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Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	768
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	113
Number of Gates	12000
Voltage - Supply	2.25V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/a54sx08a-1tq144

General Description

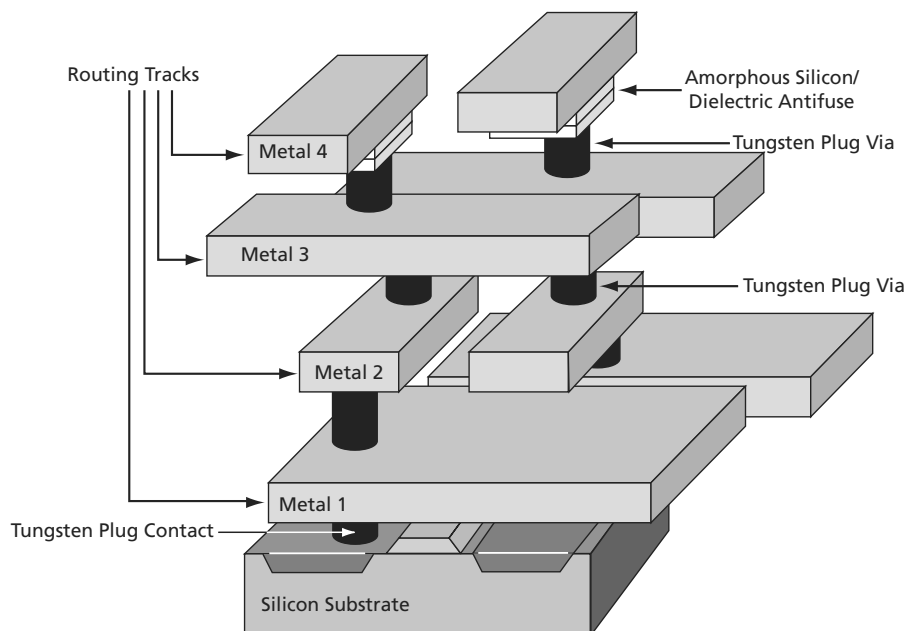
Introduction

The Actel SX-A family of FPGAs offers a cost-effective, single-chip solution for low-power, high-performance designs. Fabricated on 0.22 μm / 0.25 μm CMOS antifuse technology and with the support of 2.5 V, 3.3 V and 5 V I/Os, the SX-A is a versatile platform to integrate designs while significantly reducing time-to-market.

SX-A Family Architecture

The SX-A family's device architecture provides a unique approach to module organization and chip routing that satisfies performance requirements and delivers the most optimal register/logic mix for a wide variety of applications.

Interconnection between these logic modules is achieved using Actel's patented metal-to-metal programmable antifuse interconnect elements (Figure 1-1). The antifuses are normally open circuit and, when programmed, form a permanent low-impedance connection.



Note: The A54SX72A device has four layers of metal with the antifuse between Metal 3 and Metal 4. The A54SX08A, A54SX16A, and A54SX32A devices have three layers of metal with the antifuse between Metal 2 and Metal 3.

Figure 1-1 • SX-A Family Interconnect Elements

Logic Module Design

The SX-A family architecture is described as a “sea-of-modules” architecture because the entire floor of the device is covered with a grid of logic modules with virtually no chip area lost to interconnect elements or routing. The Actel SX-A family provides two types of logic modules: the register cell (R-cell) and the combinatorial cell (C-cell).

The R-cell contains a flip-flop featuring asynchronous clear, asynchronous preset, and clock enable, using the S0 and S1 lines control signals (Figure 1-2). The R-cell registers feature programmable clock polarity selectable on a register-by-register basis. This provides additional flexibility while allowing mapping of synthesized functions into the SX-A FPGA. The clock source for the R-cell can be chosen from either the hardwired clock, the routed clocks, or internal logic.

The C-cell implements a range of combinatorial functions of up to five inputs (Figure 1-3). Inclusion of the DB input and its associated inverter function allows up to 4,000

different combinatorial functions to be implemented in a single module. An example of the flexibility enabled by the inversion capability is the ability to integrate a 3-input exclusive-OR function into a single C-cell. This facilitates construction of 9-bit parity-tree functions with 1.9 ns propagation delays.

Module Organization

All C-cell and R-cell logic modules are arranged into horizontal banks called Clusters. There are two types of Clusters: Type 1 contains two C-cells and one R-cell, while Type 2 contains one C-cell and two R-cells.

Clusters are grouped together into SuperClusters (Figure 1-4 on page 1-3). SuperCluster 1 is a two-wide grouping of Type 1 Clusters. SuperCluster 2 is a two-wide group containing one Type 1 Cluster and one Type 2 Cluster. SX-A devices feature more SuperCluster 1 modules than SuperCluster 2 modules because designers typically require significantly more combinatorial logic than flip-flops.

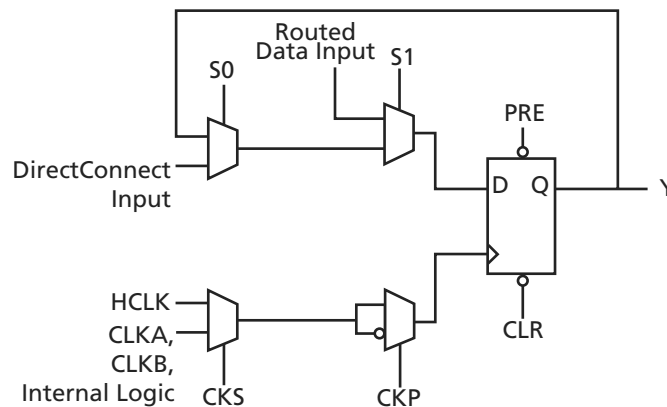


Figure 1-2 • R-Cell

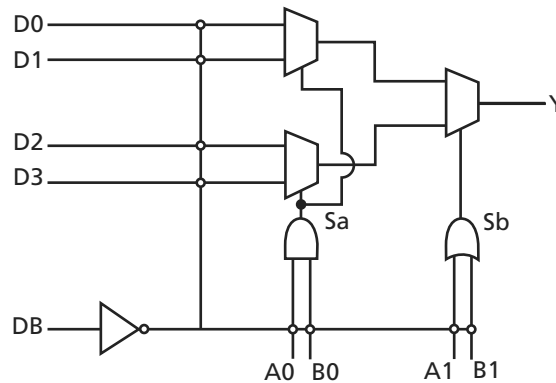


Figure 1-3 • C-Cell

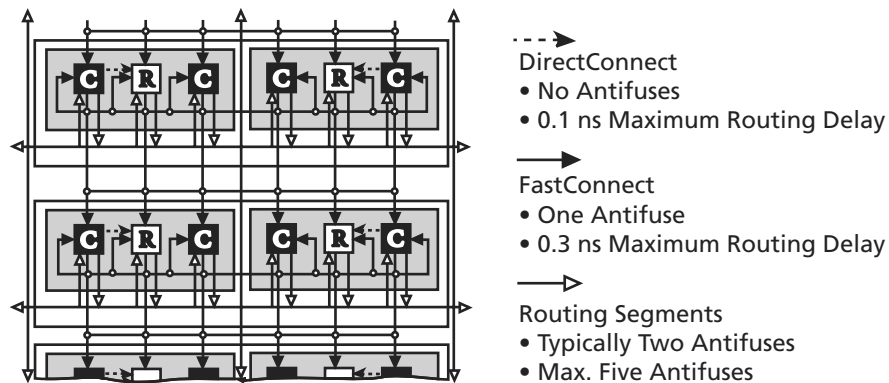


Figure 1-5 • DirectConnect and FastConnect for Type 1 SuperClusters

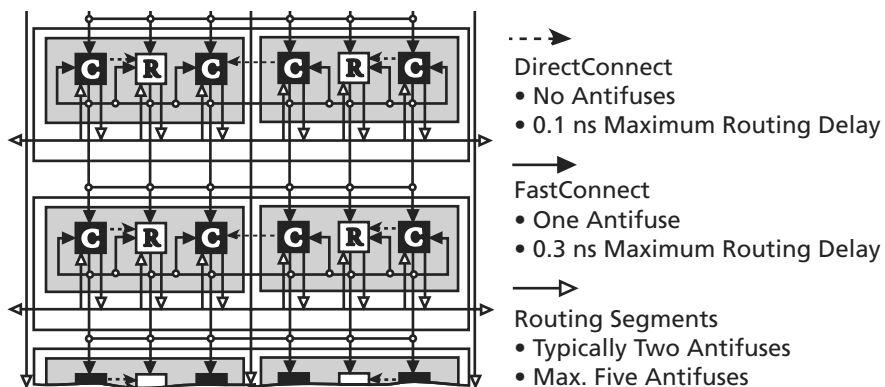


Figure 1-6 • DirectConnect and FastConnect for Type 2 SuperClusters

Boundary-Scan Testing (BST)

All SX-A devices are IEEE 1149.1 compliant and offer superior diagnostic and testing capabilities by providing Boundary Scan Testing (BST) and probing capabilities. The BST function is controlled through the special JTAG pins (TMS, TDI, TCK, TDO, and TRST). The functionality of the JTAG pins is defined by two available modes: Dedicated and Flexible. TMS cannot be employed as a user I/O in either mode.

Dedicated Mode

In Dedicated mode, all JTAG pins are reserved for BST; designers cannot use them as regular I/Os. An internal pull-up resistor is automatically enabled on both TMS and TDI pins, and the TMS pin will function as defined in the IEEE 1149.1 (JTAG) specification.

To select Dedicated mode, the user must reserve the JTAG pins in Actel's Designer software. Reserve the JTAG pins by checking the **Reserve JTAG** box in the Device Selection Wizard (Figure 1-12).

The default for the software is Flexible mode; all boxes are unchecked. Table 1-5 lists the definitions of the options in the Device Selection Wizard.

Flexible Mode

In Flexible mode, TDI, TCK, and TDO may be employed as either user I/Os or as JTAG input pins. The internal resistors on the TMS and TDI pins are not present in flexible JTAG mode.

To select the Flexible mode, uncheck the **Reserve JTAG** box in the Device Selection Wizard dialog in the Actel Designer software. In Flexible mode, TDI, TCK, and TDO pins may function as user I/Os or BST pins. The functionality is controlled by the BST Test Access Port (TAP) controller. The TAP controller receives two control inputs, TMS and TCK. Upon power-up, the TAP controller enters the Test-Logic-Reset state. In this state, TDI, TCK, and TDO function as user I/Os. The TDI, TCK, and TDO are transformed from user I/Os into BST pins when a rising edge on TCK is detected while TMS is at logic low. To return to Test-Logic Reset state, TMS must be high for at least five TCK cycles. **An external 10 k pull-up resistor to V_{CC} should be placed on the TMS pin to pull it High by default.**

Table 1-6 describes the different configuration requirements of BST pins and their functionality in different modes.

Table 1-6 • Boundary-Scan Pin Configurations and Functions

Mode	Designer "Reserve JTAG" Selection	TAP Controller State
Dedicated (JTAG)	Checked	Any
Flexible (User I/O)	Unchecked	Test-Logic-Reset
Flexible (JTAG)	Unchecked	Any EXCEPT Test-Logic-Reset

Figure 1-12 • Device Selection Wizard

Table 1-5 • Reserve Pin Definitions

Pin	Function
Reserve JTAG	Keeps pins from being used and changes the behavior of JTAG pins (no pull-up on TMS)
Reserve JTAG Test Reset	Regular I/O or JTAG reset with an internal pull-up
Reserve Probe	Keeps pins from being used or regular I/O

TRST Pin

The TRST pin functions as a dedicated Boundary-Scan Reset pin when the **Reserve JTAG Test Reset** option is selected as shown in Figure 1-12. An internal pull-up resistor is permanently enabled on the TRST pin in this mode. Actel recommends connecting this pin to ground in normal operation to keep the JTAG state controller in the Test-Logic-Reset state. When JTAG is being used, it can be left floating or can be driven high.

When the **Reserve JTAG Test Reset** option is not selected, this pin will function as a regular I/O. If unused as an I/O in the design, it will be configured as a tristated output.

Probing Capabilities

SX-A devices also provide an internal probing capability that is accessed with the JTAG pins. The Silicon Explorer II diagnostic hardware is used to control the TDI, TCK, TMS, and TDO pins to select the desired nets for debugging. The user assigns the selected internal nets in Actel Silicon Explorer II software to the PRA/PRB output pins for observation. Silicon Explorer II automatically places the device into JTAG mode. However, probing functionality is only activated when the TRST pin is driven high or left floating, allowing the internal pull-up resistor to pull TRST High. If the TRST pin is held Low, the TAP controller remains in the Test-Logic-Reset state so no probing can be performed. However, the user must drive the TRST pin High or allow the internal pull-up resistor to pull TRST High.

When selecting the **Reserve Probe Pin** box as shown in Figure 1-12 on page 1-9, direct the layout tool to reserve the PRA and PRB pins as dedicated outputs for probing. This **Reserve** option is merely a guideline. If the designer assigns user I/Os to the PRA and PRB pins and selects the **Reserve Probe Pin** option, Designer Layout will override the **Reserve Probe Pin** option and place the user I/Os on those pins.

To allow probing capabilities, the security fuse must not be programmed. Programming the security fuse disables the JTAG and probe circuitry. Table 1-9 summarizes the possible device configurations for probing once the device leaves the Test-Logic-Reset JTAG state.

Table 1-9 • Device Configuration Options for Probe Capability (TRST Pin Reserved)

JTAG Mode	TRST ¹	Security Fuse Programmed	PRA, PRB ²	TDI, TCK, TDO ²
Dedicated	Low	No	User I/O ³	JTAG Disabled
	High	No	Probe Circuit Outputs	JTAG I/O
Flexible	Low	No	User I/O ³	User I/O ³
	High	No	Probe Circuit Outputs	JTAG I/O
		Yes	Probe Circuit Secured	Probe Circuit Secured

Notes:

1. If the TRST pin is not reserved, the device behaves according to TRST = High as described in the table.
2. Avoid using the TDI, TCK, TDO, PRA, and PRB pins as input or bidirectional ports. Since these pins are active during probing, input signals will not pass through these pins and may cause contention.
3. If no user signal is assigned to these pins, they will behave as unused I/Os in this mode. Unused pins are automatically tristated by the Designer software.

Design Environment

The SX-A family of FPGAs is fully supported by both Actel Libero® Integrated Design Environment (IDE) and Designer FPGA development software. Actel Libero IDE is a design management environment, seamlessly integrating design tools while guiding the user through the design flow, managing all design and log files, and passing necessary design data among tools. Additionally, Libero IDE allows users to integrate both schematic and HDL synthesis into a single flow and verify the entire design in a single environment. Libero IDE includes Synplify® for Actel from Synplicity®, ViewDraw® for Actel from Mentor Graphics®, ModelSim® HDL Simulator from Mentor Graphics, WaveFormer Lite™ from SynaptiCAD™, and Designer software from Actel. Refer to the *Libero IDE flow* diagram for more information (located on the Actel website).

Actel Designer software is a place-and-route tool and provides a comprehensive suite of backend support tools for FPGA development. The Designer software includes timing-driven place-and-route, and a world-class integrated static timing analyzer and constraints editor. With the Designer software, a user can select and lock package pins while only minimally impacting the results of place-and-route. Additionally, the back-annotation flow is compatible with all the major simulators and the simulation results can be cross-probed with Silicon Explorer II, Actel's integrated verification and logic analysis tool. Another tool included in the Designer software is the SmarGen core generator, which easily creates popular and commonly used logic functions for implementation in your schematic or HDL design. Actel's Designer software is compatible with the most popular FPGA design entry and verification tools from companies such as Mentor Graphics, Synplicity, Synopsys, and Cadence Design Systems. The Designer software is available for both the Windows and UNIX operating systems.

Programming

Device programming is supported through Silicon Sculptor series of programmers. In particular, Silicon Sculptor is compact, robust, single-site and multi-site device programmer for the PC.

With standalone software, Silicon Sculptor allows concurrent programming of multiple units from the same PC, ensuring the fastest programming times possible. Each fuse is subsequently verified by Silicon Sculptor II to insure correct programming. In addition, integrity tests ensure that no extra fuses are programmed. Silicon Sculptor also provides extensive hardware self-testing capability.

The procedure for programming an SX-A device using Silicon Sculptor is as follows:

1. Load the .AFM file
2. Select the device to be programmed
3. Begin programming

When the design is ready to go to production, Actel offers device volume-programming services either through distribution partners or via in-house programming from the factory.

For detailed information on programming, read the following documents *Programming Antifuse Devices* and *Silicon Sculptor User's Guide*.

Detailed Specifications

Operating Conditions

Table 2-1 • Absolute Maximum Ratings

Symbol	Parameter	Limits	Units
V_{CCI}	DC Supply Voltage for I/Os	−0.3 to +6.0	V
V_{CCA}	DC Supply Voltage for Arrays	−0.3 to +3.0	V
V_I	Input Voltage	−0.5 to +5.75	V
V_O	Output Voltage	−0.5 to + V_{CCI} + 0.5	V
T_{STG}	Storage Temperature	−65 to +150	°C

Note: *Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum rated conditions for extended periods may affect device reliability. Devices should not be operated outside the "Recommended Operating Conditions".

Table 2-2 • Recommended Operating Conditions

Parameter	Commercial	Industrial	Units
Temperature Range	0 to +70	−40 to +85	°C
2.5 V Power Supply Range (V_{CCA} and V_{CCI})	2.25 to 2.75	2.25 to 2.75	V
3.3 V Power Supply Range (V_{CCI})	3.0 to 3.6	3.0 to 3.6	V
5 V Power Supply Range (V_{CCI})	4.75 to 5.25	4.75 to 5.25	V

Typical SX-A Standby Current

Table 2-3 • Typical Standby Current for SX-A at 25°C with $V_{CCA} = 2.5$ V

Product	$V_{CCI} = 2.5$ V	$V_{CCI} = 3.3$ V	$V_{CCI} = 5$ V
A54SX08A	0.8 mA	1.0 mA	2.9 mA
A54SX16A	0.8 mA	1.0 mA	2.9 mA
A54SX32A	0.9 mA	1.0 mA	3.0 mA
A54SX72A	3.6 mA	3.8 mA	4.5 mA

Table 2-4 • Supply Voltages

V_{CCA}	V_{CCI}^*	Maximum Input Tolerance	Maximum Output Drive
2.5 V	2.5 V	5.75 V	2.7 V
2.5 V	3.3 V	5.75 V	3.6 V
2.5 V	5 V	5.75 V	5.25 V

Note: *3.3 V PCI is not 5 V tolerant due to the clamp diode, but instead is 3.3 V tolerant.

Figure 2-1 shows the 5 V PCI V/I curve and the minimum and maximum PCI drive characteristics of the SX-A family.

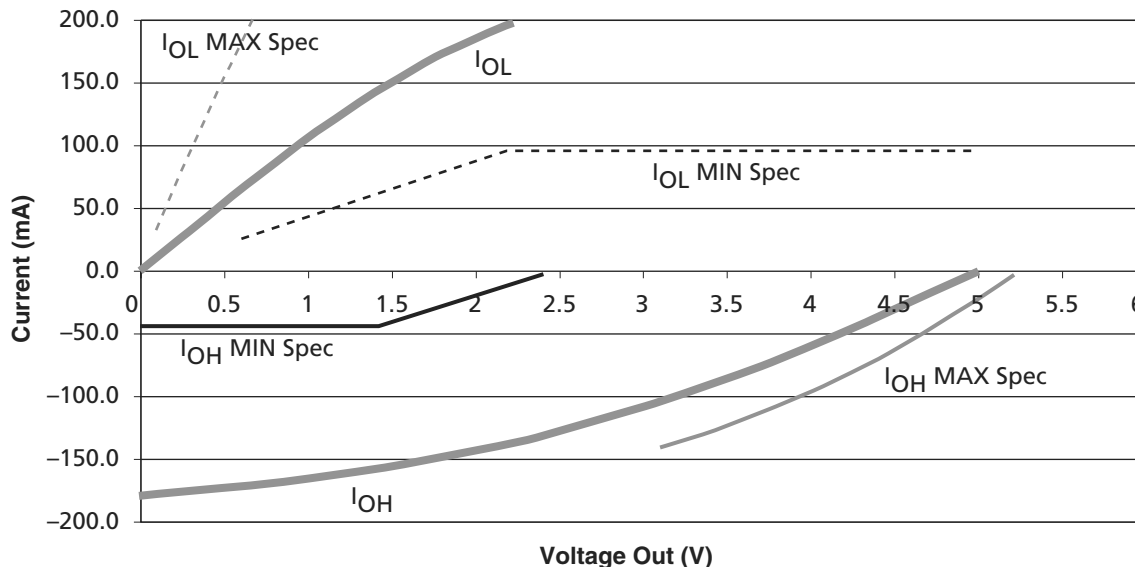


Figure 2-1 • 5 V PCI V/I Curve for SX-A Family

$$I_{OH} = 11.9 * (V_{OUT} - 5.25) * (V_{OUT} + 2.45)$$

for $V_{CCI} > V_{OUT} > 3.1V$

EQ 2-1

$$I_{OL} = 78.5 * V_{OUT} * (4.4 - V_{OUT})$$

for $0V < V_{OUT} < 0.71V$

EQ 2-2

Table 2-9 • DC Specifications (3.3 V PCI Operation)

Symbol	Parameter	Condition	Min.	Max.	Units
V_{CCA}	Supply Voltage for Array		2.25	2.75	V
V_{CCI}	Supply Voltage for I/Os		3.0	3.6	V
V_{IH}	Input High Voltage		$0.5V_{CCI}$	$V_{CCI} + 0.5$	V
V_{IL}	Input Low Voltage		-0.5	$0.3V_{CCI}$	V
I_{IPU}	Input Pull-up Voltage ¹		$0.7V_{CCI}$	–	V
I_{IL}	Input Leakage Current ²	$0 < V_{IN} < V_{CCI}$	-10	+10	μA
V_{OH}	Output High Voltage	$I_{OUT} = -500 \mu A$	$0.9V_{CCI}$	–	V
V_{OL}	Output Low Voltage	$I_{OUT} = 1,500 \mu A$		$0.1V_{CCI}$	V
C_{IN}	Input Pin Capacitance ³		–	10	pF
C_{CLK}	CLK Pin Capacitance		5	12	pF

Notes:

1. This specification should be guaranteed by design. It is the minimum voltage to which pull-up resistors are calculated to pull a floated network. Designers should ensure that the input buffer is conducting minimum current at this input voltage in applications sensitive to static power utilization.
2. Input leakage currents include hi-Z output leakage for all bidirectional buffers with tristate outputs.
3. Absolute maximum pin capacitance for a PCI input is 10 pF (except for CLK).

Table 2-10 • AC Specifications (3.3 V PCI Operation)

Symbol	Parameter	Condition	Min.	Max.	Units
$I_{OH(AC)}$	Switching Current High	$0 < V_{OUT} \leq 0.3V_{CCI}^1$	$-12V_{CCI}$	–	mA
		$0.3V_{CCI} \leq V_{OUT} < 0.9V_{CCI}^1$	$(-17.1(V_{CCI} - V_{OUT}))$	–	mA
		$0.7V_{CCI} < V_{OUT} < V_{CCI}^{1,2}$	–	EQ 2-3 on page 2-7	–
	(Test Point)	$V_{OUT} = 0.7V_{CC}^2$	–	$-32V_{CCI}$	mA
$I_{OL(AC)}$	Switching Current Low	$V_{CCI} > V_{OUT} \geq 0.6V_{CCI}^1$	$16V_{CCI}$	–	mA
		$0.6V_{CCI} > V_{OUT} > 0.1V_{CCI}^1$	$(26.7V_{OUT})$	–	mA
		$0.18V_{CCI} > V_{OUT} > 0^{1,2}$	–	EQ 2-4 on page 2-7	–
	(Test Point)	$V_{OUT} = 0.18V_{CC}^2$	–	$38V_{CCI}$	mA
I_{CL}	Low Clamp Current	$-3 < V_{IN} \leq -1$	$-25 + (V_{IN} + 1)/0.015$	–	mA
I_{CH}	High Clamp Current	$V_{CCI} + 4 > V_{IN} \geq V_{CCI} + 1$	$25 + (V_{IN} - V_{CCI} - 1)/0.015$	–	mA
$slew_R$	Output Rise Slew Rate	$0.2V_{CCI} - 0.6V_{CCI}$ load ³	1	4	V/ns
$slew_F$	Output Fall Slew Rate	$0.6V_{CCI} - 0.2V_{CCI}$ load ³	1	4	V/ns

Notes:

1. Refer to the *VII* curves in Figure 2-2 on page 2-7. Switching current characteristics for REQ# and GNT# are permitted to be one half of that specified here; i.e., half size output drivers may be used on these signals. This specification does not apply to CLK and RST#, which are system outputs. "Switching Current High" specifications are not relevant to SERR#, INTA#, INTB#, INTC#, and INTD#, which are open drain outputs.
2. Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (C and D) are provided with the respective diagrams in Figure 2-2 on page 2-7. The equation defined maximum should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.
3. This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per the latest revision of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is required (the maximum is no longer simply a guideline). Rise slew rate does not apply to open drain outputs.

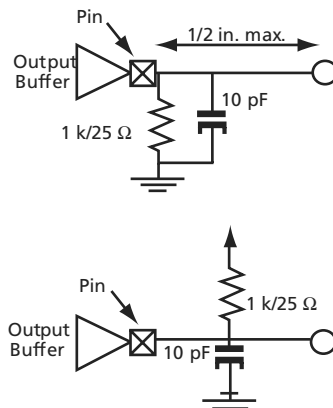


Figure 2-2 shows the 3.3 V PCI V/I curve and the minimum and maximum PCI drive characteristics of the SX-A family.

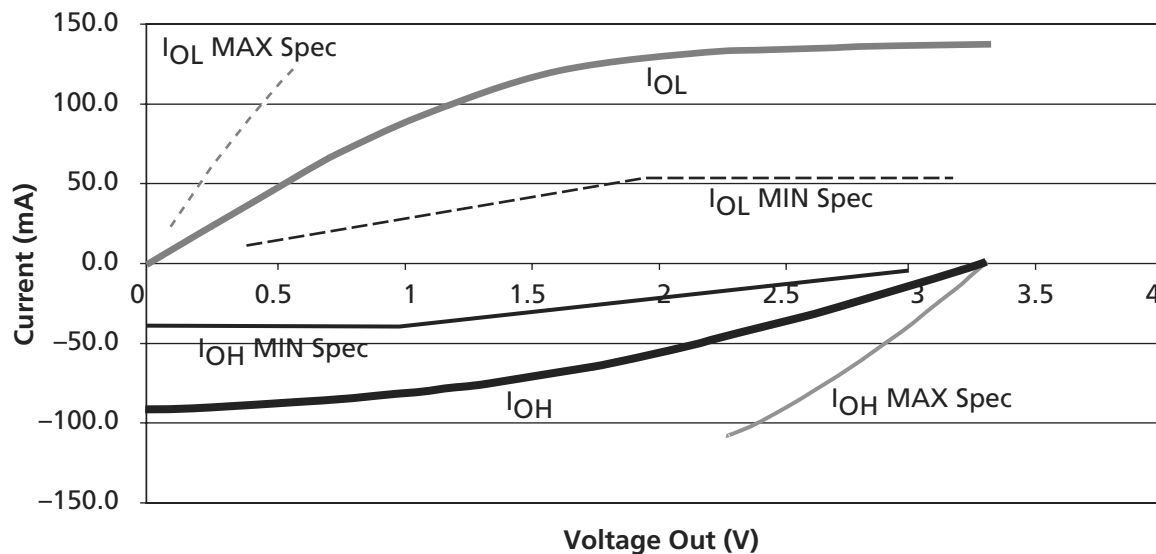


Figure 2-2 • 3.3 V PCI V/I Curve for SX-A Family

$$I_{OH} = (98.0/V_{CCI}) * (V_{OUT} - V_{CCI}) * (V_{OUT} + 0.4V_{CCI})$$

for $0.7 V_{CCI} < V_{OUT} < V_{CCI}$

EQ 2-3

$$I_{OL} = (256/V_{CCI}) * V_{OUT} * (V_{CCI} - V_{OUT})$$

for $0V < V_{OUT} < 0.18 V_{CCI}$

EQ 2-4

Theta-JA

Junction-to-ambient thermal resistance (θ_{JA}) is determined under standard conditions specified by JESD-51 series but has little relevance in actual performance of the product in real application. It should be employed with caution but is useful for comparing the thermal performance of one package to another.

A sample calculation to estimate the absolute maximum power dissipation allowed (worst case) for a 329-pin PBGA package at still air is as follows. i.e.:

$$\theta_{JA} = 17.1^{\circ}\text{C/W} \text{ is taken from Table 2-12 on page 2-11}$$

$$T_A = 125^{\circ}\text{C} \text{ is the maximum limit of ambient (from the datasheet)}$$

$$\text{Max. Allowed Power} = \frac{\text{Max Junction Temp} - \text{Max. Ambient Temp}}{\theta_{JA}} = \frac{150^{\circ}\text{C} - 125^{\circ}\text{C}}{17.1^{\circ}\text{C/W}} = 1.46 \text{ W}$$

EQ 2-11

The device's power consumption must be lower than the calculated maximum power dissipation by the package.

The power consumption of a device can be calculated using the Actel power calculator. If the power consumption is higher than the device's maximum allowable power dissipation, then a heat sink can be attached on top of the case or the airflow inside the system must be increased.

Theta-JC

Junction-to-case thermal resistance (θ_{JC}) measures the ability of a device to dissipate heat from the surface of the chip to the top or bottom surface of the package. It is applicable for packages used with external heat sinks and only applies to situations where all or nearly all of the heat is dissipated through the surface in consideration. If the power consumption is higher than the calculated maximum power dissipation of the package, then a heat sink is required.

Calculation for Heat Sink

For example, in a design implemented in a FG484 package, the power consumption value using the power calculator is 3.00 W. The user-dependent data T_J and T_A are given as follows:

$$T_J = 110^{\circ}\text{C}$$

$$T_A = 70^{\circ}\text{C}$$

From the datasheet:

$$\theta_{JA} = 18.0^{\circ}\text{C/W}$$

$$\theta_{JC} = 3.2^{\circ}\text{C/W}$$

$$P = \frac{\text{Max Junction Temp} - \text{Max. Ambient Temp}}{\theta_{JA}} = \frac{110^{\circ}\text{C} - 70^{\circ}\text{C}}{18.0^{\circ}\text{C/W}} = 2.22 \text{ W}$$

EQ 2-12

The 2.22 W power is less than then required 3.00 W; therefore, the design requires a heat sink or the airflow where the device is mounted should be increased. The design's junction-to-air thermal resistance requirement can be estimated by:

$$\theta_{JA} = \frac{\text{Max Junction Temp} - \text{Max. Ambient Temp}}{P} = \frac{110^{\circ}\text{C} - 70^{\circ}\text{C}}{3.00 \text{ W}} = 13.33^{\circ}\text{C/W}$$

EQ 2-13

Table 2-20 • A54SX08A Timing Characteristics

(Worst-Case Commercial Conditions $V_{CCA} = 2.25\text{ V}$, $V_{CCI} = 4.75\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
5 V PCI Output Module Timing ¹										
t _{DLH}	Data-to-Pad Low to High		2.4		2.8		3.2		4.5	ns
t _{DHL}	Data-to-Pad High to Low		3.2		3.6		4.2		5.9	ns
t _{ENZL}	Enable-to-Pad, Z to L		1.5		1.7		2.0		2.8	ns
t _{ENZH}	Enable-to-Pad, Z to H		2.4		2.8		3.2		4.5	ns
t _{ENLZ}	Enable-to-Pad, L to Z		3.5		3.9		4.6		6.4	ns
t _{ENHZ}	Enable-to-Pad, H to Z		3.2		3.6		4.2		5.9	ns
d _{TLH} ²	Delta Low to High		0.016		0.02		0.022		0.032	ns/pF
d _{THL} ²	Delta High to Low		0.03		0.032		0.04		0.052	ns/pF
5 V TTL Output Module Timing ³										
t _{DLH}	Data-to-Pad Low to High		2.4		2.8		3.2		4.5	ns
t _{DHL}	Data-to-Pad High to Low		3.2		3.6		4.2		5.9	ns
t _{DHLS}	Data-to-Pad High to Low—low slew		7.6		8.6		10.1		14.2	ns
t _{ENZL}	Enable-to-Pad, Z to L		2.4		2.7		3.2		4.5	ns
t _{ENZLS}	Enable-to-Pad, Z to L—low slew		8.4		9.5		11.0		15.4	ns
t _{ENZH}	Enable-to-Pad, Z to H		2.4		2.8		3.2		4.5	ns
t _{ENLZ}	Enable-to-Pad, L to Z		4.2		4.7		5.6		7.8	ns
t _{ENHZ}	Enable-to-Pad, H to Z		3.2		3.6		4.2		5.9	ns
d _{TLH}	Delta Low to High		0.017		0.017		0.023		0.031	ns/pF
d _{THL}	Delta High to Low		0.029		0.031		0.037		0.051	ns/pF
d _{THLS}	Delta High to Low—low slew		0.046		0.057		0.066		0.089	ns/pF

Notes:

- Delays based on 50 pF loading.
- To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation:

$$\text{Slew Rate [V/ns]} = (0.1 * V_{CCI} - 0.9 * V_{CCI}) / (C_{load} * d_{T[HL|HL|HLS]})$$
where C_{load} is the load capacitance driven by the I/O in pF
 $d_{T[HL|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.
- Delays based on 35 pF loading.

Table 2-25 • A54SX16A Timing Characteristics

(Worst-Case Commercial Conditions $V_{CCA} = 2.25\text{ V}$, $V_{CCI} = 2.25\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	–3 Speed ¹		–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
2.5 V LVCMOS Output Module Timing ^{2, 3}												
t _{DLH}	Data-to-Pad Low to High		3.4		3.9		4.5		5.2		7.3	ns
t _{DHL}	Data-to-Pad High to Low		2.6		3.0		3.3		3.9		5.5	ns
t _{DHLS}	Data-to-Pad High to Low—low slew		11.6		13.4		15.2		17.9		25.0	ns
t _{ENZL}	Enable-to-Pad, Z to L		2.4		2.8		3.2		3.7		5.2	ns
t _{ENZLS}	Data-to-Pad, Z to L—low slew		11.8		13.7		15.5		18.2		25.5	ns
t _{ENZH}	Enable-to-Pad, Z to H		3.4		3.9		4.5		5.2		7.3	ns
t _{ENLZ}	Enable-to-Pad, L to Z		2.1		2.5		2.8		3.3		4.7	ns
t _{ENHZ}	Enable-to-Pad, H to Z		2.6		3.0		3.3		3.9		5.5	ns
d _{TLH} ⁴	Delta Low to High		0.031		0.037		0.043		0.051		0.071	ns/pF
d _{THL} ⁴	Delta High to Low		0.017		0.017		0.023		0.023		0.037	ns/pF
d _{THLS} ⁴	Delta High to Low—low slew		0.057		0.06		0.071		0.086		0.117	ns/pF

Note:

1. All –3 speed grades have been discontinued.
2. Delays based on 35 pF loading.
3. The equivalent IO Attribute settings for 2.5 V LVCMOS is 2.5 V LVTTTL in the software.
4. To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation:

$$\text{Slew Rate [V/ns]} = (0.1 * V_{CCI} - 0.9 * V_{CCI}) / (C_{load} * d_{T[LH|HL|HLS]})$$
 where C_{load} is the load capacitance driven by the I/O in pF
 $d_{T[LH|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.

Table 2-35 • A54SX72A Timing Characteristics
(Worst-Case Commercial Conditions, $V_{CCA} = 2.25\text{ V}$, $V_{CCI} = 3.0\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	-3 Speed ¹		-2 Speed		-1 Speed		Std. Speed		-F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
C-Cell Propagation Delays ²												
t _{PD}	Internal Array Module	1.0		1.1		1.3		1.5		2.0		ns
Predicted Routing Delays ³												
t _{DC}	FO = 1 Routing Delay, Direct Connect	0.1		0.1		0.1		0.1		0.1		ns
t _{FC}	FO = 1 Routing Delay, Fast Connect	0.3		0.3		0.3		0.4		0.6		ns
t _{RD1}	FO = 1 Routing Delay	0.3		0.3		0.4		0.5		0.7		ns
t _{RD2}	FO = 2 Routing Delay	0.4		0.5		0.6		0.7		1		ns
t _{RD3}	FO = 3 Routing Delay	0.5		0.7		0.8		0.9		1.3		ns
t _{RD4}	FO = 4 Routing Delay	0.7		0.9		1		1.1		1.5		ns
t _{RD8}	FO = 8 Routing Delay	1.2		1.5		1.7		2.1		2.9		ns
t _{RD12}	FO = 12 Routing Delay	1.7		2.2		2.5		3		4.2		ns
R-Cell Timing												
t _{RCO}	Sequential Clock-to-Q	0.7		0.8		0.9		1.1		1.5		ns
t _{CLR}	Asynchronous Clear-to-Q	0.6		0.7		0.7		0.9		1.2		ns
t _{PRESET}	Asynchronous Preset-to-Q	0.7		0.8		0.8		1.0		1.4		ns
t _{SUD}	Flip-Flop Data Input Set-Up	0.7		0.8		0.9		1.0		1.4		ns
t _{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		0.0		ns
t _{WASYN}	Asynchronous Pulse Width	1.3		1.5		1.7		2.0		2.8		ns
t _{RECASYN}	Asynchronous Recovery Time	0.3		0.4		0.4		0.5		0.7		ns
t _{HASYN}	Asynchronous Hold Time	0.3		0.3		0.3		0.4		0.6		ns
t _{MPW}	Clock Minimum Pulse Width	1.5		1.7		2.0		2.3		3.2		ns
Input Module Propagation Delays												
t _{INYH}	Input Data Pad to Y High 2.5 V LVCMOS	0.6		0.7		0.8		0.9		1.3		ns
t _{INYL}	Input Data Pad to Y Low 2.5 V LVCMOS	0.8		1.0		1.1		1.3		1.7		ns
t _{INYH}	Input Data Pad to Y High 3.3 V PCI	0.6		0.7		0.7		0.9		1.2		ns
t _{INYL}	Input Data Pad to Y Low 3.3 V PCI	0.7		0.8		0.9		1.0		1.4		ns
t _{INYH}	Input Data Pad to Y High 3.3 V LVTTTL	0.7		0.7		0.8		1.0		1.4		ns
t _{INYL}	Input Data Pad to Y Low 3.3 V LVTTTL	1.0		1.2		1.3		1.5		2.1		ns

Notes:

1. All -3 speed grades have been discontinued.
2. For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.
3. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual performance.

Table 2-37 • A54SX72A Timing Characteristics (Continued)
(Worst-Case Commercial Conditions $V_{CCA} = 2.25\text{ V}$, $V_{CCI} = 3.0\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	–3 Speed*		–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t_{QCKH}	Input Low to High (100% Load) (Pad to R-cell Input)		1.7		1.9		2.2		2.5		3.5	ns
t_{QCHKL}	Input High to Low (100% Load) (Pad to R-cell Input)		1.7		2		2.2		2.6		3.6	ns
t_{QPWH}	Minimum Pulse Width High	1.5		1.7		2.0		2.3		3.2		ns
t_{QPWL}	Minimum Pulse Width Low	1.5		1.7		2.0		2.3		3.2		ns
t_{QCKSW}	Maximum Skew (Light Load)		0.2		0.3		0.3		0.3		0.5	ns
t_{QCKSW}	Maximum Skew (50% Load)		0.4		0.5		0.5		0.6		0.9	ns
t_{QCKSW}	Maximum Skew (100% Load)		0.4		0.5		0.5		0.6		0.9	ns

Note: *All –3 speed grades have been discontinued.

Table 2-40 • A54SX72A Timing Characteristics
(Worst-Case Commercial Conditions $V_{CCA} = 2.25\text{ V}$, $V_{CCI} = 3.0\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	–3 Speed ¹		–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
3.3 V PCI Output Module Timing ²												
t _{DLH}	Data-to-Pad Low to High	2.3		2.7		3.0		3.6		5.0		ns
t _{DHL}	Data-to-Pad High to Low	2.5		2.9		3.2		3.8		5.3		ns
t _{ENZL}	Enable-to-Pad, Z to L	1.4		1.7		1.9		2.2		3.1		ns
t _{ENZH}	Enable-to-Pad, Z to H	2.3		2.7		3.0		3.6		5.0		ns
t _{ENLZ}	Enable-to-Pad, L to Z	2.5		2.8		3.2		3.8		5.3		ns
t _{ENHZ}	Enable-to-Pad, H to Z	2.5		2.9		3.2		3.8		5.3		ns
d _{TLH} ³	Delta Low to High	0.025		0.03		0.03		0.04		0.045		ns/pF
d _{THL} ³	Delta High to Low	0.015		0.015		0.015		0.015		0.025		ns/pF
3.3 V LVTTTL Output Module Timing ⁴												
t _{DLH}	Data-to-Pad Low to High	3.2		3.7		4.2		5.0		6.9		ns
t _{DHL}	Data-to-Pad High to Low	3.2		3.7		4.2		4.9		6.9		ns
t _{DHLS}	Data-to-Pad High to Low—low slew	10.3		11.9		13.5		15.8		22.2		ns
t _{ENZL}	Enable-to-Pad, Z to L	2.2		2.6		2.9		3.4		4.8		ns
t _{ENZLS}	Enable-to-Pad, Z to L—low slew	15.8		18.9		21.3		25.4		34.9		ns
t _{ENZH}	Enable-to-Pad, Z to H	3.2		3.7		4.2		5.0		6.9		ns
t _{ENLZ}	Enable-to-Pad, L to Z	2.9		3.3		3.7		4.4		6.2		ns
t _{ENHZ}	Enable-to-Pad, H to Z	3.2		3.7		4.2		4.9		6.9		ns
d _{TLH} ³	Delta Low to High	0.025		0.03		0.03		0.04		0.045		ns/pF
d _{THL} ³	Delta High to Low	0.015		0.015		0.015		0.015		0.025		ns/pF
d _{THLS} ³	Delta High to Low—low slew	0.053		0.053		0.067		0.073		0.107		ns/pF

Notes:

1. All –3 speed grades have been discontinued.
2. Delays based on 10 pF loading and 25 Ω resistance.
3. To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation:

$$\text{Slew Rate [V/ns]} = (0.1 * V_{CCI} - 0.9 * V_{CCI}) / (C_{load} * d_{T[HLH|HLS]})$$
 where C_{load} is the load capacitance driven by the I/O in pF
 $d_{T[HLH|HLS]}$ is the worst case delta value from the datasheet in ns/pF.
4. Delays based on 35 pF loading.

208-Pin PQFP

Pin Number	A54SX08A Function	A54SX16A Function	A54SX32A Function	A54SX72A Function
141	NC	I/O	I/O	I/O
142	I/O	I/O	I/O	I/O
143	NC	I/O	I/O	I/O
144	I/O	I/O	I/O	I/O
145	V _{CCA}	V _{CCA}	V _{CCA}	V _{CCA}
146	GND	GND	GND	GND
147	I/O	I/O	I/O	I/O
148	V _{CCI}	V _{CCI}	V _{CCI}	V _{CCI}
149	I/O	I/O	I/O	I/O
150	I/O	I/O	I/O	I/O
151	I/O	I/O	I/O	I/O
152	I/O	I/O	I/O	I/O
153	I/O	I/O	I/O	I/O
154	I/O	I/O	I/O	I/O
155	NC	I/O	I/O	I/O
156	NC	I/O	I/O	I/O
157	GND	GND	GND	GND
158	I/O	I/O	I/O	I/O
159	I/O	I/O	I/O	I/O
160	I/O	I/O	I/O	I/O
161	I/O	I/O	I/O	I/O
162	I/O	I/O	I/O	I/O
163	I/O	I/O	I/O	I/O
164	V _{CCI}	V _{CCI}	V _{CCI}	V _{CCI}
165	I/O	I/O	I/O	I/O
166	I/O	I/O	I/O	I/O
167	NC	I/O	I/O	I/O
168	I/O	I/O	I/O	I/O
169	I/O	I/O	I/O	I/O
170	NC	I/O	I/O	I/O
171	I/O	I/O	I/O	I/O
172	I/O	I/O	I/O	I/O
173	NC	I/O	I/O	I/O
174	I/O	I/O	I/O	I/O
175	I/O	I/O	I/O	I/O

208-Pin PQFP

Pin Number	A54SX08A Function	A54SX16A Function	A54SX32A Function	A54SX72A Function
176	NC	I/O	I/O	I/O
177	I/O	I/O	I/O	I/O
178	I/O	I/O	I/O	QCLKD
179	I/O	I/O	I/O	I/O
180	CLKA	CLKA	CLKA	CLKA
181	CLKB	CLKB	CLKB	CLKB
182	NC	NC	NC	NC
183	GND	GND	GND	GND
184	V _{CCA}	V _{CCA}	V _{CCA}	V _{CCA}
185	GND	GND	GND	GND
186	PRA, I/O	PRA, I/O	PRA, I/O	PRA, I/O
187	I/O	I/O	I/O	V _{CCI}
188	I/O	I/O	I/O	I/O
189	NC	I/O	I/O	I/O
190	I/O	I/O	I/O	QCLKC
191	I/O	I/O	I/O	I/O
192	NC	I/O	I/O	I/O
193	I/O	I/O	I/O	I/O
194	I/O	I/O	I/O	I/O
195	NC	I/O	I/O	I/O
196	I/O	I/O	I/O	I/O
197	I/O	I/O	I/O	I/O
198	NC	I/O	I/O	I/O
199	I/O	I/O	I/O	I/O
200	I/O	I/O	I/O	I/O
201	V _{CCI}	V _{CCI}	V _{CCI}	V _{CCI}
202	NC	I/O	I/O	I/O
203	NC	I/O	I/O	I/O
204	I/O	I/O	I/O	I/O
205	NC	I/O	I/O	I/O
206	I/O	I/O	I/O	I/O
207	I/O	I/O	I/O	I/O
208	TCK, I/O	TCK, I/O	TCK, I/O	TCK, I/O

144-Pin TQFP			
Pin Number	A54SX08A Function	A54SX16A Function	A54SX32A Function
1	GND	GND	GND
2	TDI, I/O	TDI, I/O	TDI, I/O
3	I/O	I/O	I/O
4	I/O	I/O	I/O
5	I/O	I/O	I/O
6	I/O	I/O	I/O
7	I/O	I/O	I/O
8	I/O	I/O	I/O
9	TMS	TMS	TMS
10	V _{CCI}	V _{CCI}	V _{CCI}
11	GND	GND	GND
12	I/O	I/O	I/O
13	I/O	I/O	I/O
14	I/O	I/O	I/O
15	I/O	I/O	I/O
16	I/O	I/O	I/O
17	I/O	I/O	I/O
18	I/O	I/O	I/O
19	NC	NC	NC
20	V _{CCA}	V _{CCA}	V _{CCA}
21	I/O	I/O	I/O
22	TRST, I/O	TRST, I/O	TRST, I/O
23	I/O	I/O	I/O
24	I/O	I/O	I/O
25	I/O	I/O	I/O
26	I/O	I/O	I/O
27	I/O	I/O	I/O
28	GND	GND	GND
29	V _{CCI}	V _{CCI}	V _{CCI}
30	V _{CCA}	V _{CCA}	V _{CCA}
31	I/O	I/O	I/O
32	I/O	I/O	I/O
33	I/O	I/O	I/O
34	I/O	I/O	I/O
35	I/O	I/O	I/O
36	GND	GND	GND
37	I/O	I/O	I/O

144-Pin TQFP			
Pin Number	A54SX08A Function	A54SX16A Function	A54SX32A Function
38	I/O	I/O	I/O
39	I/O	I/O	I/O
40	I/O	I/O	I/O
41	I/O	I/O	I/O
42	I/O	I/O	I/O
43	I/O	I/O	I/O
44	V _{CCI}	V _{CCI}	V _{CCI}
45	I/O	I/O	I/O
46	I/O	I/O	I/O
47	I/O	I/O	I/O
48	I/O	I/O	I/O
49	I/O	I/O	I/O
50	I/O	I/O	I/O
51	I/O	I/O	I/O
52	I/O	I/O	I/O
53	I/O	I/O	I/O
54	PRB, I/O	PRB, I/O	PRB, I/O
55	I/O	I/O	I/O
56	V _{CCA}	V _{CCA}	V _{CCA}
57	GND	GND	GND
58	NC	NC	NC
59	I/O	I/O	I/O
60	HCLK	HCLK	HCLK
61	I/O	I/O	I/O
62	I/O	I/O	I/O
63	I/O	I/O	I/O
64	I/O	I/O	I/O
65	I/O	I/O	I/O
66	I/O	I/O	I/O
67	I/O	I/O	I/O
68	V _{CCI}	V _{CCI}	V _{CCI}
69	I/O	I/O	I/O
70	I/O	I/O	I/O
71	TDO, I/O	TDO, I/O	TDO, I/O
72	I/O	I/O	I/O
73	GND	GND	GND
74	I/O	I/O	I/O

176-Pin TQFP

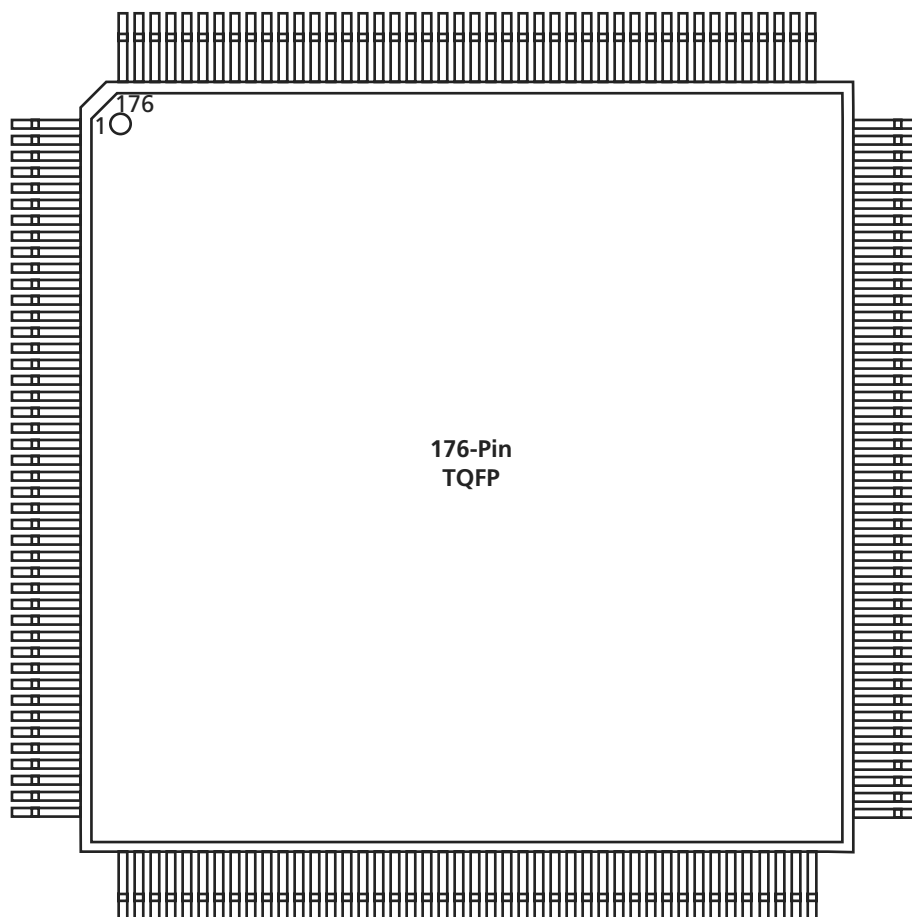


Figure 3-4 • 176-Pin TQFP (Top View)

Note

For Package Manufacturing and Environmental information, visit Resource center at <http://www.actel.com/products/rescenter/package/index.html>.

484-Pin FBGA		
Pin Number	A54SX32A Function	A54SX72A Function
AD18	I/O	I/O
AD19	I/O	I/O
AD20	I/O	I/O
AD21	I/O	I/O
AD22	I/O	I/O
AD23	V _{CCI}	V _{CCI}
AD24	NC*	I/O
AD25	NC*	I/O
AD26	NC*	I/O
AE1	NC*	NC
AE2	I/O	I/O
AE3	NC*	I/O
AE4	NC*	I/O
AE5	NC*	I/O
AE6	NC*	I/O
AE7	I/O	I/O
AE8	I/O	I/O
AE9	I/O	I/O
AE10	I/O	I/O
AE11	NC*	I/O
AE12	I/O	I/O
AE13	I/O	I/O
AE14	I/O	I/O
AE15	NC*	I/O
AE16	NC*	I/O
AE17	I/O	I/O
AE18	I/O	I/O
AE19	I/O	I/O
AE20	I/O	I/O
AE21	NC*	I/O
AE22	NC*	I/O
AE23	NC*	I/O
AE24	NC*	I/O
AE25	NC*	NC
AE26	NC*	NC

484-Pin FBGA		
Pin Number	A54SX32A Function	A54SX72A Function
AF1	NC*	NC
AF2	NC*	NC
AF3	NC	I/O
AF4	NC*	I/O
AF5	NC*	I/O
AF6	NC*	I/O
AF7	I/O	I/O
AF8	I/O	I/O
AF9	I/O	I/O
AF10	I/O	I/O
AF11	NC*	I/O
AF12	NC*	NC
AF13	HCLK	HCLK
AF14	I/O	QCLKB
AF15	NC*	I/O
AF16	NC*	I/O
AF17	I/O	I/O
AF18	I/O	I/O
AF19	I/O	I/O
AF20	NC*	I/O
AF21	NC*	I/O
AF22	NC*	I/O
AF23	NC*	I/O
AF24	NC*	I/O
AF25	NC*	NC
AF26	NC*	NC
B1	NC*	NC
B2	NC*	NC
B3	NC*	I/O
B4	NC*	I/O
B5	NC*	I/O
B6	I/O	I/O
B7	I/O	I/O
B8	I/O	I/O
B9	I/O	I/O

484-Pin FBGA		
Pin Number	A54SX32A Function	A54SX72A Function
B10	I/O	I/O
B11	NC*	I/O
B12	NC*	I/O
B13	V _{CCI}	V _{CCI}
B14	CLKA	CLKA
B15	NC*	I/O
B16	NC*	I/O
B17	I/O	I/O
B18	V _{CCI}	V _{CCI}
B19	I/O	I/O
B20	I/O	I/O
B21	NC*	I/O
B22	NC*	I/O
B23	NC*	I/O
B24	NC*	I/O
B25	I/O	I/O
B26	NC*	NC
C1	NC*	I/O
C2	NC*	I/O
C3	NC*	I/O
C4	NC*	I/O
C5	I/O	I/O
C6	V _{CCI}	V _{CCI}
C7	I/O	I/O
C8	I/O	I/O
C9	V _{CCI}	V _{CCI}
C10	I/O	I/O
C11	I/O	I/O
C12	I/O	I/O
C13	PRA, I/O	PRA, I/O
C14	I/O	I/O
C15	I/O	QCLKD
C16	I/O	I/O
C17	I/O	I/O
C18	I/O	I/O

Note: *These pins must be left floating on the A54SX32A device.