline{SHD0}$ and $\overline{SHD1}$ can be driven by multiple bus masters beginning two cycles after $\overline{TS}$. Timing is the same as $\overline{ARTRY}$, that is, the signal is high impedance for a fraction of a cycle, then negated for up to an entire cycle (crossing a bus cycle boundary) before being three-stated again. The nominal precharge width for $\overline{SHD0}$ and $\overline{SHD1}$ is 1.0 t_{SYSCLK}. The edges of the precharge vary depending on the programmed ratio of core to bus (PLL configurations).
8. $\overline{BMODE}$[0:1] and BVSEL[0:1] are mode select inputs. $\overline{BMODE}$[0:1] are sampled before and after $\overline{HRESET}$ negation. BVSEL[0:1] are sampled before $\overline{HRESET}$ negation. These parameters represent the input setup and hold times for each sample. These values are guaranteed by design and not tested. $\overline{BMODE}$[0:1] must remain stable after the second sample; BVSEL[0:1] must remain stable after the first (and only) sample. See Figure 5 for sample timing.

5.2.3 IEEE Std. 1149.1 AC Timing Specifications

Table 10 provides the IEEE Std. 1149.1 (JTAG) AC timing specifications as defined in Figure 8 through Figure 11.

Table 10. JTAG AC Timing Specifications (Independent of SYSCLK)[1]

At recommended operating conditions. See Table 4.

Parameter	Symbol	Min	Max	Unit	Notes
TCK frequency of operation	f_{TCLK}	0	33.3	MHz	
TCK cycle time	t_{TCLK}	30	—	ns	
TCK clock pulse width measured at 1.4 V	t_{JHJL}	15	—	ns	
TCK rise and fall times	t_{JR} and t_{JF}	—	2	ns	
$\overline{\text{TRST}}$ assert time	t_{TRST}	25	—	ns	2
Input setup times: Boundary-scan data TMS, TDI	 t_{DVJH} t_{IVJH}	 4 0	 — —	ns	3
Input hold times: Boundary-scan data TMS, TDI	 t_{DXJH} t_{IXJH}	 20 25	 — —	ns	3
Valid times: Boundary-scan data TDO	 t_{JLDV} t_{JLOV}	 4 4	 20 25	ns	4
Output hold times: Boundary-scan data TDO	 t_{JLDX} t_{JLOX}	 30 30	 — —	ns	4
TCK to output high impedance: Boundary-scan data TDO	 t_{JLDZ} t_{JLOZ}	 3 3	 19 9	ns	4, 5

Notes:

1. All outputs are measured from the midpoint voltage of the falling/rising edge of TCLK to the midpoint of the signal in question. The output timings are measured at the pins. All output timings assume a purely resistive 50-Ω load (see Figure 7). Time-of-flight delays must be added for trace lengths, vias, and connectors in the system.
2. $\overline{\text{TRST}}$ is an asynchronous level sensitive signal. The time is for test purposes only.
3. Non-JTAG signal input timing with respect to TCK.
4. Non-JTAG signal output timing with respect to TCK.
5. Guaranteed by design and characterization.

Figure 7 provides the AC test load for TDO and the boundary-scan outputs of the MPC7448.

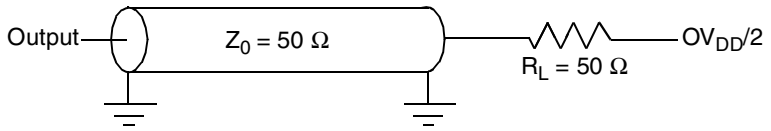


Figure 7. Alternate AC Test Load for the JTAG Interface

Figure 8 provides the JTAG clock input timing diagram.

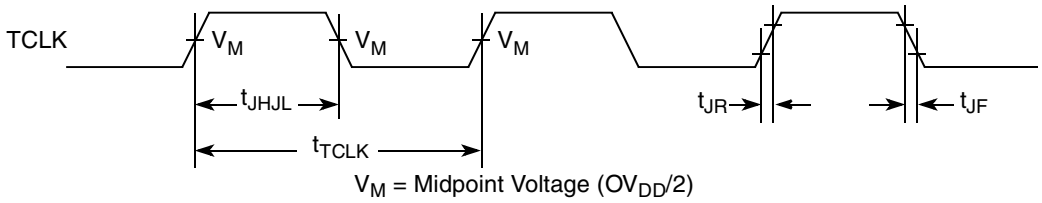


Figure 8. JTAG Clock Input Timing Diagram

Figure 9 provides the $\overline{\text{TRST}}$ timing diagram.

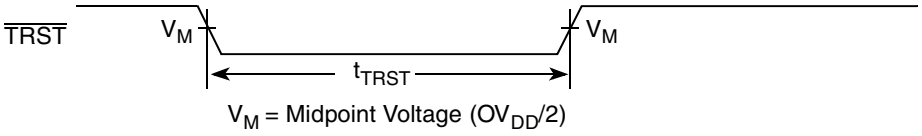


Figure 9. $\overline{\text{TRST}}$ Timing Diagram

Figure 10 provides the boundary-scan timing diagram.

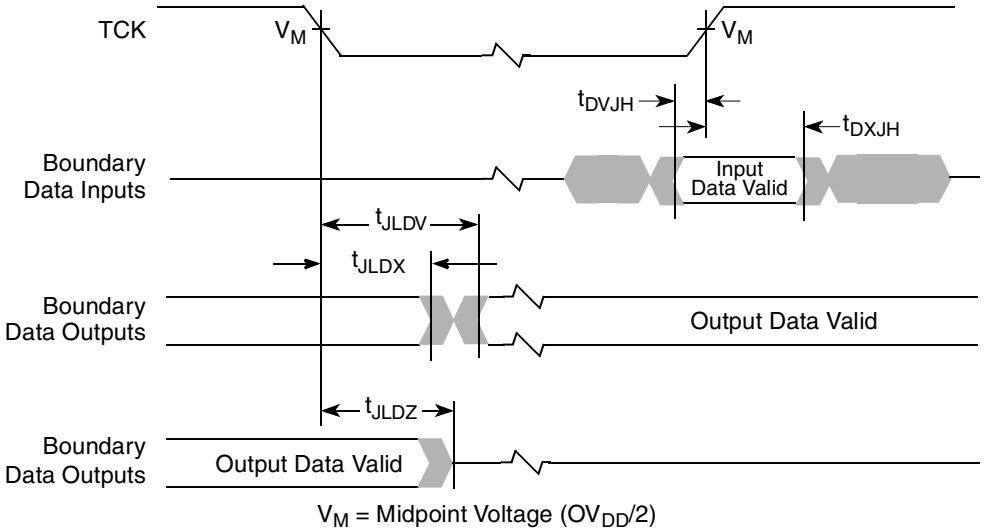


Figure 10. Boundary-Scan Timing Diagram

7 Pinout Listings

Table 11 provides the pinout listing for the MPC7448, 360 HCTE package. The pinouts of the MPC7448 and MPC7447A are compatible, but the requirements regarding the use of the additional power and ground pins have changed. The MPC7448 requires these pins be connected to the appropriate power or ground plane to achieve high core frequencies; see Section 9.3, “Connection Recommendations,” for additional information. As a result, these pins should be connected in all new designs.

Additionally, the MPC7448 may be populated on a board designed for a MPC7447 (or MPC7445 or MPC7441), provided the core voltage can be made to match the requirements in Table 4 and all pins defined as ‘no connect’ for the MPC7447 are unterminated, as required by the *MPC7457 RISC Microprocessor Hardware Specifications*. The MPC7448 uses pins previously marked ‘no connect’ for the temperature diode pins and for additional power and ground connections. The additional power and ground pins are required to achieve high core frequencies and core frequency will be limited if they are not connected; see Section 9.3, “Connection Recommendations,” for additional information. Because these ‘no connect’ pins in the MPC7447 360 pin package are not driven in functional mode, an MPC7447 can be populated in an MPC7448 board.

NOTE

Caution must be exercised when performing boundary scan test operations on a board designed for an MPC7448, but populated with an MPC7447 or earlier device. This is because in the MPC7447 it is possible to drive the latches associated with the former ‘no connect’ pins in the MPC7447, potentially causing contention on those pins. To prevent this, ensure that these pins are not connected on the board or, if they are connected, ensure that the states of internal MPC7447 latches do not cause these pins to be driven during board testing.

For the MPC7448, pins that were defined as the TEST[0:4] factory test signal group on the MPC7447A and earlier devices have been assigned new functions. For most of these, the termination recommendations for the TEST[0:4] pins of the MPC7447A are compatible with the MPC7448 and will allow correct operation with no performance loss. The exception is BVSEL1 (TEST3 on the MPC7447A and earlier devices), which may require a different termination depending which I/O voltage mode is desired; see Table 3 for more information.

NOTE

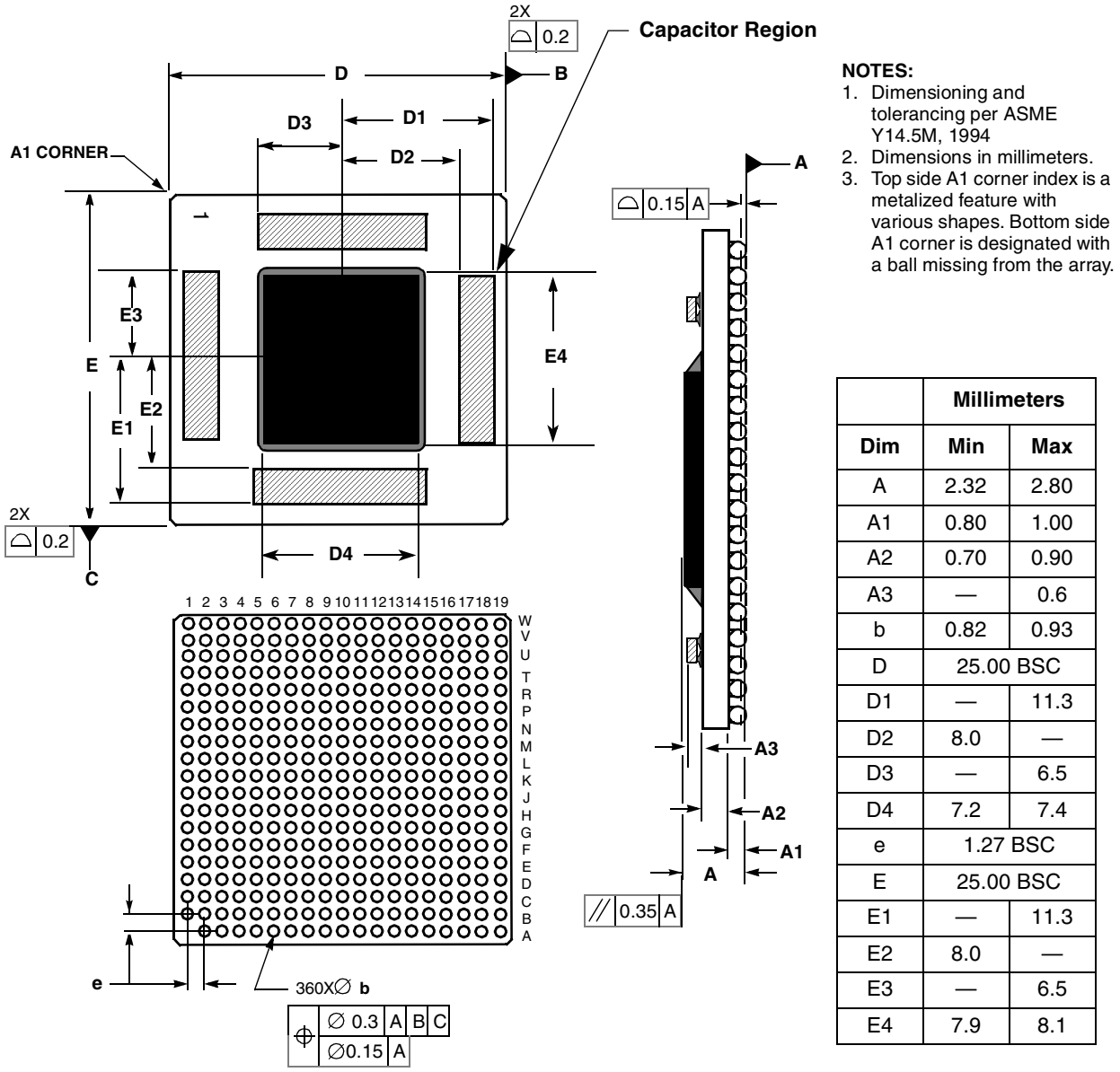
This pinout is not compatible with the MPC750, MPC7400, or MPC7410 360 BGA package.

Table 11. Pinout Listing for the MPC7448, 360 HCTE Package (continued)

Signal Name	Pin Number	Active	I/O	Notes
$\overline{\text{LVRAM}}$	B10	—	—	12, 20, 22
NC (no connect)	A6, A14, A15, B14, B15, C14, C15, C16, C17, C18, C19, D14, D15, D16, D17, D18, D19, E14, E15, F14, F15, G14, G15, H15, H16, J15, J16, J17, J18, J19, K15, K16, K17, K18, K19, L15, L16, L17, L18, L19	—	—	11
$\overline{\text{LSSD\_MODE}}$	E8	Low	Input	6, 12
$\overline{\text{MCP}}$	C9	Low	Input	
OV_{DD}	B4, C2, C12, D5, F2, H3, J5, K2, L5, M3, N6, P2, P8, P11, R4, R13, R16, T6, T9, U2, U12, U16, V4, V7, V10, V14	—	—	
OVDD_SENSE	E18, G18	—	—	16
PLL_CFG[0:4]	B8, C8, C7, D7, A7	High	Input	
PLL_CFG[5]	D10	High	Input	9, 20
$\overline{\text{PMON\_IN}}$	D9	Low	Input	13
$\overline{\text{PMON\_OUT}}$	A9	Low	Output	
$\overline{\text{QACK}}$	G5	Low	Input	
$\overline{\text{QREQ}}$	P4	Low	Output	
$\overline{\text{SHD}}$[0:1]	E4, H5	Low	I/O	3
$\overline{\text{SMI}}$	F9	Low	Input	
$\overline{\text{SRESET}}$	A2	Low	Input	
SYSCLK	A10	—	Input	
$\overline{\text{TA}}$	K6	Low	Input	
TBEN	E1	High	Input	
$\overline{\text{TBST}}$	F11	Low	Output	
TCK	C6	High	Input	
TDI	B9	High	Input	6
TDO	A4	High	Output	
$\overline{\text{TEA}}$	L1	Low	Input	
TEMP_ANODE	N18	—	—	17
TEMP_CATHODE	N19	—	—	17
TMS	F1	High	Input	6
$\overline{\text{TRST}}$	A5	Low	Input	6, 14
$\overline{\text{TS}}$	L4	Low	I/O	3
TSIZ[0:2]	G6, F7, E7	High	Output	
TT[0:4]	E5, E6, F6, E9, C5	High	I/O	
$\overline{\text{WT}}$	D3	Low	Output	
V_{DD}	H8, H10, H12, J7, J9, J11, J13, K8, K10, K12, K14, L7, L9, L11, L13, M8, M10, M12	—	—	
V_{DD}	A13, A16, A18, B17, B19, C13, E13, E16, F12, F17, F19, G11, G16, H14, H17, H19, M14, M16, M18, N15, N17, P16, P18	—	—	15

8.2 Mechanical Dimensions for the MPC7448, 360 HCTE BGA

Figure 13 provides the mechanical dimensions and bottom surface nomenclature for the MPC7448, 360 HCTE BGA package.



NOTES:
1. Dimensioning and tolerancing per ASME Y14.5M, 1994
2. Dimensions in millimeters.
3. Top side A1 corner index is a metalized feature with various shapes. Bottom side A1 corner is designated with a ball missing from the array.

	Millimeters	
Dim	Min	Max
A	2.32	2.80
A1	0.80	1.00
A2	0.70	0.90
A3	—	0.6
b	0.82	0.93
D	25.00 BSC	
D1	—	11.3
D2	8.0	—
D3	—	6.5
D4	7.2	7.4
e	1.27 BSC	
E	25.00 BSC	
E1	—	11.3
E2	8.0	—
E3	—	6.5
E4	7.9	8.1

Figure 13. Mechanical Dimensions and Bottom Surface Nomenclature for the MPC7448, 360 HCTE BGA Package

8.3 Package Parameters for the MPC7448, 360 HCTE LGA

The package parameters are as provided in the following list. The package type is 25 × 25 mm, 360 pin high coefficient of thermal expansion ceramic land grid array (HCTE).

Package outline	25 × 25 mm
Interconnects	360 (19 × 19 ball array – 1)
Pitch	1.27 mm (50 mil)
Minimum module height	1.52 mm
Maximum module height	1.80 mm
Pad diameter	0.89 mm (35 mil)
Coefficient of thermal expansion	12.3 ppm/°C

9 System Design Information

This section provides system and thermal design requirements and recommendations for successful application of the MPC7448.

9.1 Clocks

The following sections provide more detailed information regarding the clocking of the MPC7448.

9.1.1 PLL Configuration

The MPC7448 PLL is configured by the PLL_CFG[0:5] signals. For a given SYSCLK (bus) frequency, the PLL configuration signals set the internal CPU and VCO frequency of operation. The PLL configuration for the MPC7448 is shown in Table 12. In this example, shaded cells represent settings that, for a given SYSCLK frequency, result in core and/or VCO frequencies that do not comply with Table 8. When enabled, dynamic frequency switching (DFS) also affects the core frequency by halving or quartering the bus-to-core multiplier; see Section 9.7.5, "Dynamic Frequency Switching (DFS)," for more information. Note that when DFS is enabled the resulting core frequency must meet the adjusted minimum core frequency requirements (f_{core_DFS}) described in Table 8. Note that the PLL_CFG[5] is currently used for factory test only and should be tied low, and that the MPC7448 PLL configuration settings are compatible with the MPC7447A PLL configuration settings when PLL_CFG[5] = 0.

Table 12. MPC7448 Microprocessor PLL Configuration Example

PLL_CFG[0:5]	Example Core and VCO Frequency in MHz									
	Bus-to-Core Multiplier [5]	Core-to-VCO Multiplier [5]	Bus (SYSCLK) Frequency							
			33.3 MHz	50 MHz	66.6 MHz	75 MHz	83 MHz	100 MHz	133 MHz	167 MHz	200 MHz
010000	2x [6]	1x									
100000	3x [6]	1x									600
101000	4x [6]	1x								667	800
101100	5x	1x							667	835	1000
100100	5.5x	1x							733	919	1100
110100	6x	1x						600	800	1002	1200
010100	6.5x	1x						650	866	1086	1300
001000	7x	1x						700	931	1169	1400
000100	7.5x	1x					623	750	1000	1253	1500
110000	8x	1x				600	664	800	1064	1336	1600
011000	8.5x	1x				638	706	850	1131	1417	1700
011110	9x	1x			600	675	747	900	1197	1500	
011100	9.5x	1x			633	712	789	950	1264	1583	
101010	10x	1x			667	750	830	1000	1333	1667	
100010	10.5x	1x			700	938	872	1050	1397		

9.2.4 Decoupling Recommendations

Due to the MPC7448 dynamic power management feature, large address and data buses, and high operating frequencies, the MPC7448 can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. This noise must be prevented from reaching other components in the MPC7448 system, and the MPC7448 itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer use sufficient decoupling capacitors, typically one capacitor for every V_{DD} pin, and a similar amount for the OV_{DD} pins, placed as close as possible to the power pins of the MPC7448. It is also recommended that these decoupling capacitors receive their power from separate V_{DD}, OV_{DD}, and GND power planes in the PCB, using short traces to minimize inductance.

These capacitors should have a value of 0.01 or 0.1 µF. Only ceramic surface mount technology (SMT) capacitors should be used to minimize lead inductance. Orientations where connections are made along the length of the part, such as 0204, are preferable but not mandatory. Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993) and contrary to previous recommendations for decoupling Freescale microprocessors, multiple small capacitors of equal value are recommended over using multiple values of capacitance.

In addition, it is recommended that there be several bulk storage capacitors distributed around the PCB, feeding the V_{DD} and OV_{DD} planes, to enable quick recharging of the smaller chip capacitors. These bulk capacitors should have a low equivalent series resistance (ESR) rating to ensure the quick response time necessary. They should also be connected to the power and ground planes through two vias to minimize inductance. Suggested bulk capacitors are 100–330 µF (AVX TPS tantalum or Sanyo OSCON).

9.3 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. Unless otherwise noted, unused active low inputs should be tied to OV_{DD} and unused active high inputs should be connected to GND. All NC (no connect) signals must remain unconnected.

Power and ground connections must be made to all external V_{DD}, OV_{DD}, and GND pins in the MPC7448. For backward compatibility with the MPC7447, MPC7445, and MP7441, or for migrating a system originally designed for one of these devices to the MPC7448, the new power and ground signals (formerly NC, see Table 11) may be left unconnected if the core frequency is 1 GHz or less. Operation above 1 GHz requires that these additional power and ground signals be connected, and it is strongly recommended that all new designs include the additional connections. See also Section 7, “Pinout Listings,” for additional information.

The MPC7448 provides VDD_SENSE, OVDD_SENSE, and GND_SENSE pins. These pins connect directly to the power/ground planes in the device package and are intended to allow an external device to measure the voltage present on the V_{DD}, OV_{DD} and GND planes in the device package. The most common use for these signals is as a feedback signal to a power supply regulator to allow it to compensate for board losses and supply the correct voltage at the device. (Note that all voltage parameters are specified at the pins of the device.) If not used for this purpose, it is recommended that these signals be connected to test points that can be used in the event that an accurate measurement of the voltage at the device is needed during system debug. Otherwise, these signals should be connected to the appropriate power/ground planes on the circuit board or left unconnected.

to fully control the processor. If the target system has independent reset sources, such as voltage monitors, watchdog timers, power supply failures, or push-button switches, then the COP reset signals must be merged into these signals with logic.

The arrangement shown in Figure 21 allows the COP port to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$, while ensuring that the target can drive $\overline{\text{HRESET}}$ as well. If the JTAG interface and COP header will not be used, $\overline{\text{TRST}}$ should be tied to $\overline{\text{HRESET}}$ through a 0-Ω isolation resistor so that it is asserted when the system reset signal ($\overline{\text{HRESET}}$) is asserted, ensuring that the JTAG scan chain is initialized during power-on. Although Freescale recommends that the COP header be designed into the system as shown in Figure 21, if this is not possible, the isolation resistor will allow future access to $\overline{\text{TRST}}$ in the case where a JTAG interface may need to be wired onto the system in debug situations.

The COP header shown in Figure 21 adds many benefits—breakpoints, watchpoints, register and memory examination/modification, and other standard debugger features are possible through this interface—and can be as inexpensive as an unpopulated footprint for a header to be added when needed.

The COP interface has a standard header for connection to the target system, based on the 0.025" square-post, 0.100" centered header assembly (often called a Berg header). The connector typically has pin 14 removed as a connector key.

There is no standardized way to number the COP header shown in Figure 21; consequently, many different pin numbers have been observed from emulator vendors. Some are numbered top-to-bottom then left-to-right, while others use left-to-right then top-to-bottom, while still others number the pins counter clockwise from pin 1 (as with an IC). Regardless of the numbering, the signal placement recommended in Figure 21 is common to all known emulators.

The $\overline{\text{QACK}}$ signal shown in Figure 21 is usually connected to the bridge chip or other system control logic in a system and is an input to the MPC7448 informing it that it can go into the quiescent state. Under normal operation this occurs during a low-power mode selection. In order for COP to work, the MPC7448 must see this signal asserted (pulled down). While shown on the COP header, not all emulator products drive this signal. If the product does not, a pull-down resistor can be populated to assert this signal. Additionally, some emulator products implement open-drain type outputs and can only drive $\overline{\text{QACK}}$ asserted; for these tools, a pull-up resistor can be implemented to ensure this signal is negated when it is not being driven by the tool. Note that the pull-up and pull-down resistors on the $\overline{\text{QACK}}$ signal are mutually exclusive and it is never necessary to populate both in a system. To preserve correct power-down operation, $\overline{\text{QACK}}$ should be merged through logic so that it also can be driven by the bridge or system logic.

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Laird Technologies - Thermal (formerly Thermagon Inc.) 4707 Detroit Ave. Cleveland, OH 44102 Internet: www.lairdtech.com	888-246-905

The following section provides a heat sink selection example using one of the commercially available heat sinks.

9.7.3 Heat Sink Selection Example

For preliminary heat sink sizing, the die-junction temperature can be expressed as follows:

$$T_j = T_i + T_r + (R_{\theta JC} + R_{\theta int} + R_{\theta sa}) \times P_d$$

where:

T_j is the die-junction temperature
T_i is the inlet cabinet ambient temperature
T_r is the air temperature rise within the computer cabinet
$R_{\theta JC}$ is the junction-to-case thermal resistance
$R_{\theta int}$ is the adhesive or interface material thermal resistance
$R_{\theta sa}$ is the heat sink base-to-ambient thermal resistance
P_d is the power dissipated by the device

During operation, the die-junction temperatures (T_j) should be maintained less than the value specified in Table 4. The temperature of air cooling the component greatly depends on the ambient inlet air temperature and the air temperature rise within the electronic cabinet. An electronic cabinet inlet-air temperature (T_i) may range from 30□ to 40□C. The air temperature rise within a cabinet (T_r) may be in the range of 5□ to 10□C. The thermal resistance of the thermal interface material ($R_{\theta int}$) is typically about 1.1 □C/W. For example, assuming a T_i of 30□C, a T_r of 5□C, an HCTE package $R_{\theta JC}$ = 0.1, and a power consumption (P_d) of 25.6 W, the following expression for T_j is obtained:

Die-junction temperature: $T_j = 30□C + 5□C + (0.1□C/W + 1.1□C/W + \theta_{sa}) \times 25.6$

For this example, a $R_{\theta sa}$value of 1.53 □C/W or less is required to maintain the die junction temperature below the maximum value of Table 4.

Though the die junction-to-ambient and the heat sink-to-ambient thermal resistances are a common figure-of-merit used for comparing the thermal performance of various microelectronic packaging technologies, one should exercise caution when only using this metric in determining thermal management because no single parameter can adequately describe three-dimensional heat flow. The final die-junction operating temperature is not only a function of the component-level thermal resistance, but the system-level design and its operating conditions. In addition to the component's power consumption, a number of factors affect the final operating die-junction temperature—airflow, board population (local heat flux of adjacent components), heat sink efficiency, heat sink attach, heat sink placement, next-level interconnect technology, system air temperature rise, altitude, and so on.

Table 16. Valid Divide Ratio Configurations

DFS mode disabled		DFS divide-by-2 mode enabled (HID1[DFS2] = 1 or $\overline{\text{DFS2}}$ = 0)		DFS divide-by-4 mode enabled (HID1[DFS4] = 1 or $\overline{\text{DFS4}}$ = 0)	
Bus-to-Core Multiplier Configured by PLL_CFG[0:5] (see Table 12)	HID1[PC0-5] [3]	Bus-to-Core Multiplier	HID1[PC0-5] [3]	Bus-to-Core Multiplier	HID1[PC0-5] [3]
2x [4]	010000	N/A (unchanged) [1]	unchanged [1]	N/A (unchanged) [1]	unchanged [1]
3x [4]	100000	N/A (unchanged) [1]	unchanged [1]	N/A (unchanged) [1]	unchanged [1]
4x [4]	101000	2x [4]	010000	N/A (unchanged) [1]	unchanged [1]
5x	101100	2.5x [4]	010101	N/A (unchanged) [1]	unchanged [1]
5.5x	100100	2.75x [4]	110101 [2]	N/A (unchanged) [1]	unchanged [1]
6x	110100	3x [4]	100000	N/A (unchanged) [1]	unchanged [1]
6.5x	010100	3.25x [4]	100000 [2]	N/A (unchanged) [1]	unchanged [1]
7x	001000	3.5x [4]	110101	N/A (unchanged) [1]	unchanged [1]
7.5x	000100	3.75x [4]	110101 [2]	N/A (unchanged) [1]	unchanged [1]
8x	110000	4x [4]	101000 [4]	2x [4]	010000
8.5x	011000	4.25x [4]	101000 [2]	N/A (unchanged) [1]	unchanged [1]
9x	011110	4.5x [4]	011101	2.25x [4]	010000 [2]
9.5x	011100	4.75x [4]	011101 [2]	N/A (unchanged) [1]	unchanged [1]
10x	101010	5x	101100	2.5x [4]	010101
10.5x	100010	5.25x	101100 [2]	N/A (unchanged) [1]	unchanged [1]
11x	100110	5.5x	100100	2.75x [4]	010101 [2]
11.5x	000000	5.75x	100100 [2]	N/A (unchanged) [1]	unchanged [1]
12x	101110	6x	110100	3x [4]	100000
12.5x	111110	6.25x	110100 [2]	N/A (unchanged) [1]	unchanged [1]
13x	010110	6.5x	010100	3.25x [4]	100000 [2]
13.5x	111000	6.75	010100 [2]	N/A (unchanged) [1]	unchanged [1]
14x	110010	7x	001000	3.5x [4]	110101
15x	000110	7.5x	000100	3.75x [4]	110101 [2]
16x	110110	8x	110000	4x [4]	101000
17x	000010	8.5x	011000	4.25x [4]	101000 [2]
18x	001010	9x	011110	4.5x [4]	011101
20x	001110	10x	101010	5x	101100
21x	010010	10.5x	100010	5.25x	101100 [2]

Table 20. Part Numbers Addressed by MC7448T*xxnnnn*N*x* Series Hardware Specification Addendum (Document Order No. MPC7448ECS02AD)

xx	**7448**	**T**	***xx***	***nnnn***	**N**	***x***

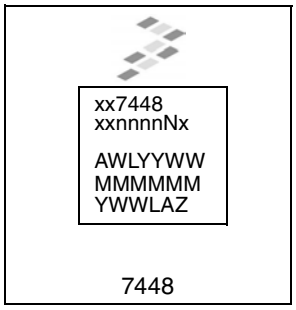
Product Code	Part Identifier	Specification Modifier	Package	Processor Frequency	Application Modifier	Revision Level
MC PPC[1]	7448	T = Extended Temperature Device	HX = HCTE BGA	1400	N: 1.15 V ± 50 mV – 40 to 105 °C	C: 2.1; PVR = 0x8004_0201 D: 2.2; PVR = 0x8004_0202
				1267 Revision C only	N: 1.1 V ± 50 mV – 40 to 105 °C	
				1267 Revision D only	N: 1.05 V ± 50 mV – 40 to 105 °C	
				1000	N: 1.0 V ± 50 mV – 40 to 105 °C	

Notes:

1. The P prefix in a Freescale part number designates a “Pilot Production Prototype” as defined by Freescale SOP 3-13. These parts have only preliminary reliability and characterization data. Before pilot production prototypes can be shipped, written authorization from the customer must be on file in the applicable sales office acknowledging the qualification status and the fact that product changes may still occur as pilot production prototypes are shipped.

11.3 Part Marking

Parts are marked as the example shown in Figure 27.



BGA/LGA

Notes:

AWLYYWW is the test code, where YYWW is the date code (YY = year, WW = work week)
MMMMMM is the M00 (mask) number.
YWWLAZ is the assembly traceability code.

Figure 27. Part Marking for BGA and LGA Device

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