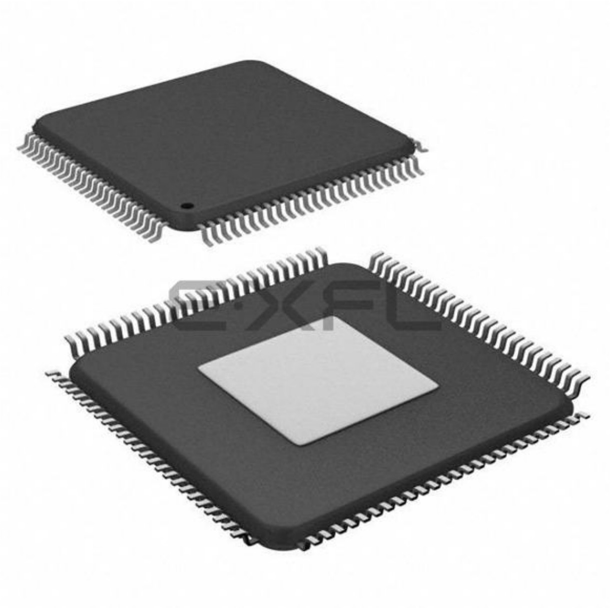
Infineon Technologies - XE164HN16F80LAAFXUMA1 Datasheet



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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	C166SV2
Core Size	16-Bit
Speed	80MHz
Connectivity	EBI/EMI, I²C, LINbus, SPI, SSC, UART/USART, USI
Peripherals	I²S, POR, PWM, WDT
Number of I/O	75
Program Memory Size	128KB (128K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	18K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 16x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP Exposed Pad
Supplier Device Package	PG-LQFP-100-8
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/xe164hn16f80laafxuma1

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16-Bit Single-Chip
Real Time Signal Controller

XE164xN (XE166 Family)

1 Summary of Features

For a quick overview and easy reference, the features of the XE164xN are summarized here.

- High-performance CPU with five-stage pipeline and MPU
  - 12.5 ns instruction cycle @ 80 MHz CPU clock (single-cycle execution)
  - One-cycle 32-bit addition and subtraction with 40-bit result
  - One-cycle multiplication (16 × 16 bit)
  - Background division (32 / 16 bit) in 21 cycles
  - One-cycle multiply-and-accumulate (MAC) instructions
  - Enhanced Boolean bit manipulation facilities
  - Zero-cycle jump execution
  - Additional instructions to support HLL and operating systems
  - Register-based design with multiple variable register banks
  - Fast context switching support with two additional local register banks
  - 16 Mbytes total linear address space for code and data
  - 1,024 Bytes on-chip special function register area (C166 Family compatible)
  - Integrated Memory Protection Unit (MPU)
- Interrupt system with 16 priority levels providing 96 interrupt nodes
  - Selectable external inputs for interrupt generation and wake-up
  - Fastest sample-rate 12.5 ns
- Eight-channel interrupt-driven single-cycle data transfer with Peripheral Event Controller (PEC), 24-bit pointers cover total address space
- Clock generation from internal or external clock sources, using on-chip PLL or prescaler
- Hardware CRC-Checker with Programmable Polynomial to Supervise On-Chip Memory Areas
- On-chip memory modules
  - 8 Kbytes on-chip stand-by RAM (SBRAM)
  - 2 Kbytes on-chip dual-port RAM (DPRAM)
  - Up to 16 Kbytes on-chip data SRAM (DSRAM)
  - Up to 16 Kbytes on-chip program/data SRAM (PSRAM)
  - Up to 320 Kbytes on-chip program memory (Flash memory)
  - Memory content protection through Error Correction Code (ECC)

The XE164xN types are offered with several SRAM memory sizes. **Figure 1** shows the allocation rules for PSRAM and DSRAM. Note that the rules differ:

- PSRAM allocation starts from the **lower** address
- DSRAM allocation starts from the **higher** address

For example 8 Kbytes of PSRAM will be allocated at E0’0000h-E0’1FFFh and 8 Kbytes of DSRAM will be at 00’C000h-00’DFFFh.

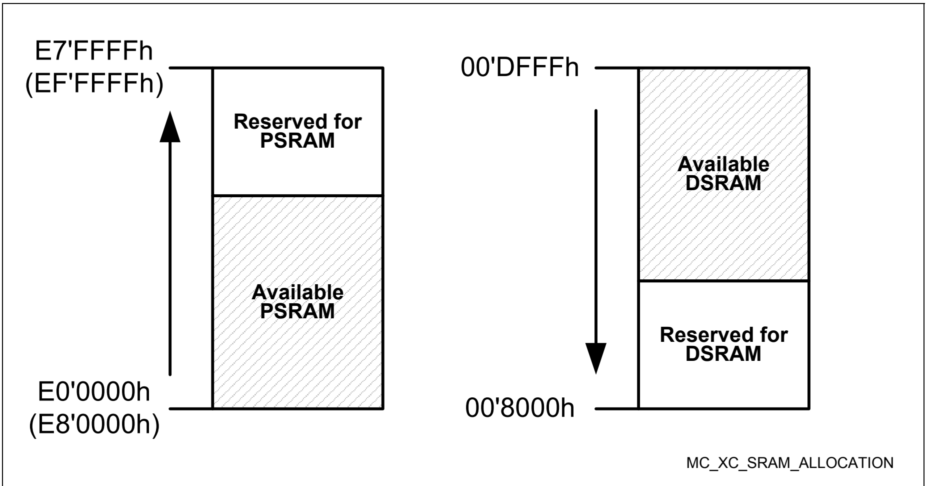

Figure 1 SRAM Allocation

Table 5 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
39	P2.0	O0 / I	St/B	**Bit 0 of Port 2, General Purpose Input/Output**
	AD13	OH / IH	St/B	**External Bus Interface Address/Data Line 13**
	RxDC0C	I	St/B	**CAN Node 0 Receive Data Input**
	T5INB	I	St/B	**GPT12E Timer T5 Count/Gate Input**
40	P2.1	O0 / I	St/B	**Bit 1 of Port 2, General Purpose Input/Output**
	TxDC0	O1	St/B	**CAN Node 0 Transmit Data Output**
	AD14	OH / IH	St/B	**External Bus Interface Address/Data Line 14**
	T5EUDB	I	St/B	**GPT12E Timer T5 External Up/Down Control Input**
	ESR1_5	I	St/B	**ESR1 Trigger Input 5**
41	P2.2	O0 / I	St/B	**Bit 2 of Port 2, General Purpose Input/Output**
	TxDC1	O1	St/B	**CAN Node 1 Transmit Data Output**
	AD15	OH / IH	St/B	**External Bus Interface Address/Data Line 15**
	ESR2_5	I	St/B	**ESR2 Trigger Input 5**
42	P4.0	O0 / I	St/B	**Bit 0 of Port 4, General Purpose Input/Output**
	CC2_CC24	O3 / I	St/B	**CAPCOM2 CC24IO Capture Inp./ Compare Out.**
	$\overline{\text{CS0}}$	OH	St/B	**External Bus Interface Chip Select 0 Output**
43	P2.3	O0 / I	St/B	**Bit 3 of Port 2, General Purpose Input/Output**
	U0C0_DOUT	O1	St/B	**USIC0 Channel 0 Shift Data Output**
	CC2_CC16	O3 / I	St/B	**CAPCOM2 CC16IO Capture Inp./ Compare Out.**
	A16	OH	St/B	**External Bus Interface Address Line 16**
	ESR2_0	I	St/B	**ESR2 Trigger Input 0**
	U0C0_DX0E	I	St/B	**USIC0 Channel 0 Shift Data Input**
	U0C1_DX0D	I	St/B	**USIC0 Channel 1 Shift Data Input**
	RxDC0A	I	St/B	**CAN Node 0 Receive Data Input**

Table 5 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
92	P1.5	O0 / I	St/B	**Bit 5 of Port 1, General Purpose Input/Output**
	U1C1_SELO3	O2	St/B	**USIC1 Channel 1 Select/Control 3 Output**
	$\overline{\text{BRKOUT}}$	O3	St/B	**OCDS Break Signal Output**
	A13	OH	St/B	**External Bus Interface Address Line 13**
	U2C0_DX0C	I	St/B	**USIC2 Channel 0 Shift Data Input**
93	P1.6	O0 / I	St/B	**Bit 6 of Port 1, General Purpose Input/Output**
	U1C1_SELO2	O2	St/B	**USIC1 Channel 1 Select/Control 2 Output**
	U2C0_DOUT	O3	St/B	**USIC2 Channel 0 Shift Data Output**
	A14	OH	St/B	**External Bus Interface Address Line 14**
	U2C0_DX0D	I	St/B	**USIC2 Channel 0 Shift Data Input**
94	P1.7	O0 / I	St/B	**Bit 7 of Port 1, General Purpose Input/Output**
	U1C1_MCLKOUT	O2	St/B	**USIC1 Channel 1 Master Clock Output**
	U2C0_SCLKOUT	O3	St/B	**USIC2 Channel 0 Shift Clock Output**
	A15	OH	St/B	**External Bus Interface Address Line 15**
	U2C0_DX1C	I	St/B	**USIC2 Channel 0 Shift Clock Input**
95	XTAL2	O	Sp/M	**Crystal Oscillator Amplifier Output**
96	XTAL1	I	Sp/M	**Crystal Oscillator Amplifier Input** To clock the device from an external source, drive XTAL1, while leaving XTAL2 unconnected. Voltages on XTAL1 must comply to the core supply voltage V_{DDIM}.
	ESR2_9	I	St/B	**ESR2 Trigger Input 9**
97	$\overline{\text{PORST}}$	I	In/B	**Power On Reset Input** A low level at this pin resets the XE164xN completely. A spike filter suppresses input pulses <10 ns. Input pulses >100 ns safely pass the filter. The minimum duration for a safe recognition should be 120 ns. An internal pull-up device will hold this pin high when nothing is driving it.

2.2 Identification Registers

The identification registers describe the current version of the XE164xN and of its modules.

Table 6 XE164xN Identification Registers

Short Name	Value	Address	Notes
SCU_IDMANUF	1820_H	$00'F07E_H$	
SCU_IDCHIP	3001_H	$00'F07C_H$	marking EES-AA or ES-AA
	3002_H	$00'F07C_H$	marking AA, AB
SCU_IDMEM	$304F_H$	$00'F07A_H$	
SCU_IDPROG	1313_H	$00'F078_H$	
JTAG_ID	$0018'B083_H$	---	marking EES-AA or ES-AA
	$1018'B083_H$	---	marking AA, AB

3 Functional Description

The architecture of the XE164xN combines advantages of RISC, CISC, and DSP processors with an advanced peripheral subsystem in a well-balanced design. On-chip memory blocks allow the design of compact systems-on-silicon with maximum performance suited for computing, control, and communication.

The on-chip memory blocks (program code memory and SRAM, dual-port RAM, data SRAM) and the generic peripherals are connected to the CPU by separate high-speed buses. Another bus, the LXBus, connects additional on-chip resources and external resources. This bus structure enhances overall system performance by enabling the concurrent operation of several subsystems of the XE164xN.

The block diagram gives an overview of the on-chip components and the advanced internal bus structure of the XE164xN.

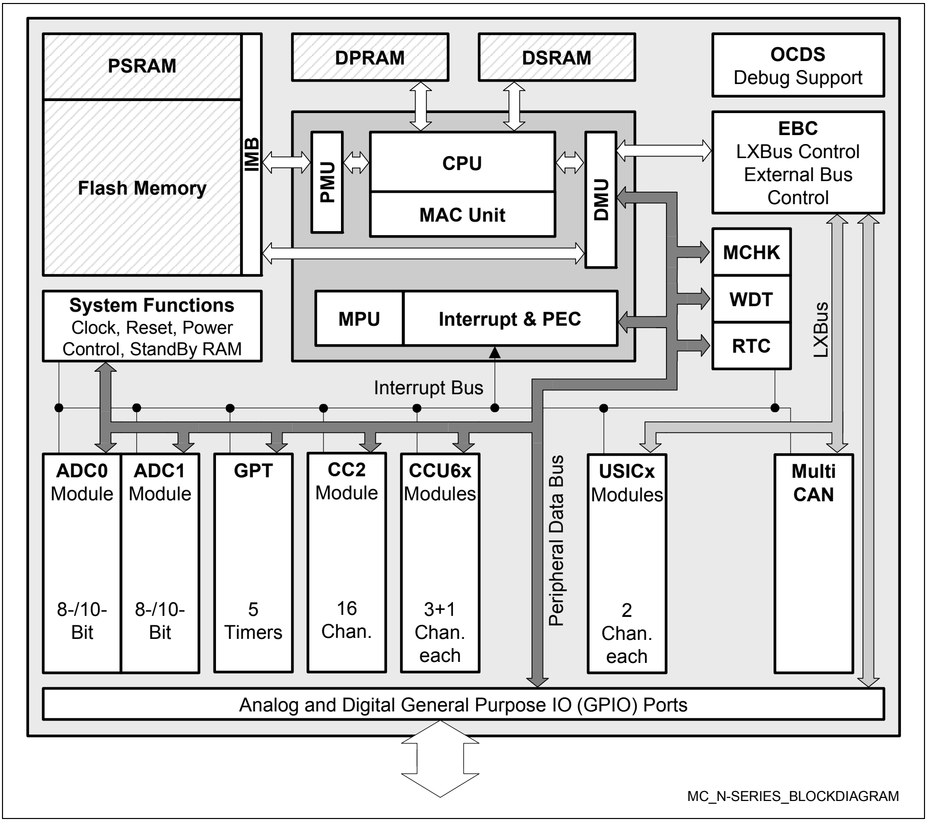

Figure 4 Block Diagram

Memory Content Protection

The contents of on-chip memories can be protected against soft errors (induced e.g. by radiation) by activating the parity mechanism or the Error Correction Code (ECC).
The parity mechanism can detect a single-bit error and prevent the software from using incorrect data or executing incorrect instructions.
The ECC mechanism can detect and automatically correct single-bit errors. This supports the stable operation of the system.

It is strongly recommended to activate the ECC mechanism wherever possible because this dramatically increases the robustness of an application against such soft errors.

Table 8 Compare Modes (cont'd)

Compare Modes	Function
Mode 2	Interrupt-only compare mode; Only one compare interrupt per timer period is generated
Mode 3	Pin set '1' on match; pin reset '0' on compare timer overflow; Only one compare event per timer period is generated
Double Register Mode	Two registers operate on one pin; Pin toggles on each compare match; Several compare events per timer period are possible
Single Event Mode	Generates single edges or pulses; Can be used with any compare mode

When a capture/compare register has been selected for capture mode, the current contents of the allocated timer will be latched ('captured') into the capture/compare register in response to an external event at the port pin associated with this register. In addition, a specific interrupt request for this capture/compare register is generated. Either a positive, a negative, or both a positive and a negative external signal transition at the pin can be selected as the triggering event.

The contents of all registers selected for one of the five compare modes are continuously compared with the contents of the allocated timers.

When a match occurs between the timer value and the value in a capture/compare register, specific actions will be taken based on the compare mode selected.

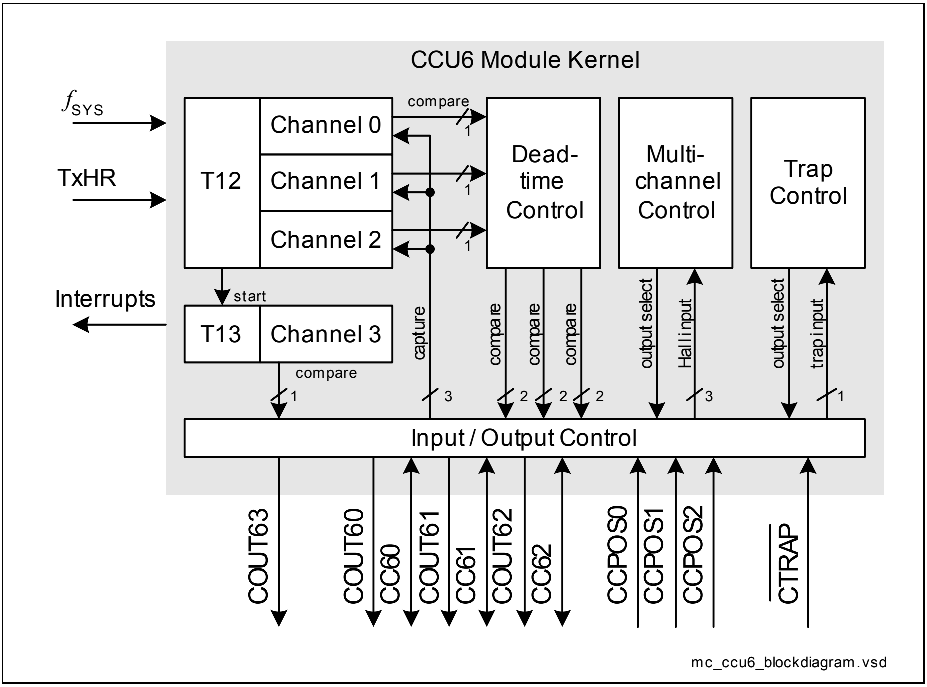

Figure 7 CCU6 Block Diagram

Timer T12 can work in capture and/or compare mode for its three channels. The modes can also be combined. Timer T13 can work in compare mode only. The multi-channel control unit generates output patterns that can be modulated by timer T12 and/or timer T13. The modulation sources can be selected and combined for signal modulation.

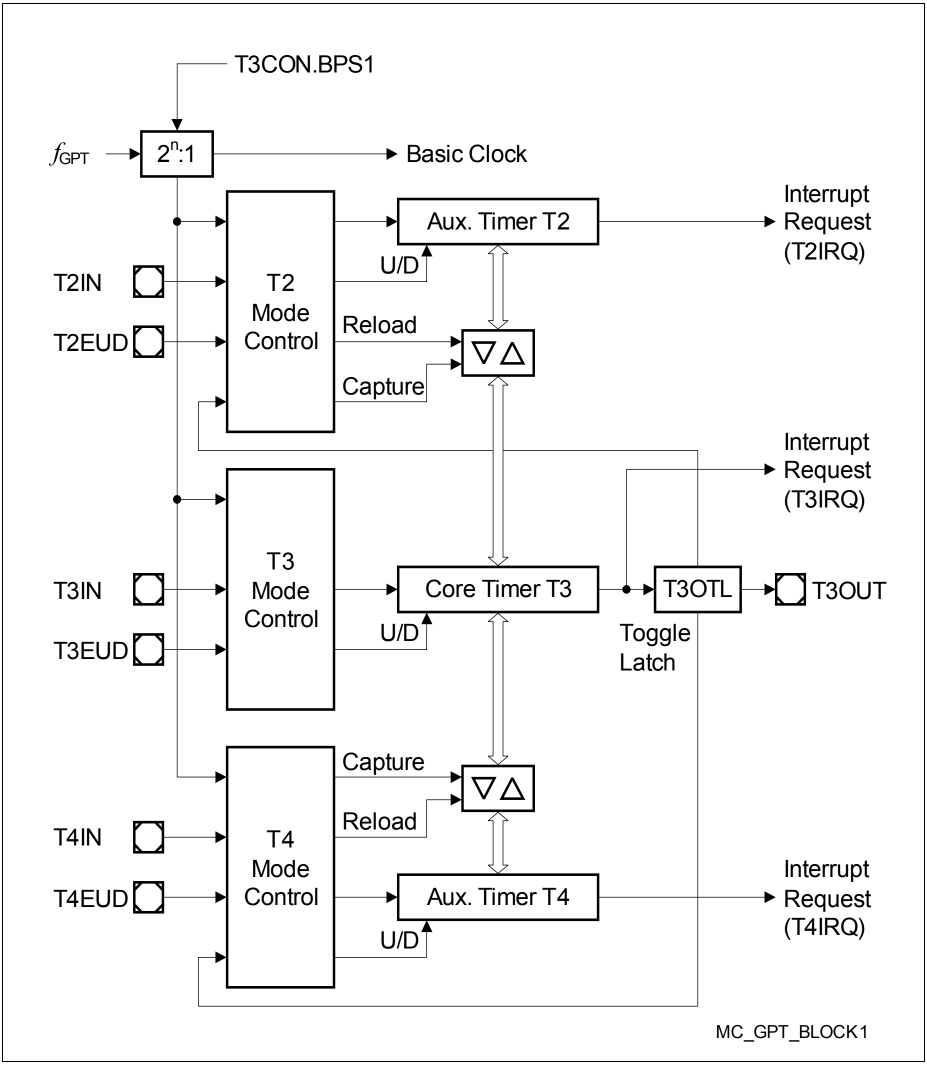

Figure 8 Block Diagram of GPT1

3.11 Real Time Clock

The Real Time Clock (RTC) module of the XE164xN can be clocked with a clock signal selected from internal sources or external sources (pins).

The RTC basically consists of a chain of divider blocks:

- Selectable 32:1 and 8:1 dividers (on - off)
- The reloadable 16-bit timer T14
- The 32-bit RTC timer block (accessible via registers RTCH and RTCL) consisting of:
  – a reloadable 10-bit timer
  – a reloadable 6-bit timer
  – a reloadable 6-bit timer
  – a reloadable 10-bit timer

All timers count up. Each timer can generate an interrupt request. All requests are combined to a common node request.

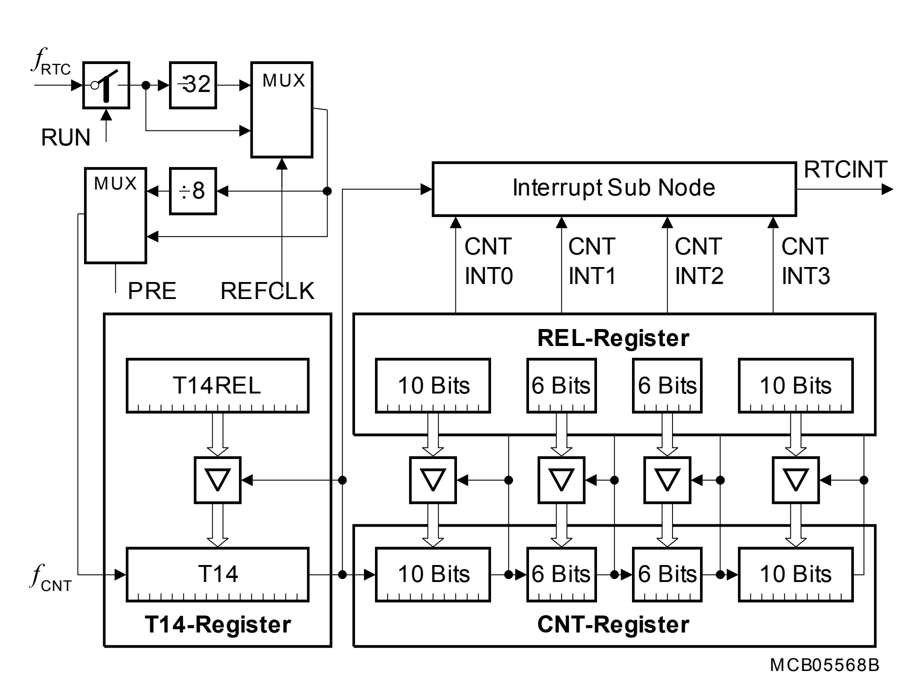

Figure 10 RTC Block Diagram

Note: The registers associated with the RTC are only affected by a power reset.

3.14 MultiCAN Module

The MultiCAN module contains independently operating CAN nodes with Full-CAN functionality which are able to exchange Data and Remote Frames using a gateway function. Transmission and reception of CAN frames is handled in accordance with CAN specification V2.0 B (active). Each CAN node can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers.

All CAN nodes share a common set of message objects. Each message object can be individually allocated to one of the CAN nodes. Besides serving as a storage container for incoming and outgoing frames, message objects can be combined to build gateways between the CAN nodes or to set up a FIFO buffer.

Note: The number of CAN nodes and message objects depends on the selected device type.

The message objects are organized in double-chained linked lists, where each CAN node has its own list of message objects. A CAN node stores frames only into message objects that are allocated to its own message object list and it transmits only messages belonging to this message object list. A powerful, command-driven list controller performs all message object list operations.

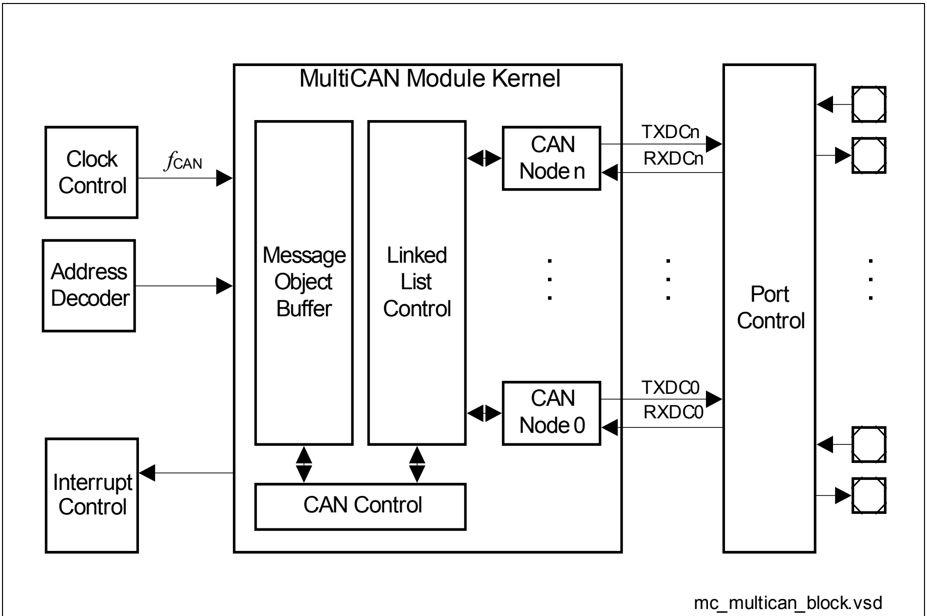

Figure 12 Block Diagram of MultiCAN Module

Table 10 Instruction Set Summary (cont'd)

Mnemonic	Description	Bytes
NOP	Null operation	2
CoMUL/CoMAC	Multiply (and accumulate)	4
CoADD/CoSUB	Add/Subtract	4
Co(A)SHR	(Arithmetic) Shift right	4
CoSHL	Shift left	4
CoLOAD/STORE	Load accumulator/Store MAC register	4
CoCMP	Compare	4
CoMAX/MIN	Maximum/Minimum	4
CoABS/CoRND	Absolute value/Round accumulator	4
CoMOV	Data move	4
CoNEG/NOP	Negate accumulator/Null operation	4

1) The Enter Power Down Mode instruction is not used in the XE164xN, due to the enhanced power control scheme. PWRDN will be correctly decoded, but will trigger no action.

4.1.1 Operating Conditions

The following operating conditions must not be exceeded to ensure correct operation of the XE164xN. All parameters specified in the following sections refer to these operating conditions, unless otherwise noticed.

Note: Typical parameter values refer to room temperature and nominal supply voltage, minimum/maximum parameter values also include conditions of minimum/maximum temperature and minimum/maximum supply voltage. Additional details are described where applicable.

Table 12 Operating Conditions

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Voltage Regulator Buffer Capacitance for DMP_M	C_{EVRM} SR	1.0	–	4.7	µF	[1]
Voltage Regulator Buffer Capacitance for DMP_1	C_{EVR1} SR	0.47	–	2.2	µF	[2][1]
External Load Capacitance	C_L SR	–	20[3]	–	pF	pin out driver= default [4]
System frequency	f_{SYS} SR	–	–	80	MHz	[5]
Overload current for analog inputs[6]	I_{OVA} SR	-2	–	5	mA	not subject to production test
Overload current for digital inputs[6]	I_{OVD} SR	-5	–	5	mA	not subject to production test
Overload current coupling factor for analog inputs[7]	K_{OVA} CC	–	2.5 x 10^{-4}	1.5 x 10^{-3}	-	I_{OV}< 0 mA; not subject to production test
		–	1.0 x 10^{-6}	1.0 x 10^{-4}	-	I_{OV}> 0 mA; not subject to production test

4.3.2 DC Parameters for Lower Voltage Area

Keeping signal levels within the limits specified in this table ensures operation without overload conditions. For signal levels outside these specifications, also refer to the specification of the overload current I_{OV}.

Note: Operating Conditions apply.

Table 16 is valid under the following conditions: $V_{DDP} \geq 3.0$ V; V_{DDP} typ. 3.3 V; $V_{DDP} \leq 4.5$ V

Table 16 DC Characteristics for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Pin capacitance (digital inputs/outputs). To be doubled for double bond pins.[1]	C_{IO} CC	–	–	10	pF	not subject to production test
Input Hysteresis[2]	HYS CC	0.07 x V_{DDP}	–	–	V	R_S = 0 Ohm
Absolute input leakage current on pins of analog ports[3]	$\lvert I_{OZ1}\rvert$ CC	–	10	200	nA	$V_{IN} > V_{SS}$; $V_{IN} < V_{DDP}$
Absolute input leakage current for all other pins. To be doubled for double bond pins.[3)1)4]	$\lvert I_{OZ2}\rvert$ CC	–	0.2	2.5	µA	$T_J \leq 110$ °C; $V_{IN} > V_{SS}$; $V_{IN} < V_{DDP}$
		–	0.2	8	µA	$T_J \leq 150$ °C; $V_{IN} > V_{SS}$; $V_{IN} < V_{DDP}$
Pull Level Force Current[5]	$\lvert I_{PLF}\rvert$ SR	150	–	–	µA	$V_{IN} \geq V_{IHmin}$ *(pull down)* ; $V_{IN} \leq V_{ILmax}$ *(pull up)*
Pull Level Keep Current[6]	$\lvert I_{PLK}\rvert$ SR	–	–	10	µA	$V_{IN} \geq V_{IHmin}$ *(pull up)* ; $V_{IN} \leq V_{ILmax}$ *(pull down)*
Input high voltage (all except XTAL1)	V_{IH} SR	0.7 x V_{DDP}	–	V_{DDP} + 0.3	V	
Input low voltage (all except XTAL1)	V_{IL} SR	-0.3	–	0.3 x V_{DDP}	V	

Table 16 DC Characteristics for Lower Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Output High voltage[7)]	V_{OH} CC	V_{DDP} - 1.0	–	–	V	$I_{OH} \geq I_{OHmax}$
		V_{DDP} - 0.4	–	–	V	$I_{OH} \geq I_{OHnom}$ [8)]
Output Low Voltage[7)]	V_{OL} CC	–	–	0.4	V	$I_{OL} \leq I_{OLnom}$ [8)]
		–	–	1.0	V	$I_{OL} \leq I_{OLmax}$

1) Because each double bond pin is connected to two pads (standard pad and high-speed pad), it has twice the normal value. For a list of affected pins refer to the pin definitions table in chapter 2.

2) Not subject to production test - verified by design/characterization. Hysteresis is implemented to avoid metastable states and switching due to internal ground bounce. It cannot suppress switching due to external system noise under all conditions.

3) If the input voltage exceeds the respective supply voltage due to ground bouncing ($V_{IN} < V_{SS}$) or supply ripple ($V_{IN} > V_{DDP}$), a certain amount of current may flow through the protection diodes. This current adds to the leakage current. An additional error current (I_{INJ}) will flow if an overload current flows through an adjacent pin. Please refer to the definition of the overload coupling factor K_{OV}.

4) The given values are worst-case values. In production test, this leakage current is only tested at 125 °C; other values are ensured by correlation. For derating, please refer to the following descriptions: Leakage derating depending on temperature (T_J = junction temperature [°C]): $I_{OZ} = 0.05 \times e^{(1.5 + 0.028 \times TJ>)}$ [µA]. For example, at a temperature of 95 °C the resulting leakage current is 3.2 µA. Leakage derating depending on voltage level (DV = V_{DDP} - V_{PIN} [V]): $I_{OZ} = I_{OZtempmax}$ - (1.6 x DV) (µA]. This voltage derating formula is an approximation which applies for maximum temperature.

5) Drive the indicated minimum current through this pin to change the default pin level driven by the enabled pull device: $V_{PIN} <= V_{IL}$ for a pullup; $V_{PIN} >= V_{IH}$ for a pulldown.

6) Limit the current through this pin to the indicated value so that the enabled pull device can keep the default pin level: $V_{PIN} >= V_{IH}$ for a pullup; $V_{PIN} <= V_{IL}$ for a pulldown.

7) The maximum deliverable output current of a port driver depends on the selected output driver mode. This specification is not valid for outputs which are switched to open drain mode. In this case the respective output will float and the voltage is determined by the external circuit.

8) As a rule, with decreasing output current the output levels approach the respective supply level (V_{OL}->V_{SS}, V_{OH}->V_{DDP}). However, only the levels for nominal output currents are verified.

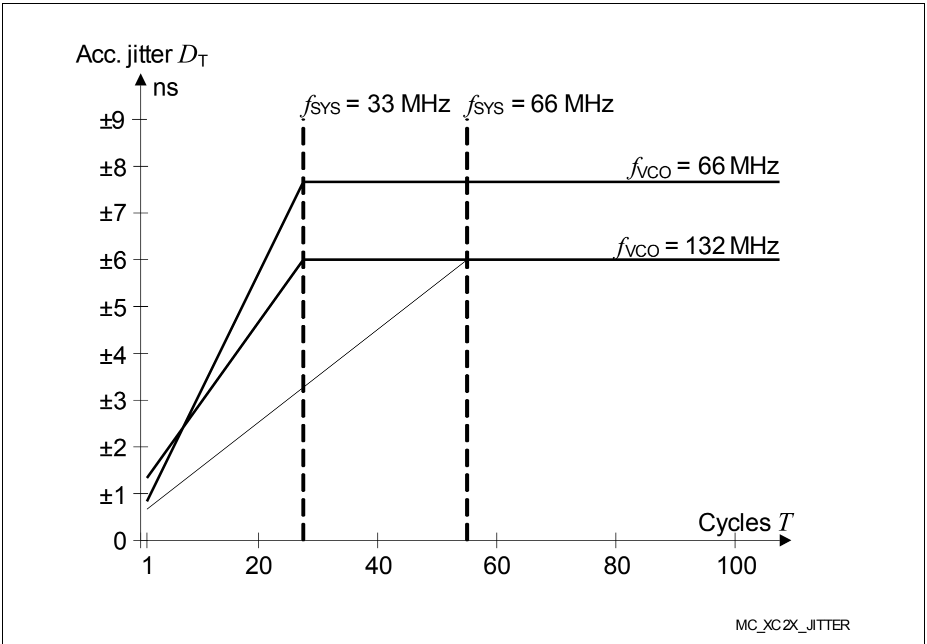

Figure 20 Approximated Accumulated PLL Jitter

Note: The specified PLL jitter values are valid if the capacitive load per pin does not exceed C_L = 20 pF.
The maximum peak-to-peak noise on the pad supply voltage (measured between V_{DDPB} pin 100 and V_{SS} pin 1) is limited to a peak-to-peak voltage of V_{PP} = 50 mV. This can be achieved by appropriate blocking of the supply voltage as close as possible to the supply pins and using PCB supply and ground planes.

PLL frequency band selection

Different frequency bands can be selected for the VCO so that the operation of the PLL can be adjusted to a wide range of input and output frequencies:

4.7.5 External Bus Timing

The following parameters specify the behavior of the XE164xN bus interface.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 29 Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
CLKOUT Cycle Time[1)]	t_5 CC	–	$1 / f_{SYS}$	–	ns	
CLKOUT high time	t_6 CC	3	–	–		
CLKOUT low time	t_7 CC	3	–	–		
CLKOUT rise time	t_8 CC	–	–	3	ns	
CLKOUT fall time	t_9 CC	–	–	3		

1) The CLKOUT cycle time is influenced by PLL jitter. For longer periods the relative deviation decreases (see PLL deviation formula).

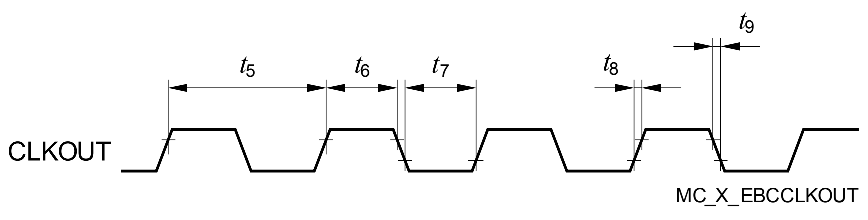

Figure 22 CLKOUT Signal Timing

Note: The term CLKOUT refers to the reference clock output signal which is generated by selecting f_{SYS} as the source signal for the clock output signal EXTCLK on pin P2.8 and by enabling the high-speed clock driver on this pin.

4.7.7 Debug Interface Timing

The debugger can communicate with the XE164xN either via the 2-pin DAP interface or via the standard JTAG interface.

Debug via DAP

The following parameters are applicable for communication through the DAP debug interface.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 37 is valid under the following conditions: C_L= 20 pF; voltage_range= upper

Table 37 DAP Interface Timing for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
DAP0 clock period[1)]	t_{11} SR	25	–	–	ns	
DAP0 high time	t_{12} SR	8	–	–	ns	
DAP0 low time[1)]	t_{13} SR	8	–	–	ns	
DAP0 clock rise time	t_{14} SR	–	–	4	ns	
DAP0 clock fall time	t_{15} SR	–	–	4	ns	
DAP1 setup to DAP0 rising edge	t_{16} SR	6	–	–	ns	
DAP1 hold after DAP0 rising edge	t_{17} SR	6	–	–	ns	
DAP1 valid per DAP0 clock period[2)]	t_{19} CC	17	20	–	ns	

1) See the DAP chapter for clock rate restrictions in the Active::IDLE protocol state.

2) The Host has to find a suitable sampling point by analyzing the sync telegram response.

Table 38 is valid under the following conditions: C_L= 20 pF; voltage_range= lower