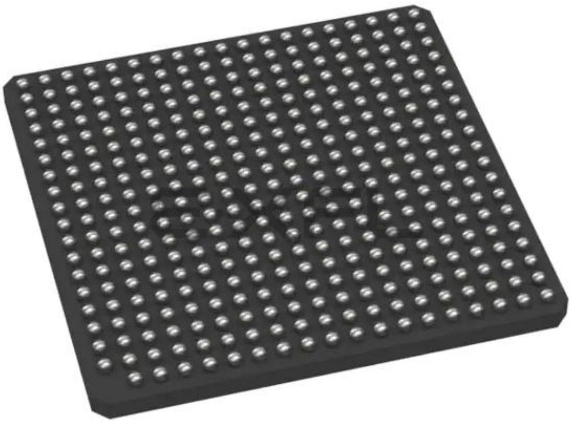

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC G4
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	667MHz
Co-Processors/DSP	Multimedia; SIMD
RAM Controllers	-
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	-
SATA	-
USB	-
Voltage - I/O	1.5V, 1.8V, 2.5V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	360-BCBGA, FCCBGA
Supplier Device Package	360-FCCBGA (25x25)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc7448hx667nc

4 General Parameters

The following list summarizes the general parameters of the MPC7448:

Technology	90 nm CMOS SOI, nine-layer metal	
Die size	8.0 mm × 7.3 mm	
Transistor count	90 million	
Logic design	Mixed static and dynamic	
Packages	Surface mount 360 ceramic ball grid array (HCTE)	
	Surface mount 360 ceramic land grid array (HCTE)	
	Surface mount 360 ceramic ball grid array with lead-free spheres (HCTE)	
Core power supply	1.30 V	(1700 MHz device)
	1.25 V	(1600 MHz device)
	1.20 V	(1420 MHz device)
	1.15 V	(1000 MHz device)
I/O power supply	1.5 V, 1.8 V, or 2.5 V	

5 Electrical and Thermal Characteristics

This section provides the AC and DC electrical specifications and thermal characteristics for the MPC7448.

5.1 DC Electrical Characteristics

The tables in this section describe the MPC7448 DC electrical characteristics. Table 2 provides the absolute maximum ratings. See Section 9.2, “Power Supply Design and Sequencing,” for power sequencing requirements.

Table 2. Absolute Maximum Ratings [1]

Characteristic		Symbol	Maximum Value	Unit	Notes
Core supply voltage		V_{DD}	–0.3 to 1.4	V	2
PLL supply voltage		AV_{DD}	–0.3 to 1.4	V	2
Processor bus supply voltage	I/O Voltage Mode = 1.5 V	OV_{DD}	–0.3 to 1.8	V	3
	I/O Voltage Mode = 1.8 V		–0.3 to 2.2		3
	I/O Voltage Mode = 2.5 V		–0.3 to 3.0		3
Input voltage	Processor bus	V_{in}	–0.3 to OV_{DD} + 0.3	V	4
	JTAG signals	V_{in}	–0.3 to OV_{DD} + 0.3	V	
Storage temperature range		T_{stg}	– 55 to 150	•C	

Notes:

1. Functional and tested operating conditions are given in Table 4. Absolute maximum ratings are stress ratings only and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect device reliability or cause permanent damage to the device.
2. See Section 9.2, “Power Supply Design and Sequencing” for power sequencing requirements.
3. Bus must be configured in the corresponding I/O voltage mode; see Table 3.
4. **Caution**: V_{in} must not exceed OV_{DD} by more than 0.3 V at any time including during power-on reset except as allowed by the overshoot specifications. V_{in} may overshoot/undershoot to a voltage and for a maximum duration as shown in Figure 2.

Table 4 provides the recommended operating conditions for the MPC7448 part numbers described by this document; see Section 11.1, “Part Numbers Fully Addressed by This Document,” for more information. See Section 9.2, “Power Supply Design and Sequencing” for power sequencing requirements.

Table 4. Recommended Operating Conditions[1]

Characteristic		Symbol	Recommended Value								Unit	Notes
			1000 MHz		1420 MHz		1600 MHz		1700 MHz			
			Min	Max	Min	Max	Min	Max	Min	Max		
Core supply voltage		V_{DD}	1.15 V ± 50 mV		1.2 V ± 50 mV		1.25 V ± 50 mV		1.3 V +20/ – 50 mV		V	3, 4, 5
PLL supply voltage		AV_{DD}	1.15 V ± 50 mV		1.2 V ± 50 mV		1.25 V ± 50 mV		1.3 V +20/ – 50 mV		V	2, 3, 4
Processor bus supply voltage	I/O Voltage Mode = 1.5 V	OV_{DD}	1.5 V ± 5%		1.5 V ± 5%		1.5 V ± 5%		1.5 V ± 5%		V	4
	I/O Voltage Mode = 1.8 V		1.8 V ± 5%		1.8 V ± 5%		1.8 V ± 5%		1.8 V ± 5%			4
	I/O Voltage Mode = 2.5 V		2.5 V ± 5%		2.5 V ± 5%		2.5 V ± 5%		2.5 V ± 5%			4
Input voltage	Processor bus	V_{in}	GND	OV_{DD}	GND	OV_{DD}	GND	OV_{DD}	GND	OV_{DD}	V	
	JTAG signals	V_{in}	GND	OV_{DD}	GND	OV_{DD}	GND	OV_{DD}	GND	OV_{DD}		
Die-junction temperature		T_j	0	105	0	105	0	105	0	105	•C	6

Notes:

1. These are the recommended and tested operating conditions.
2. This voltage is the input to the filter discussed in Section 9.2.2, “PLL Power Supply Filtering,” and not necessarily the voltage at the AV_{DD} pin, which may be reduced from V_{DD} by the filter.
3. Some early devices supported voltage and frequency derating whereby VDD (and AVDD) could be reduced to reduce power consumption. This feature has been superseded and is no longer supported. See Section 5.3, “Voltage and Frequency Derating,” for more information.
4. **Caution**: Power sequencing requirements must be met; see Section 9.2, “Power Supply Design and Sequencing”.
5. **Caution**: See Section 9.2.3, “Transient Specifications” for information regarding transients on this power supply.
6. For information on extended temperature devices, see Section 11.2, “Part Numbers Not Fully Addressed by This Document.”

Table 6. DC Electrical Specifications (continued)

At recommended operating conditions. See Table 4.

Characteristic	Nominal Bus Voltage [1]	Symbol	Min	Max	Unit	Notes
High-impedance (off-state) leakage current: V_{in} = OV_{DD} V_{in} = GND	—	I_{TSI}	—	 50 – 50	μA	2, 3, 4
Output high voltage @ I_{OH} = –5 mA	1.5	V_{OH}	OV_{DD} – 0.45	—	V	
	1.8		OV_{DD} – 0.45	—		
	2.5		1.8	—		
Output low voltage @ I_{OL} = 5 mA	1.5	V_{OL}	—	0.45	V	
	1.8		—	0.45		
	2.5		—	0.6		
Capacitance, V_{in} = 0 V, f = 1 MHz — All inputs		C_{in}	—	8.0	pF	5

Notes:

1. Nominal voltages; see Table 4 for recommended operating conditions.
2. All I/O signals are referenced to OV_{DD}.
3. Excludes test signals and IEEE Std. 1149.1 boundary scan (JTAG) signals
4. The leakage is measured for nominal OV_{DD} and V_{DD}, or both OV_{DD} and V_{DD} must vary in the same direction (for example, both OV_{DD} and V_{DD} vary by either +5% or –5%).
5. Capacitance is periodically sampled rather than 100% tested.
6. These pins have internal pull-up resistors.

Table 7 provides the power consumption for the MPC7448 part numbers described by this document; see Section 11.1, "Part Numbers Fully Addressed by This Document," for information regarding which part numbers are described by this document. Freescale also offers MPC7448 part numbers that meet lower power consumption specifications by adhering to lower core voltage and core frequency specifications. For more information on these devices, including references to the MPC7448 Hardware Specification Addenda that describe these devices, see Section 11.2, "Part Numbers Not Fully Addressed by This Document."

The power consumptions provided in Table 7 represent the power consumption of each speed grade when operated at the rated maximum core frequency (see Table 8). Freescale sorts devices by power as well as by core frequency, and power limits for each speed grade are independent of each other. Each device is tested at its maximum core frequency only. (Note that Deep Sleep Mode power consumption is independent of clock frequency.) Operating a device at a frequency lower than its rated maximum is fully supported provided the clock frequencies are within the specifications given in Table 8, and a device operated below its rated maximum will have lower power consumption. However, inferences should not be made about a device's power consumption based on the power specifications of another (lower) speed grade. For example, a 1700 MHz device operated at 1420 MHz may not exhibit the same power consumption as a 1420 MHz device operated at 1420 MHz.

For all MPC7448 devices, the following guidelines on the use of these parameters for system design are suggested. The Full-Power Mode–Typical value represents the sustained power consumption of the device

Figure 3 provides the SYSCLK input timing diagram.

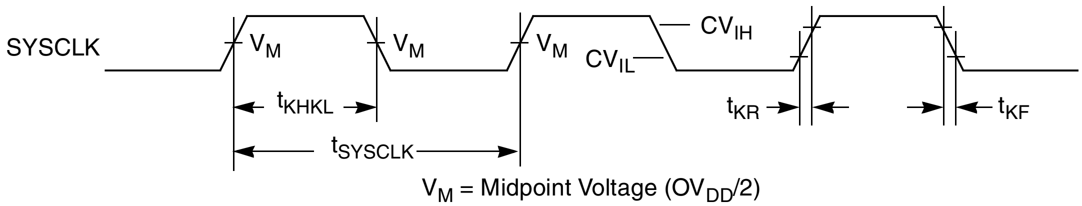


Figure 3. SYSCLK Input Timing Diagram

5.2.2 Processor Bus AC Specifications

Table 9 provides the processor bus AC timing specifications for the MPC7448 as defined in Figure 4 and Figure 5.

Table 9. Processor Bus AC Timing Specifications[1]

At recommended operating conditions. See Table 4.

Parameter	Symbol [2]	All Speed Grades		Unit	Notes
		Min	Max		
Input setup times:				ns	
A[0:35], AP[0:4]	t_{AVKH}	1.5	—		—
D[0:63], DP[0:7]	t_{DVKH}	1.5	—		—
AACK, ARTRY, BG, CKSTP_IN, DBG, DTI[0:3], GBL, TT[0:4], QACK, TA, TBEN, TEA, TS, EXT_QUAL, PMON_IN, SHD[0:1]	t_{IVKH}	1.5	—		—
BMODE[0:1], BVSEL[0:1]	t_{MVKH}	1.5	—		8
Input hold times:				ns	
A[0:35], AP[0:4]	t_{AXKH}	0	—		—
D[0:63], DP[0:7]	t_{DXKH}	0	—		—
AACK, ARTRY, BG, CKSTP_IN, DBG, DTI[0:3], GBL, TT[0:4], QACK, TA, TBEN, TEA, TS, EXT_QUAL, PMON_IN, SHD[0:1]	t_{IXKH}	0	—		—
					—
BMODE[0:1], BVSEL[0:1]	t_{MXKH}	0	—		8
Output valid times:				ns	
A[0:35], AP[0:4]	t_{KHAV}	—	1.8		
D[0:63], DP[0:7]	t_{KHDV}	—	1.8		
BR, CI, DRDY, GBL, HIT, PMON_OUT, QREQ, TBST, TSIZ[0:2], TT[0:4], WT	t_{KHOV}	—	1.8		
TS	t_{KHTSV}	—	1.8		
ARTRY, SHD[0:1]	t_{KHARV}	—	1.8		
Output hold times:				ns	
A[0:35], AP[0:4]	t_{KHAX}	0.5	—		
D[0:63], DP[0:7]	t_{KHDX}	0.5	—		
BR, CI, DRDY, GBL, HIT, PMON_OUT, QREQ, TBST, TSIZ[0:2], TT[0:4], WT	t_{KHOX}	0.5	—		
TS	t_{KHTSX}	0.5	—		
ARTRY, SHD[0:1]	t_{KHARX}	0.5	—		
SYSCLK to output enable	t_{KHOE}	0.5	—	ns	5

Table 9. Processor Bus AC Timing Specifications[1] (continued)

At recommended operating conditions. See Table 4.

Parameter	Symbol [2]	All Speed Grades		Unit	Notes
		Min	Max		
SYSCLK to output high impedance (all except $\overline{TS}$, $\overline{ARTRY}$, $\overline{SHD0}$, $\overline{SHD1}$)	t_{KHOZ}	—	1.8	ns	5
SYSCLK to $\overline{TS}$ high impedance after precharge	t_{KHTSPZ}	—	1	t_{SYSCLK}	3, 4, 5
Maximum delay to $\overline{ARTRY}$/$\overline{SHD0}$/$\overline{SHD1}$ precharge	t_{KHARP}	—	1	t_{SYSCLK}	3, 5, 6, 7
SYSCLK to $\overline{ARTRY}$/$\overline{SHD0}$/$\overline{SHD1}$ high impedance after precharge	t_{KHARPZ}	—	2	t_{SYSCLK}	3, 5, 6, 7

Notes:

1. All input specifications are measured from the midpoint of the signal in question to the midpoint of the rising edge of the input SYSCLK. All output specifications are measured from the midpoint of the rising edge of SYSCLK to the midpoint of the signal in question. All output timings assume a purely resistive 50-Ω load (see Figure 4). Input and output timings are measured at the pin; time-of-flight delays must be added for trace lengths, vias, and connectors in the system.
2. The symbology used for timing specifications herein follows the pattern of $t_{(signal)(state)(reference)(state)}$ for inputs and $t_{(reference)(state)(signal)(state)}$ for outputs. For example, t_{IVKH} symbolizes the time input signals (I) reach the valid state (V) relative to the SYSCLK reference (K) going to the high (H) state or input setup time. And t_{KHOV} symbolizes the time from SYSCLK(K) going high (H) until outputs (O) are valid (V) or output valid time. Input hold time can be read as the time that the input signal (I) went invalid (X) with respect to the rising clock edge (KH) (note the position of the reference and its state for inputs) and output hold time can be read as the time from the rising edge (KH) until the output went invalid (OX).
3. t_{sysclk} is the period of the external clock (SYSCLK) in ns. The numbers given in the table must be multiplied by the period of SYSCLK to compute the actual time duration (in ns) of the parameter in question.
4. According to the bus protocol, $\overline{TS}$ is driven only by the currently active bus master. It is asserted low and precharged high before returning to high impedance, as shown in Figure 6. The nominal precharge width for $\overline{TS}$ is t_{SYSCLK}, that is, one clock period. Since no master can assert $\overline{TS}$ on the following clock edge, there is no concern regarding contention with the precharge. Output valid and output hold timing is tested for the signal asserted. Output valid time is tested for precharge.The high-impedance behavior is guaranteed by design.
5. Guaranteed by design and not tested
6. According to the bus protocol, $\overline{ARTRY}$ can be driven by multiple bus masters through the clock period immediately following $\overline{AACK}$. Bus contention is not an issue because any master asserting $\overline{ARTRY}$ will be driving it low. Any master asserting it low in the first clock following $\overline{AACK}$ will then go to high impedance for a fraction of a cycle, then negated for up to an entire cycle (crossing a bus cycle boundary) before being three-stated again. The nominal precharge width for $\overline{ARTRY}$ is 1.0 t_{SYSCLK}; that is, it should be high impedance as shown in Figure 6 before the first opportunity for another master to assert $\overline{ARTRY}$. Output valid and output hold timing is tested for the signal asserted.The high-impedance behavior is guaranteed by design.
7. According to the MPX bus protocol, $\overline{SHD0}$ and $\overline{SHD1}$ can be driven by multiple bus masters beginning two cycles after $\overline{TS}$. Timing is the same as $\overline{ARTRY}$, that is, the signal is high impedance for a fraction of a cycle, then negated for up to an entire cycle (crossing a bus cycle boundary) before being three-stated again. The nominal precharge width for $\overline{SHD0}$ and $\overline{SHD1}$ is 1.0 t_{SYSCLK}. The edges of the precharge vary depending on the programmed ratio of core to bus (PLL configurations).
8. $\overline{BMODE}$[0:1] and BVSEL[0:1] are mode select inputs. $\overline{BMODE}$[0:1] are sampled before and after $\overline{HRESET}$ negation. BVSEL[0:1] are sampled before $\overline{HRESET}$ negation. These parameters represent the input setup and hold times for each sample. These values are guaranteed by design and not tested. $\overline{BMODE}$[0:1] must remain stable after the second sample; BVSEL[0:1] must remain stable after the first (and only) sample. See Figure 5 for sample timing.

Figure 6 provides the input/output timing diagram for the MPC7448.

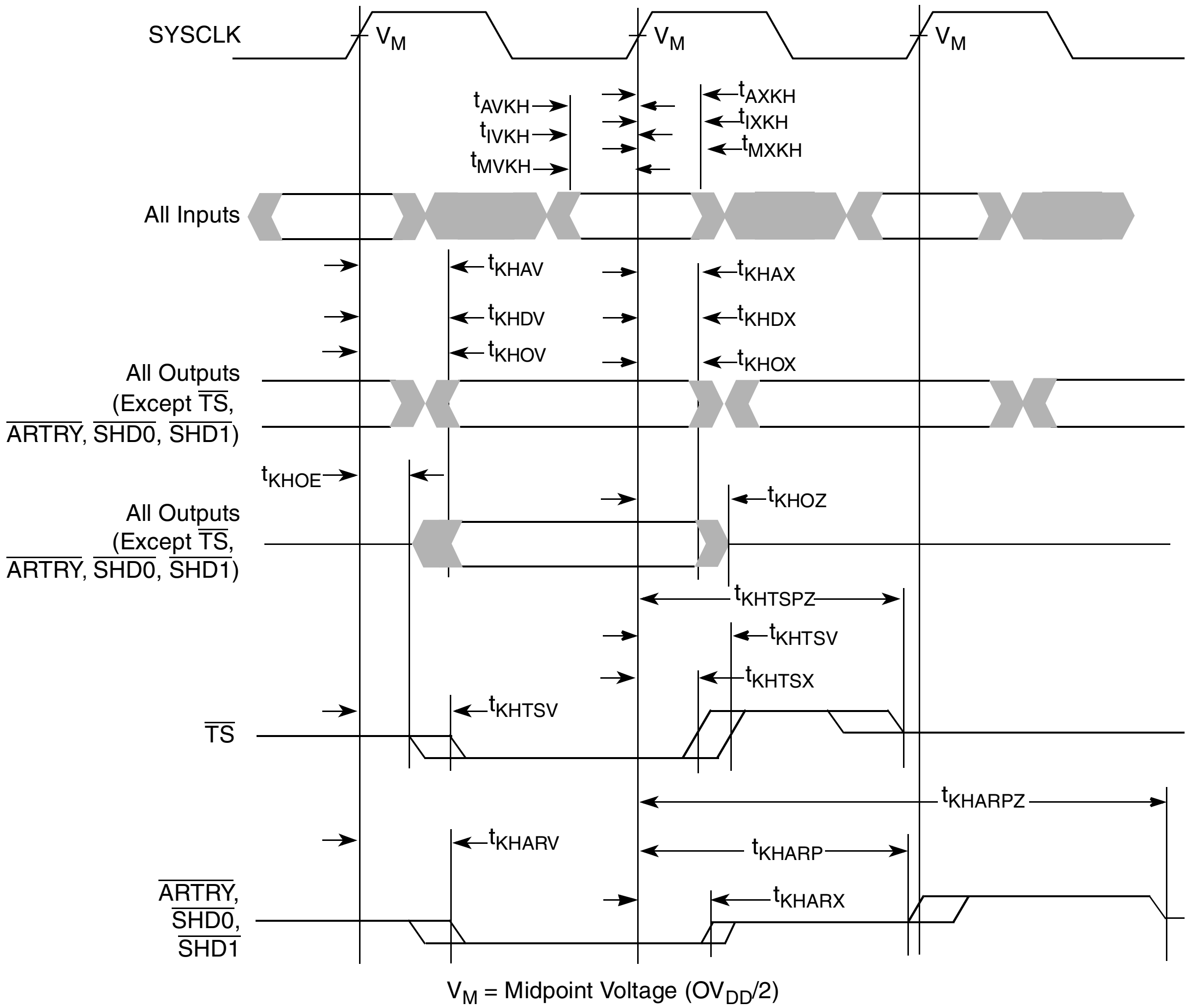


Figure 6. Input/Output Timing Diagram

Figure 7 provides the AC test load for TDO and the boundary-scan outputs of the MPC7448.

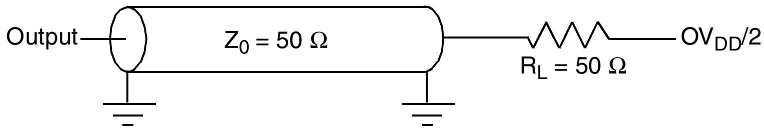


Figure 7. Alternate AC Test Load for the JTAG Interface

Figure 8 provides the JTAG clock input timing diagram.

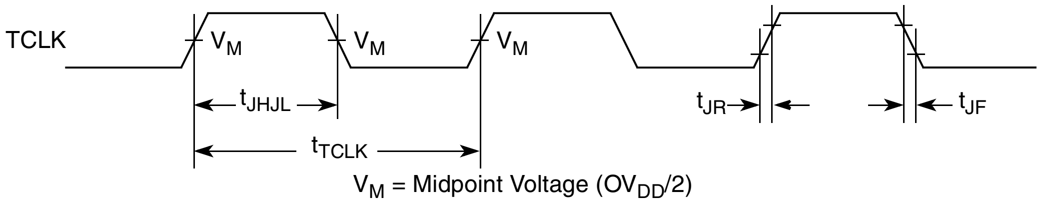


Figure 8. JTAG Clock Input Timing Diagram

Figure 9 provides the $\overline{\text{TRST}}$ timing diagram.

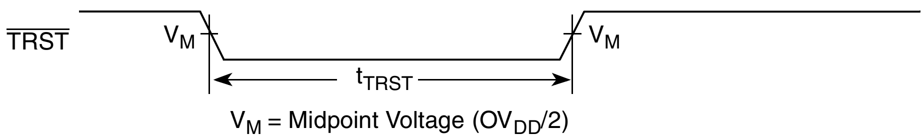


Figure 9. $\overline{\text{TRST}}$ Timing Diagram

Figure 10 provides the boundary-scan timing diagram.

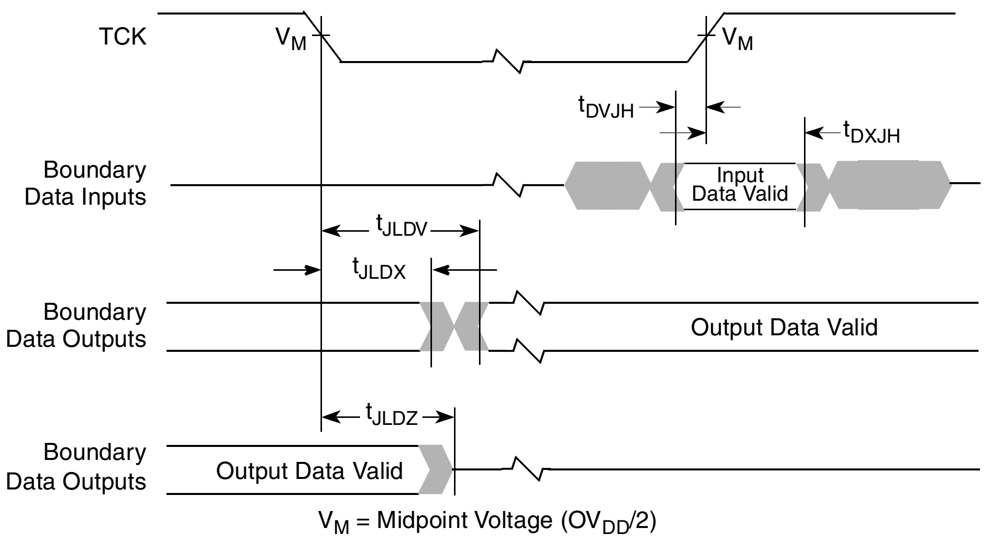


Figure 10. Boundary-Scan Timing Diagram

6 Pin Assignments

Figure 12 (in Part A) shows the pinout of the MPC7448, 360 high coefficient of thermal expansion ceramic ball grid array (HCTE) package as viewed from the top surface. Part B shows the side profile of the HCTE package to indicate the direction of the top surface view.

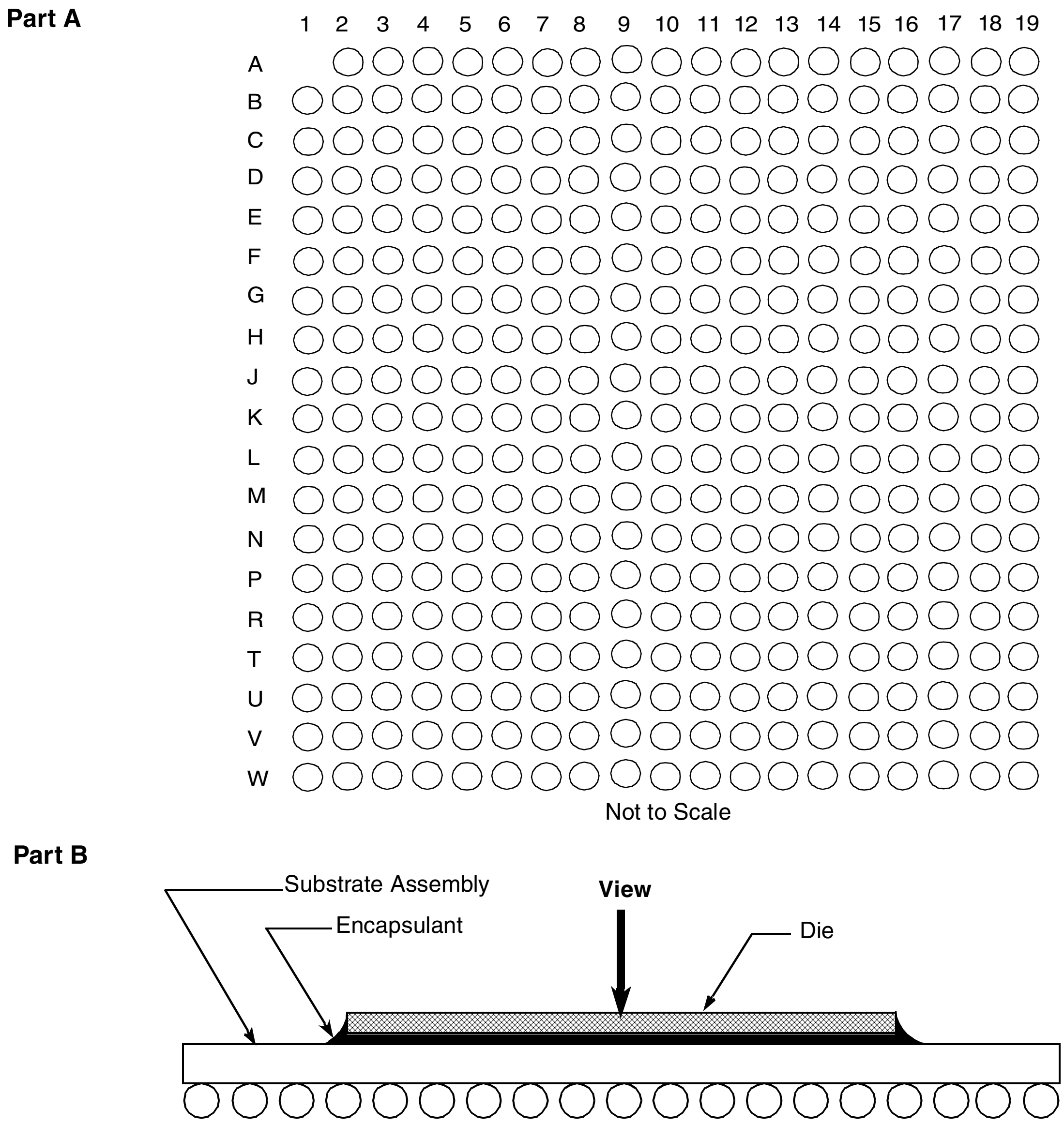


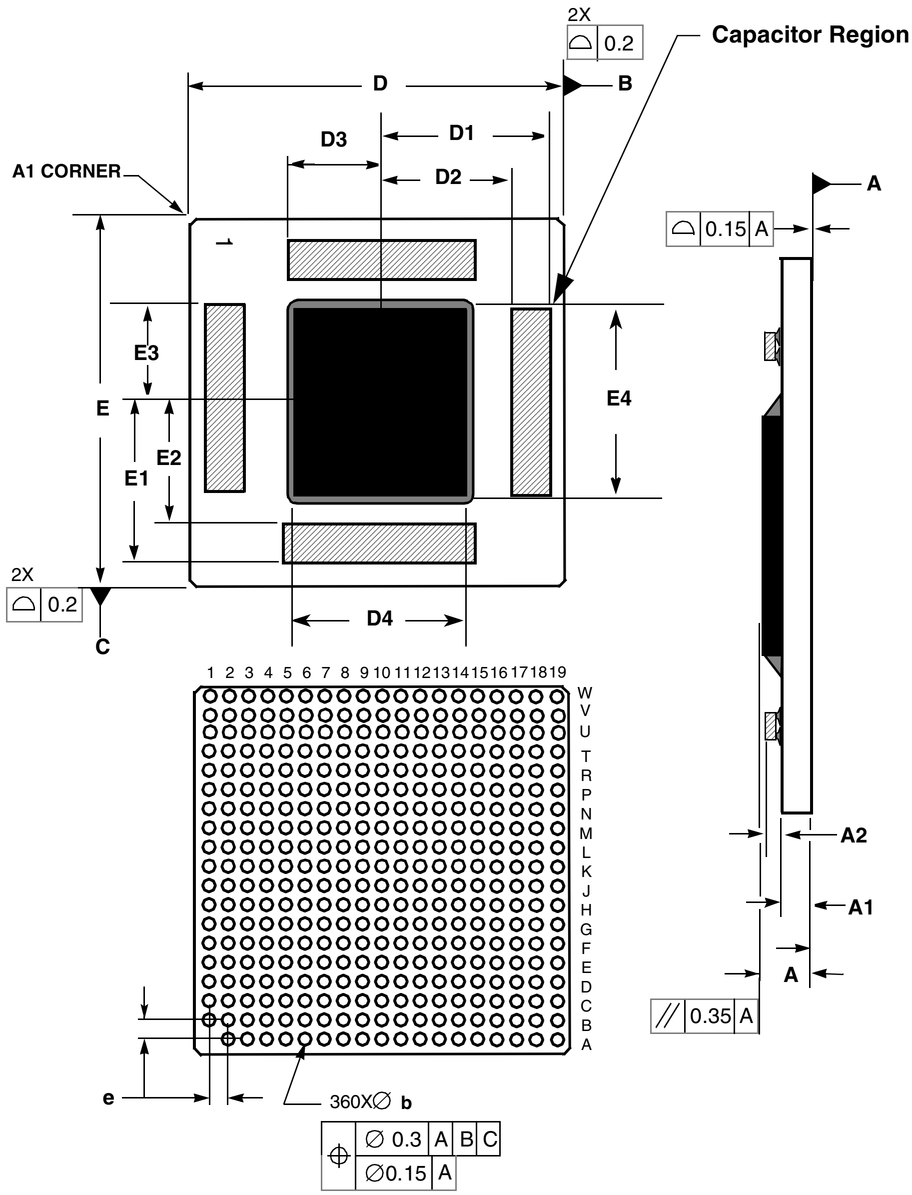
Figure 12. Pinout of the MPC7448, 360 HCTE Package as Viewed from the Top Surface

Table 11. Pinout Listing for the MPC7448, 360 HCTE Package (continued)

Signal Name	Pin Number	Active	I/O	Notes
$\overline{\text{LVRAM}}$	B10	—	—	12, 20, 22
NC (no connect)	A6, A14, A15, B14, B15, C14, C15, C16, C17, C18, C19, D14, D15, D16, D17, D18, D19, E14, E15, F14, F15, G14, G15, H15, H16, J15, J16, J17, J18, J19, K15, K16, K17, K18, K19, L15, L16, L17, L18, L19	—	—	11
$\overline{\text{LSSD\_MODE}}$	E8	Low	Input	6, 12
$\overline{\text{MCP}}$	C9	Low	Input	
OV_{DD}	B4, C2, C12, D5, F2, H3, J5, K2, L5, M3, N6, P2, P8, P11, R4, R13, R16, T6, T9, U2, U12, U16, V4, V7, V10, V14	—	—	
OVDD_SENSE	E18, G18	—	—	16
PLL_CFG[0:4]	B8, C8, C7, D7, A7	High	Input	
PLL_CFG[5]	D10	High	Input	9, 20
$\overline{\text{PMON\_IN}}$	D9	Low	Input	13
$\overline{\text{PMON\_OUT}}$	A9	Low	Output	
$\overline{\text{QACK}}$	G5	Low	Input	
$\overline{\text{QREQ}}$	P4	Low	Output	
$\overline{\text{SHD}}$[0:1]	E4, H5	Low	I/O	3
$\overline{\text{SMI}}$	F9	Low	Input	
$\overline{\text{SRESET}}$	A2	Low	Input	
SYSCLK	A10	—	Input	
$\overline{\text{TA}}$	K6	Low	Input	
TBEN	E1	High	Input	
$\overline{\text{TBST}}$	F11	Low	Output	
TCK	C6	High	Input	
TDI	B9	High	Input	6
TDO	A4	High	Output	
$\overline{\text{TEA}}$	L1	Low	Input	
TEMP_ANODE	N18	—	—	17
TEMP_CATHODE	N19	—	—	17
TMS	F1	High	Input	6
$\overline{\text{TRST}}$	A5	Low	Input	6, 14
$\overline{\text{TS}}$	L4	Low	I/O	3
TSIZ[0:2]	G6, F7, E7	High	Output	
TT[0:4]	E5, E6, F6, E9, C5	High	I/O	
$\overline{\text{WT}}$	D3	Low	Output	
V_{DD}	H8, H10, H12, J7, J9, J11, J13, K8, K10, K12, K14, L7, L9, L11, L13, M8, M10, M12	—	—	
V_{DD}	A13, A16, A18, B17, B19, C13, E13, E16, F12, F17, F19, G11, G16, H14, H17, H19, M14, M16, M18, N15, N17, P16, P18	—	—	15

8.4 Mechanical Dimensions for the MPC7448, 360 HCTE LGA

Figure 13 provides the mechanical dimensions and bottom surface nomenclature for the MPC7448, 360 HCTE LGA package.



NOTES:
1. Dimensioning and tolerancing per ASME Y14.5M, 1994
2. Dimensions in millimeters
3. Top side A1 corner index is a metalized feature with various shapes. Bottom side A1 corner is designated with a pad missing from the array.

	Millimeters	
Dim	**Min**	**Max**
A	1.52	1.80
A1	0.70	0.90
A2	—	0.6
b	0.82	0.93
D	25.00 BSC	
D1	—	11.3
D2	8.0	—
D3	—	6.5
D4	7.2	7.4
e	1.27 BSC	
E	25.00 BSC	
E1	—	11.3
E2	8.0	—
E3	—	6.5
E4	7.9	8.1

Figure 14. Mechanical Dimensions and Bottom Surface Nomenclature for the MPC7448, 360 HCTE LGA Package

9 System Design Information

This section provides system and thermal design requirements and recommendations for successful application of the MPC7448.

9.1 Clocks

The following sections provide more detailed information regarding the clocking of the MPC7448.

9.1.1 PLL Configuration

The MPC7448 PLL is configured by the PLL_CFG[0:5] signals. For a given SYSCLK (bus) frequency, the PLL configuration signals set the internal CPU and VCO frequency of operation. The PLL configuration for the MPC7448 is shown in Table 12. In this example, shaded cells represent settings that, for a given SYSCLK frequency, result in core and/or VCO frequencies that do not comply with Table 8. When enabled, dynamic frequency switching (DFS) also affects the core frequency by halving or quartering the bus-to-core multiplier; see Section 9.7.5, "Dynamic Frequency Switching (DFS)," for more information. Note that when DFS is enabled the resulting core frequency must meet the adjusted minimum core frequency requirements (f_{core_DFS}) described in Table 8. Note that the PLL_CFG[5] is currently used for factory test only and should be tied low, and that the MPC7448 PLL configuration settings are compatible with the MPC7447A PLL configuration settings when PLL_CFG[5] = 0.

Table 12. MPC7448 Microprocessor PLL Configuration Example

PLL_CFG[0:5]	Example Core and VCO Frequency in MHz										
	Bus-to-Core Multiplier [5]	Core-to-VCO Multiplier [5]	Bus (SYSCLK) Frequency								
			33.3 MHz	50 MHz	66.6 MHz	75 MHz	83 MHz	100 MHz	133 MHz	167 MHz	200 MHz
010000	2x [6]	1x									
100000	3x [6]	1x									600
101000	4x [6]	1x								667	800
101100	5x	1x							667	835	1000
100100	5.5x	1x							733	919	1100
110100	6x	1x						600	800	1002	1200
010100	6.5x	1x						650	866	1086	1300
001000	7x	1x						700	931	1169	1400
000100	7.5x	1x					623	750	1000	1253	1500
110000	8x	1x				600	664	800	1064	1336	1600
011000	8.5x	1x				638	706	850	1131	1417	1700
011110	9x	1x			600	675	747	900	1197	1500	
011100	9.5x	1x			633	712	789	950	1264	1583	
101010	10x	1x			667	750	830	1000	1333	1667	
100010	10.5x	1x			700	938	872	1050	1397		

9.4 Output Buffer DC Impedance

The MPC7448 processor bus drivers are characterized over process, voltage, and temperature. To measure Z_0, an external resistor is connected from the chip pad to OV_{DD} or GND. The value of each resistor is varied until the pad voltage is $OV_{DD}/2$. Figure 20 shows the driver impedance measurement.

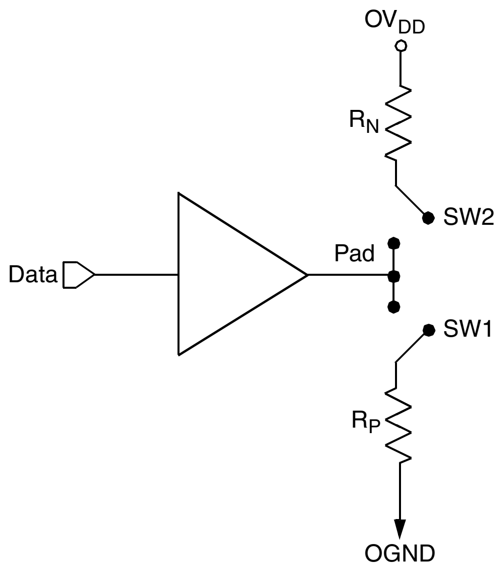


Figure 20. Driver Impedance Measurement

The output impedance is the average of two components—the resistances of the pull-up and pull-down devices. When data is held low, SW2 is closed (SW1 is open), and R_N is trimmed until the voltage at the pad equals $OV_{DD}/2$. R_N then becomes the resistance of the pull-down devices. When data is held high, SW1 is closed (SW2 is open), and R_P is trimmed until the voltage at the pad equals $OV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

Table 15 summarizes the signal impedance results. The impedance increases with junction temperature and is relatively unaffected by bus voltage.

Table 15. Impedance Characteristics

At recommended operating conditions. See Table 4

Impedance		Processor Bus	Unit
Z_0	Typical	33–42	Ω
	Maximum	31–51	Ω

9.5 Pull-Up/Pull-Down Resistor Requirements

The MPC7448 requires high-resistive (weak: 4.7-KΩ) pull-up resistors on several control pins of the bus interface to maintain the control signals in the negated state after they have been actively negated and released by the MPC7448 or other bus masters. These pins are: $\overline{\text{TS}}$, $\overline{\text{ARTRY}}$, $\overline{\text{SHDO}}$, and $\overline{\text{SHD1}}$.

Some pins designated as being factory test pins must be pulled up to OV_{DD} or down to GND to ensure proper device operation. The pins that must be pulled up to OV_{DD} are $\overline{\text{LSSD\_MODE}}$ and TEST[0:3]; the pins that must be pulled down to GND are L1_TSTCLK and TEST[4]. The $\overline{\text{CKSTP\_IN}}$ signal should likewise be pulled up through a pull-up resistor (weak or stronger: 4.7–1 KΩ) to prevent erroneous assertions of this signal.

In addition, the MPC7448 has one open-drain style output that requires a pull-up resistor (weak or stronger: 4.7–1 KΩ) if it is used by the system. This pin is $\overline{\text{CKSTP\_OUT}}$.

BVSEL0 and BVSEL1 should not be allowed to float, and should be configured either via pull-up or pull-down resistors or actively driven by external logic. If pull-down resistors are used to configure BVSEL0 or BVSEL1, the resistors should be less than 250 Ω (see Table 11). Because PLL_CFG[0:5] must remain stable during normal operation, strong pull-up and pull-down resistors (1 KΩ or less) are recommended to configure these signals in order to protect against erroneous switching due to ground bounce, power supply noise, or noise coupling.

During inactive periods on the bus, the address and transfer attributes may not be driven by any master and may, therefore, float in the high-impedance state for relatively long periods of time. Because the MPC7448 must continually monitor these signals for snooping, this float condition may cause excessive power draw by the input receivers on the MPC7448 or by other receivers in the system. These signals can be pulled up through weak (10-KΩ) pull-up resistors by the system, address bus driven mode enabled (see the *MPC7450 RISC Microprocessor Family Users' Manual* for more information on this mode), or they may be otherwise driven by the system during inactive periods of the bus to avoid this additional power draw. Preliminary studies have shown the additional power draw by the MPC7448 input receivers to be negligible and, in any event, none of these measures are necessary for proper device operation. The snooped address and transfer attribute inputs are: A[0:35], AP[0:4], TT[0:4], $\overline{\text{CI}}$, $\overline{\text{WT}}$, and $\overline{\text{GBL}}$.

If address or data parity is not used by the system, and respective parity checking is disabled through HID1, the input receivers for those pins are disabled and do not require pull-up resistors, therefore they may be left unconnected by the system. If extended addressing is not used (HID0[XAEN] = 0), A[0:3] are unused and must be pulled low to GND through weak pull-down resistors; additionally, if address parity checking is enabled (HID1[EBA] = 1) and extended addressing is not used, AP[0] must be pulled up to OV_{DD} through a weak pull-up resistor. If the MPC7448 is in 60x bus mode, DTI[0:3] must be pulled low to GND through weak pull-down resistors.

The data bus input receivers are normally turned off when no read operation is in progress and, therefore, do not require pull-up resistors on the bus. Other data bus receivers in the system, however, may require pull-ups or require that those signals be otherwise driven by the system during inactive periods. The data bus signals are D[0:63] and DP[0:7].

9.6 JTAG Configuration Signals

Boundary-scan testing is enabled through the JTAG interface signals. The $\overline{\text{TRST}}$ signal is optional in the IEEE 1149.1 standard specification, but is typically provided on all processors that implement the PowerPC architecture. While it is possible to force the TAP controller to the reset state using only the TCK and TMS signals, more reliable power-on reset performance will be obtained if the $\overline{\text{TRST}}$ signal is asserted during power-on reset. Because the JTAG interface is also used for accessing the common on-chip processor (COP) function, simply tying $\overline{\text{TRST}}$ to $\overline{\text{HRESET}}$ is not practical.

The COP function of these processors allows a remote computer system (typically a PC with dedicated hardware and debugging software) to access and control the internal operations of the processor. The COP interface connects primarily through the JTAG port of the processor, with some additional status monitoring signals. The COP port requires the ability to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$ in order

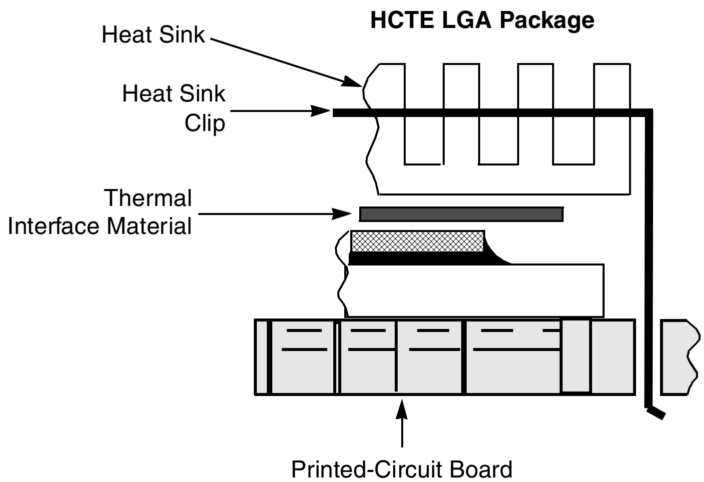


Figure 23. LGA Package Exploded Cross-Sectional View with Several Heat Sink Options

There are several commercially-available heat sinks for the MPC7448 provided by the following vendors:

Aavid Thermalloy 80 Commercial St. Concord, NH 03301 Internet: www.aavidthermalloy.com	603-224-9988
Alpha Novatech 473 Sapena Ct. #12 Santa Clara, CA 95054 Internet: www.alphanovatech.com	408-567-8082
Calgreg Thermal Solutions 60 Alhambra Road, Suite 1 Warwick, RI 02886 Internet: www.calgregthermalsolutions.com	888-732-6100
International Electronic Research Corporation (IERC) 413 North Moss St. Burbank, CA 91502 Internet: www.ctscorp.com	818-842-7277
Tyco Electronics Chip Coolers™ P.O. Box 3668 Harrisburg, PA 17105-3668 Internet: www.tycoelectronics.com	800-522-6752
Wakefield Engineering 33 Bridge St. Pelham, NH 03076 Internet: www.wakefield.com	603-635-2800

Ultimately, the final selection of an appropriate heat sink depends on many factors, such as thermal performance at a given air velocity, spatial volume, mass, attachment method, assembly, and cost.

of any thermal interface material depends on many factors—thermal performance requirements, manufacturability, service temperature, dielectric properties, cost, and so on.

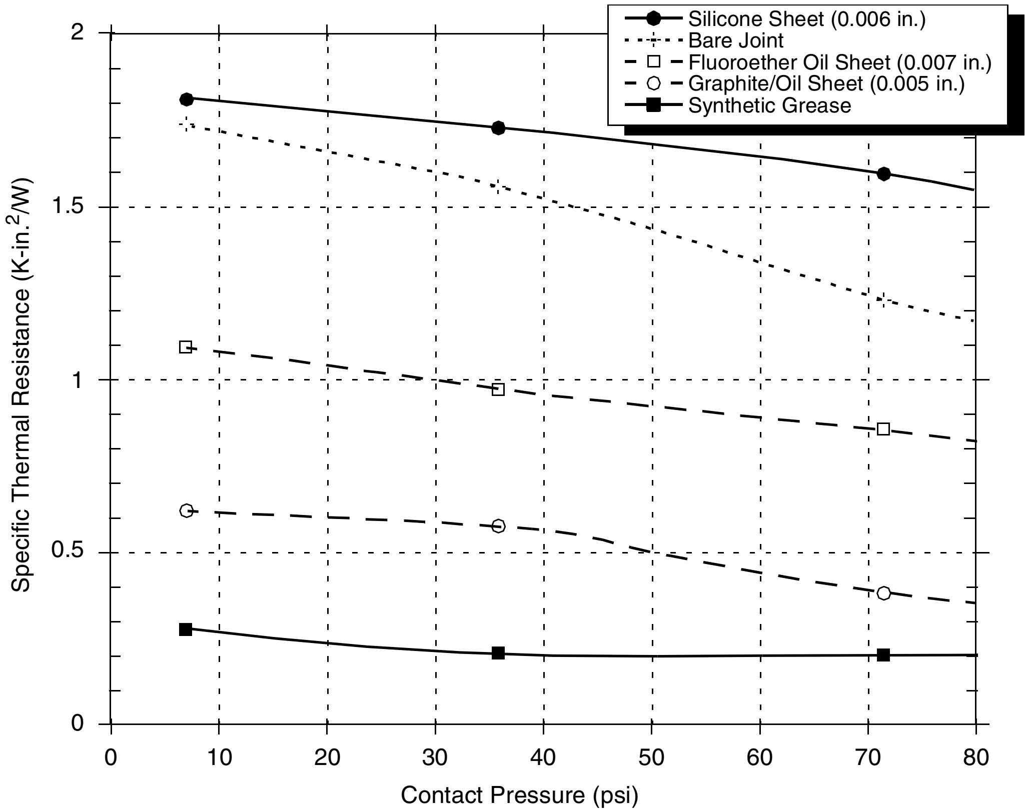


Figure 25. Thermal Performance of Select Thermal Interface Material

The board designer can choose between several types of thermal interfaces. Heat sink adhesive materials should be selected based on high conductivity and mechanical strength to meet equipment shock/vibration requirements. There are several commercially available thermal interfaces and adhesive materials provided by the following vendors:

The Bergquist Company 18930 West 78th St. Chanhassen, MN 55317 Internet: www.bergquistcompany.com	800-347-4572
Chomerics, Inc. 77 Dragon Ct. Woburn, MA 01801 Internet: www.chomerics.com	781-935-4850
Dow-Corning Corporation Corporate Center P.O. Box 994. Midland, MI 48686-0994 Internet: www.dowcorning.com	800-248-2481

Due to the complexity and variety of system-level boundary conditions for today's microelectronic equipment, the combined effects of the heat transfer mechanisms (radiation, convection, and conduction) may vary widely. For these reasons, we recommend using conjugate heat transfer models for the board as well as system-level designs.

For system thermal modeling, the MPC7448 thermal model is shown in Figure 26. Four volumes represent this device. Two of the volumes, solder ball-air and substrate, are modeled using the package outline size of the package. The other two, die and bump-underfill, have the same size as the die. The silicon die should be modeled $8.0 \times 7.3 \times 0.86$ mm^3 with the heat source applied as a uniform source at the bottom of the volume. The bump and underfill layer is modeled as $8.0 \times 7.3 \times 0.07$ mm^{3}collapsed in the z-direction with a thermal conductivity of 5.0 W/(m • K) in the z-direction. The substrate volume is $25 \times 25 \times 1.14$ mm^3 and has 9.9 W/(m • K) isotropic conductivity in the xy-plane and 2.95 W/(m • K) in the direction of the z-axis. The solder ball and air layer are modeled with the same horizontal dimensions as the substrate and is 0.8 mm thick. For the LGA package the solder and air layer is 0.1 mm thick, but the material properties are the same. It can also be modeled as a collapsed volume using orthotropic material properties: 0.034 W/(m • K) in the xy-plane direction and 11.2 W/(m • K) in the direction of the z-axis.

Conductivity	Value	Unit
Die ($8.0 \times 7.3 \times 0.86$ mm^3)		
Silicon	Temperature-dependent	W/(m • K)
Bump and Underfill ($8.0 \times 7.3 \times 0.07$ mm^3)		
k_z	5.0	W/(m • K)
Substrate ($25 \times 25 \times 1.14$ mm^3)		
k_x	9.9	W/(m • K)
k_y	9.9	
k_z	2.95	
Solder Ball and Air ($25 \times 25 \times 0.8$ mm^3)		
k_x	0.034	W/(m • K)
k_y	0.034	
k_z	11.2	

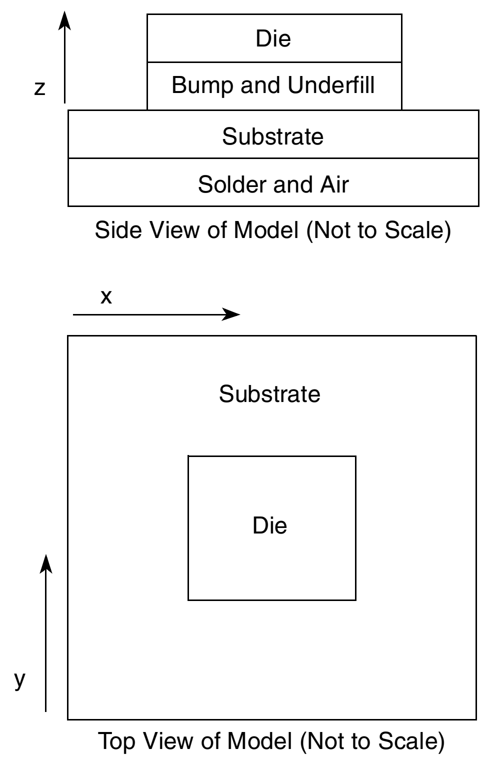


Figure 26. Recommended Thermal Model of MPC7448

9.7.4 Temperature Diode

The MPC7448 has a temperature diode on the microprocessor that can be used in conjunction with other system temperature monitoring devices (such as Analog Devices, ADT7461™). These devices use the negative temperature coefficient of a diode operated at a constant current to determine the temperature of the microprocessor and its environment. For proper operation, the monitoring device used should auto-calibrate the device by canceling out the V_{BE} variation of each MPC7448's internal diode.

The following are the specifications of the MPC7448 on-board temperature diode:

$V_f > 0.40$ V

$V_f < 0.90$ V

Operating range 2–300 µA

Diode leakage < 10 nA @ 125°C

Ideality factor over 5–150 µA at 60°C: n = 1.0275 ± 0.9%

Ideality factor is defined as the deviation from the ideal diode equation:

$$I_{fw} = I_s \left[e^{\frac{qV_f}{nKT}} - 1 \right]$$

Another useful equation is:

$$V_H - V_L = n\frac{KT}{q}\left[\ln\frac{I_H}{I_L} \right] - 1$$

Where:

I_{fw} = Forward current

I_s = Saturation current

V_d = Voltage at diode

V_f = Voltage forward biased

V_H = Diode voltage while I_H is flowing

V_L = Diode voltage while I_L is flowing

I_H = Larger diode bias current

I_L = Smaller diode bias current

q = Charge of electron (1.6×10^{-19} C)

n = Ideality factor (normally 1.0)

K = Boltzman's constant (1.38×10^{-23} Joules/K)

T = Temperature (Kelvins)

The ratio of I_H to I_L is usually selected to be 10:1. The previous equation simplifies to the following:

$$V_H - V_L = 1.986 \times 10^{-4} \times nT$$

Table 16. Valid Divide Ratio Configurations

DFS mode disabled		DFS divide-by-2 mode enabled (HID1[DFS2] = 1 or $\overline{DFS2}$ = 0)		DFS divide-by-4 mode enabled (HID1[DFS4] = 1 or $\overline{DFS4}$ = 0)	
Bus-to-Core Multiplier Configured by PLL_CFG[0:5] (see Table 12)	HID1[PC0-5] [3]	Bus-to-Core Multiplier	HID1[PC0-5] [3]	Bus-to-Core Multiplier	HID1[PC0-5] [3]
2x [4]	010000	N/A (unchanged) [1]	unchanged [1]	N/A (unchanged) [1]	unchanged [1]
3x [4]	100000	N/A (unchanged) [1]	unchanged [1]	N/A (unchanged) [1]	unchanged [1]
4x [4]	101000	2x [4]	010000	N/A (unchanged) [1]	unchanged [1]
5x	101100	2.5x [4]	010101	N/A (unchanged) [1]	unchanged [1]
5.5x	100100	2.75x [4]	110101 [2]	N/A (unchanged) [1]	unchanged [1]
6x	110100	3x [4]	100000	N/A (unchanged) [1]	unchanged [1]
6.5x	010100	3.25x [4]	100000 [2]	N/A (unchanged) [1]	unchanged [1]
7x	001000	3.5x [4]	110101	N/A (unchanged) [1]	unchanged [1]
7.5x	000100	3.75x [4]	110101 [2]	N/A (unchanged) [1]	unchanged [1]
8x	110000	4x [4]	101000 [4]	2x [4]	010000
8.5x	011000	4.25x [4]	101000 [2]	N/A (unchanged) [1]	unchanged [1]
9x	011110	4.5x [4]	011101	2.25x [4]	010000 [2]
9.5x	011100	4.75x [4]	011101 [2]	N/A (unchanged) [1]	unchanged [1]
10x	101010	5x	101100	2.5x [4]	010101
10.5x	100010	5.25x	101100 [2]	N/A (unchanged) [1]	unchanged [1]
11x	100110	5.5x	100100	2.75x [4]	010101 [2]
11.5x	000000	5.75x	100100 [2]	N/A (unchanged) [1]	unchanged [1]
12x	101110	6x	110100	3x [4]	100000
12.5x	111110	6.25x	110100 [2]	N/A (unchanged) [1]	unchanged [1]
13x	010110	6.5x	010100	3.25x [4]	100000 [2]
13.5x	111000	6.75	010100 [2]	N/A (unchanged) [1]	unchanged [1]
14x	110010	7x	001000	3.5x [4]	110101
15x	000110	7.5x	000100	3.75x [4]	110101 [2]
16x	110110	8x	110000	4x [4]	101000
17x	000010	8.5x	011000	4.25x [4]	101000 [2]
18x	001010	9x	011110	4.5x [4]	011101
20x	001110	10x	101010	5x	101100
21x	010010	10.5x	100010	5.25x	101100 [2]

Table 16. Valid Divide Ratio Configurations (continued)

DFS mode disabled		DFS divide-by-2 mode enabled (HID1[DFS2] = 1 or $\overline{DFS2}$ = 0)		DFS divide-by-4 mode enabled (HID1[DFS4] = 1 or $\overline{DFS4}$ = 0)	
Bus-to-Core Multiplier Configured by PLL_CFG[0:5] (see Table 12)	HID1[PC0-5] [3]	Bus-to-Core Multiplier	HID1[PC0-5] [3]	Bus-to-Core Multiplier	HID1[PC0-5] [3]
24x	011010	12x	101110	6x	110100
28x	111010	14x	110010	7x	001000

Notes:
1. DFS mode is not supported for this combination of DFS mode and PLL_CFG[0:5] setting. As a result, the processor will ignore these settings and remain at the previous multiplier, as reflected by the HID1[PC0-PC5] bits.
2. Though supported by the MPC7448 clock circuitry, multipliers of *n*.25x and *n*.75x cannot be expressed as valid PLL configuration codes. As a result, the values displayed in HID1[PC0-PC5] are rounded down to the nearest valid PLL configuration code. However, the actual bus-to-core multiplier is as stated in this table.
3. Note that in the HID1 register of the MPC7448, the PC0, PC1, PC2, PC3, PC4, and PC5 bits are bits 15, 16, 17, 18, 19, and 14 (respectively). See the *MPC7450 RISC Microprocessor Reference Manual* for more information.
4. Special considerations regarding snooped transactions must be observed for bus-to-core multipliers less than 5x. See the *MPC7450 RISC Microprocessor Reference Manual* for more information.

9.7.5.3 Minimum Core Frequency Requirements with DFS

In many systems, enabling DFS can result in very low processor core frequencies. However, care must be taken to ensure that the resulting processor core frequency is within the limits specified in Table 8. Proper operation of the device is not guaranteed at core frequencies below the specified minimum f_{core}.

10 Document Revision History

Table 17 provides a revision history for this hardware specification.

Table 17. Document Revision History

Revision	Date	Substantive Change(s)
4	3/2007	Table 19: Added 800 MHz processor frequency.
3	10/2006	Section 9.7, "Power and Thermal Management Information": Updated contact information. Table 18, Table 20, and Table 19: Added Revision D PVR. Table 19: Added 600 processor frequency, additional product codes, date codes for 1400 processor frequency, and footnotes 1 and 2. Table 20: Added PPC product code and footnote 1. Table 19 and Table 20: Added Revision D information for 1267 processor frequency.

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