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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	ARM® Cortex®-M4F
Core Size	32-Bit Single-Core
Speed	168MHz
Connectivity	CANbus, FlexIO, I²C, SPI, UART/USART
Peripherals	DMA, LVD, PWM, WDT
Number of I/O	58
Program Memory Size	512KB (512K x 8)
Program Memory Type	FLASH
EEPROM Size	68K x 8
RAM Size	64K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 16x12b; D/A 1x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LQFP
Supplier Device Package	64-LQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mke16f512vlh16

specific module is not reset by the corresponding Reset source.

Table 3. Reset source

Reset sources	Descriptions	Modules									
		PMC	SIM	SMC	RCM	Reset pin is negated	WDOG	SCG	RTC	LPTMR	Others
POR reset	Power-on reset (POR)	Y	Y	Y	Y	Y	Y	Y	Y	Y	Y
System resets	Low-voltage detect (LVD)	Y[1]	Y	Y	Y	Y	Y	Y	N	Y	Y
	External pin reset ($\overline{\text{RESET}}$)	Y[1]	Y[2]	Y[3]	Y[4]	Y	Y[5]	Y[6]	N	N	Y
	Watchdog (WDOG) reset	Y[1]	Y[2]	Y[3]	Y[4]	Y	Y[5]	Y[6]	N	N	Y
	Multipurpose clock generator loss of clock (LOC) reset	Y[1]	Y[2]	Y[3]	Y[4]	Y	Y[5]	Y[6]	N	N	Y
	Multipurpose clock generator loss of lock (LOL) reset	Y[1]	Y[2]	Y[3]	Y[4]	Y	Y[5]	Y[6]	N	N	Y
	Stop mode acknowledge error (SACKERR)	Y[1]	Y[2]	Y[3]	Y[4]	Y	Y[5]	Y[6]	N	N	Y
	Software reset (SW)	Y[1]	Y[2]	Y[3]	Y[4]	Y	Y[5]	Y[6]	N	N	Y
	Lockup reset (LOCKUP)	Y[1]	Y[2]	Y[3]	Y[4]	Y	Y[5]	Y[6]	N	N	Y
	MDM DAP system reset	Y[1]	Y[2]	Y[3]	Y[4]	Y	Y[5]	Y[6]	N	N	Y
Debug reset	Debug reset	Y[1]	Y[2]	Y[3]	Y[4]	Y	Y[5]	Y[6]	N	N	Y

1. Except PMC_LVDSC1[LVDV] and PMC_LVDSC2[LVWV]
2. Except SIM_SOPT1
3. Except SMC_PMPROT, SMC_PMCTRL_RUM, SMC_PMCTRL_STOPM, SMC_STOPCTRL, SMC_PMSTAT
4. Except RCM_RPC, RCM_MR, RCM_FM, RCM_SRIE, RCM_SRS, RCM_SSRS
5. Except WDOG_CS[TST]
6. Except SCG_CSR and SCG_FIRCSTAT

This device supports booting from:

- internal flash
- boot ROM

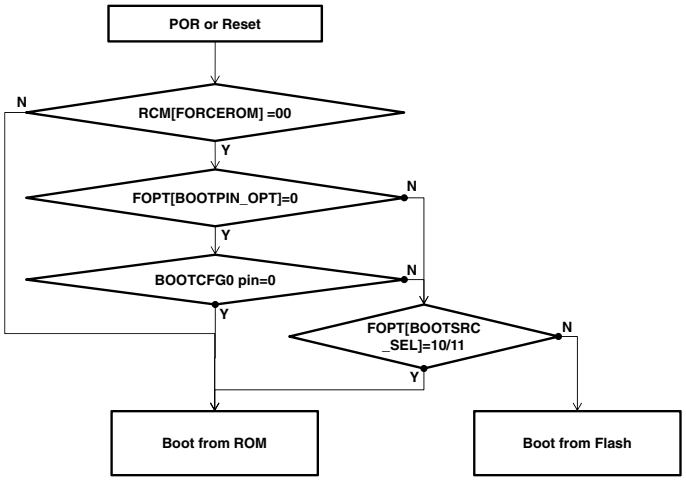

Figure 3. Boot flow chart

The blank chip is default to boot from ROM and remaps the vector table to ROM base address, otherwise, it remaps to flash address.

2.1.6 Clock options

The SCG module controls which clock source is used to derive the system clocks. The clock generation logic divides the selected clock source into a variety of clock domains, including the clocks for the system bus masters, system bus slaves, and flash memory . The clock generation logic also implements module-specific clock gating to allow granular shutoff of modules.

The following figure is a high level block diagram of the clock generation. For more details on the clock operation and configuration, see the Clocking chapter in the Reference Manual.

segments is controlled as the FAC provides a cycle-by-cycle evaluation of the access rights for each transaction routed to the on-chip flash memory. Configurability allows an increasing number of protected segments while supporting two levels of vendors adding their proprietary software to a device.

2.1.7.2 Error-correcting code (ECC)

The ECC detection is also supported on Flash and SRAM memories. It supports auto correction of one-bit error and reporting more than one-bit error.

2.1.8 Power management

The Power Management Controller (PMC) expands upon ARM's operational modes of Run, Sleep, and Deep Sleep, to provide multiple configurable modes. These modes can be used to optimize current consumption for a wide range of applications. The WFI or WFE instruction invokes a Wait or a Stop mode, depending on the current configuration. For more information on ARM's operational modes, See the ARM® Cortex® User Guide.

The PMC provides High Speed Run (HSRUN), Normal Run (RUN), and Very Low Power Run (VLPR) configurations in ARM's Run operation mode. In these modes, the MCU core is active and can access all peripherals. The difference between the modes is the maximum clock frequency of the system and therefore the power consumption. The configuration that matches the power versus performance requirements of the application can be selected.

The PMC provides Wait (Wait) and Very Low Power Wait (VLPW) configurations in ARM's Sleep operation mode. In these modes, even though the MCU core is inactive, all of the peripherals can be enabled and operate as programmed. The difference between the modes is the maximum clock frequency of the system and therefore the power consumption.

The PMC provides Stop (Stop), Very Low Power Stop (VLPS) configurations in ARM's Deep Sleep operational mode. In these modes, the MCU core and most of the peripherals are disabled. Depending on the requirements of the application, different portions of the analog, logic, and memory can be retained or disabled to conserve power.

2.2.1 eDMA and DMAMUX

The eDMA is a highly programmable data-transfer engine optimized to minimize any required intervention from the host processor. It is intended for use in applications where the data size to be transferred is statically known and not defined within the transferred data itself. The DMA controller in this device implements 16 channels which can be routed from up to 63 DMA request sources through DMA MUX module.

Main features of eDMA are listed below:

- All data movement via dual-address transfers: read from source, write to destination
- 16-channel implementation that performs complex data transfers with minimal intervention from a host processor
- Transfer control descriptor (TCD) organized to support two-deep, nested transfer operations
- Channel activation via one of three methods
- Fixed-priority and round-robin channel arbitration
- Channel completion reported via programmable interrupt requests
- Programmable support for scatter/gather DMA processing
- Support for complex data structures

2.2.2 FTM

This device contains four FlexTimer modules.

The FlexTimer module (FTM) is a two-to-eight channel timer that supports input capture, output compare, and the generation of PWM signals to control electric motor and power management applications. The FTM time reference is a 16-bit counter that can be used as an unsigned or signed counter.

Several key enhancements of this module are made:

- Signed up counter
- Deadtime insertion hardware
- Fault control inputs
- Enhanced triggering functionality
- Initialization and polarity control

- Static operation in Normal Stop mode
- 16-word data buffer supported with multiple operation modes
- DMA support

2.2.5 CMP

There are three analog comparators on this device.

- Each CMP has its own independent 8-bit DAC.
- Each CMP supports up to 7 analog inputs from external pins.
- Each CMP is able to convert an internal reference from the bandgap.
- Each CMP supports internal reference from the on-chip 12-bit DAC out.
- Each CMP supports the round-robin sampling scheme. In summary, this allow the CMP to operate independently in VLPS and Stop modes, whilst being triggered periodically to sample up to 8 inputs. Only if an input changes state is a full wakeup generated.

The CMP has the following features:

- Inputs may range from rail to rail
- Programmable hysteresis control
- Selectable interrupt on rising-edge, falling-edge, or both rising and falling edges of the comparator output
- Selectable inversion on comparator output
- Capability to produce a wide range of outputs such as sampled, windowed, or digitally filtered
- External hysteresis can be used at the same time that the output filter is used for internal functions
- Two software selectable performance levels: Shorter propagation delay at the expense of higher power, and Low power with longer propagation delay
- DMA transfer support
- Functional in all power modes available on this MCU
- The window and filter functions are not available in STOP mode
- Integrated 8-bit DAC with selectable supply reference source and can be power down to conserve power

2.2.6 RTC

The RTC is an always powered-on block that remains active in all low power modes. The time counter within the RTC is clocked by a 32.768 kHz clock sourced from an external crystal using the oscillator, or clock directly from RTC_CLKIN pin.

- Transmit and receive baud rate can operate asynchronous to the bus clock and can be configured independently of the bus clock frequency, support operation in Stop mode
- Interrupt, DMA or polled operation
- Hardware parity generation and checking
- Programmable 8-bit, 9-bit or 10-bit character length
- Programmable 1-bit or 2-bit stop bits
- Three receiver wakeup methods
  - Idle line wakeup
  - Address mark wakeup
  - Receive data match
- Automatic address matching to reduce ISR overhead:
  - Address mark matching
  - Idle line address matching
  - Address match start, address match end
- Optional 13-bit break character generation / 11-bit break character detection
- Configurable idle length detection supporting 1, 2, 4, 8, 16, 32, 64 or 128 idle characters
- Selectable transmitter output and receiver input polarity

2.2.12 LPSPI

This device contains two LPSPI modules. The LPSPI is a low power Serial Peripheral Interface (SPI) module that supports an efficient interface to an SPI bus as a master and/or a slave. The LPSPI can continue operating in stop modes provided an appropriate clock is available and is designed for low CPU overhead with DMA offloading of FIFO register accesses.

The LPSPI modules have the following features:

- Command/transmit FIFO of 4 words
- Receive FIFO of 4 words
- Host request input can be used to control the start time of an SPI bus transfer

2.2.13 FlexCAN

This device contains two FlexCAN modules. The FlexCAN module is a communication controller implementing the CAN protocol according to the ISO 11898-1 standard and CAN 2.0 B protocol specifications.

100 LQFP	64 LQFP	Pin Name	Default	ALT0	ALT1	ALT2	ALT3	ALT4	ALT5	ALT6	ALT7
41	—	PTD9	ACMP1_IN5	ACMP1_IN5	PTD9	LPI2C1_SCL		FTM2_FLT3		FTM1_CH5	
42	—	PTD8	DISABLED		PTD8	LPI2C1_SDA		FTM2_FLT2		FTM1_CH4	
43	27	PTC17	ADC0_SE15	ADC0_SE15	PTC17	FTM1_FLT3		LPI2C1_SCLS			
44	28	PTC16	ADC0_SE14	ADC0_SE14	PTC16	FTM1_FLT2		LPI2C1_SDAS			
45	29	PTC15	ADC0_SE13/ ACMP2_IN4	ADC0_SE13/ ACMP2_IN4	PTC15	FTM1_CH3					
46	30	PTC14	ADC0_SE12/ ACMP2_IN5	ADC0_SE12/ ACMP2_IN5	PTC14	FTM1_CH2					
47	31	PTB3	ADC0_SE7	ADC0_SE7	PTB3	FTM1_CH1	LPSPI0_SIN	FTM1_QD_ PHA		TRGMUX_IN2	
48	32	PTB2	ADC0_SE6	ADC0_SE6	PTB2	FTM1_CH0	LPSPI0_SCK	FTM1_QD_ PHB		TRGMUX_IN3	
49	—	PTC13	DISABLED		PTC13	FTM3_CH7	FTM2_CH7				
50	—	PTC12	DISABLED		PTC12	FTM3_CH6	FTM2_CH6				
51	—	PTC11	DISABLED		PTC11	FTM3_CH5					
52	—	PTC10	DISABLED		PTC10	FTM3_CH4					
53	33	PTB1	ADC0_SE5	ADC0_SE5	PTB1	LPUART0_TX	LPSPI0_SOUT	TCLK0			
54	34	PTB0	ADC0_SE4	ADC0_SE4	PTB0	LPUART0_RX	LPSPI0_PCS0	LPTMR0_ ALT3	PWT_IN3		
55	35	PTC9	ADC2_SE15	ADC2_SE15	PTC9	LPUART1_TX	FTM1_FLT1			LPUART0_ RTS	
56	36	PTC8	ADC2_SE14	ADC2_SE14	PTC8	LPUART1_RX	FTM1_FLT0			LPUART0_ CTS	
57	37	PTA7	ADC0_SE3/ ACMP1_IN1	ADC0_SE3/ ACMP1_IN1	PTA7	FTM0_FLT2		RTC_CLKIN		LPUART1_ RTS	
58	38	PTA6	ADC0_SE2/ ACMP1_IN0	ADC0_SE2/ ACMP1_IN0	PTA6	FTM0_FLT1	LPSPI1_PCS1			LPUART1_ CTS	
59	39	PTE7	ADC2_SE2/ ACMP2_IN6	ADC2_SE2/ ACMP2_IN6	PTE7	FTM0_CH7	FTM3_FLT0				
60	40	VSS	VSS	VSS							
61	41	VDD	VDD	VDD							
62	—	PTA17	DISABLED		PTA17	FTM0_CH6	FTM3_FLT0	EWM_OUT_b			
63	—	PTB17	ADC2_SE3	ADC2_SE3	PTB17	FTM0_CH5	LPSPI1_PCS3				
64	—	PTB16	ADC1_SE15	ADC1_SE15	PTB16	FTM0_CH4	LPSPI1_SOUT				
65	—	PTB15	ADC1_SE14	ADC1_SE14	PTB15	FTM0_CH3	LPSPI1_SIN				
66	—	PTB14	ADC1_SE9	ADC1_SE9	PTB14	FTM0_CH2	LPSPI1_SCK				
67	42	PTB13	ADC1_SE8	ADC1_SE8	PTB13	FTM0_CH1	FTM3_FLT1				
68	43	PTB12	ADC1_SE7	ADC1_SE7	PTB12	FTM0_CH0	FTM3_FLT2				
69	44	PTD4	ADC1_SE6/ ACMP1_IN6	ADC1_SE6/ ACMP1_IN6	PTD4	FTM0_FLT3	FTM3_FLT3				
70	45	PTD3	NMI_b	ADC1_SE3	PTD3	FTM3_CH5	LPSPI1_PCS0	FXIO_D5		TRGMUX_IN4	NMI_b
71	46	PTD2	ADC1_SE2	ADC1_SE2	PTD2	FTM3_CH4	LPSPI1_SOUT	FXIO_D4		TRGMUX_IN5	
72	47	PTA3	ADC1_SE1	ADC1_SE1	PTA3	FTM3_CH1	LPI2C0_SCL	EWM_IN		LPUART0_TX	

5.2.2 Moisture handling ratings

Symbol	Description	Min.	Max.	Unit	Notes
MSL	Moisture sensitivity level	—	3	—	1

1. Determined according to IPC/JEDEC Standard J-STD-020, *Moisture/Reflow Sensitivity Classification for Nonhermetic Solid State Surface Mount Devices*.

5.2.3 ESD handling ratings

Symbol	Description	Min.	Max.	Unit	Notes
V_{HBM}	Electrostatic discharge voltage, human body model	– 6000	6000	V	1
V_{CDM}	Electrostatic discharge voltage, charged-device model				2
	All pins except the corner pins	– 500	500	V	
	Corner pins only	– 750	750	V	
I_{LAT}	Latch-up current at ambient temperature upper limit	– 100	100	mA	3

1. Determined according to JEDEC Standard JESD22-A114, *Electrostatic Discharge (ESD) Sensitivity Testing Human Body Model (HBM)*.
2. Determined according to JEDEC Standard JESD22-C101, *Field-Induced Charged-Device Model Test Method for Electrostatic-Discharge-Withstand Thresholds of Microelectronic Components*.
3. Determined according to JEDEC Standard JESD78, *IC Latch-Up Test*.

5.2.4 Voltage and current operating ratings

Table 26. Voltage and current operating ratings

Symbol	Description	Min.	Max.	Unit
V_{DD}	Supply voltage	2.7	5.5	V
I_{DD}	Digital supply current	—	80	mA
V_{IO}	IO pin input voltage	V_{SS} – 0.3	V_{DD} + 0.3	V
I_D	Instantaneous maximum current single pin limit (applies to all port pins)	–25	25	mA
V_{DDA}	Analog supply voltage	V_{DD} – 0.1	V_{DD} + 0.1	V

5.3 General

Table 32. Power consumption operating behaviors (continued)

Mode	Symbol	Clock Configuration	Description	Temperature	Min	Typ	Max[1]	Units
			Core@2MHz, bus @2MHz, flash @1MHz, VDD=5V					
		IRC2M	Very Low Power Run While(1) loop in Flash all peripheral clock enabled. Core@2MHz, bus @2MHz, flash @1MHz, VDD=5V	25 °C	—	1.28	1.39	
WAIT	I_{DD_WAIT}	PLL	core disabled, system@120MHz, bus @60MHz, flash disabled (**flash doze enabled**), VDD=5 V, all peripheral clocks disabled	25 °C	—	14.13	14.78	mA
		IRC48M	core disabled, system@48 MHz, bus @48MHz, flash disabled (**flash doze enabled**), VDD=5 V, all peripheral clocks disabled	25 °C	—	8.50	9.15	
VLPW	I_{DD_VLPW}	IRC8M	Very Low Power Wait current, core disabled system@4MHz, bus@4Mhz and flash@1MHz, all peripheral clocks disabled, VDD=5V	25 °C	—	0.84	0.94	mA
		IRC2M	Very Low Power Wait current, core disabled system@2MHz, bus@2Mhz and flash@1MHz, all peripheral clocks disabled, VDD=5V	25 °C	—	1.08	1.18	
STOP	I_{DD_STOP}	-	Stop mode current, VDD=5V, bias disabled [2]	25 °C and blew	—	175	484	µA
				50 °C	—	438	1014	
				85 °C	—	1433	2864	
				105 °C	—	2860	5263	
STOP	I_{DD_STOP}	-	Stop mode current, VDD=5V, bias enabled [2]	25 °C and blew	—	92	299	µA
				50 °C	—	211	530	
				85 °C	—	671	1397	
				105 °C	—	1287	2502	
VLPS	I_{DD_VLPS}	-	Very Low Power Stop current, VDD=5V, bias disabled [2]	25 °C and blew	—	175	483	µA
				50 °C	—	424	998	
				85 °C	—	1367	2792	
				105 °C	—	2864	5258	
VLPS	I_{DD_VLPS}	-	Very Low Power Stop current, VDD=5V, bias enabled [2]	25 °C and blew	—	91	298	µA
				50 °C	—	208	525	
				85 °C	—	656	1378	
				105 °C	—	1305	2514	

5.3.2.5 AC specifications at 5 V range

Table 37. Functional pad AC specifications

Characteristic	Symbol	Min	Typ	Max	Unit
I/O Supply Voltage	Vdd [1]	4		5.5	V

1. Max power supply ramp rate is 500 V/ms.

Name	Prop Delay (ns) [1]	Rise/Fall Edge (ns) [2]		Drive Load (pF)
	Max	Min	Max	
Normal drive I/O pad	12	3.6	10	25
	18	8	17	50
High drive I/O pad	13	3.6	10	25
	19	8	19	50
CMOS Input [3]	3	1.2	2.8	0.5

1. As measured from 50% of core side input to 50% of the output.
2. Edges measured using 20% and 80% of the VDD supply.
3. Input slope = 2 ns.

NOTE

All measurements were taken accounting for 150 mV drop across VDD and VSS.

5.3.3 Thermal specifications

5.3.3.1 Thermal operating requirements

Table 38. Thermal operating requirements

Symbol	Description	Min.	Max.	Unit	Notes
T_J	Die junction temperature	–40	125	°C	
T_A	Ambient temperature	–40	105	°C	1

1. Maximum T_A can be exceeded only if the user ensures that T_J does not exceed maximum T_J. The simplest method to determine T_J is: $T_J = T_A + R_{\Theta JA} \times$ chip power dissipation.

5.3.3.2 Thermal attributes

5.3.3.2.1 Description

The tables in the following sections describe the thermal characteristics of the device.

NOTE

Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting side (board) temperature, ambient temperature, air flow, power dissipation or other components on the board, and board thermal resistance.

5.3.3.2.2 Thermal characteristics for the 64-pin LQFP package

Table 39. Thermal characteristics for the 64-pin LQFP package

Rating	Conditions	Symbol	Value	Unit
Thermal resistance, Junction to Ambient (Natural Convection)[1, 2]	Single layer board (1s)	$R_{\theta JA}$	60	°C/W
Thermal resistance, Junction to Ambient (Natural Convection)[1, 2]	Four layer board (2s2p)	$R_{\theta JA}$	42	°C/W
Thermal resistance, Junction to Ambient (@200 ft/min)[1, 3]	Single layer board (1s)	$R_{\theta JMA}$	49	°C/W
Thermal resistance, Junction to Ambient (@200 ft/min)[1, 3]	Four layer board (2s2p)	$R_{\theta JMA}$	36	°C/W
Thermal resistance, Junction to Board[4]	—	$R_{\theta JB}$	24	°C/W
Thermal resistance, Junction to Case [5]	—	$R_{\theta JC}$	12	°C/W
Thermal resistance, Junction to Package Top[6]	Natural Convection	Ψ_{JT}	2	°C/W

1. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.
2. Per JEDEC JESD51-2 with natural convection for horizontally oriented board. Board meets JESD51-9 specification for 1s or 2s2p board, respectively.
3. Per JEDEC JESD51-6 with forced convection for horizontally oriented board. Board meets JESD51-9 specification for 1s or 2s2p board, respectively.
4. Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
5. Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1).
6. Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2.

5.3.3.2.3 Thermal characteristics for the 100-pin LQFP package

Table 40. Thermal characteristics for the 100-pin LQFP package

Rating	Conditions	Symbol	Value	Unit
Thermal resistance, Junction to Ambient (Natural Convection)[1, 2]	Single layer board (1s)	$R_{\theta JA}$	57	°C/W

Table continues on the next page...

Table 40. Thermal characteristics for the 100-pin LQFP package (continued)

Rating	Conditions	Symbol	Value	Unit
Thermal resistance, Junction to Ambient (Natural Convection)[1, 2]	Four layer board (2s2p)	$R_{\theta JA}$	44	°C/W
Thermal resistance, Junction to Ambient (@200 ft/min)[1, 3]	Single layer board (1s)	$R_{\theta JMA}$	47	°C/W
Thermal resistance, Junction to Ambient (@200 ft/min)[1, 3]	Four layer board (2s2p)	$R_{\theta JMA}$	38	°C/W
Thermal resistance, Junction to Board[4]	—	$R_{\theta JB}$	30	°C/W
Thermal resistance, Junction to Case [5]	—	$R_{\theta JC}$	14	°C/W
Thermal resistance, Junction to Package Top[6]	Natural Convection	ψ_{JT}	2	°C/W

1. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.
2. Per JEDEC JESD51-2 with natural convection for horizontally oriented board. Board meets JESD51-9 specification for 1s or 2s2p board, respectively.
3. Per JEDEC JESD51-6 with forced convection for horizontally oriented board. Board meets JESD51-9 specification for 1s or 2s2p board, respectively.
4. Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
5. Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1).
6. Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2.

5.3.3.2.4 General notes for specifications at maximum junction temperature

An estimation of the chip junction temperature, T_J, can be obtained from this equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

- T_A = ambient temperature for the package (°C)
- $R_{\theta JA}$ = junction to ambient thermal resistance (°C/W)
- P_D = power dissipation in the package (W)

The junction to ambient thermal resistance is an industry standard value that provides a quick and easy estimation of thermal performance. Unfortunately, there are two values in common usage: the value determined on a single layer board and the value obtained on a board with two planes. For packages such as the PBGA, these values can be different by a factor of two. Which value is closer to the application depends on the power dissipated by other components on the board. The value obtained on a single layer board is appropriate for the tightly packed printed circuit board. The value obtained on the board with the internal planes is usually appropriate if the board has low power dissipation and the components are well separated.

When a heat sink is used, the thermal resistance is expressed in the following equation as the sum of a junction-to-case thermal resistance and a case-to-ambient thermal resistance:

$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$

where:

- $R_{\theta JA}$ = junction to ambient thermal resistance (°C/W)
- $R_{\theta JC}$ = junction to case thermal resistance (°C/W)
- $R_{\theta CA}$ = case to ambient thermal resistance (°C/W)

$R_{\theta JC}$ is device related and cannot be influenced by the user. The user controls the thermal environment to change the case to ambient thermal resistance, $R_{\theta CA}$. For instance, the user can change the size of the heat sink, the air flow around the device, the interface material, the mounting arrangement on printed circuit board, or change the thermal dissipation on the printed circuit board surrounding the device.

To determine the junction temperature of the device in the application when heat sinks are not used, the Thermal Characterization Parameter (Ψ_{JT}) can be used to determine the junction temperature with a measurement of the temperature at the top center of the package case using this equation:

$T_J = T_T + (\Psi_{JT} \times P_D)$

where:

- T_T = thermocouple temperature on top of the package (°C)
- Ψ_{JT} = thermal characterization parameter (°C/W)
- P_D = power dissipation in the package (W)

The thermal characterization parameter is measured per JESD51-2 specification using a 40 gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned so that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over about 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by cooling effects of the thermocouple wire.

5.4 Peripheral operating requirements and behaviors

5.4.5.2 CMP with 8-bit DAC electrical specifications

Table 55. Comparator with 8-bit DAC electrical specifications

Symbol	Description	Min.	Typ. [1]	Max.	Unit
V_{DD}	Supply voltage	2.7	—	5.5	V
I_{DDHS}	Supply current, High-speed mode[2]				µA
	within ambient temperature range	—	145	200	
I_{DDLS}	Supply current, Low-speed mode[2]				µA
	within ambient temperature range	—	5	10	
V_{AIN}	Analog input voltage	0	0 - V_{DDX}	V_{DDX}	V
V_{AIO}	Analog input offset voltage, High-speed mode				mV
	within ambient temperature range	-25	±1	25	
V_{AIO}	Analog input offset voltage, Low-speed mode				mV
	within ambient temperature range	-40	±4	40	
t_{DHSB}	Propagation delay, High-speed mode[3]				ns
	within ambient temperature range	—	30	200	
t_{DLSB}	Propagation delay, Low-speed mode[3]				µs
	within ambient temperature range	—	0.5	2	
t_{DHSS}	Propagation delay, High-speed mode[4]				ns
	within ambient temperature range	—	70	400	
t_{DLSS}	Propagation delay, Low-speed mode[4]				µs
	within ambient temperature range	—	1	5	
t_{IDHS}	Initialization delay, High-speed mode [3]				µs
	within ambient temperature range	—	1.5	3	
t_{IDLS}	Initialization delay, Low-speed mode[3]				µs
	within ambient temperature range	—	10	30	
V_{HYST0}	Analog comparator hysteresis, Hyst0 (V_{AIO})				mV
	within ambient temperature range	—	0	—	
V_{HYST1}	Analog comparator hysteresis, Hyst1, High-speed mode				mV
	within ambient temperature range	—	16	53	
	Analog comparator hysteresis, Hyst1, Low-speed mode				
	within ambient temperature range	—	11	30	
V_{HYST2}	Analog comparator hysteresis, Hyst2, High-speed mode				mV
	within ambient temperature range	—	32	90	
	Analog comparator hysteresis, Hyst2, Low-speed mode				
	within ambient temperature range	—	22	53	
V_{HYST3}	Analog comparator hysteresis, Hyst3, High-speed mode				mV

Table continues on the next page...

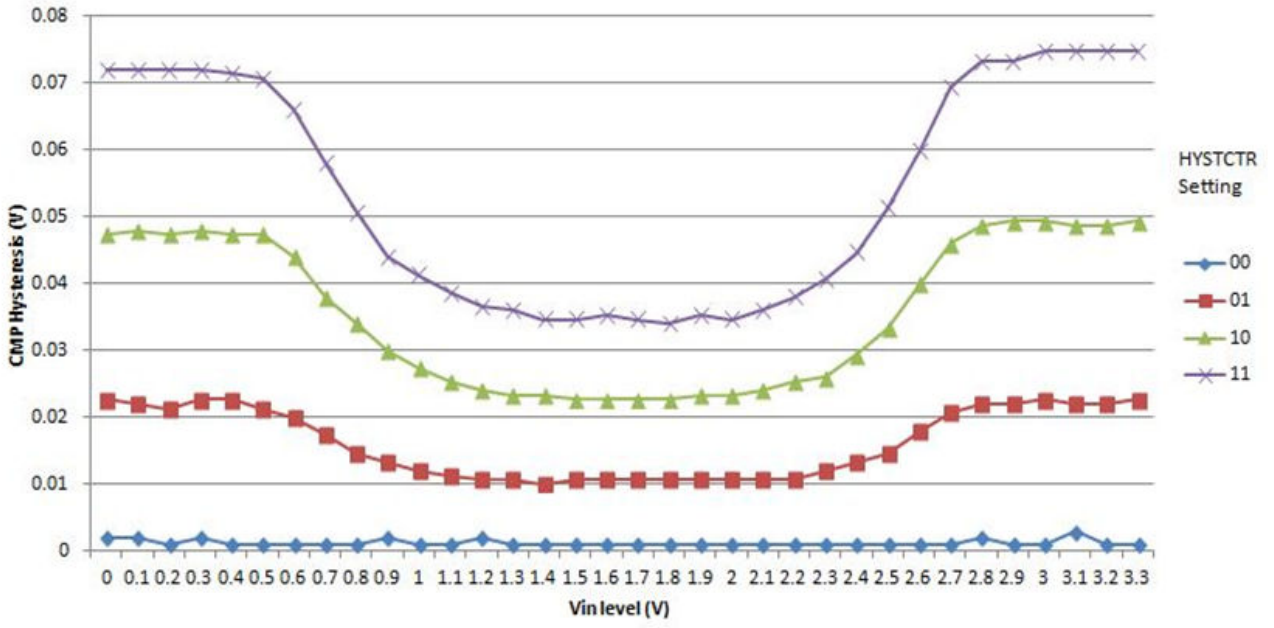

Figure 21. Typical hysteresis vs. Vin level (VDD = 3.3 V, PMODE = 1)

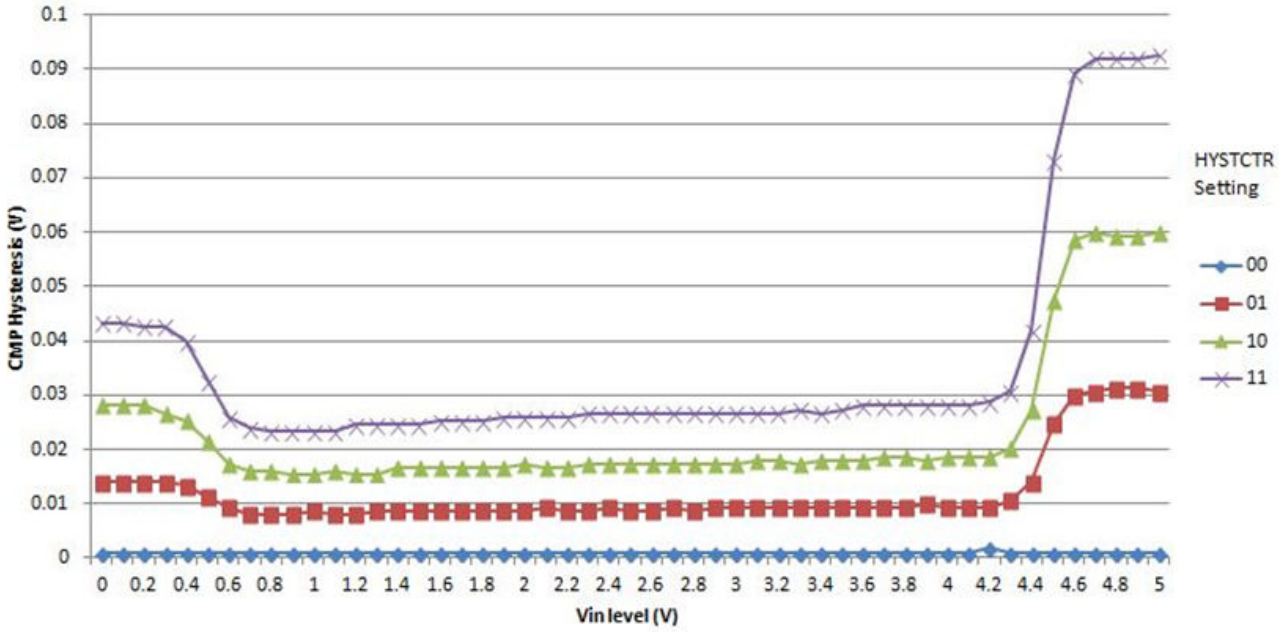

Figure 22. Typical hysteresis vs. Vin level (VDD = 5 V, PMODE = 0)

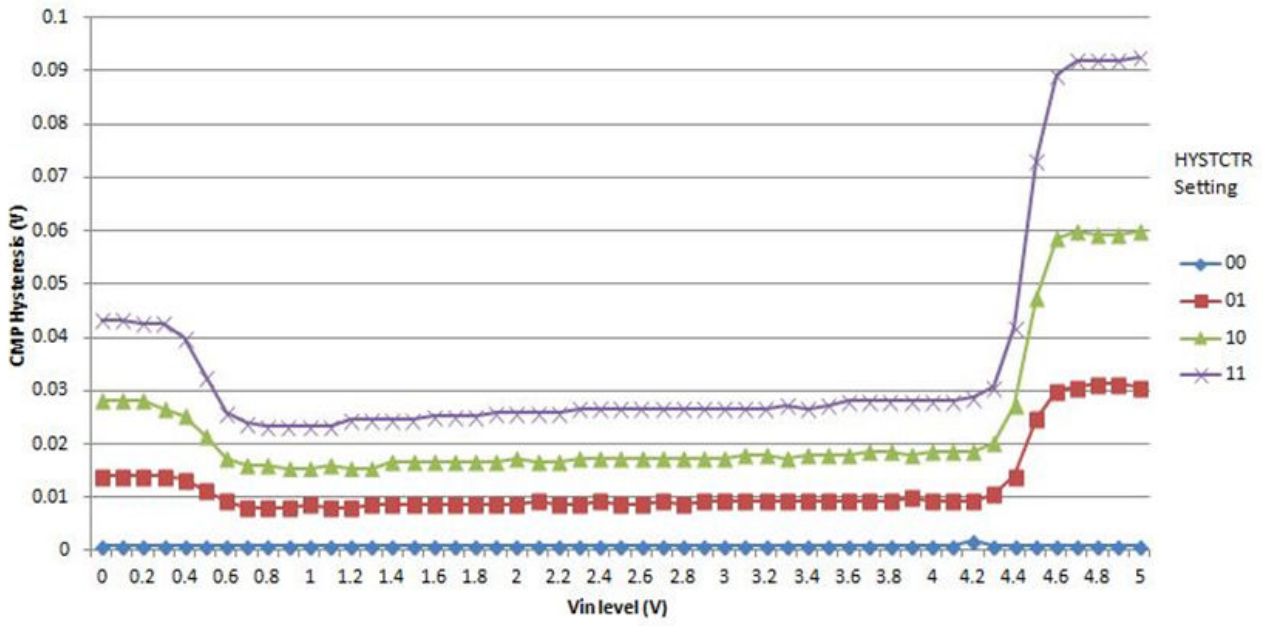

Figure 23. Typical hysteresis vs. Vin level (VDD = 5 V, PMODE = 1)

5.4.5.3 12-bit DAC electrical characteristics

5.4.5.3.1 12-bit DAC operating requirements

Table 56. 12-bit DAC operating requirements

Symbol	Desciption	Min.	Max.	Unit	Notes
V_{DDA}	Supply voltage	2.7	5.5	V	
V_{DACR}	Reference voltage	2.7	5.5	V	1
C_L	Output load capacitance	20	100	pF	2
I_L	Output load current	—	1	mA	3

1. The DAC reference can be selected to be V_{DDA} or V_{REFH}.
2. A small load capacitance can improve the bandwidth performance of the DAC.
3. Output range is from ground + 0.2 to V_{DACR} - 0.2

5.4.5.3.2 12-bit DAC operating behaviors

Table 57. 12-bit DAC operating behaviors

Symbol	Description	Min.	Typ.	Max.	Unit	Notes
I_{DDA_DACLP}	Supply current — low-power mode	—	—	330	µA	
I_{DDA_DACHP}	Supply current — high-power mode	—	—	1200	µA	

Table continues on the next page...

5.4.6.2 LPSPI electrical specifications

The Low Power Serial Peripheral Interface (LPSPI) provides a synchronous serial bus with master and slave operations. Many of the transfer attributes are programmable. The following tables provide timing characteristics for classic LPSPI timing modes.

All timing is shown with respect to 20% V_{DD} and 80% V_{DD} thresholds, unless noted, as well as input signal transitions of 3 ns and a 30 pF maximum load on all LPSPI pins.

Table 58. LPSPI master mode timing

Num.	Symbol	Description	Min.	Max.	Unit	Note
1	f_{op}	Frequency of operation	$f_{periph}/2048$	$f_{periph}/2$	Hz	1
2	t_{SPSCK}	SPSCK period	2 x t_{periph}	2048 x t_{periph}	ns	2
3	t_{Lead}	Enable lead time	1/2	—	t_{SPSCK}	—
4	t_{Lag}	Enable lag time	1/2	—	t_{SPSCK}	—
5	t_{WSPSCK}	Clock (SPSCK) high or low time	t_{periph} - 30	1024 x t_{periph}	ns	—
6	t_{SU}	Data setup time (inputs)	18	—	ns	—
7	t_{HI}	Data hold time (inputs)	0	—	ns	—
8	t_{v}	Data valid (after SPSCK edge)	—	15	ns	—
9	t_{HO}	Data hold time (outputs)	0	—	ns	—
10	t_{RI}	Rise time input	—	t_{periph} - 25	ns	—
	t_{FI}	Fall time input				
11	t_{RO}	Rise time output	—	25	ns	—
	t_{FO}	Fall time output				

1. f_{periph} = LPSPI peripheral clock
2. $t_{periph} = 1/f_{periph}$

NOTE

High drive pin should be used for fast bit rate.

5.4.7.1 SWD electricals

Table 61. SWD full voltage range electricals

Symbol	Description	Min.	Max.	Unit
V_{DDA}	Operating voltage	2.7	5.5	V
S1	SWD_CLK frequency of operation	0	25	MHz
S2	SWD_CLK cycle period	1/S1	—	ns
S3	SWD_CLK clock pulse width	15	—	ns
S4	SWD_CLK rise and fall times	—	3	ns
S9	SWD_DIO input data setup time to SWD_CLK rise	8	—	ns
S10	SWD_DIO input data hold time after SWD_CLK rise	1.4	—	ns
S11	SWD_CLK high to SWD_DIO data valid	—	25	ns
S12	SWD_CLK high to SWD_DIO high-Z	5	—	ns

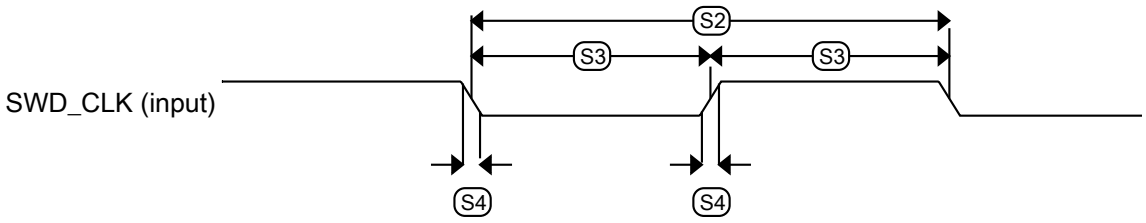

Figure 28. Serial wire clock input timing

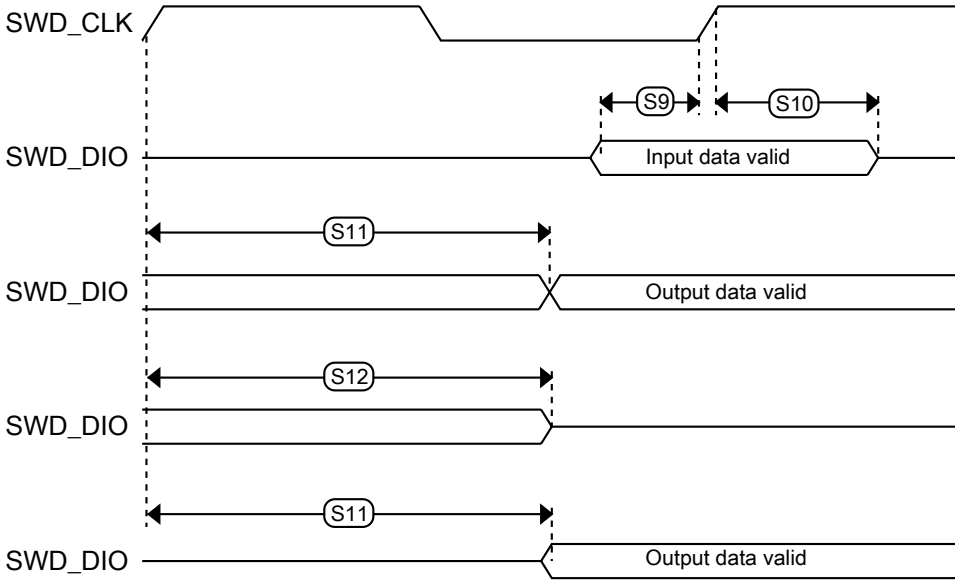

Figure 29. Serial wire data timing

6.1.1 Printed circuit board recommendations

- Place connectors or cables on one edge of the board and do not place digital circuits between connectors.
- Drivers and filters for I/O functions must be placed as close to the connectors as possible. Connect TVS devices at the connector to a good ground. Connect filter capacitors at the connector to a good ground. Consider to add ferrite bead or inductor to some sensitive lines.
- Physically isolate analog circuits from digital circuits if possible.
- Place input filter capacitors as close to the MCU as possible.
- For best EMC performance, route signals as transmission lines; use a ground plane directly under LQFP packages; and solder the exposed pad (EP) to ground directly under QFN packages.

6.1.2 Power delivery system

Consider the following items in the power delivery system:

- Use a plane for ground.
- Use a plane for MCU VDD supply if possible.
- Always route ground first, as a plane or continuous surface, and never as sequential segments.
- Always route the power net as star topology, and make each power trace loop as minimum as possible.
- Route power next, as a plane or traces that are parallel to ground traces.
- Place bulk capacitance, 10 μF or more, at the entrance of the power plane.
- Place bypass capacitors for MCU power domain as close as possible to each VDD/VSS pair, including VDDA/VSSA and VREFH/VREFL.
- The minimum bypass requirement is to place 0.1 μF capacitors positioned as near as possible to the package supply pins.

6.1.3 Analog design

Each ADC input must have an RC filter as shown in the following figure. The maximum value of R must be RAS max if fast sampling and high resolution are required. The value of C must be chosen to ensure that the RC time constant is very small compared to the sample period.

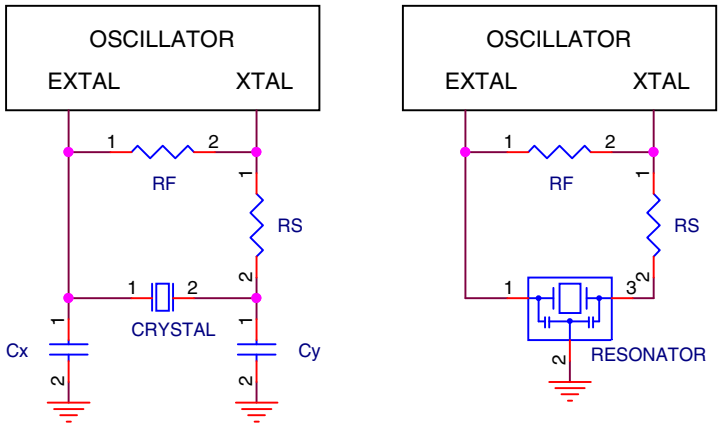

Figure 42. Crystal connection – Diagram 3

NOTE

For PCB layout, the user could consider to add the guard ring to the crystal oscillator circuit.

6.2 Software considerations

All Kinetis MCUs are supported by comprehensive NXP and third-party hardware and software enablement solutions, which can reduce development costs and time to market. Featured software and tools are listed below. Visit http://www.nxp.com/kinetis/sw for more information and supporting collateral.

Evaluation and Prototyping Hardware

- Tower System Development Platform: http://www.nxp.com/tower

IDEs for Kinetis MCUs

- Kinetis Design Studio IDE: http://www.nxp.com/kds
- Partner IDEs: http://www.nxp.com/kide

Run-time Software

- Kinetis SDK: http://www.nxp.com/ksdk
- Kinetis Bootloader: http://www.nxp.com/kboot
- ARM mbed Development Platform: http://www.nxp.com/mbed

For all other partner-developed software and tools, visit http://www.nxp.com/partners.