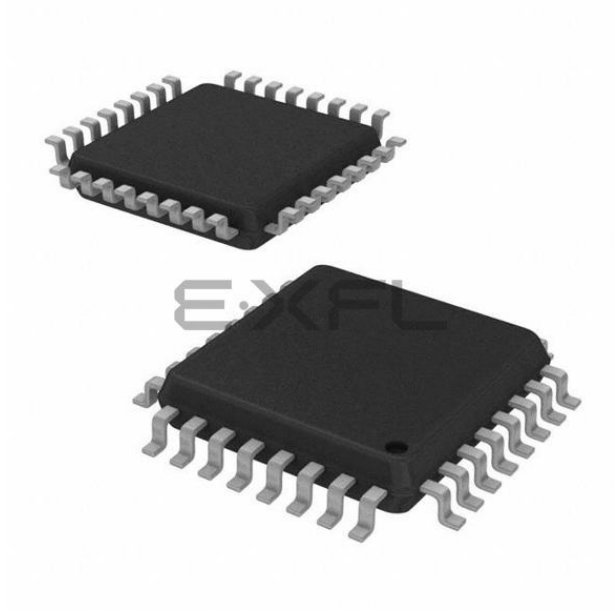

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	8051
Core Size	8-Bit
Speed	25MHz
Connectivity	SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, WDT
Number of I/O	22
Program Memory Size	8KB (8K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 3.6V
Data Converters	A/D 22x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	32-LQFP
Supplier Device Package	32-LQFP (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/c8051f221

SFR Definition 5.1. AMX0SL: AMUX Channel Select

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
-	-	AMXEN	PRTSL1	PRTSL0	PINSL2	PINSL1	PINSL0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xBB

Bits 7–6: UNUSED. Read = 00b; Write = don't care
Bit 5: AMXEN enable
0: AMXEN disabled and port pins are unavailable for analog use.
1: AMXEN enabled to use/select port pins for analog use.
Bits 4–3: PRTSL1–0: Port Select Bits .
00: Port0 select to configure pin for analog input from this port.
01: Port1 select to configure pin for analog input from this port.
10: Port2 select to configure pin for analog input from this port.
11: Port3 select to configure pin for analog input from this port.
Bits 2–0:PINSL2–0: Pin Select Bits
000: Pin 0 of selected port (above) to be used for analog input.
001: Pin 1 of selected port (above) to be used for analog input.
010: Pin 2 of selected port (above) to be used for analog input.
011: Pin 3 of selected port (above) to be used for analog input.
100: Pin 4 of selected port (above) to be used for analog input.
101: Pin 5 of selected port (above) to be used for analog input.
110: Pin 6 of selected port (above) to be used for analog input.
111: Pin 7 of selected port (above) to be used for analog input.

* Selecting a port for analog input does NOT default all pins of that port as analog input. After selecting a port for analog input, a pin must be selected using pin select bits (PINSL2–0). For example, after setting the AMXEN to ‘1’, setting PRTSL1–0 to “11”, and setting PINSL2–0 to “100” P3.4 is configured as analog input. All other Port 3 pins remain as GPIO pins. Also note that in order to use a port pin as analog input, its input mode should be set to *analog*. Please see section 14.2.

SFR Definition 6.4. ADC0H: ADC Data Word MSB (C8051F206)

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xBF

Bits7–0: ADC Data Word Bits
For ADLJST = 1: Upper 8-bits of the 12-bit ADC Data Word.
For ADLJST = 0: Bits7–4 are the sign extension of Bit3. Bits 3–0 are the upper 4-bits of the 12-bit ADC Data Word.

SFR Definition 6.5. ADC0L: ADC Data Word LSB (C8051F206)

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xBE

Bits7–0: ADC Data Word Bits
For ADLJST = 1: Bits7–4 are the lower 4-bits of the 12-bit ADC Data Word. Bits3–0 will always read 0.
For ADLJST = 0: Bits7–0 are the lower 8-bits of the 12-bit ADC Data Word.

NOTE: Resulting 12-bit ADC Data Word appears in the ADC Data Word Registers as follows:
ADC0H[3:0]:ADC0L[7:0], if ADLJST = 0
(ADC0H[7:4] will be sign extension of ADC0H.3 if a differential reading, otherwise = 0000b)

ADC0H[7:0]:ADC0L[7:4], if ADLJST = 1
(ADC0L[3:0] = 0000b)

EXAMPLE: ADC Data Word Conversion Map, AIN0 Input in Single-Ended Mode
(AMX0CF=0x00, AMX0SL=0x00)

AIN0 – AGND (Volts)	ADC0H:ADC0L (ADLJST = 0)	ADC0H:ADC0L (ADLJST = 1)
REF x (4095/4096)	0x0FFF	0xFFF0
REF x ½	0x0800	0x8000
REF x (2047/4096)	0x07FF	0x7FF0
0	0x0000	0x0000

6.3. ADC Programmable Window Detector

The ADC programmable window detector is very useful in many applications. It continuously compares the ADC output to user-programmed limits and notifies the system when an out-of-band condition is detected. This is especially effective in an interrupt-driven system, saving code space and CPU bandwidth while delivering faster system response times. The window detector interrupt flag (ADWINT in ADC0CN) can also be used in polled mode. The high and low bytes of the reference words are loaded into the ADC Greater-Than and ADC Less-Than registers (ADC0GTH, ADC0GTL, ADC0LTH, and ADC0LTL). Figure 6.3 and Figure 6.4 show example comparisons for reference. Notice that the window detector flag can be asserted when the measured data is inside or outside the user-programmed limits, depending on the programming of the ADC0GTx and ADC0LTx registers.

SFR Definition 6.6. ADC0GTH: ADC Greater-Than Data High Byte (C8051F206)

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								11111111
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xC5

Bits7–0:
The high byte of the ADC Greater-Than Data Word.

SFR Definition 6.7. ADC0GTL: ADC Greater-Than Data Low Byte (C8051F206)

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								11111111
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xC4

Bits7–0:
The low byte of the ADC Greater-Than Data Word.

Definition: ADC Greater-Than Data Word = ADC0GTH:ADC0GTL

SFR Definition 6.8. ADC0LTH: ADC Less-Than Data High Byte (C8051F206)

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								11111111
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xC4

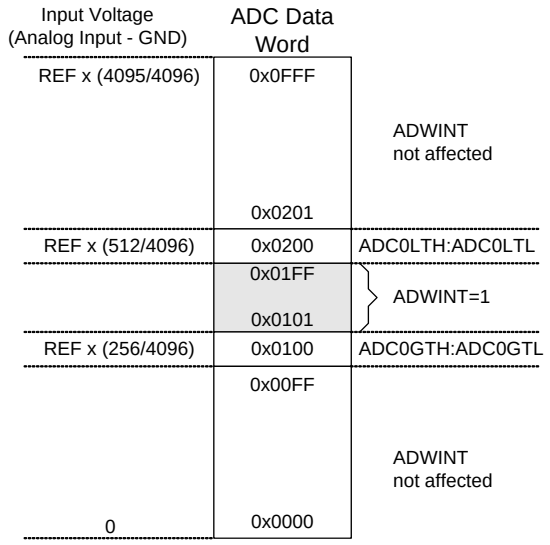
Bits7–0:
The high byte of the ADC Less-Than Data Word.

SFR Definition 6.9. ADC0LTL: ADC Less-Than Data Low Byte (C8051F206)

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								11111111
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xC4

Bits7–0:
These bits are the low byte of the ADC Less-Than Data Word.

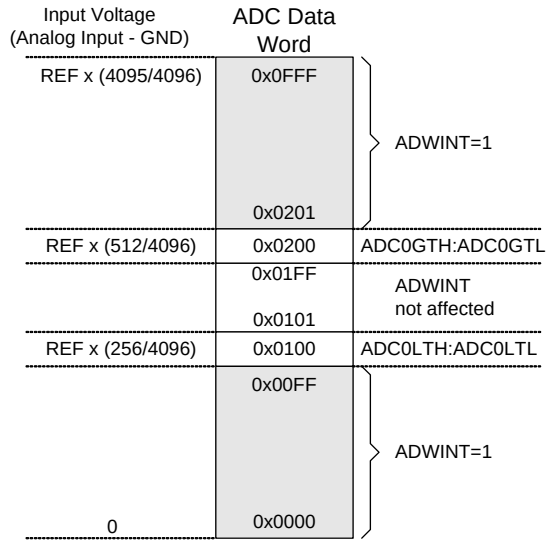
Definition:
ADC Less-Than Data Word = ADC0LTH:ADC0LTL



Given:

AMX0SL = 0x00, AMX0CF = 0x00, ADLJST = 0, ADC0LTH:ADC0LTL = 0x0200, ADC0GTH:ADC0GTL = 0x0100.

An ADC End of Conversion will cause an ADC Window Compare Interrupt (ADWINT=1) if the resulting ADC Data Word is < 0x0200 and > 0x0100.



Given:

AMX0SL = 0x00, AMX0CF = 0x00, ADLJST = 0, ADC0LTH:ADC0LTL = 0x0100, ADC0GTH:ADC0GTL = 0x0200.

An ADC End of Conversion will cause an ADC Window Compare Interrupt (ADWINT=1) if the resulting ADC Data Word is < 0x0100 or > 0x0200.

Figure 6.3. 12-Bit ADC Window Interrupt Examples, Right Justified Data

9.2.5. Stack

A programmer's stack can be located anywhere in the 256-byte data memory. The stack area is designated using the Stack Pointer (SP, 0x81) SFR. The SP will point to the last location used. The next value pushed on the stack is placed at SP+1 and then SP is incremented. A reset initializes the stack pointer to location 0x07. Therefore, the first value pushed on the stack is placed at location 0x08, which is also the first register (R0) of register bank 1. Thus, if more than one register bank is to be used, the SP should be initialized to a location in the data memory not being used for data storage. The stack depth can extend up to 256 bytes.

The MCU also has built-in hardware for a stack record. The stack record is a 32-bit shift register, where each Push or increment SP pushes one record bit onto the register, and each Call pushes two record bits onto the register. (A Pop or decrement SP pops one record bit, and a Return pops two record bits, also.) The stack record circuitry can also detect an overflow or underflow on the 32-bit shift register, and can notify the emulator software even with the MCU running full-speed debug.

Table 9.4. Interrupt Summary (Continued)

Interrupt Source	Interrupt Vector	Priority Order	Interrupt-Pending Flag	Enable
ADC0 End of Conversion	0x007B	15	ADCINT (ADC0CN.5)	EADC0 (EIE2.1)
Software Controlled Interrupt 0	0x0083	16	SCI0 (SWCINT.4)	ESCI0 (EIE2.2)
Software Controlled Interrupt 1	0x008B	17	SCI1 (SWCINT.5)	ESCI1 (EIE2.3)
Software Controlled Interrupt 2	0x0093	18	SCI2 (SWCINT.6)	ESCI2 (EIE2.4)
Software Controlled Interrupt 3	0x009B	19	SCI3 (SWCINT.7)	ESCI3 (EIE2.5)
Unused Interrupt Location	0x00A3	20	None	Reserved (EIE2.6)
External Crystal OSC Ready	0x00AB	21	XTLVLD (OSCXCN.7)	EXVLD (EIE2.7)

9.4.4. Interrupt Priorities

Each interrupt source can be individually programmed to one of two priority levels: low or high. A low priority interrupt service routine can be preempted by a high priority interrupt. A high priority interrupt cannot be preempted. Each interrupt has an associated interrupt priority bit in an SFR (IP–EIP2) used to configure its priority level. Low priority is the default. If two interrupts are recognized simultaneously, the interrupt with the higher priority is serviced first. If both interrupts have the same priority level, a fixed priority order is used to arbitrate.

9.4.5. Interrupt Latency

Interrupt response time depends on the state of the CPU when the interrupt occurs. Pending interrupts are sampled and priority decoded each system clock cycle. Therefore, the fastest possible response time is 5 system clock cycles: 1 clock cycle to detect the interrupt and 4 clock cycles to complete the LCALL to the ISR. If an interrupt is pending when a RETI is executed, a single instruction is executed before an LCALL is made to service the pending interrupt. Therefore, the maximum response time for an interrupt (when no other interrupt is currently being serviced or the new interrupt is of greater priority) occurs when the CPU is performing an RETI instruction followed by a DIV as the next instruction. In this case, the response time is 18 system clock cycles: 1 clock cycle to detect the interrupt, 5 clock cycles to execute the RETI, 8 clock cycles to complete the DIV instruction and 4 clock cycles to execute the LCALL to the ISR. NOTE: If a Flash write or erase is performed, the MCU is stalled during the operation and interrupts will not be serviced until the operation is complete. If the CPU is executing an ISR for an interrupt with equal or higher priority, the new interrupt will not be serviced until the current ISR completes, including the RETI and following instruction.

12.1. Power-on Reset

The CIP-51 incorporates a power supply monitor that holds the MCU in the reset state until V_{DD} rises above the VRST level during power-up. (See Figure 12.2 for timing diagram, and refer to Table 12.1 for the Electrical Characteristics of the power supply monitor circuit.) The $\overline{RST}$ pin is asserted (low) until the end of the 100msec V_{DD} Monitor timeout in order to allow the V_{DD} supply to become stable. On 48-pin packages, the V_{DD} monitor is enabled by pulling the MONEN pin high and is disabled by pulling the MONEN pin low. The MONEN pin should never be left floating. On 32-pin packages, the V_{DD} monitor is always enabled and cannot be disabled.

On exit from a power-on reset, the PORSF flag (RSTSRC.1) is set by hardware to logic 1. All of the other reset flags in the RSTSRC Register are indeterminate. PORSF is cleared by all other resets. Since all resets cause program execution to begin at the same location (0x0000), software can read the PORSF flag to determine if a power-up was the cause of reset. The content of internal data memory should be assumed to be undefined after a power-on reset.

12.2. Software Forced Reset

Writing a 1 to the PORSF bit forces a Power-On Reset as described in Section 12.1.

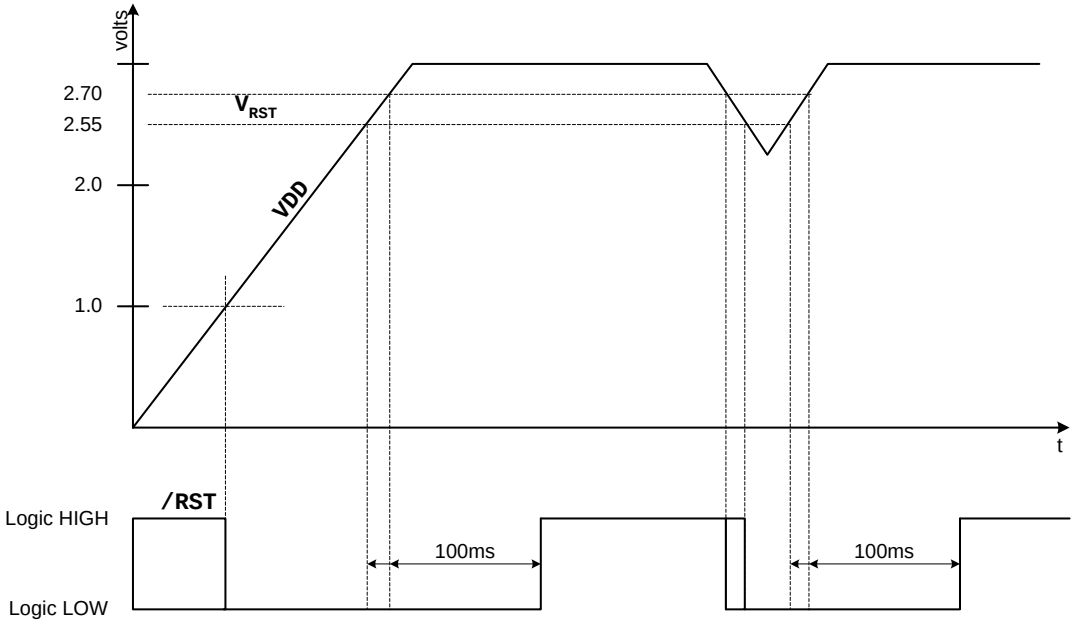


Figure 12.2. V_{DD} Monitor Timing Diagram

12.3. Power-fail Reset

When the V_{DD} monitor is enabled, the MONEN pin (not on C8051F221/F231 32 pin parts) is "pulled high", and power-down transition or power irregularity causes V_{DD} to drop below VRST, the power supply monitor will drive the $\overline{RST}$ pin low and return the CIP-51 to the reset state (see Figure 12.2). When V_{DD} returns to a level above VRST, the CIP-51 will leave the reset state in the same manner as that for the power-on reset. Note that even though internal data memory contents are not altered by the power-fail reset, it is impossible to determine if V_{DD} dropped below the level required for data retention. If the PORSF flag is set, the data may no longer be valid.

Disable WDT

Writing 0xDE followed by 0xAD to the WDTCN register disables the WDT. The following code segment illustrates disabling the WDT.

```
CLR    EA            ; disable all interrupts
MOV    WDTCN,#0DEh   ; disable watchdog timer
MOV    WDTCN,#0ADh   ;
SETB   EA            ; re-enable interrupts
```

The writes of 0xDE and 0xAD must occur within 4 clock cycles of each other, or the disable operation is ignored. Interrupts should be disabled during this procedure to avoid delay between the two writes.

Disable WDT Lockout

Writing 0xFF to WDTCN locks out the disable feature. Once locked out, the disable operation is ignored until the next system reset. Writing 0xFF does not enable or reset the watchdog timer. Applications alays intending to use the watchdog should write 0xFF to WDTCN in their initialization code.

Setting WDT Interval

WDTCN.[2:0] control the watchdog timeout interval. The interval is given by the following equation:

$$4^{3+WDTCN[2:0]} \times T_{SYSCLK}\text{ , (where } T_{SYSCLK} \text{ is the system clock period).}$$

For a 2.0 MHz system clock, this provides an interval range of 32msec to 524msec. WDTCN.7 must be written as 0 when setting this interval. Reading WDTCN returns the programmed interval. WDTCN.[2:0] is 111b after a system reset.

SFR Definition 12.1. WDTCN: Watchdog Timer Control

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								xxxxx111
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xFF

Bits7–0: WDT Control
Writing 0xA5 both enables and reloads the WDT.
Writing 0xDE followed within 4 clocks by 0xAD disables the WDT.
Writing 0xFF locks out the disable feature.

Bit4: Watchdog Status Bit (when Read)
Reading the WDTCN.[4] bit indicates the Watchdog Timer Status.
0: WDT is inactive
1: WDT is active

Bits2–0: Watchdog Timeout Interval Bits
The WDTCN.[2:0] bits set the Watchdog Timeout Interval. When writing these bits, WDTCN.7 must be set to 0.

Table 12.1. V_{DD} Monitor Electrical Characteristics

–40 to +85 ×C unless otherwise specified.

Parameter	Conditions	Min	Typ	Max	Units
$\overline{RST}$ Output Low Voltage	I_{OL} = 8.5 mA, V_{DD} = 2.7 to 3.6 V	—	—	0.6	V
$\overline{RST}$ Input High Voltage		0.8 x V_{DD}	—	—	V
$\overline{RST}$ Input Low Voltage		—	—	0.2 x V_{DD}	V
$\overline{RST}$ Input Leakage Current	$\overline{RST}$ = 0.0 V	—	—	50	µA
V_{DD} for $\overline{RST}$ Output Valid		1.0	—	—	V
Reset Threshold (Vrst)		2.40	2.55	2.70	V
Reset Time Delay	$\overline{RST}$ rising edge after crossing reset threshold	80	100	120	ms
Missing Clock Detector Timeout	Time from last system clock to reset generation	100	220	500	µs

13.1. External Crystal Example

If a crystal were used to generate the system clock for the MCU, the circuit would be as shown in Figure 13.1, Option 1. For an ECS-110.5-20-4 crystal, the resonate frequency is 11.0592 MHz, the intrinsic capacitance is 7 pF, and the ESR is 60 W. The compensation capacitors should be 33 pF each, and the PWB parasitic capacitance is estimated to be 2 pF. The appropriate External Oscillator Frequency Control value (XFCN) from the Crystal column in the table in SFR Definition 13.2 (OSCXCN Register) should be 111b.

The Crystal Oscillator Valid Flag (XTLVLD in register OSCXCN) is set to logic 1 by hardware when the external oscillator is running and stable. The XTLVLD detection circuit requires a startup time of at least 1ms between enabling the oscillator and checking the XTLVLD flag. Switching to the external oscillator before 1ms can result in unpredictable behavior. The recommend procedure is:

1. Enable the external oscillator
2. Wait 1 ms
3. Poll for XTLVLD '0' ==> '1'
4. Switch to the external oscillator

Switching to the external oscillator before the crystal oscillator has stabilized could result in unpredictable behavior.

NOTE: Crystal oscillator circuits are quite sensitive to PCB layout. The crystal should be placed as close as possible to the XTAL pins on the device, keeping the traces as short as possible and shielded with ground plane from any other traces which could introduce noise or interference.

13.2. External RC Example

If an external RC network were used to generate the system clock for the MCU, the circuit would be as shown in Figure 13.1, Option 2. The capacitor must be no greater than 100 pF, but using a very small capacitor will increase the frequency drift due to the PWB parasitic capacitance. To determine the required External Oscillator Frequency Control value (XFCN) in the OSCXCN Register, first select the RC network value to produce the desired frequency of oscillation. If the frequency desired is 100 kHz, let R = 246 kW and C = 50 pF:

f = 1.23(103)/RC = 1.23(103) / [246 x 50] = 0.1 MHz = 100 kHz

XFCN ³ log2(f/25 kHz)
XFCN ³ log2(100 kHz/25 kHz) = log2(4)
XFCN ³ 2, or code 010

13.3. External Capacitor Example

If an external capacitor were used to generate the system clock for the MCU, the circuit would be as shown in Figure 13.1, Option 3. The capacitor must be no greater than 100 pF, but using a very small capacitor will increase the frequency drift due to the PWB parasitic capacitance. To determine the required External Oscillator Frequency Control value (XFCN) in the OSCXCN Register, select the capacitor to be used and find the frequency of oscillation from the equations below. Assume V_{DD} = 3.0 V and C = 50 pF:

f = KF / (C x V_{DD}) = KF / (50 x 3)
f = KF / 150

If a frequency of roughly 90kHz is desired, select the K Factor from the table in SFR Definition 13.2 as KF = 13:

f = 13 /150 = 0.087 MHz, or 87 kHz

Therefore, the XFCN value to use in this example is 011.

SFR Definition 14.1. PRT0MX: Port I/O MUX Register 0

R/W	R/W	R/W	R/W	R/W	R/W	R	R/W	Reset Value
T2EXE	T2E	T1E	T0E	INT1E	INT0E	-	UARTEN	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xE1

Bit7: T2EXE: T2EX Enable Bit
0: T2EX unavailable at Port pin.
1: T2EX routed to Port Pin.

Bit6: T2E: T2 Enable Bit
0: T2 unavailable at Port pin.
1: T2 routed to Port Pin.

Bit5: T1E: T1 Enable Bit
0: T1 unavailable at Port pin.
1: T1 routed to Port Pin.

Bit4: T0E: T0 Enable Bit
0: T0 unavailable at Port pin.
1: T0 routed to Port Pin.

Bit3: INT1E: /INT1 Enable Bit
0: /INT1 unavailable at Port pin.
1: /INT1 routed to port pin.

Bit2: INT0E: /INT0 Enable Bit
0: /INT0 unavailable at Port pin.
1: /INT0 routed to Port Pin.

Bit1: UNUSED. Read = 0, Write = don't care.

Bit0: UARTEN: UART I/O Enable
0: UART I/O unavailable at port pins.
1: TX, RX routed to pins P0.0 and P0.1, respectively.

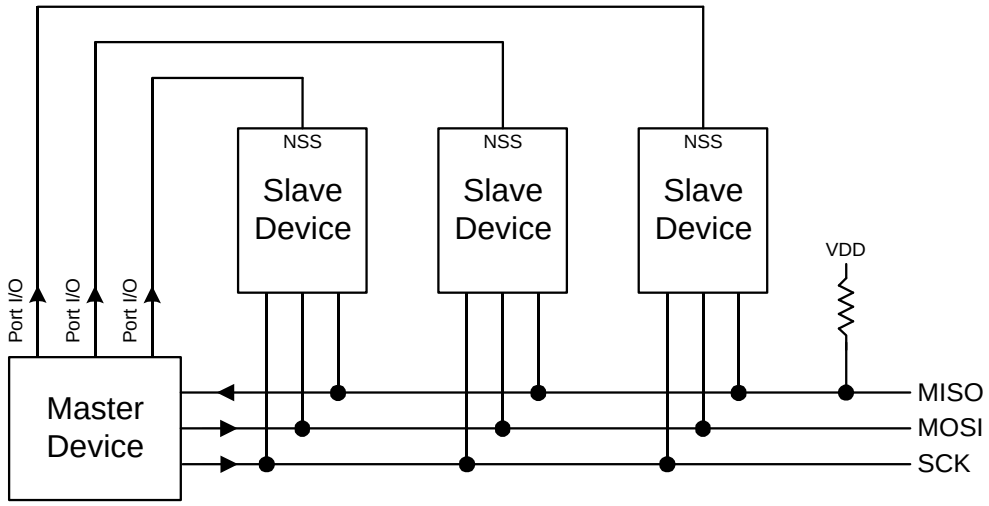


Figure 15.2. SPI Block Diagram

15.1. Signal Descriptions

The four signals used by the SPI (MOSI, MISO, SCK, NSS) are described below.

15.1.1. Master Out, Slave In

The master-out, slave-in (MOSI) signal is an output from a master device and an input to slave devices. It is used to serially transfer data from the master to the slave. Data is transferred most-significant bit first.

15.1.2. Master In, Slave Out

The master-in, slave-out (MISO) signal is an output from a slave device and an input to the master device. It is used to serially transfer data from the slave to the master. Data is transferred most-significant bit first. A SPI slave places the MISO pin in a high-impedance state when the slave is not selected.

15.1.3. Serial Clock

The serial clock (SCK) signal is an output from the master device and an input to slave devices. It is used to synchronize the transfer of data between the master and slave on the MOSI and MISO lines.

15.1.4. Slave Select

The slave select (NSS) signal is an input used to select the SPI module when in slave mode by a master, or to disable the SPI module when in master mode. When in slave mode, it is pulled low to initiate a data transfer and remains low for the duration of the transfer.

15.2. Operation

Only a SPI master device can initiate a data transfer. The SPI is placed in master mode by setting the Master Enable flag (MSTEN, SPI0CN.1). Writing a byte of data to the SPI data register (SPI0DAT) when in Master Mode starts a data transfer. The SPI master immediately shifts out the data serially on the MOSI line while providing the serial clock on SCK. The SPIF (SPI0CN.7) flag is set to logic 1 at the end of the transfer. If interrupts are enabled, an interrupt request is generated when the SPIF flag is set. The SPI master can be configured to shift in/out from one to eight bits in a transfer operation in order to accommodate slave devices with different word lengths. The SPIFRS bits in the SPI Configuration Register (SPI0CFG.[2:0]) are used to select the number of bits to shift in/out in a transfer operation.

While the SPI master transfers data to a slave on the MOSI line, the addressed SPI slave device simultaneously transfers the contents of its shift register to the SPI master on the MISO line in a full-duplex operation. The data byte received from the slave replaces the data in the master's data register. Therefore, the SPIF flag serves as both a transmit-complete and receive-data-ready flag. The data transfer in both directions is synchronized with the serial clock generated by the master. Figure 15.3 illustrates the full-duplex operation of an SPI master and an addressed slave.

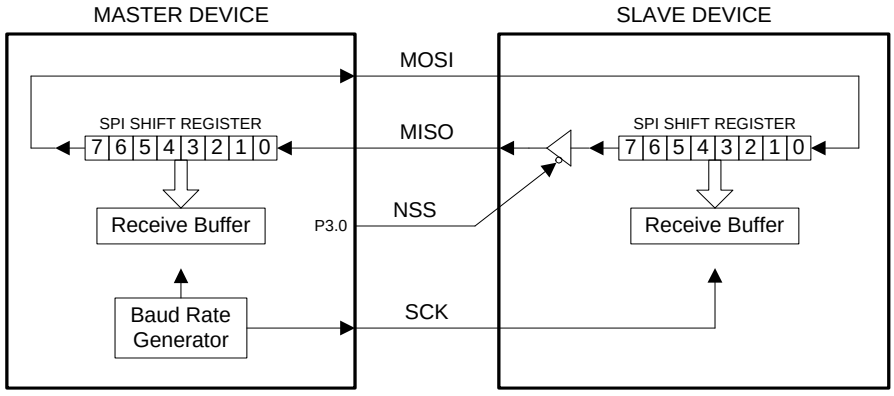


Figure 15.3. Full Duplex Operation

The SPI data register is double buffered on reads, but not on a write. If a write to SPI0DAT is attempted during a data transfer, the WCOL flag (SPI0CN.6) will be set to logic 1 and the write is ignored. The current data transfer will continue uninterrupted. A read of the SPI data register by the system controller actually reads the receive buffer. If the receive buffer still holds unread data from a previous transfer when the last bit of the current transfer is shifted into the SPI shift register, a receive overrun occurs and the RXOVRN flag (SPI0CN.4) is set to logic 1. The new data is not transferred to the receive buffer, allowing the previously received data byte to be read. The data byte causing the overrun is lost.

When the SPI is enabled and not configured as a master, it will operate as an SPI slave. Another SPI device acting as a master will initiate a transfer by driving the NSS signal low. The master then shifts data out of the shift register on the MOSI pin using the its serial clock. The SPIF flag is set to logic 1 at the end of a data transfer (when the NSS signal goes high). The slave can load its shift register for the next data transfer by writing to the SPI data register. The slave must make the write to the data register at least one SPI serial clock cycle before the master starts the next transmission. Otherwise, the byte of data already in the slave's shift register will be transferred.

Multiple masters may reside on the same bus. A Mode Fault flag (MODF, SPI0CN.5) is set to logic 1 when the SPI is configured as a master (MSTEN = 1) and its slave select signal NSS is pulled low. When the Mode Fault flag is set, the MSTEN and SPIEN bits of the SPI control register are cleared by hardware, thereby placing the SPI module in an "off-line" state. In a multiple-master environment, the system controller should check the state of the SLVSEL flag (SPI0CN.2) to ensure the bus is free before setting the MSTEN bit and initiating a data transfer.

15.2. Serial Clock Timing

As shown in Figure 15.4, four combinations of serial clock phase and polarity can be selected using the clock control bits in the SPI Configuration Register (SPI0CFG). The CKPHA bit (SPI0CFG.7) selects one of two clock phases (edge used to latch the data). The CKPOL bit (SPI0CFG.6) selects between an active-high or active-low clock. Both master and slave devices must be configured to use the same clock phase and polarity. Note: the SPI should be disabled (by clearing the SPIEN bit, SPI0CN.0) while changing the clock phase and polarity.

The SPI Clock Rate Register (SPI0CKR) as shown in SFR Definition 15.3 controls the master mode serial clock frequency. This register is ignored when operating in slave mode.

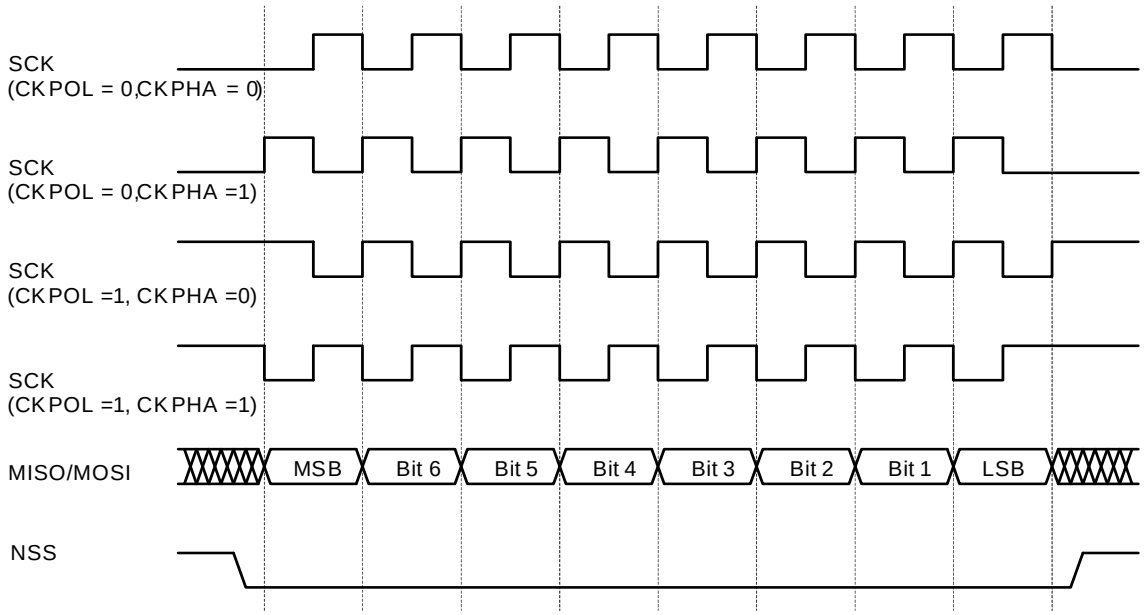


Figure 15.4. Full Duplex Operation

15.3. SPI Special Function Registers

The SPI is accessed and controlled through four special function registers in the system controller: SPI0CN Control Register, SPI0DAT Data Register, SPI0CFG Configuration Register, and SPI0CKR Clock Rate Register. The four special function registers related to the operation of the SPI Bus are described in the following section.

16.1. UART Operational Modes

The UART provides four operating modes (one synchronous and three asynchronous) selected by setting configuration bits in the SCON register. These four modes offer different baud rates and communication protocols. The four modes are summarized in Table 16.1 below. Detailed descriptions follow.

Table 16.1. UART Modes

Mode	Synchronization	Baud Clock	Data Bits	Start/Stop Bits
0	Synchronous	SYSCLK/12	8	None
1	Asynchronous	Timer 1 or Timer 2 Overflow	8	1 Start, 1 Stop
2	Asynchronous	SYSCLK/32 or SYSCLK/64	9	1 Start, 1 Stop
3	Asynchronous	Timer 1 or Timer 2 Overflow	9	1 Start, 1 Stop

16.1.1. Mode 0: Synchronous Mode

Mode 0 provides synchronous, half-duplex communication. Serial data is transmitted and received on the RX pin. The TX pin provides the shift clock for both transmit and receive. The MCU must be the master since it generates the shift clock for transmission in both directions (see the interconnect diagram in Figure 16.2).

Eight data bits are transmitted/received, LSB first (see the timing diagram in Figure 16.3). Data transmission begins when an instruction writes a data byte to the SBUF register. The TI Transmit Interrupt Flag (SCON.1) is set at the end of the eighth bit time. Data reception begins when the REN Receive Enable bit (SCON.4) is set to logic 1 and the RI Receive Interrupt Flag (SCON.0) is cleared. One cycle after the eighth bit is shifted in, the RI flag is set and reception stops until software clears the RI bit. An interrupt will occur if enabled when either TI or RI are set.

The Mode 0 baud rate is system clock frequency divided by twelve.

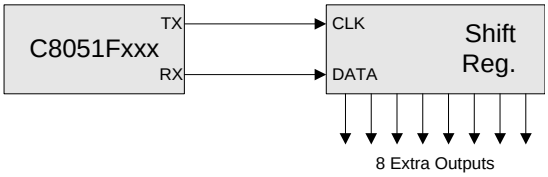


Figure 16.2. UART Mode 0 Interconnect

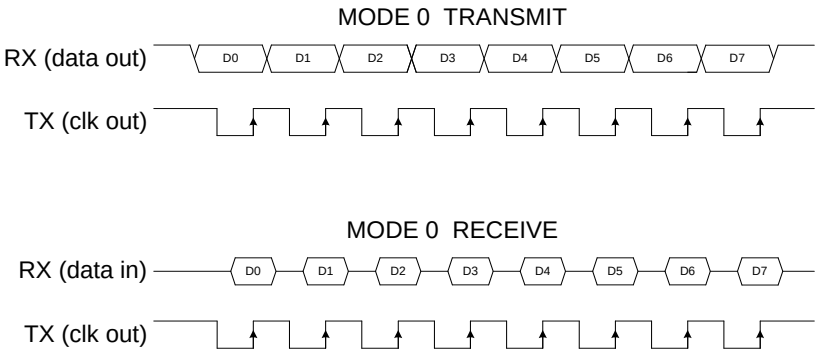


Figure 16.3. UART Mode 0 Timing Diagram

If T1M (CKCON.4) is logic 1, then the above equation becomes:

T1_OVERFLOWRATE = (SYSCLK) / (256 – TH1).

If T1M (CKCON.4) is logic 0, then the above equation becomes:

T1_OVERFLOWRATE = (SYSCLK/12) / (256 – TH1).

The Timer 2 overflow rate, when in Baud Rate Generator Mode and using an internal clock source, is determined solely by the Timer 2 16-bit reload value (RCAP2H:RCAP2L). The Timer 2 clock source is fixed at SYSCLK/2. The Timer 2 overflow rate can be calculated as follows:

T2_OVERFLOWRATE = (SYSCLK/2) / (65536 – [RCAP2H:RCAP2L]).

Timer 2 can be selected as the baud rate generator for RX and/or TX by setting RCLK (T2CON.5) and/or TCLK (T2CON.4), respectively. When either RCLK or TCLK is set to logic 1, Timer 2 interrupts are automatically disabled and the timer is forced into Baud Rate Generator Mode with SYSCLK/2 as its clock source. If a different timebase is required, setting the C/T2 bit (T2CON.1) to logic 1 will allow Timer 2 to be clocked from the external input pin T2. See the Timers section for complete timer configuration details.

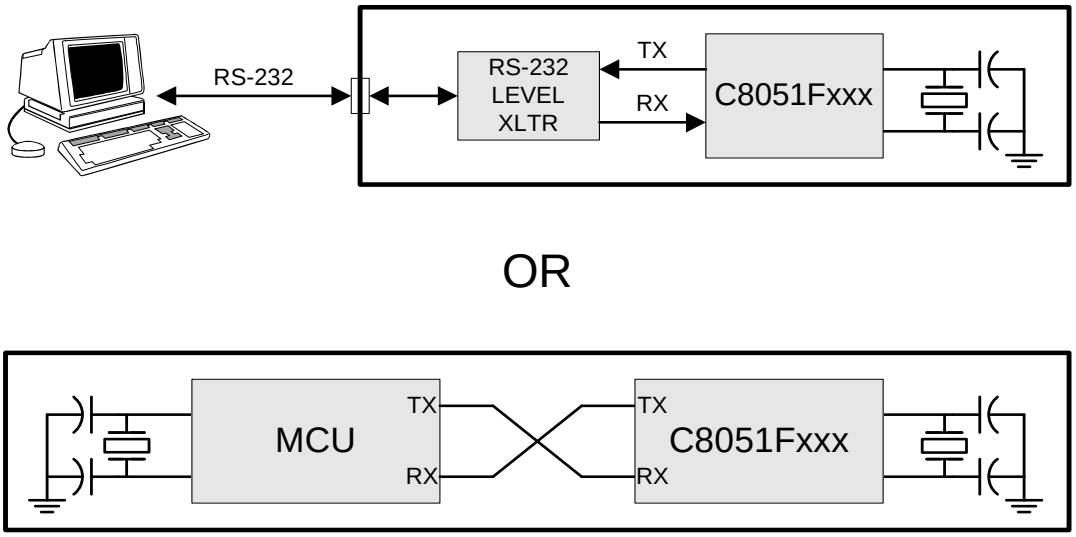


Figure 16.5. UART Modes 1, 2, and 3 Interconnect Diagram

16.2. Multiprocessor Communications

Modes 2 and 3 support multiprocessor communication between a master processor and one or more slave processors by special use of the ninth data bit. When a master processor wants to transmit to one or more slaves, it first sends an address byte to select the target(s). An address byte differs from a data byte in that its ninth bit is logic 1; in a data byte, the ninth bit is always set to logic 0.

Setting the SM2 bit (SCON.5) of a slave processor configures its UART such that when a stop bit is received, the UART will generate an interrupt only if the ninth bit is logic one (RB8 = 1) signifying an address byte has been received. In the UART's interrupt handler, software will compare the received address with the slave's own assigned 8-bit address. If the addresses match, the slave will clear its SM2 bit to enable interrupts on the reception of the following data byte(s). Slaves that weren't addressed leave their SM2 bits set and do not generate interrupts on the reception of the following data bytes, thereby ignoring the data. Once the entire message is received, the addressed slave resets its SM2 bit to ignore all transmissions until it receives the next address byte.

Multiple addresses can be assigned to a single slave and/or a single address can be assigned to multiple slaves, thereby enabling "broadcast" transmissions to more than one slave simultaneously. The master processor can be configured to receive all transmissions or a protocol can be implemented such that the master/slave role is temporarily reversed to enable half-duplex transmission between the original master and slave(s).

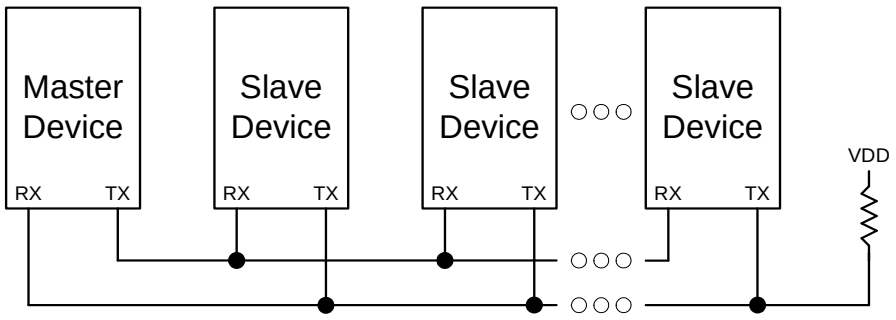


Figure 16.7. UART Multi-Processor Mode Interconnect Diagram

Table 16.2. Oscillator Frequencies for Standard Baud Rates

Oscillator Frequency (MHz)	Divide Factor	Timer 1 Load Value*	Resulting Baud Rate**
24.0	208	0xF3	115200 (115384)
23.592	205	0xF3	115200 (113423)
22.1184	192	0xF4	115200
18.432	160	0xF6	115200
16.5888	144	0xF7	115200
14.7456	128	0xF8	115200
12.9024	112	0xF9	115200
11.0592	96	0xFA	115200
9.216	80	0xFB	115200

17.1.4. Mode 3: Two 8-bit Counter/Timers (Timer 0 Only)

Timer 0 and Timer 1 behave differently in Mode 3. Timer 0 is configured as two separate 8-bit counter/timers held in TL0 and TH0. The counter/timer in TL0 is controlled using the Timer 0 control/status bits in TCON and TMOD: TR0, C/T0, GATE0 and TF0. It can use either the system clock or an external input signal as its time base. The TH0 register is restricted to a timer function sourced by the system clock. TH0 is enabled using the Timer 1 run control bit TR1. TH0 sets the Timer 1 overflow flag TF1 on overflow and thus controls the Timer 1 interrupt.

Timer 1 is inactive in Mode 3, so with Timer 0 in Mode 3, Timer 1 can be turned off and on by switching it into and out of its Mode 3. When Timer 0 is in Mode 3, Timer 1 can be operated in Modes 0, 1 or 2, but cannot be clocked by external signals nor set the TF1 flag and generate an interrupt. However, the Timer 1 overflow can be used for baud rate generation. Refer to Section 16 (UART) for information on configuring Timer 1 for baud rate generation.

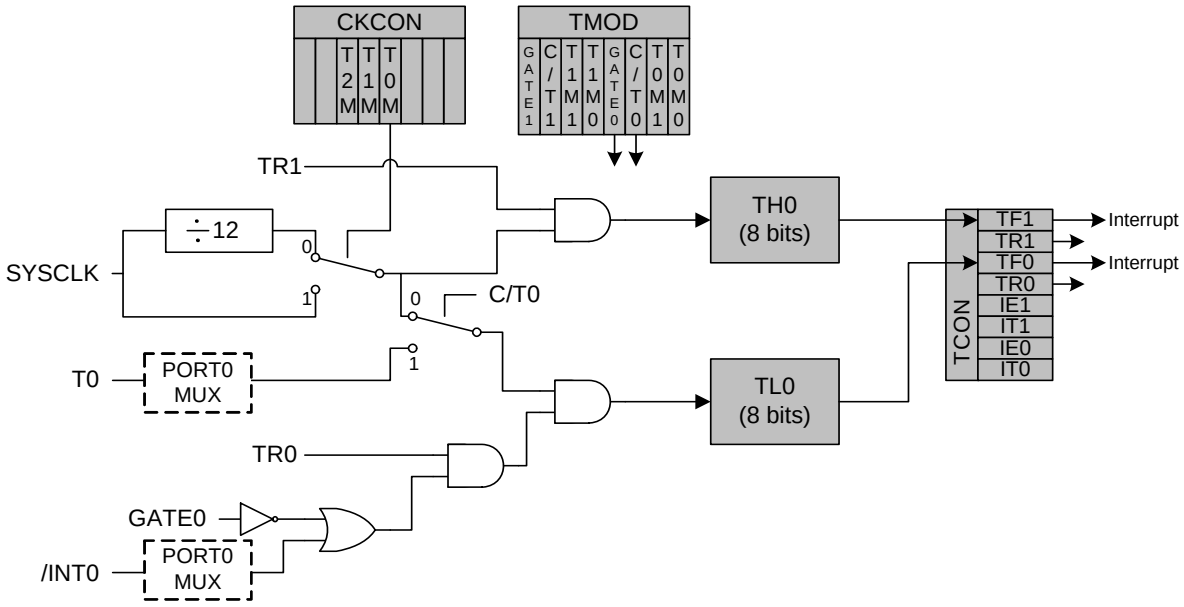


Figure 17.3. T0 Mode 3 Block Diagram

SFR Definition 17.9. RCAP2L: Timer 2 Capture Register Low Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xCA

Bits 7–0: RCAP2L: Timer 2 Capture Register Low Byte.
The RCAP2L register captures the low byte of Timer 2 when Timer 2 is configured in capture mode. When Timer 2 is configured in auto-reload mode, it holds the low byte of the reload value.

SFR Definition 17.10. RCAP2H: Timer 2 Capture Register High Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xCB

Bits 7–0: RCAP2H: Timer 2 Capture Register High Byte.
The RCAP2H register captures the high byte of Timer 2 when Timer 2 is configured in capture mode. When Timer 2 is configured in auto-reload mode, it holds the high byte of the reload value.

SFR Definition 17.11. TL2: Timer 2 Low Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xCC

Bits 7–0: TL2: Timer 2 Low Byte.
The TL2 register contains the low byte of the 16-bit Timer 2.

SFR Definition 17.12. TH2: Timer 2 High Byte

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xCD

Bits 7–0: TH2: Timer 2 High Byte.
The TH2 register contains the high byte of the 16-bit Timer 2.

JTAG Register Definition 18.4. FLASHDAT: JTAG Flash Data

										Reset Value
DATA7	DATA6	DATA5	DATA4	DATA3	DATA2	DATA1	DATA0	FAIL	BUSY	0000000000
Bit9	Bit8	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	

This register is used to read or write data to the Flash memory across the JTAG interface.

Bits9–2: DATA7–0: Flash Data Byte.
Bit1: FAIL: Flash Fail Bit.
- 0: Previous Flash memory operation was successful.
- 1: Previous Flash memory operation failed. Usually indicates the associated memory location was locked.

Bit0: BUSY: Flash Busy Bit.
- 0: Flash interface logic is not busy.
- 1: Flash interface logic is processing a request. Reads or writes while BUSY = 1 will not initiate another operation

JTAG Register Definition 18.5. FLASHSCL: JTAG Flash Scale

								Reset Value
FOSE	FRAE	-	-	FLSCL3	FLSCL2	FLSCL1	FLSCL0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	

This register controls the Flash read timing circuit and the prescaler required to generate the correct timing for Flash operations.

Bit7: FOSE: Flash One-Shot Enable Bit.
- 0: Flash read strobe is a full clock-cycle wide.
- 1: Flash read strobe is 50nsec.

Bit6: FRAE: Flash Read Always Bit.
- 0: The Flash output enable and sense amplifier enable are on only when needed to read the Flash memory.
- 1: The Flash output enable and sense amplifier enable are always on. This can be used to limit the variations in digital supply current due to switching the sense amplifiers, thereby reducing digitally induced noise.

Bits5–4: UNUSED. Read = 00b, Write = don't care.

Bits3–0: FLSCL3–0: Flash Prescaler Control Bits.
The FLSCL3–0 bits control the prescaler used to generate timing signals for Flash operations. Its value should be written before any Flash write or erase operations are initiated. The value written should be the smallest integer for which:

FLSCL[3:0] > log2(fSYSCLK / 50kHz)

Where fSYSCLK is the system clock frequency. All Flash read/write/erase operations are disallowed when FLSCL[3:0] = 1111b.