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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	768
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	111
Number of Gates	12000
Voltage - Supply	2.25V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	144-LBGA
Supplier Device Package	144-FPBGA (13x13)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a54sx08a-ffg144

Logic Module Design

The SX-A family architecture is described as a “sea-of-modules” architecture because the entire floor of the device is covered with a grid of logic modules with virtually no chip area lost to interconnect elements or routing. The Actel SX-A family provides two types of logic modules: the register cell (R-cell) and the combinatorial cell (C-cell).

The R-cell contains a flip-flop featuring asynchronous clear, asynchronous preset, and clock enable, using the S0 and S1 lines control signals (Figure 1-2). The R-cell registers feature programmable clock polarity selectable on a register-by-register basis. This provides additional flexibility while allowing mapping of synthesized functions into the SX-A FPGA. The clock source for the R-cell can be chosen from either the hardwired clock, the routed clocks, or internal logic.

The C-cell implements a range of combinatorial functions of up to five inputs (Figure 1-3). Inclusion of the DB input and its associated inverter function allows up to 4,000

different combinatorial functions to be implemented in a single module. An example of the flexibility enabled by the inversion capability is the ability to integrate a 3-input exclusive-OR function into a single C-cell. This facilitates construction of 9-bit parity-tree functions with 1.9 ns propagation delays.

Module Organization

All C-cell and R-cell logic modules are arranged into horizontal banks called Clusters. There are two types of Clusters: Type 1 contains two C-cells and one R-cell, while Type 2 contains one C-cell and two R-cells.

Clusters are grouped together into SuperClusters (Figure 1-4 on page 1-3). SuperCluster 1 is a two-wide grouping of Type 1 Clusters. SuperCluster 2 is a two-wide group containing one Type 1 Cluster and one Type 2 Cluster. SX-A devices feature more SuperCluster 1 modules than SuperCluster 2 modules because designers typically require significantly more combinatorial logic than flip-flops.

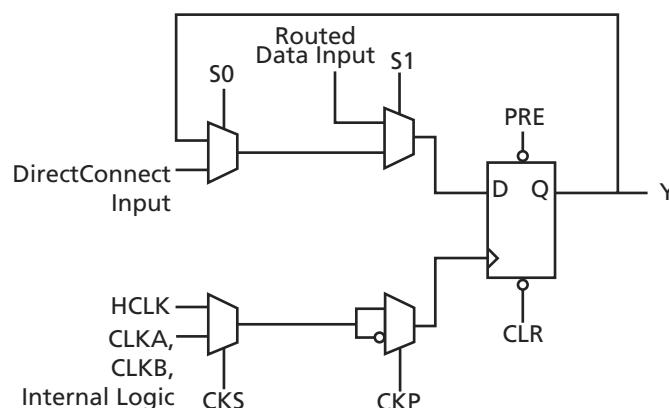


Figure 1-2 • R-Cell

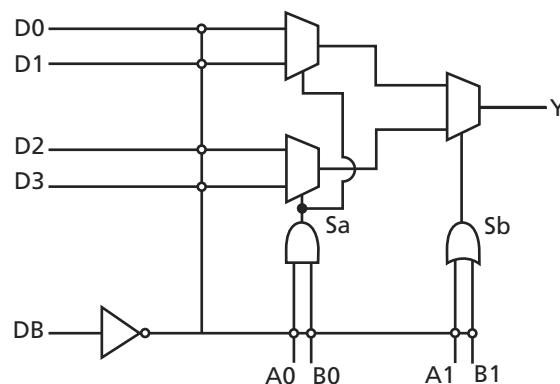


Figure 1-3 • C-Cell

Clock Resources

Actel's high-drive routing structure provides three clock networks (Table 1-1). The first clock, called HCLK, is hardwired from the HCLK buffer to the clock select multiplexor (MUX) in each R-cell. HCLK cannot be connected to combinatorial logic. This provides a fast propagation path for the clock signal. If not used, this pin must be set as Low or High on the board. It must not be left floating. Figure 1-7 describes the clock circuit used for the constant load HCLK and the macros supported.

HCLK does not function until the fourth clock cycle each time the device is powered up to prevent false output levels due to any possible slow power-on-reset signal and fast start-up clock circuit. To activate HCLK from the first cycle, the TRST pin must be reserved in the Design software and the pin must be tied to GND on the board.

Two additional clocks (CLKA, CLKB) are global clocks that can be sourced from external pins or from internal logic signals within the SX-A device. CLKA and CLKB may be connected to sequential cells or to combinational logic. If CLKA or CLKB pins are not used or sourced from signals, these pins must be set as Low or High on the board. They must not be left floating. Figure 1-8 describes the CLKA

and CLKB circuit used and the macros supported in SX-A devices with the exception of A54SX72A.

In addition, the A54SX72A device provides four quadrant clocks (QCLKA, QCLKB, QCLKC, and QCLKD—corresponding to bottom-left, bottom-right, top-left, and top-right locations on the die, respectively), which can be sourced from external pins or from internal logic signals within the device. Each of these clocks can individually drive up to an entire quadrant of the chip, or they can be grouped together to drive multiple quadrants (Figure 1-9 on page 1-6). QCLK pins can function as user I/O pins. If not used, the QCLK pins must be tied Low or High on the board and must not be left floating.

For more information on how to use quadrant clocks in the A54SX72A device, refer to the *Global Clock Networks in Actel's Antifuse Devices* and *Using A54SX72A and RT54SX72S Quadrant Clocks* application notes.

The CLKA, CLKB, and QCLK circuits for A54SX72A as well as the macros supported are shown in Figure 1-10 on page 1-6. Note that bidirectional clock buffers are only available in A54SX72A. For more information, refer to the "Pin Description" section on page 1-15.

Table 1-1 • SX-A Clock Resources

	A54SX08A	A54SX16A	A54SX32A	A54SX72A
Routed Clocks (CLKA, CLKB)	2	2	2	2
Hardwired Clocks (HCLK)	1	1	1	1
Quadrant Clocks (QCLKA, QCLKB, QCLKC, QCLKD)	0	0	0	4

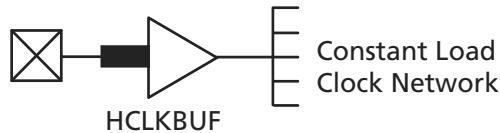


Figure 1-7 • SX-A HCLK Clock Buffer

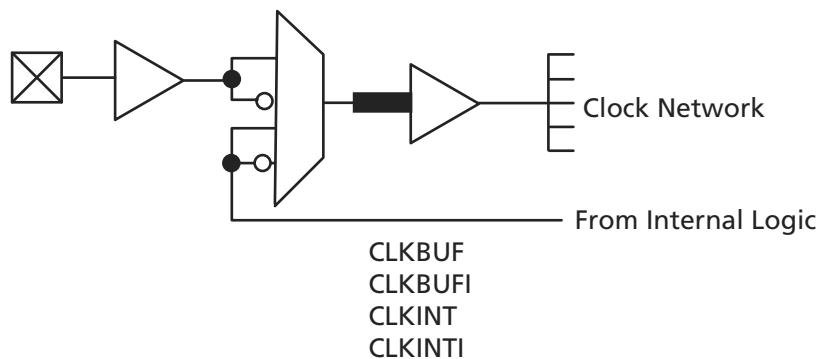


Figure 1-8 • SX-A Routed Clock Buffer

SX-A Probe Circuit Control Pins

SX-A devices contain internal probing circuitry that provides built-in access to every node in a design, enabling 100% real-time observation and analysis of a device's internal logic nodes without design iteration. The probe circuitry is accessed by Silicon Explorer II, an easy to use, integrated verification and logic analysis tool that can sample data at 100 MHz (asynchronous) or 66 MHz (synchronous). Silicon Explorer II attaches to a PC's standard COM port, turning the PC into a fully functional 18-channel logic analyzer. Silicon Explorer II allows designers to complete the design verification process at their desks and reduces verification time from several hours per cycle to a few seconds.

The Silicon Explorer II tool uses the boundary-scan ports (TDI, TCK, TMS, and TDO) to select the desired nets for verification. The selected internal nets are assigned to the

PRA/PRB pins for observation. Figure 1-13 illustrates the interconnection between Silicon Explorer II and the FPGA to perform in-circuit verification.

Design Considerations

In order to preserve device probing capabilities, users should avoid using the TDI, TCK, TDO, PRA, and PRB pins as input or bidirectional ports. Since these pins are active during probing, critical input signals through these pins are not available. In addition, the security fuse must not be programmed to preserve probing capabilities. Actel recommends that you use a $70\ \Omega$ series termination resistor on every probe connector (TDI, TCK, TMS, TDO, PRA, PRB). The $70\ \Omega$ series termination is used to prevent data transmission corruption during probing and reading back the checksum.

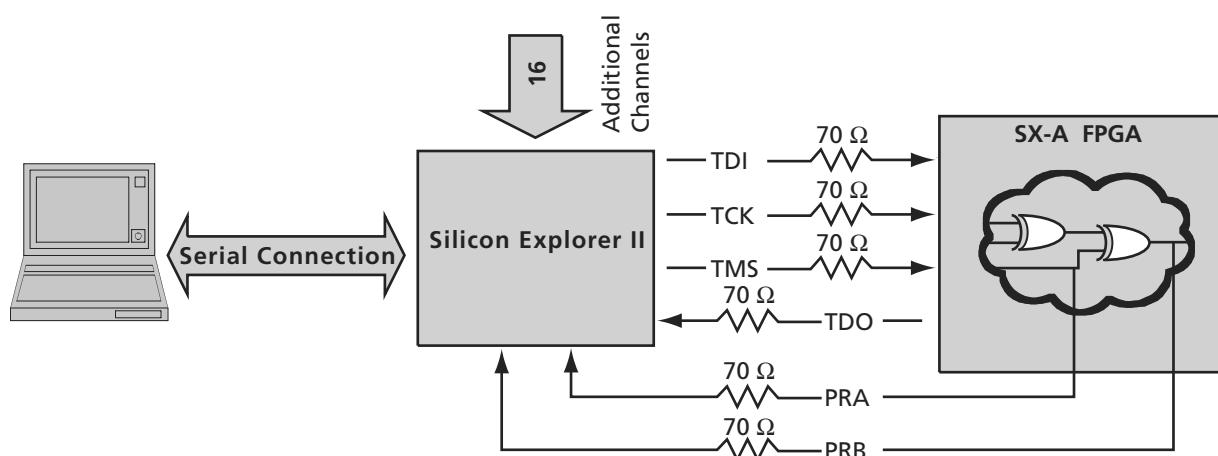


Figure 1-13 • Probe Setup

Electrical Specifications

Table 2-5 • 3.3 V LVTTL and 5 V TTL Electrical Specifications

Symbol	Parameter	Commercial		Industrial		Units	
		Min.	Max.	Min.	Max.		
V_{OH}	$V_{CCI} = \text{Minimum}$ $V_I = V_{IH} \text{ or } V_{IL}$	($I_{OH} = -1 \text{ mA}$)	0.9 V_{CCI}	0.9 V_{CCI}		V	
	$V_{CCI} = \text{Minimum}$ $V_I = V_{IH} \text{ or } V_{IL}$	($I_{OH} = -8 \text{ mA}$)	2.4	2.4		V	
V_{OL}	$V_{CCI} = \text{Minimum}$ $V_I = V_{IH} \text{ or } V_{IL}$	($I_{OL} = 1 \text{ mA}$)	0.4	0.4		V	
	$V_{CCI} = \text{Minimum}$ $V_I = V_{IH} \text{ or } V_{IL}$	($I_{OL} = 12 \text{ mA}$)	0.4	0.4		V	
V_{IL}	Input Low Voltage		0.8	0.8		V	
V_{IH}	Input High Voltage		2.0	5.75	2.0	5.75	V
I_{IL}/I_{IH}	Input Leakage Current, $V_{IN} = V_{CCI} \text{ or GND}$		-10	10	-10	10	μA
I_{OZ}	Tristate Output Leakage Current		-10	10	-10	10	μA
t_R, t_F	Input Transition Time t_R, t_F		10	10		ns	
C_{IO}	I/O Capacitance		10	10		pF	
I_{CC}	Standby Current		10	20		mA	
IV Curve*	Can be derived from the IBIS model on the web.						

Note: *The IBIS model can be found at <http://www.actel.com/download/ibis/default.aspx>.

Table 2-6 • 2.5 V LVCMS2 Electrical Specifications

Symbol	Parameter	Commercial		Industrial		Units	
		Min.	Max.	Min.	Max.		
V_{OH}	$V_{DD} = \text{MIN},$ $V_I = V_{IH} \text{ or } V_{IL}$	($I_{OH} = -100 \mu\text{A}$)	2.1	2.1		V	
	$V_{DD} = \text{MIN},$ $V_I = V_{IH} \text{ or } V_{IL}$	($I_{OH} = -1 \text{ mA}$)	2.0	2.0		V	
	$V_{DD} = \text{MIN},$ $V_I = V_{IH} \text{ or } V_{IL}$	($I_{OH} = -2 \text{ mA}$)	1.7	1.7		V	
V_{OL}	$V_{DD} = \text{MIN},$ $V_I = V_{IH} \text{ or } V_{IL}$	($I_{OL} = 100 \mu\text{A}$)	0.2	0.2		V	
	$V_{DD} = \text{MIN},$ $V_I = V_{IH} \text{ or } V_{IL}$	($I_{OL} = 1 \text{ mA}$)	0.4	0.4		V	
	$V_{DD} = \text{MIN},$ $V_I = V_{IH} \text{ or } V_{IL}$	($I_{OL} = 2 \text{ mA}$)	0.7	0.7		V	
V_{IL}	Input Low Voltage, $V_{OUT} \leq V_{VOL(\text{max})}$		-0.3	0.7	-0.3	0.7	V
V_{IH}	Input High Voltage, $V_{OUT} \geq V_{VOH(\text{min})}$		1.7	5.75	1.7	5.75	V
I_{IL}/I_{IH}	Input Leakage Current, $V_{IN} = V_{CCI} \text{ or GND}$		-10	10	-10	10	μA
I_{OZ}	Tristate Output Leakage Current, $V_{OUT} = V_{CCI} \text{ or GND}$		-10	10	-10	10	μA
t_R, t_F	Input Transition Time t_R, t_F		10	10		ns	
C_{IO}	I/O Capacitance		10	10		pF	
I_{CC}	Standby Current		10	20		mA	
IV Curve*	Can be derived from the IBIS model on the web.						

Note: *The IBIS model can be found at <http://www.actel.com/download/ibis/default.aspx>.

Table 2-8 • AC Specifications (5 V PCI Operation)

Symbol	Parameter	Condition	Min.	Max.	Units
$I_{OH(AC)}$	Switching Current High	$0 < V_{OUT} \leq 1.4$ ¹	-44	-	mA
		$1.4 \leq V_{OUT} < 2.4$ ^{1, 2}	(-44 + ($V_{OUT} - 1.4$)/0.024)	-	mA
		$3.1 < V_{OUT} < V_{CCI}$ ^{1, 3}	-	EQ 2-1 on page 2-5	-
	(Test Point)	$V_{OUT} = 3.1$ ³	-	-142	mA
$I_{OL(AC)}$	Switching Current Low	$V_{OUT} \geq 2.2$ ¹	95	-	mA
		$2.2 > V_{OUT} > 0.55$ ¹	($V_{OUT}/0.023$)	-	mA
		$0.71 > V_{OUT} > 0$ ^{1, 3}	-	EQ 2-2 on page 2-5	-
	(Test Point)	$V_{OUT} = 0.71$ ³	-	206	mA
I_{CL}	Low Clamp Current	$-5 < V_{IN} \leq -1$	-25 + ($V_{IN} + 1$)/0.015	-	mA
$slew_R$	Output Rise Slew Rate	0.4 V to 2.4 V load ⁴	1	5	V/ns
$slew_F$	Output Fall Slew Rate	2.4 V to 0.4 V load ⁴	1	5	V/ns

Notes:

1. Refer to the V/I curves in Figure 2-1 on page 2-5. Switching current characteristics for $REQ\#$ and $GNT\#$ are permitted to be one half of that specified here; i.e., half size output drivers may be used on these signals. This specification does not apply to CLK and $RST\#$, which are system outputs. "Switching Current High" specifications are not relevant to $SERR\#$, $INTA\#$, $INTB\#$, $INTC\#$, and $INTD\#$, which are open drain outputs.
2. Note that this segment of the minimum current curve is drawn from the AC drive point directly to the DC drive point rather than toward the voltage rail (as is done in the pull-down curve). This difference is intended to allow for an optional N-channel pull-up.
3. Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (A and B) are provided with the respective diagrams in Figure 2-1 on page 2-5. The equation defined maximum should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.
4. This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per revision 2.0 of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is now required (the maximum is no longer simply a guideline). Since adherence to the maximum slew rate was not required prior to revision 2.1 of the specification, there may be components in the market for some time that have faster edge rates; therefore, motherboard designers must bear in mind that rise and fall times faster than this specification could occur and should ensure that signal integrity modeling accounts for this. Rise slew rate does not apply to open drain outputs.

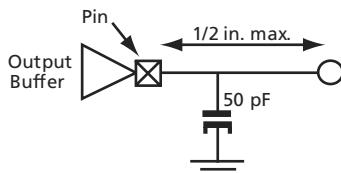


Figure 2-1 shows the 5 V PCI V/I curve and the minimum and maximum PCI drive characteristics of the SX-A family.

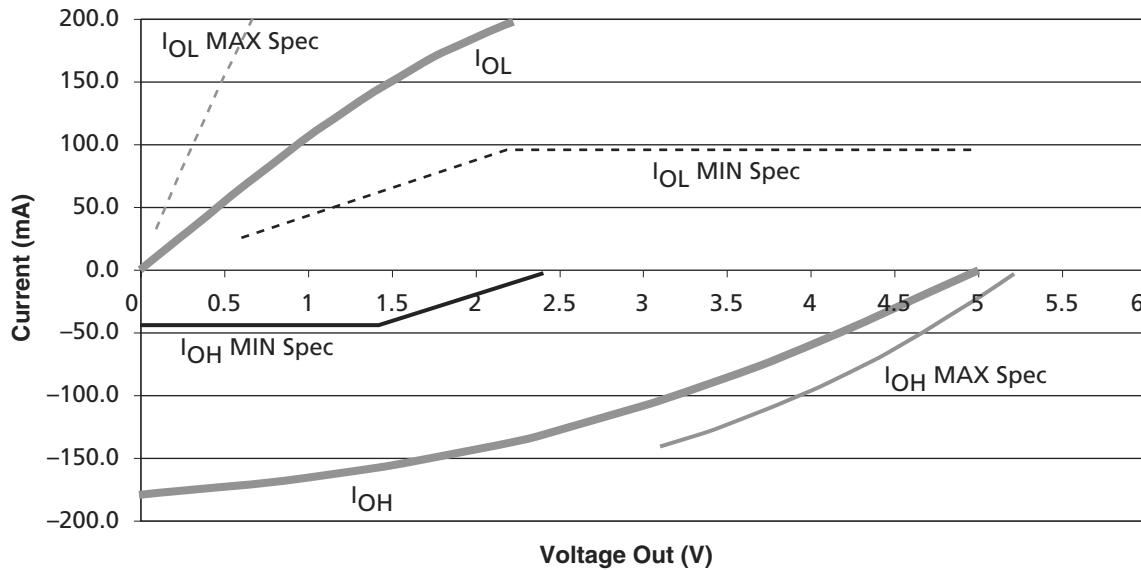


Figure 2-1 • 5 V PCI V/I Curve for SX-A Family

$$I_{OH} = 11.9 * (V_{OUT} - 5.25) * (V_{OUT} + 2.45)$$

for $V_{CCI} > V_{OUT} > 3.1V$

EQ 2-1

$$I_{OL} = 78.5 * V_{OUT} * (4.4 - V_{OUT})$$

for $0V < V_{OUT} < 0.71V$

EQ 2-2

Table 2-9 • DC Specifications (3.3 V PCI Operation)

Symbol	Parameter	Condition	Min.	Max.	Units
V_{CCA}	Supply Voltage for Array		2.25	2.75	V
V_{CCI}	Supply Voltage for I/Os		3.0	3.6	V
V_{IH}	Input High Voltage		$0.5V_{CCI}$	$V_{CCI} + 0.5$	V
V_{IL}	Input Low Voltage		-0.5	$0.3V_{CCI}$	V
I_{IPU}	Input Pull-up Voltage ¹		$0.7V_{CCI}$	-	V
I_{IL}	Input Leakage Current ²	$0 < V_{IN} < V_{CCI}$	-10	+10	μA
V_{OH}	Output High Voltage	$I_{OUT} = -500 \mu A$	$0.9V_{CCI}$	-	V
V_{OL}	Output Low Voltage	$I_{OUT} = 1,500 \mu A$		$0.1V_{CCI}$	V
C_{IN}	Input Pin Capacitance ³		-	10	pF
C_{CLK}	CLK Pin Capacitance		5	12	pF

Notes:

1. This specification should be guaranteed by design. It is the minimum voltage to which pull-up resistors are calculated to pull a floated network. Designers should ensure that the input buffer is conducting minimum current at this input voltage in applications sensitive to static power utilization.
2. Input leakage currents include hi-Z output leakage for all bidirectional buffers with tristate outputs.
3. Absolute maximum pin capacitance for a PCI input is 10 pF (except for CLK).

Where:

C_{EQCM} = Equivalent capacitance of combinatorial modules (C-cells) in pF

C_{EQSM} = Equivalent capacitance of sequential modules (R-Cells) in pF

C_{EQI} = Equivalent capacitance of input buffers in pF

C_{EQO} = Equivalent capacitance of output buffers in pF

C_{EQCR} = Equivalent capacitance of CLKA/B in pF

C_{EQHV} = Variable capacitance of HCLK in pF

C_{EQHF} = Fixed capacitance of HCLK in pF

C_L = Output lead capacitance in pF

f_m = Average logic module switching rate in MHz

f_n = Average input buffer switching rate in MHz

f_p = Average output buffer switching rate in MHz

f_{q1} = Average CLKA rate in MHz

f_{q2} = Average CLKB rate in MHz

f_{s1} = Average HCLK rate in MHz

m = Number of logic modules switching at f_m

n = Number of input buffers switching at f_n

p = Number of output buffers switching at f_p

q_1 = Number of clock loads on CLKA

q_2 = Number of clock loads on CLKB

r_1 = Fixed capacitance due to CLKA

r_2 = Fixed capacitance due to CLKB

s_1 = Number of clock loads on HCLK

x = Number of I/Os at logic low

y = Number of I/Os at logic high

Table 2-11 • CEQ Values for SX-A Devices

	A54SX08A	A54SX16A	A54SX32A	A54SX72A
Combinatorial modules (C_{EQCM})	1.70 pF	2.00 pF	2.00 pF	1.80 pF
Sequential modules (C_{EQCM})	1.50 pF	1.50 pF	1.30 pF	1.50 pF
Input buffers (C_{EQI})	1.30 pF	1.30 pF	1.30 pF	1.30 pF
Output buffers (C_{EQO})	7.40 pF	7.40 pF	7.40 pF	7.40 pF
Routed array clocks (C_{EQCR})	1.05 pF	1.05 pF	1.05 pF	1.05 pF
Dedicated array clocks – variable (C_{EQHV})	0.85 pF	0.85 pF	0.85 pF	0.85 pF
Dedicated array clocks – fixed (C_{EQHF})	30.00 pF	55.00 pF	110.00 pF	240.00 pF
Routed array clock A (r_1)	35.00 pF	50.00 pF	90.00 pF	310.00 pF

Thermal Characteristics

Introduction

The temperature variable in Actel Designer software refers to the junction temperature, not the ambient, case, or board temperatures. This is an important distinction because dynamic and static power consumption will cause the chip's junction to be higher than the ambient, case, or board temperatures. EQ 2-9 and EQ 2-10 give the relationship between thermal resistance, temperature gradient and power.

$$\theta_{JA} = \frac{T_J - T_A}{P}$$

EQ 2-9

$$\theta_{JC} = \frac{T_C - T_A}{P}$$

EQ 2-10

Where:

θ_{JA} = Junction-to-air thermal resistance

θ_{JC} = Junction-to-case thermal resistance

T_J = Junction temperature

T_A = Ambient temperature

T_C = Case temperature

P = total power dissipated by the device

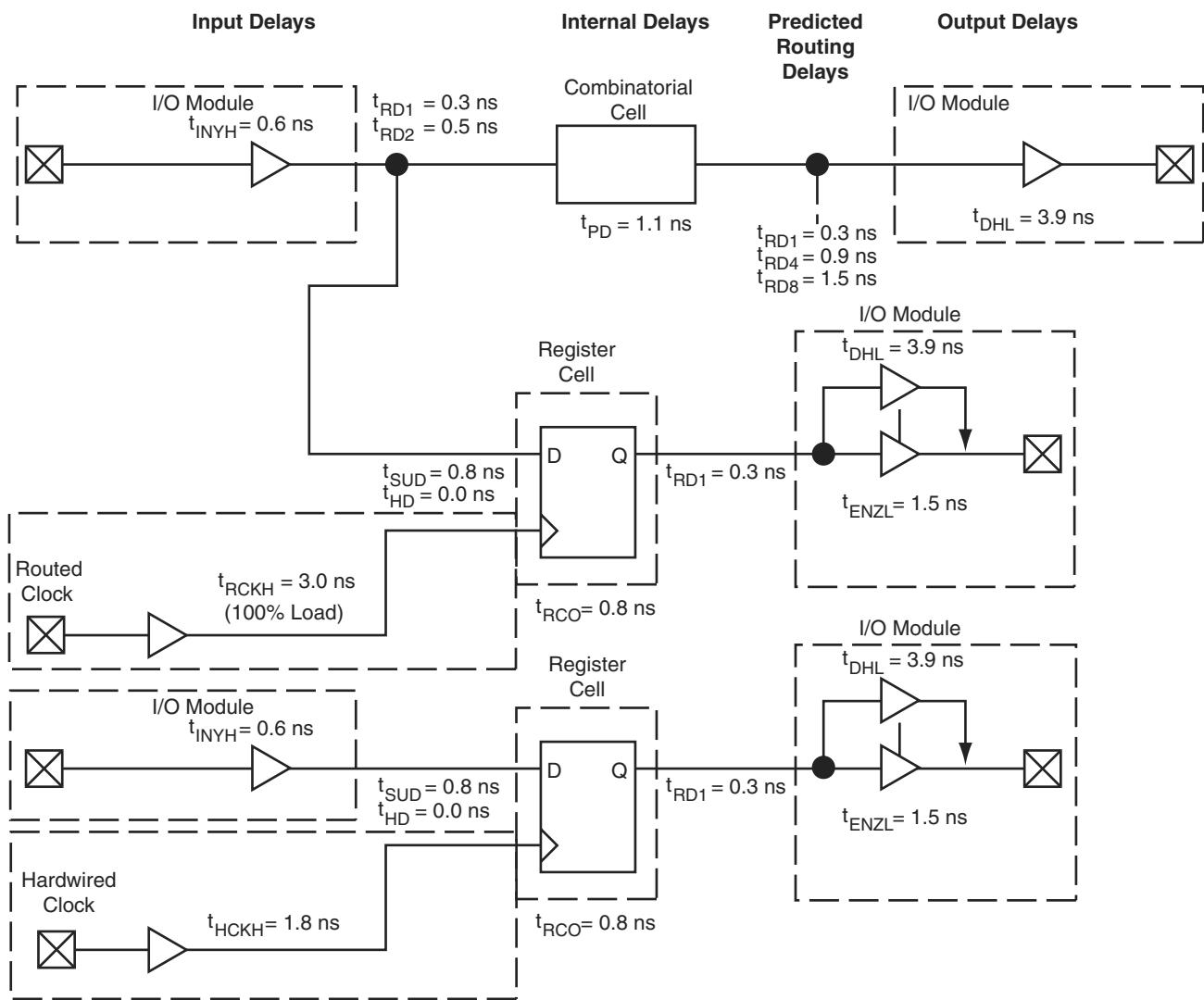
Table 2-12 • Package Thermal Characteristics

Package Type	Pin Count	θ_{JC}	θ_{JA}			Units
			Still Air	1.0 m/s 200 ft./min.	2.5 m/s 500 ft./min.	
Thin Quad Flat Pack (TQFP)	100	14	33.5	27.4	25	°C/W
Thin Quad Flat Pack (TQFP)	144	11	33.5	28	25.7	°C/W
Thin Quad Flat Pack (TQFP)	176	11	24.7	19.9	18	°C/W
Plastic Quad Flat Pack (PQFP) ¹	208	8	26.1	22.5	20.8	°C/W
Plastic Quad Flat Pack (PQFP) with Heat Spreader ²	208	3.8	16.2	13.3	11.9	°C/W
Plastic Ball Grid Array (PBGA)	329	3	17.1	13.8	12.8	°C/W
Fine Pitch Ball Grid Array (FBGA)	144	3.8	26.9	22.9	21.5	°C/W
Fine Pitch Ball Grid Array (FBGA)	256	3.8	26.6	22.8	21.5	°C/W
Fine Pitch Ball Grid Array (FBGA)	484	3.2	18	14.7	13.6	°C/W

Notes:

1. The A54SX08A PQ208 has no heat spreader.
2. The SX-A PQ208 package has a heat spreader for A54SX16A, A54SX32A, and A54SX72A.

SX-A Timing Model



Note: *Values shown for A54SX72A, -2, worst-case commercial conditions at 5 V PCI with standard place-and-route.

Figure 2-3 • SX-A Timing Model

Sample Path Calculations

Hardwired Clock

$$\begin{aligned}
 \text{External Setup} &= (t_{INYH} + t_{RD1} + t_{SUD}) - t_{HCKH} \\
 &= 0.6 + 0.3 + 0.8 - 1.8 = -0.1 \text{ ns} \\
 \text{Clock-to-Out (Pad-to-Pad)} &= t_{HCKH} + t_{RCO} + t_{RD1} + t_{DHL} \\
 &= 1.8 + 0.8 + 0.3 + 3.9 = 6.8 \text{ ns}
 \end{aligned}$$

Routed Clock

$$\begin{aligned}
 \text{External Setup} &= (t_{INYH} + t_{RD1} + t_{SUD}) - t_{RCKH} \\
 &= 0.6 + 0.3 + 0.8 - 3.0 = -1.3 \text{ ns} \\
 \text{Clock-to-Out (Pad-to-Pad)} &= t_{RCKH} + t_{RCO} + t_{RD1} + t_{DHL} \\
 &= 3.0 + 0.8 + 0.3 + 3.9 = 8.0 \text{ ns}
 \end{aligned}$$

Output Buffer Delays

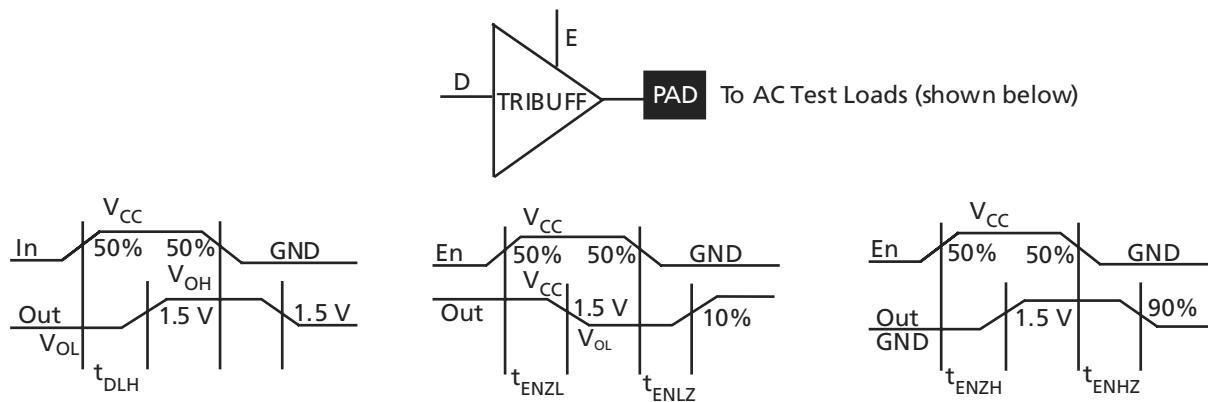


Figure 2-4 • Output Buffer Delays

AC Test Loads

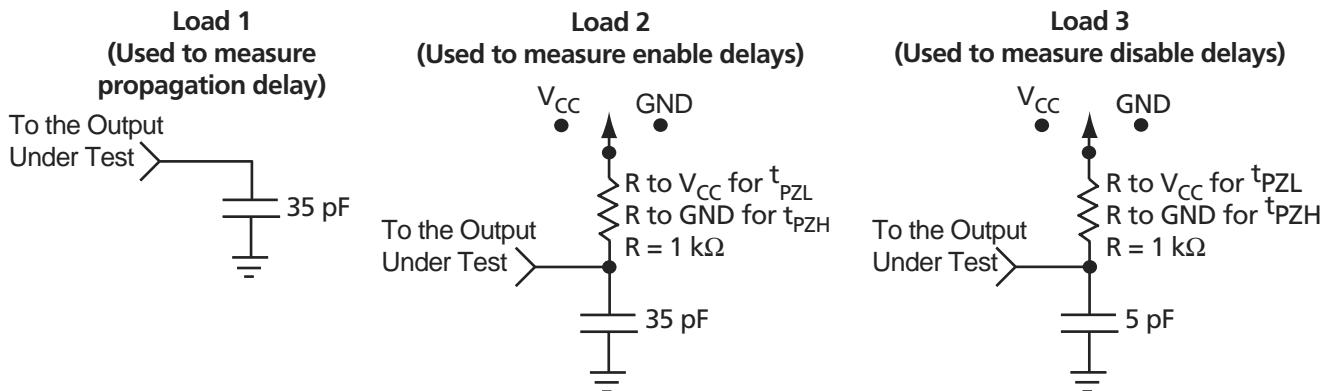


Figure 2-5 • AC Test Loads

Table 2-14 • A54SX08A Timing Characteristics (Continued)
 (Worst-Case Commercial Conditions, $V_{CCA} = 2.25\text{ V}$, $V_{CCI} = 3.0\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	-2 Speed	-1 Speed	Std. Speed	-F Speed	Units
		Min.	Max.	Min.	Max.	
t_{INYH}	Input Data Pad to Y High 5 V PCI	0.5	0.6	0.7	0.9	ns
t_{INYL}	Input Data Pad to Y Low 5 V PCI	0.8	0.9	1.1	1.5	ns
t_{INYH}	Input Data Pad to Y High 5 V TTL	0.5	0.6	0.7	0.9	ns
t_{INYL}	Input Data Pad to Y Low 5 V TTL	0.8	0.9	1.1	1.5	ns
Input Module Predicted Routing Delays²						
t_{IRD1}	FO = 1 Routing Delay	0.3	0.3	0.4	0.6	ns
t_{IRD2}	FO = 2 Routing Delay	0.5	0.5	0.6	0.8	ns
t_{IRD3}	FO = 3 Routing Delay	0.6	0.7	0.8	1.1	ns
t_{IRD4}	FO = 4 Routing Delay	0.8	0.9	1	1.4	ns
t_{IRD8}	FO = 8 Routing Delay	1.4	1.5	1.8	2.5	ns
t_{IRD12}	FO = 12 Routing Delay	2	2.2	2.6	3.6	ns

Notes:

1. For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.
2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual performance.

Table 2-19 • A54SX08A Timing Characteristics
 (Worst-Case Commercial Conditions $V_{CCA} = 2.25\text{ V}$, $V_{CCI} = 3.0\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	-2 Speed		-1 Speed		Std. Speed	-F Speed	Units
		Min.	Max.	Min.	Max.	Min.	Max.	
3.3 V PCI Output Module Timing¹								
t_{DLH}	Data-to-Pad Low to High	2.2	2.4	2.9	4.0	ns		
t_{DHL}	Data-to-Pad High to Low	2.3	2.6	3.1	4.3	ns		
t_{ENZL}	Enable-to-Pad, Z to L	1.7	1.9	2.2	3.1	ns		
t_{ENZH}	Enable-to-Pad, Z to H	2.2	2.4	2.9	4.0	ns		
t_{ENLZ}	Enable-to-Pad, L to Z	2.8	3.2	3.8	5.3	ns		
t_{ENHZ}	Enable-to-Pad, H to Z	2.3	2.6	3.1	4.3	ns		
d_{TLH}^2	Delta Low to High	0.03	0.03	0.04	0.045	ns/pF		
d_{THL}^2	Delta High to Low	0.015	0.015	0.015	0.025	ns/pF		
3.3 V LVTTL Output Module Timing³								
t_{DLH}	Data-to-Pad Low to High	3.0	3.4	4.0	5.6	ns		
t_{DHL}	Data-to-Pad High to Low	3.0	3.3	3.9	5.5	ns		
t_{DHLS}	Data-to-Pad High to Low—low slew	10.4	11.8	13.8	19.3	ns		
t_{ENZL}	Enable-to-Pad, Z to L	2.6	2.9	3.4	4.8	ns		
t_{ENZLS}	Enable-to-Pad, Z to L—low slew	18.9	21.3	25.4	34.9	ns		
t_{ENZH}	Enable-to-Pad, Z to H	3	3.4	4	5.6	ns		
t_{ENLZ}	Enable-to-Pad, L to Z	3.3	3.7	4.4	6.2	ns		
t_{ENHZ}	Enable-to-Pad, H to Z	3	3.3	3.9	5.5	ns		
d_{TLH}^2	Delta Low to High	0.03	0.03	0.04	0.045	ns/pF		
d_{THL}^2	Delta High to Low	0.015	0.015	0.015	0.025	ns/pF		
d_{THLS}^2	Delta High to Low—low slew	0.053	0.067	0.073	0.107	ns/pF		

Notes:

1. Delays based on 10 pF loading and 25 Ω resistance.
2. To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation:

$$\text{Slew Rate [V/ns]} = (0.1 * V_{CCI} - 0.9 * V_{CCI}) / (C_{load} * d_{T[|LH|HL|HLS]})$$
 where C_{load} is the load capacitance driven by the I/O in pF
 $d_{T[|LH|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.
3. Delays based on 35 pF loading.

Table 2-28 • A54SX32A Timing Characteristics
 (Worst-Case Commercial Conditions, $V_{CCA} = 2.25\text{ V}$, $V_{CCI} = 3.0\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	-3 Speed¹	-2 Speed	-1 Speed	Std. Speed	-F Speed	Units
		Min.	Max.	Min.	Max.	Min.	
C-Cell Propagation Delays²							
t_{PD}	Internal Array Module	0.8	0.9	1.1	1.2	1.7	ns
Predicted Routing Delays³							
t_{DC}	FO = 1 Routing Delay, Direct Connect	0.1	0.1	0.1	0.1	0.1	ns
t_{FC}	FO = 1 Routing Delay, Fast Connect	0.3	0.3	0.3	0.4	0.6	ns
t_{RD1}	FO = 1 Routing Delay	0.3	0.3	0.4	0.5	0.6	ns
t_{RD2}	FO = 2 Routing Delay	0.4	0.5	0.5	0.6	0.8	ns
t_{RD3}	FO = 3 Routing Delay	0.5	0.6	0.7	0.8	1.1	ns
t_{RD4}	FO = 4 Routing Delay	0.7	0.8	0.9	1.0	1.4	ns
t_{RD8}	FO = 8 Routing Delay	1.2	1.4	1.5	1.8	2.5	ns
t_{RD12}	FO = 12 Routing Delay	1.7	2.0	2.2	2.6	3.6	ns
R-Cell Timing							
t_{RCO}	Sequential Clock-to-Q	0.6	0.7	0.8	0.9	1.3	ns
t_{CLR}	Asynchronous Clear-to-Q	0.5	0.6	0.6	0.8	1.0	ns
t_{PRESET}	Asynchronous Preset-to-Q	0.6	0.7	0.7	0.9	1.2	ns
t_{SUD}	Flip-Flop Data Input Set-Up	0.6	0.7	0.8	0.9	1.2	ns
t_{HD}	Flip-Flop Data Input Hold	0.0	0.0	0.0	0.0	0.0	ns
t_{WASYN}	Asynchronous Pulse Width	1.2	1.4	1.5	1.8	2.5	ns
$t_{RECASYN}$	Asynchronous Recovery Time	0.3	0.4	0.4	0.5	0.7	ns
t_{HASYN}	Asynchronous Removal Time	0.3	0.3	0.3	0.4	0.6	ns
t_{MPW}	Clock Pulse Width	1.4	1.6	1.8	2.1	2.9	ns
Input Module Propagation Delays							
t_{INYH}	Input Data Pad to Y High 2.5 V LVC MOS	0.6	0.7	0.8	0.9	1.2	ns
t_{INYL}	Input Data Pad to Y Low 2.5 V LVC MOS	1.2	1.3	1.5	1.8	2.5	ns
t_{INYH}	Input Data Pad to Y High 3.3 V PCI	0.5	0.6	0.6	0.7	1.0	ns
t_{INYL}	Input Data Pad to Y Low 3.3 V PCI	0.6	0.7	0.8	0.9	1.3	ns
t_{INYH}	Input Data Pad to Y High 3.3 V LV TTL	0.8	0.9	1.0	1.2	1.6	ns
t_{INYL}	Input Data Pad to Y Low 3.3 V LV TTL	1.4	1.6	1.8	2.2	3.0	ns

Notes:

1. All -3 speed grades have been discontinued.
2. For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.
3. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual performance.

Table 2-37 • A54SX72A Timing Characteristics
 (Worst-Case Commercial Conditions $V_{CCA} = 2.25\text{ V}$, $V_{CCI} = 3.0\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	-3 Speed*		-2 Speed		-1 Speed		Std. Speed	-F Speed	Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Dedicated (Hardwired) Array Clock Networks										
t_{HCKH}	Input Low to High (Pad to R-cell Input)	1.6		1.9		2.1		2.5		3.8 ns
t_{HCKL}	Input High to Low (Pad to R-cell Input)		1.7		1.9		2.1		2.5	3.8 ns
t_{HPWH}	Minimum Pulse Width High	1.5		1.7		2.0		2.3		3.2 ns
t_{HPWL}	Minimum Pulse Width Low	1.5		1.7		2.0		2.3		3.2 ns
t_{HCKSW}	Maximum Skew		1.4		1.6		1.8		2.1	3.3 ns
t_{HP}	Minimum Period	3.0		3.4		4.0		4.6		6.4 ns
f_{HMAX}	Maximum Frequency		333		294		250		217	156 MHz
Routed Array Clock Networks										
t_{RCKH}	Input Low to High (Light Load) (Pad to R-cell Input)	2.2		2.6		2.9		3.4		4.8 ns
t_{RCKL}	Input High to Low (Light Load) (Pad to R-cell Input)		2.8		3.3		3.7		4.3	6.0 ns
t_{RCKH}	Input Low to High (50% Load) (Pad to R-cell Input)	2.4		2.8		3.2		3.7		5.2 ns
t_{RCKL}	Input High to Low (50% Load) (Pad to R-cell Input)		2.9		3.4		3.8		4.5	6.2 ns
t_{RCKH}	Input Low to High (100% Load) (Pad to R-cell Input)	2.6		3.0		3.4		4.0		5.6 ns
t_{RCKL}	Input High to Low (100% Load) (Pad to R-cell Input)		3.1		3.6		4.1		4.8	6.7 ns
t_{RPWH}	Minimum Pulse Width High	1.5		1.7		2.0		2.3		3.2 ns
t_{RPWL}	Minimum Pulse Width Low	1.5		1.7		2.0		2.3		3.2 ns
t_{RCKSW}	Maximum Skew (Light Load)		1.9		2.2		2.5		3	4.1 ns
t_{RCKSW}	Maximum Skew (50% Load)	1.9		2.1		2.4		2.8		3.9 ns
t_{RCKSW}	Maximum Skew (100% Load)	1.9		2.1		2.4		2.8		3.9 ns
Quadrant Array Clock Networks										
t_{QCKH}	Input Low to High (Light Load) (Pad to R-cell Input)	1.3		1.5		1.7		1.9		2.7 ns
t_{QCHKL}	Input High to Low (Light Load) (Pad to R-cell Input)		1.3		1.5		1.7		2	2.8 ns
t_{QCKH}	Input Low to High (50% Load) (Pad to R-cell Input)	1.5		1.7		1.9		2.2		3.1 ns
t_{QCHKL}	Input High to Low (50% Load) (Pad to R-cell Input)	1.5		1.8		2		2.3		3.2 ns

Note: *All -3 speed grades have been discontinued.

100-TQFP			
Pin Number	A54SX08A Function	A54SX16A Function	A54SX32A Function
1	GND	GND	GND
2	TDI, I/O	TDI, I/O	TDI, I/O
3	I/O	I/O	I/O
4	I/O	I/O	I/O
5	I/O	I/O	I/O
6	I/O	I/O	I/O
7	TMS	TMS	TMS
8	V _{CCI}	V _{CCI}	V _{CCI}
9	GND	GND	GND
10	I/O	I/O	I/O
11	I/O	I/O	I/O
12	I/O	I/O	I/O
13	I/O	I/O	I/O
14	I/O	I/O	I/O
15	I/O	I/O	I/O
16	TRST, I/O	TRST, I/O	TRST, I/O
17	I/O	I/O	I/O
18	I/O	I/O	I/O
19	I/O	I/O	I/O
20	V _{CCI}	V _{CCI}	V _{CCI}
21	I/O	I/O	I/O
22	I/O	I/O	I/O
23	I/O	I/O	I/O
24	I/O	I/O	I/O
25	I/O	I/O	I/O
26	I/O	I/O	I/O
27	I/O	I/O	I/O
28	I/O	I/O	I/O
29	I/O	I/O	I/O
30	I/O	I/O	I/O
31	I/O	I/O	I/O
32	I/O	I/O	I/O
33	I/O	I/O	I/O
34	PRB, I/O	PRB, I/O	PRB, I/O
35	V _{CCA}	V _{CCA}	V _{CCA}

100-TQFP			
Pin Number	A54SX08A Function	A54SX16A Function	A54SX32A Function
36	GND	GND	GND
37	NC	NC	NC
38	I/O	I/O	I/O
39	HCLK	HCLK	HCLK
40	I/O	I/O	I/O
41	I/O	I/O	I/O
42	I/O	I/O	I/O
43	I/O	I/O	I/O
44	V _{CCI}	V _{CCI}	V _{CCI}
45	I/O	I/O	I/O
46	I/O	I/O	I/O
47	I/O	I/O	I/O
48	I/O	I/O	I/O
49	TDO, I/O	TDO, I/O	TDO, I/O
50	I/O	I/O	I/O
51	GND	GND	GND
52	I/O	I/O	I/O
53	I/O	I/O	I/O
54	I/O	I/O	I/O
55	I/O	I/O	I/O
56	I/O	I/O	I/O
57	V _{CCA}	V _{CCA}	V _{CCA}
58	V _{CCI}	V _{CCI}	V _{CCI}
59	I/O	I/O	I/O
60	I/O	I/O	I/O
61	I/O	I/O	I/O
62	I/O	I/O	I/O
63	I/O	I/O	I/O
64	I/O	I/O	I/O
65	I/O	I/O	I/O
66	I/O	I/O	I/O
67	V _{CCA}	V _{CCA}	V _{CCA}
68	GND	GND	GND
69	GND	GND	GND
70	I/O	I/O	I/O

100-TQFP			
Pin Number	A54SX08A Function	A54SX16A Function	A54SX32A Function
71	I/O	I/O	I/O
72	I/O	I/O	I/O
73	I/O	I/O	I/O
74	I/O	I/O	I/O
75	I/O	I/O	I/O
76	I/O	I/O	I/O
77	I/O	I/O	I/O
78	I/O	I/O	I/O
79	I/O	I/O	I/O
80	I/O	I/O	I/O
81	I/O	I/O	I/O
82	V _{CCI}	V _{CCI}	V _{CCI}
83	I/O	I/O	I/O
84	I/O	I/O	I/O
85	I/O	I/O	I/O
86	I/O	I/O	I/O
87	CLKA	CLKA	CLKA
88	CLKB	CLKB	CLKB
89	NC	NC	NC
90	V _{CCA}	V _{CCA}	V _{CCA}
91	GND	GND	GND
92	PRA, I/O	PRA, I/O	PRA, I/O
93	I/O	I/O	I/O
94	I/O	I/O	I/O
95	I/O	I/O	I/O
96	I/O	I/O	I/O
97	I/O	I/O	I/O
98	I/O	I/O	I/O
99	I/O	I/O	I/O
100	TCK, I/O	TCK, I/O	TCK, I/O

144-Pin TQFP			
Pin Number	A54SX08A Function	A54SX16A Function	A54SX32A Function
75	I/O	I/O	I/O
76	I/O	I/O	I/O
77	I/O	I/O	I/O
78	I/O	I/O	I/O
79	V _{CCA}	V _{CCA}	V _{CCA}
80	V _{CCI}	V _{CCI}	V _{CCI}
81	GND	GND	GND
82	I/O	I/O	I/O
83	I/O	I/O	I/O
84	I/O	I/O	I/O
85	I/O	I/O	I/O
86	I/O	I/O	I/O
87	I/O	I/O	I/O
88	I/O	I/O	I/O
89	V _{CCA}	V _{CCA}	V _{CCA}
90	NC	NC	NC
91	I/O	I/O	I/O
92	I/O	I/O	I/O
93	I/O	I/O	I/O
94	I/O	I/O	I/O
95	I/O	I/O	I/O
96	I/O	I/O	I/O
97	I/O	I/O	I/O
98	V _{CCA}	V _{CCA}	V _{CCA}
99	GND	GND	GND
100	I/O	I/O	I/O
101	GND	GND	GND
102	V _{CCI}	V _{CCI}	V _{CCI}
103	I/O	I/O	I/O
104	I/O	I/O	I/O
105	I/O	I/O	I/O
106	I/O	I/O	I/O
107	I/O	I/O	I/O
108	I/O	I/O	I/O
109	GND	GND	GND
110	I/O	I/O	I/O

144-Pin TQFP			
Pin Number	A54SX08A Function	A54SX16A Function	A54SX32A Function
111	I/O	I/O	I/O
112	I/O	I/O	I/O
113	I/O	I/O	I/O
114	I/O	I/O	I/O
115	V _{CCI}	V _{CCI}	V _{CCI}
116	I/O	I/O	I/O
117	I/O	I/O	I/O
118	I/O	I/O	I/O
119	I/O	I/O	I/O
120	I/O	I/O	I/O
121	I/O	I/O	I/O
122	I/O	I/O	I/O
123	I/O	I/O	I/O
124	I/O	I/O	I/O
125	CLKA	CLKA	CLKA
126	CLKB	CLKB	CLKB
127	NC	NC	NC
128	GND	GND	GND
129	V _{CCA}	V _{CCA}	V _{CCA}
130	I/O	I/O	I/O
131	PRA, I/O	PRA, I/O	PRA, I/O
132	I/O	I/O	I/O
133	I/O	I/O	I/O
134	I/O	I/O	I/O
135	I/O	I/O	I/O
136	I/O	I/O	I/O
137	I/O	I/O	I/O
138	I/O	I/O	I/O
139	I/O	I/O	I/O
140	V _{CCI}	V _{CCI}	V _{CCI}
141	I/O	I/O	I/O
142	I/O	I/O	I/O
143	I/O	I/O	I/O
144	TCK, I/O	TCK, I/O	TCK, I/O

176-Pin TQFP

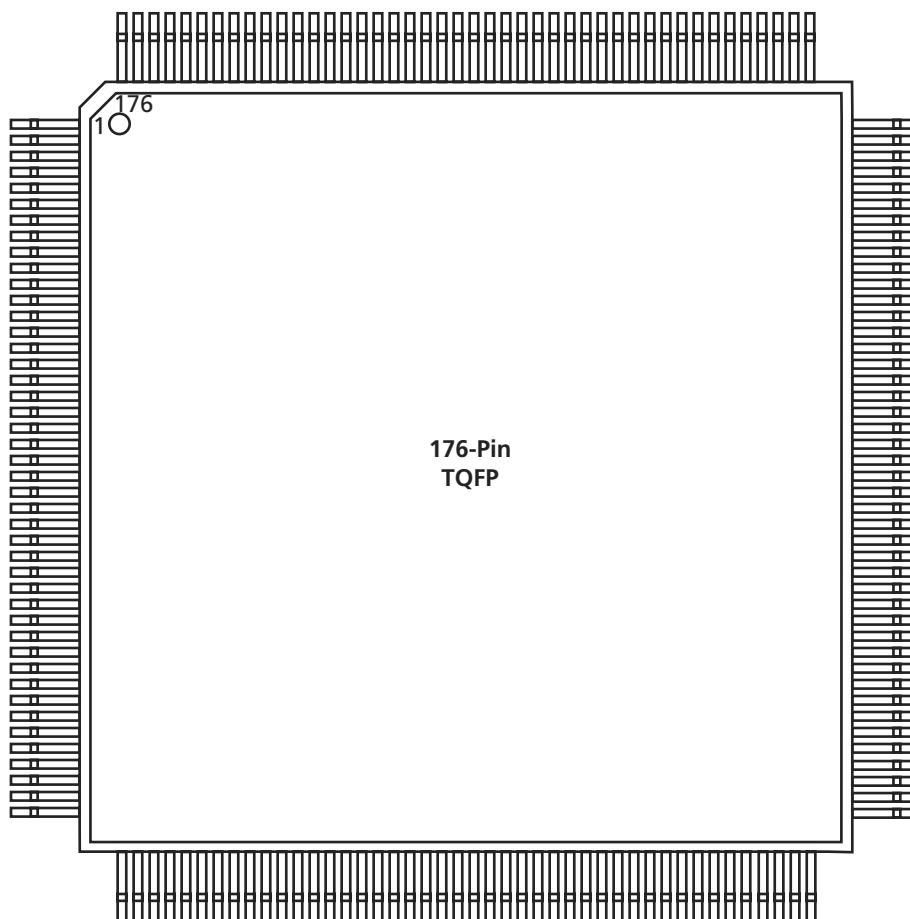


Figure 3-4 • 176-Pin TQFP (Top View)

Note

For Package Manufacturing and Environmental information, visit Resource center at
<http://www.actel.com/products/rescenter/package/index.html>.

484-Pin FBGA		
Pin Number	A54SX32A Function	A54SX72A Function
T3	I/O	I/O
T4	I/O	I/O
T5	I/O	I/O
T10	GND	GND
T11	GND	GND
T12	GND	GND
T13	GND	GND
T14	GND	GND
T15	GND	GND
T16	GND	GND
T17	GND	GND
T22	I/O	I/O
T23	I/O	I/O
T24	I/O	I/O
T25	NC*	I/O
T26	NC*	I/O
U1	I/O	I/O
U2	V _{CCI}	V _{CCI}
U3	I/O	I/O
U4	I/O	I/O
U5	I/O	I/O
U10	GND	GND
U11	GND	GND
U12	GND	GND
U13	GND	GND
U14	GND	GND
U15	GND	GND
U16	GND	GND
U17	GND	GND
U22	I/O	I/O
U23	I/O	I/O
U24	I/O	I/O
U25	V _{CCI}	V _{CCI}
U26	I/O	I/O
V1	NC*	I/O

484-Pin FBGA		
Pin Number	A54SX32A Function	A54SX72A Function
V2	NC*	I/O
V3	I/O	I/O
V4	I/O	I/O
V5	I/O	I/O
V22	V _{CCA}	V _{CCA}
V23	I/O	I/O
V24	I/O	I/O
V25	NC*	I/O
V26	NC*	I/O
W1	I/O	I/O
W2	I/O	I/O
W3	I/O	I/O
W4	I/O	I/O
W5	I/O	I/O
W22	I/O	I/O
W23	V _{CCA}	V _{CCA}
W24	I/O	I/O
W25	NC*	I/O
W26	NC*	I/O
Y1	NC*	I/O
Y2	NC*	I/O
Y3	I/O	I/O
Y4	I/O	I/O
Y5	NC*	I/O
Y22	I/O	I/O
Y23	I/O	I/O
Y24	V _{CCI}	V _{CCI}
Y25	I/O	I/O
Y26	I/O	I/O

Note: *These pins must be left floating on the A54SX32A device.