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Understanding <u>Embedded - FPGAs (Field</u> <u>Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	2880
Number of Logic Elements/Cells	·
Total RAM Bits	-
Number of I/O	147
Number of Gates	48000
Voltage - Supply	2.25V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	176-LQFP
Supplier Device Package	176-TQFP (24x24)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a54sx32a-1tq176

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

Ordering Information



Notes:

1. For more information about the CQFP package options, refer to the HiRel SX-A datasheet.

2. All –3 speed grades have been discontinued.

Device Resources

		User I/Os (Including Clock Buffers)										
Device	208-Pin PQFP	100-Pin TQFP	144-Pin TQFP	176-Pin TQFP	329-Pin PBGA	144-Pin FBGA	256-Pin FBGA	484-Pin FBGA				
A54SX08A	130	81	113	-	-	111	-	-				
A54SX16A	175	81	113	-	-	111	180	-				
A54SX32A	174	81	113	147	249	111	203	249				
A54SX72A	171	-	-	-	-	_	203	360				

Notes: Package Definitions: PQFP = Plastic Quad Flat Pack, TQFP = Thin Quad Flat Pack, PBGA = Plastic Ball Grid Array, FBGA = Fine Pitch Ball Grid Array

Routing Resources

The routing and interconnect resources of SX-A devices are in the top two metal layers above the logic modules (Figure 1-1 on page 1-1), providing optimal use of silicon, thus enabling the entire floor of the device to be spanned with an uninterrupted grid of logic modules. Interconnection between these logic modules is achieved using the Actel patented metal-to-metal programmable antifuse interconnect elements. The antifuses are normally open circuits and, when programmed, form a permanent low-impedance connection.

Clusters and SuperClusters can be connected through the use of two innovative local routing resources called FastConnect and DirectConnect, which enable extremely fast and predictable interconnection of modules within Clusters and SuperClusters (Figure 1-5 on page 1-4 and Figure 1-6 on page 1-4). This routing architecture also dramatically reduces the number of antifuses required to complete a circuit, ensuring the highest possible performance, which is often required in applications such as fast counters, state machines, and data path logic. The interconnect elements (i.e., the antifuses and metal tracks) have lower capacitance and lower resistance than any other device of similar capacity, leading to the fastest signal propagation in the industry.

DirectConnect is a horizontal routing resource that provides connections from a C-cell to its neighboring R-Cell in a given SuperCluster. DirectConnect uses a hardwired signal path requiring no programmable interconnection to achieve its fast signal propagation time of less than 0.1 ns.

FastConnect enables horizontal routing between any two logic modules within a given SuperCluster, and vertical routing with the SuperCluster immediately below it. Only one programmable connection is used in a FastConnect path, delivering a maximum pin-to-pin propagation time of 0.3 ns.

In addition to DirectConnect and FastConnect, the architecture makes use of two globally oriented routing resources known as segmented routing and high-drive routing. The Actel segmented routing structure provides a variety of track lengths for extremely fast routing between SuperClusters. The exact combination of track lengths and antifuses within each path is chosen by the 100% automatic place-and-route software to minimize signal propagation delays.

The general system of routing tracks allows any logic module in the array to be connected to any other logic or I/O module. Within this system, most connections typically require three or fewer antifuses, resulting in fast and predictable performance.

The unique local and general routing structure featured in SX-A devices allows 100% pin-locking with full logic utilization, enables concurrent printed circuit board (PCB) development, reduces design time, and allows designers to achieve performance goals with minimum effort.



Figure 1-4 • Cluster Organization



Clock Resources

Actel's high-drive routing structure provides three clock networks (Table 1-1). The first clock, called HCLK, is hardwired from the HCLK buffer to the clock select multiplexor (MUX) in each R-cell. HCLK cannot be connected to combinatorial logic. This provides a fast propagation path for the clock signal. If not used, this pin must be set as Low or High on the board. It must not be left floating. Figure 1-7 describes the clock circuit used for the constant load HCLK and the macros supported.

HCLK does not function until the fourth clock cycle each time the device is powered up to prevent false output levels due to any possible slow power-on-reset signal and fast start-up clock circuit. To activate HCLK from the first cycle, the TRST pin must be reserved in the Design software and the pin must be tied to GND on the board.

Two additional clocks (CLKA, CLKB) are global clocks that can be sourced from external pins or from internal logic signals within the SX-A device. CLKA and CLKB may be connected to sequential cells or to combinational logic. If CLKA or CLKB pins are not used or sourced from signals, these pins must be set as Low or High on the board. They must not be left floating. Figure 1-8 describes the CLKA and CLKB circuit used and the macros supported in SX-A devices with the exception of A54SX72A.

In addition, the A54SX72A device provides four quadrant clocks (QCLKA, QCLKB, QCLKC, and QCLKD corresponding to bottom-left, bottom-right, top-left, and top-right locations on the die, respectively), which can be sourced from external pins or from internal logic signals within the device. Each of these clocks can individually drive up to an entire quadrant of the chip, or they can be grouped together to drive multiple quadrants (Figure 1-9 on page 1-6). QCLK pins can function as user I/O pins. If not used, the QCLK pins must be tied Low or High on the board and must not be left floating.

For more information on how to use quadrant clocks in the A54SX72A device, refer to the *Global Clock Networks in Actel's Antifuse Devices* and *Using A54SX72A and RT54SX72S Quadrant Clocks* application notes.

The CLKA, CLKB, and QCLK circuits for A54SX72A as well as the macros supported are shown in Figure 1-10 on page 1-6. Note that bidirectional clock buffers are only available in A54SX72A. For more information, refer to the "Pin Description" section on page 1-15.

Table 1-1 • SX-A Clock Resources

	A54SX08A	A54SX16A	A54SX32A	A54SX72A
Routed Clocks (CLKA, CLKB)	2	2	2	2
Hardwired Clocks (HCLK)	1	1	1	1
Quadrant Clocks (QCLKA, QCLKB, QCLKC, QCLKD)	0	0	0	4



Figure 1-7 • SX-A HCLK Clock Buffer



Figure 1-8 • SX-A Routed Clock Buffer



Probing Capabilities

SX-A devices also provide an internal probing capability that is accessed with the JTAG pins. The Silicon Explorer II diagnostic hardware is used to control the TDI, TCK, TMS, and TDO pins to select the desired nets for debugging. The user assigns the selected internal nets in Actel Silicon Explorer II software to the PRA/PRB output pins for observation. Silicon Explorer II automatically places the device into JTAG mode. However, probing functionality is only activated when the TRST pin is driven high or left floating, allowing the internal pull-up resistor to pull TRST High. If the TRST pin is held Low, the TAP controller remains in the Test-Logic-Reset state so no probing can be performed. However, the user must drive the TRST pin High or allow the internal pull-up resistor to pull TRST High. When selecting the **Reserve Probe Pin** box as shown in Figure 1-12 on page 1-9, direct the layout tool to reserve the PRA and PRB pins as dedicated outputs for probing. This **Reserve** option is merely a guideline. If the designer assigns user I/Os to the PRA and PRB pins and selects the **Reserve Probe Pin** option, Designer Layout will override the **Reserve Probe Pin** option and place the user I/Os on those pins.

To allow probing capabilities, the security fuse must not be programmed. Programming the security fuse disables the JTAG and probe circuitry. Table 1-9 summarizes the possible device configurations for probing once the device leaves the Test-Logic-Reset JTAG state.

JTAG Mode	TRST ¹	Security Fuse Programmed	PRA, PRB ²	TDI, TCK, TDO ²
Dedicated	Low	No	User I/O ³	JTAG Disabled
	High	No	Probe Circuit Outputs	JTAG I/O
Flexible	Low	No	User I/O ³	User I/O ³
	High	No	Probe Circuit Outputs	JTAG I/O
		Yes	Probe Circuit Secured	Probe Circuit Secured

Table 1-9 • Device Configuration Options for Probe Capability (TRST Pin Reserved)

Notes:

1. If the TRST pin is not reserved, the device behaves according to TRST = High as described in the table.

2. Avoid using the TDI, TCK, TDO, PRA, and PRB pins as input or bidirectional ports. Since these pins are active during probing, input signals will not pass through these pins and may cause contention.

3. If no user signal is assigned to these pins, they will behave as unused I/Os in this mode. Unused pins are automatically tristated by the Designer software.

SX-A Probe Circuit Control Pins

SX-A devices contain internal probing circuitry that provides built-in access to every node in a design, enabling 100% real-time observation and analysis of a device's internal logic nodes without design iteration. The probe circuitry is accessed by Silicon Explorer II, an easy to use, integrated verification and logic analysis tool that can sample data at 100 MHz (asynchronous) or 66 MHz (synchronous). Silicon Explorer II attaches to a PC's standard COM port, turning the PC into a fully functional 18-channel logic analyzer. Silicon Explorer II allows designers to complete the design verification process at their desks and reduces verification time from several hours per cycle to a few seconds.

The Silicon Explorer II tool uses the boundary-scan ports (TDI, TCK, TMS, and TDO) to select the desired nets for verification. The selected internal nets are assigned to the

PRA/PRB pins for observation. Figure 1-13 illustrates the interconnection between Silicon Explorer II and the FPGA to perform in-circuit verification.

Design Considerations

In order to preserve device probing capabilities, users should avoid using the TDI, TCK, TDO, PRA, and PRB pins as input or bidirectional ports. Since these pins are active during probing, critical input signals through these pins are not available. In addition, the security fuse must not be programmed to preserve probing capabilities. Actel recommends that you use a 70 Ω series termination resistor on every probe connector (TDI, TCK, TMS, TDO, PRA, PRB). The 70 Ω series termination is used to prevent data transmission corruption during probing and reading back the checksum.



Figure 1-13 • Probe Setup

Pin Description

CLKA/B, I/O Clock A and B

These pins are clock inputs for clock distribution networks. Input levels are compatible with standard TTL, LVTTL, LVCMOS2, 3.3 V PCI, or 5 V PCI specifications. The clock input is buffered prior to clocking the R-cells. When not used, this pin must be tied Low or High (NOT left floating) on the board to avoid unwanted power consumption.

For A54SX72A, these pins can also be configured as user I/Os. When employed as user I/Os, these pins offer builtin programmable pull-up or pull-down resistors active during power-up only. When not used, these pins must be tied Low or High (NOT left floating).

QCLKA/B/C/D, I/O Quadrant Clock A, B, C, and D

These four pins are the quadrant clock inputs and are only used for A54SX72A with A, B, C, and D corresponding to bottom-left, bottom-right, top-left, and top-right quadrants, respectively. They are clock inputs for clock distribution networks. Input levels are compatible with standard TTL, LVTTL, LVCMOS2, 3.3 V PCI, or 5 V PCI specifications. Each of these clock inputs can drive up to a quarter of the chip, or they can be grouped together to drive multiple quadrants. The clock input is buffered prior to clocking the R-cells. When not used, these pins must be tied Low or High on the board (NOT left floating).

These pins can also be configured as user I/Os. When employed as user I/Os, these pins offer built-in programmable pull-up or pull-down resistors active during power-up only.

GND Ground

Low supply voltage.

HCLK Dedicated (Hardwired) Array Clock

This pin is the clock input for sequential modules. Input levels are compatible with standard TTL, LVTTL, LVCMOS2, 3.3 V PCI, or 5 V PCI specifications. This input is directly wired to each R-cell and offers clock speeds independent of the number of R-cells being driven. When not used, HCLK must be tied Low or High on the board (NOT left floating). When used, this pin should be held Low or High during power-up to avoid unwanted static power consumption.

I/O Input/Output

The I/O pin functions as an input, output, tristate, or bidirectional buffer. Based on certain configurations, input and output levels are compatible with standard TTL, LVTTL, LVCMOS2, 3.3 V PCI or 5 V PCI specifications. Unused I/O pins are automatically tristated by the Designer software.

NC No Connection

This pin is not connected to circuitry within the device and can be driven to any voltage or be left floating with no effect on the operation of the device.

PRA/B, I/O Probe A/B

The Probe pin is used to output data from any userdefined design node within the device. This independent diagnostic pin can be used in conjunction with the other probe pin to allow real-time diagnostic output of any signal path within the device. The Probe pin can be used as a user-defined I/O when verification has been completed. The pin's probe capabilities can be permanently disabled to protect programmed design confidentiality.

TCK, I/O Test Clock

Test clock input for diagnostic probe and device programming. In Flexible mode, TCK becomes active when the TMS pin is set Low (refer to Table 1-6 on page 1-9). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TDI, I/O Test Data Input

Serial input for boundary scan testing and diagnostic probe. In Flexible mode, TDI is active when the TMS pin is set Low (refer to Table 1-6 on page 1-9). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TDO, I/O Test Data Output

Serial output for boundary scan testing. In flexible mode, TDO is active when the TMS pin is set Low (refer to Table 1-6 on page 1-9). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state. When Silicon Explorer II is being used, TDO will act as an output when the checksum command is run. It will return to user /IO when checksum is complete.

TMS Test Mode Select

The TMS pin controls the use of the IEEE 1149.1 Boundary Scan pins (TCK, TDI, TDO, TRST). In flexible mode when the TMS pin is set Low, the TCK, TDI, and TDO pins are boundary scan pins (refer to Table 1-6 on page 1-9). Once the boundary scan pins are in test mode, they will remain in that mode until the internal boundary scan state machine reaches the logic reset state. At this point, the boundary scan pins will be released and will function as regular I/O pins. The logic reset state is reached five TCK cycles after the TMS pin is set High. In dedicated test mode, TMS functions as specified in the IEEE 1149.1 specifications.

TRST, I/O Boundary Scan Reset Pin

Once it is configured as the JTAG Reset pin, the TRST pin functions as an active low input to asynchronously initialize or reset the boundary scan circuit. The TRST pin is equipped with an internal pull-up resistor. This pin functions as an I/O when the **Reserve JTAG Reset Pin** is not selected in Designer.

V_{CCI} Supply Voltage

Supply voltage for I/Os. See Table 2-2 on page 2-1. All V_{CCI} power pins in the device should be connected.

V_{CCA} Supply Voltage

Supply voltage for array. See Table 2-2 on page 2-1. All V_{CCA} power pins in the device should be connected.



To determine the heat sink's thermal performance, use the following equation:

$$\theta_{JA(TOTAL)} = \theta_{JC} + \theta_{CS} + \theta_{SA}$$

EQ 2-14

where:

 $\theta_{CS} = 0.37^{\circ}C/W$

 thermal resistance of the interface material between the case and the heat sink, usually provided by the thermal interface manufacturer

 θ_{SA} = thermal resistance of the heat sink in °C/W

 $\theta_{SA} = \theta_{JA(TOTAL)} - \theta_{JC} - \theta_{CS}$ EQ 2-15 $\theta_{SA} = 13.33^{\circ}C/W - 3.20^{\circ}C/W - 0.37^{\circ}C/W$

$$\theta_{SA} = 9.76^{\circ}C/W$$

A heat sink with a thermal resistance of 9.76°C/W or better should be used. Thermal resistance of heat sinks is a function of airflow. The heat sink performance can be significantly improved with the presence of airflow.

Carefully estimating thermal resistance is important in the long-term reliability of an Actel FPGA. Design engineers should always correlate the power consumption of the device with the maximum allowable power dissipation of the package selected for that device, using the provided thermal resistance data.

Note: The values may vary depending on the application.

SX-A Timing Model



Note: *Values shown for A54SX72A, –2, worst-case commercial conditions at 5 V PCI with standard place-and-route. Figure 2-3 • SX-A Timing Model

Sample Path Calculations

Hardwired Clock

External Setup	=	(t _{INYH} + t _{RD1} + t _{SUD}) – t _{HCKH}
	=	0.6 + 0.3 + 0.8 - 1.8 = - 0.1 ns
Clock-to-Out (Pad-to-Pad)	=	t _{HCKH} + t _{RCO} + t _{RD1} + t _{DHL}
	=	1.8 + 0.8 + 0.3 + 3.9 = 6.8 ns

Routed Clock

External Setup	$= (t_{INYH} + t_{RD1} + t_{SUD}) - t_{RC}$	СКН
	= 0.6 + 0.3 + 0.8 - 3.0 = -1.	3 ns
Clock-to-Out (Pad-to-Pad	$= t_{RCKH} + t_{RCO} + t_{RD1} + t_{DH}$	L
	= 3.0 + 0.8 + 0.3 + 3.9 = 8.0) ns

Input Buffer Delays



t INY **C-Cell Delays**



Figure 2-6 • Input Buffer Delays

GND

Figure 2-7 • C-Cell Delays

Cell Timing Characteristics

t_{INY}



Figure 2-8 • Flip-Flops

Table 2-19 • A54SX08A Timing Characteristics

(Worst-Case Commercial Conditions V_{CCA} = 2.25 V, V_{CCI} = 3.0 V, T_J = 70°C)

		-2 S	peed	-1 S	peed	Std. S	Speed	–F Speed		
Parameter	Description	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Units
3.3 V PCI Ou	tput Module Timing ¹									
t _{DLH}	Data-to-Pad Low to High		2.2		2.4		2.9		4.0	ns
t _{DHL}	Data-to-Pad High to Low		2.3		2.6		3.1		4.3	ns
t _{ENZL}	Enable-to-Pad, Z to L		1.7		1.9		2.2		3.1	ns
t _{ENZH}	Enable-to-Pad, Z to H		2.2		2.4		2.9		4.0	ns
t _{ENLZ}	Enable-to-Pad, L to Z		2.8		3.2		3.8		5.3	ns
t _{ENHZ}	Enable-to-Pad, H to Z		2.3		2.6		3.1		4.3	ns
d_{TLH}^2	Delta Low to High		0.03		0.03		0.04		0.045	ns/pF
d_{THL}^2	Delta High to Low		0.015		0.015		0.015		0.025	ns/pF
3.3 V LVTTL O	Dutput Module Timing ³							-		
t _{DLH}	Data-to-Pad Low to High		3.0		3.4		4.0		5.6	ns
t _{DHL}	Data-to-Pad High to Low		3.0		3.3		3.9		5.5	ns
t _{DHLS}	Data-to-Pad High to Low—low slew		10.4		11.8		13.8		19.3	ns
t _{ENZL}	Enable-to-Pad, Z to L		2.6		2.9		3.4		4.8	ns
t _{ENZLS}	Enable-to-Pad, Z to L—low slew		18.9		21.3		25.4		34.9	ns
t _{ENZH}	Enable-to-Pad, Z to H		3		3.4		4		5.6	ns
t _{ENLZ}	Enable-to-Pad, L to Z		3.3		3.7		4.4		6.2	ns
t _{ENHZ}	Enable-to-Pad, H to Z		3		3.3		3.9		5.5	ns
d_{TLH}^{2}	Delta Low to High		0.03		0.03		0.04		0.045	ns/pF
d_{THL}^2	Delta High to Low		0.015		0.015		0.015		0.025	ns/pF
d_{THLS}^2	Delta High to Low—low slew		0.053		0.067		0.073		0.107	ns/pF

Notes:

1. Delays based on 10 pF loading and 25 Ω resistance.

2. To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation: Slew Rate $[V/ns] = (0.1 * V_{CCI} - 0.9 * V_{CCI}) / (C_{load} * d_{T[LH|HL|HLS]})$ where C_{load} is the load capacitance driven by the I/O in pF

 $d_{T[LH|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.

3. Delays based on 35 pF loading.

Table 2-23 • A54SX16A Timing Characteristics

(Worst-Case Commercial Conditions V _{CC}	_A = 2.25 V, V _{CCl} = 3.0 V, T _J = 70°C)
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		-3 Sp	beed*	-2 S	peed	-1 S	peed	Std.	Speed	-F S	peed	
Parameter	Description	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Units
Dedicated	(Hardwired) Array Clock Netwo	rks										
t _{HCKH}	Input Low to High (Pad to R-cell Input)		1.2		1.4		1.6		1.8		2.8	ns
t _{HCKL}	Input High to Low (Pad to R-cell Input)		1.0		1.1		1.3		1.5		2.2	ns
t _{HPWH}	Minimum Pulse Width High	1.4		1.7		1.9		2.2		3.0		ns
t _{HPVVL}	Minimum Pulse Width Low	1.4		1.7		1.9		2.2		3.0		ns
t _{HCKSW}	Maximum Skew		0.3		0.3		0.4		0.4		0.6	ns
t _{HP}	Minimum Period	2.8		3.4		3.8		4.4		6.0		ns
f _{HMAX}	Maximum Frequency		357		294		263		227		167	MHz
Routed Arr	ay Clock Networks											
t _{RCKH}	Input Low to High (Light Load) (Pad to R-cell Input)		1.0		1.2		1.3		1.5		2.1	ns
t _{RCKL}	Input High to Low (Light Load) (Pad to R-cell Input)		1.1		1.3		1.5		1.7		2.4	ns
t _{RCKH}	Input Low to High (50% Load) (Pad to R-cell Input)		1.1		1.3		1.4		1.7		2.3	ns
t _{RCKL}	Input High to Low (50% Load) (Pad to R-cell Input)		1.1		1.3		1.5		1.7		2.4	ns
t _{RCKH}	Input Low to High (100% Load) (Pad to R-cell Input)		1.3		1.5		1.7		2.0		2.7	ns
t _{RCKL}	Input High to Low (100% Load) (Pad to R-cell Input)		1.3		1.5		1.7		2.0		2.8	ns
t _{RPWH}	Minimum Pulse Width High	1.4		1.7		1.9		2.2		3.0		ns
t _{RPWL}	Minimum Pulse Width Low	1.4		1.7		1.9		2.2		3.0		ns
t _{RCKSW}	Maximum Skew (Light Load)		0.8		0.9		1.0		1.2		1.7	ns
t _{RCKSW}	Maximum Skew (50% Load)		0.8		0.9		1.0		1.2		1.7	ns
t _{RCKSW}	Maximum Skew (100% Load)		1.0		1.1		1.3		1.5		2.1	ns

Note: *All –3 speed grades have been discontinued.

Table 2-26 • A54SX16A Timing Characteristics

(Worst-Case Commercial Condition	V _{CCA} = 2.25 V, V _{CCI} =	= 3.0 V, T _J = 70°C)
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		-3 Speed ¹	-2 Speed	-1	1 Speed	Std. Speed	-F Speed	
Parameter	Description	Min. Max.	Min. Max	. Mi	in. Max.	Min. Max.	Min. Max.	Units
3.3 V PCI O	utput Module Timing ²							
t _{DLH}	Data-to-Pad Low to High	2.0	2.3		2.6	3.1	4.3	ns
t _{DHL}	Data-to-Pad High to Low	2.2	2.5		2.8	3.3	4.6	ns
t _{ENZL}	Enable-to-Pad, Z to L	1.4	1.7		1.9	2.2	3.1	ns
t _{ENZH}	Enable-to-Pad, Z to H	2.0	2.3		2.6	3.1	4.3	ns
t _{ENLZ}	Enable-to-Pad, L to Z	2.5	2.8		3.2	3.8	5.3	ns
t _{ENHZ}	Enable-to-Pad, H to Z	2.2	2.5		2.8	3.3	4.6	ns
d _{TLH} ³	Delta Low to High	0.025	0.0	3	0.03	0.04	0.045	ns/pF
d _{THL} ³	Delta High to Low	0.015	0.01	5	0.015	0.015	0.025	ns/pF
3.3 V LVTTL	Output Module Timing ⁴						•	
t _{DLH}	Data-to-Pad Low to High	2.8	3.2		3.6	4.3	6.0	ns
t _{DHL}	Data-to-Pad High to Low	2.7	3.1		3.5	4.1	5.7	ns
t _{DHLS}	Data-to-Pad High to Low—low slew	9.5	10.9	9	12.4	14.6	20.4	ns
t _{ENZL}	Enable-to-Pad, Z to L	2.2	2.6		2.9	3.4	4.8	ns
t _{ENZLS}	Enable-to-Pad, Z to L—low slew	15.8	18.9	9	21.3	25.4	34.9	ns
t _{ENZH}	Enable-to-Pad, Z to H	2.8	3.2		3.6	4.3	6.0	ns
t _{ENLZ}	Enable-to-Pad, L to Z	2.9	3.3		3.7	4.4	6.2	ns
t _{ENHZ}	Enable-to-Pad, H to Z	2.7	3.1		3.5	4.1	5.7	ns
d _{TLH} ³	Delta Low to High	0.025	0.0	3	0.03	0.04	0.045	ns/pF
d _{THL} ³	Delta High to Low	0.015	0.01	5	0.015	0.015	0.025	ns/pF
d _{THLS} ³	Delta High to Low—low slew	0.053	0.05	3	0.067	0.073	0.107	ns/pF

Notes:

1. All –3 speed grades have been discontinued.

2. Delays based on 10 pF loading and 25 Ω resistance.

3. To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation: Slew Rate [V/ns] = (0.1* V_{CCI} – 0.9* V_{CCI} / (C_{load} * $d_{T[LH|HL|HLS]}$) where C_{load} is the load capacitance driven by the I/O in pF

 $d_{T[LH|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.

4. Delays based on 35 pF loading.

Table 2-32 A54SX32A Timing Characteristics

(Worst-Case Commercial	Conditions V _{CCA} = 2.25	$V, V_{CCI} = 2.3 V, T_{J} = 70^{\circ}C$
•		

		–3 Spee	ed ¹	-2 S	peed	–1 S	peed Std. Speed		-F Speed			
Parameter	Description	Min. M	lax.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Units
2.5 V LVCM	OS Output Module Timing ^{2,3}											
t _{DLH}	Data-to-Pad Low to High	3	3.3		3.8		4.2		5.0		7.0	ns
t _{DHL}	Data-to-Pad High to Low	2	2.5		2.9		3.2		3.8		5.3	ns
t _{DHLS}	Data-to-Pad High to Low—low slew	1	1.1		12.8		14.5		17.0		23.8	ns
t _{ENZL}	Enable-to-Pad, Z to L	2	2.4		2.8		3.2		3.7		5.2	ns
t _{ENZLS}	Data-to-Pad, Z to L—low slew	1	1.8		13.7		15.5		18.2		25.5	ns
t _{ENZH}	Enable-to-Pad, Z to H	3	3.3		3.8		4.2		5.0		7.0	ns
t _{ENLZ}	Enable-to-Pad, L to Z	2	2.1		2.5		2.8		3.3		4.7	ns
t _{ENHZ}	Enable-to-Pad, H to Z	2	2.5		2.9		3.2		3.8		5.3	ns
d_{TLH}^{4}	Delta Low to High	0.0	031		0.037		0.043		0.051		0.071	ns/pF
d_{THL}^{4}	Delta High to Low	0.0	017		0.017		0.023		0.023		0.037	ns/pF
d_{THLS}^4	Delta High to Low—low slew	0.0	057		0.06		0.071		0.086		0.117	ns/pF

Note:

1. All –3 speed grades have been discontinued.

2. Delays based on 35 pF loading.

3. The equivalent IO Attribute settings for 2.5 V LVCMOS is 2.5 V LVTTL in the software.

4. To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation: Slew Rate [V/ns] = $(0.1 * V_{CCI} - 0.9 * V_{CCI})/(C_{load} * d_{T[LH|HL|HLS]})$ where C_{load} is the load capacitance driven by the I/O in pF

 $d_{T[LH|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.

Table 2-33 • A54SX32A Timing Characteristics

(Worst-Case Commercial Conditions	V _{CCA} = 2.25 V, V _{CCI} =	= 3.0 V, T _J = 70°C)
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		-3 Speed ¹	-2 Speed	–1 Speed	Std. Speed	-F Speed	
Parameter	Description	Min. Max.	Min. Max.	Min. Max.	Min. Max.	Min. Max.	Units
3.3 V PCI O	utput Module Timing ²		•				
t _{DLH}	Data-to-Pad Low to High	1.9	2.2	2.4	2.9	4.0	ns
t _{DHL}	Data-to-Pad High to Low	2.0	2.3	2.6	3.1	4.3	ns
t _{ENZL}	Enable-to-Pad, Z to L	1.4	1.7	1.9	2.2	3.1	ns
t _{ENZH}	Enable-to-Pad, Z to H	1.9	2.2	2.4	2.9	4.0	ns
t _{ENLZ}	Enable-to-Pad, L to Z	2.5	2.8	3.2	3.8	5.3	ns
t _{ENHZ}	Enable-to-Pad, H to Z	2.0	2.3	2.6	3.1	4.3	ns
d _{TLH} ³	Delta Low to High	0.025	0.03	0.03	0.04	0.045	ns/pF
d _{THL} ³	Delta High to Low	0.015	0.015	0.015	0.015	0.025	ns/pF
3.3 V LVTTL	Output Module Timing ⁴		•		•	•	
t _{DLH}	Data-to-Pad Low to High	2.6	3.0	3.4	4.0	5.6	ns
t _{DHL}	Data-to-Pad High to Low	2.6	3.0	3.3	3.9	5.5	ns
t _{DHLS}	Data-to-Pad High to Low—low slew	9.0	10.4	11.8	13.8	19.3	ns
t _{ENZL}	Enable-to-Pad, Z to L	2.2	2.6	2.9	3.4	4.8	ns
t _{ENZLS}	Enable-to-Pad, Z to L—low slew	15.8	18.9	21.3	25.4	34.9	ns
t _{ENZH}	Enable-to-Pad, Z to H	2.6	3.0	3.4	4.0	5.6	ns
t _{ENLZ}	Enable-to-Pad, L to Z	2.9	3.3	3.7	4.4	6.2	ns
t _{ENHZ}	Enable-to-Pad, H to Z	2.6	3.0	3.3	3.9	5.5	ns
d_{TLH}^{3}	Delta Low to High	0.025	0.03	0.03	0.04	0.045	ns/pF
d _{THL} ³	Delta High to Low	0.015	0.015	0.015	0.015	0.025	ns/pF
d _{THLS} ³	Delta High to Low—low slew	0.053	0.053	0.067	0.073	0.107	ns/pF

Notes:

1. All –3 speed grades have been discontinued.

2. Delays based on 10 pF loading and 25 Ω resistance.

3. To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation: Slew Rate [V/ns] = (0.1* V_{CCI} – 0.9* V_{CCI} / (C_{load} * $d_{T[LH|HL|HLS]}$) where C_{load} is the load capacitance driven by the I/O in pF

 $d_{T[LH|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.

4. Delays based on 35 pF loading.

Table 2-36 A54SX72A Timing Characteristics

(Worst-Case Commercial Conditions V _{CCA}	_λ = 2.25 V, V _{CCI} = 2.25 V, Τ _J = 70°C
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		-3 Sp	beed*	-2 S	peed	-1 S	peed	Std. Speed		-F Speed		
Parameter	Description	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Units
Dedicated ((Hardwired) Array Clock Netwo	rks										1
t _{нскн}	Input Low to High (Pad to R-cell Input)		1.6		1.9		2.1		2.5		3.8	ns
t _{HCKL}	Input High to Low (Pad to R-cell Input)		1.6		1.9		2.1		2.5		3.8	ns
t _{HPWH}	Minimum Pulse Width High	1.5		1.7		2.0		2.3		3.2		ns
t _{HPWL}	Minimum Pulse Width Low	1.5		1.7		2.0		2.3		3.2		ns
t _{HCKSW}	Maximum Skew		1.4		1.6		1.8		2.1		3.3	ns
t _{HP}	Minimum Period	3.0		3.4		4.0		4.6		6.4		ns
f _{HMAX}	Maximum Frequency		333		294		250		217		156	MHz
Routed Arra	ay Clock Networks											
t _{rckh}	Input Low to High (Light Load) (Pad to R-cell Input)		2.3		2.6		2.9		3.4		4.8	ns
t _{rckl}	Input High to Low (Light Load) (Pad to R-cell Input)		2.8		3.2		3.7		4.3		6.0	ns
t _{rckh}	Input Low to High (50% Load) (Pad to R-cell Input)		2.4		2.8		3.2		3.7		5.2	ns
t _{rckl}	Input High to Low (50% Load) (Pad to R-cell Input)		2.9		3.3		3.8		4.5		6.2	ns
t _{rckh}	Input Low to High (100% Load) (Pad to R-cell Input)		2.6		3.0		3.4		4.0		5.6	ns
t _{rckl}	Input High to Low (100% Load) (Pad to R-cell Input)		3.1		3.6		4.0		4.7		6.6	ns
t _{RPWH}	Minimum Pulse Width High	1.5		1.7		2.0		2.3		3.2		ns
t _{RPWL}	Minimum Pulse Width Low	1.5		1.7		2.0		2.3		3.2		ns
t _{RCKSW}	Maximum Skew (Light Load)		1.9		2.2		2.5		3.0		4.1	ns
t _{RCKSW}	Maximum Skew (50% Load)		1.8		2.1		2.4		2.8		3.9	ns
t _{RCKSW}	Maximum Skew (100% Load)		1.8		2.1		2.4		2.8		3.9	ns
Quadrant A	rray Clock Networks											
t _{QCKH}	Input Low to High (Light Load) (Pad to R-cell Input)		2.6		3.0		3.4		4.0		5.6	ns
t _{QCHKL}	Input High to Low (Light Load) (Pad to R-cell Input)		2.6		3.0		3.3		3.9		5.5	ns
t _{QCKH}	Input Low to High (50% Load) (Pad to R-cell Input)		2.8		3.2		3.6		4.3		6.0	ns
t _{qchkl}	Input High to Low (50% Load) (Pad to R-cell Input)		2.8		3.2		3.6		4.2		5.9	ns

Note: *All –3 speed grades have been discontinued.



Package Pin Assignments

208-Pin PQFP



Figure 3-1 • 208-Pin PQFP (Top View)

Note

For Package Manufacturing and Environmental information, visit Resource center at http://www.actel.com/products/rescenter/package/index.html.

176-P	176-Pin TQFP		in TQFP	176-P	in TQFP	176-Pin TQFP		
Pin Number	A54SX32A Function	Pin Number	A54SX32A Function	Pin Number	A54SX32A Function	Pin Number	A54SX32A Function	
1	GND	37	I/O	73	I/O	109	V _{CCA}	
2	TDI, I/O	38	I/O	74	I/O	110	GND	
3	I/O	39	I/O	75	I/O	111	I/O	
4	I/O	40	I/O	76	I/O	112	I/O	
5	I/O	41	I/O	77	I/O	113	I/O	
6	I/O	42	I/O	78	I/O	114	I/O	
7	I/O	43	I/O	79	I/O	115	I/O	
8	I/O	44	GND	80	I/O	116	I/O	
9	I/O	45	I/O	81	I/O	117	I/O	
10	TMS	46	I/O	82	V _{CCI}	118	I/O	
11	V _{CCI}	47	I/O	83	I/O	119	I/O	
12	I/O	48	I/O	84	I/O	120	I/O	
13	I/O	49	I/O	85	I/O	121	I/O	
14	I/O	50	I/O	86	I/O	122	V _{CCA}	
15	I/O	51	I/O	87	TDO, I/O	123	GND	
16	I/O	52	V _{CCI}	88	I/O	124	V _{CCI}	
17	I/O	53	I/O	89	GND	125	I/O	
18	I/O	54	I/O	90	I/O	126	I/O	
19	I/O	55	I/O	91	I/O	127	I/O	
20	I/O	56	I/O	92	I/O	128	I/O	
21	GND	57	I/O	93	I/O	129	I/O	
22	V _{CCA}	58	I/O	94	I/O	130	I/O	
23	GND	59	I/O	95	I/O	131	I/O	
24	I/O	60	I/O	96	I/O	132	I/O	
25	TRST, I/O	61	I/O	97	I/O	133	GND	
26	I/O	62	I/O	98	V _{CCA}	134	I/O	
27	I/O	63	I/O	99	V _{CCI}	135	I/O	
28	I/O	64	PRB, I/O	100	I/O	136	I/O	
29	I/O	65	GND	101	I/O	137	I/O	
30	I/O	66	V _{CCA}	102	I/O	138	I/O	
31	I/O	67	NC	103	I/O	139	I/O	
32	V _{CCI}	68	I/O	104	I/O	140	V _{CCI}	
33	V _{CCA}	69	HCLK	105	I/O	141	I/O	
34	I/O	70	I/O	106	I/O	142	I/O	
35	I/O	71	I/O	107	I/O	143	I/O	
36	I/O	72	I/O	108	GND	144	I/O	

329-Pi	n PBGA	329-Pin PBGA		329-Pi	n PBGA	329-Pi	n PBGA
Pin Number	A54SX32A Function	Pin Number	A54SX32A Function	Pin Number	A54SX32A Function	Pin Number	A54SX32A Function
D11	V _{CCA}	H1	I/O	L14	GND	P12	GND
D12	NC	H2	I/O	L20	NC	P13	GND
D13	I/O	H3	I/O	L21	I/O	P14	GND
D14	I/O	H4	I/O	L22	I/O	P20	I/O
D15	I/O	H20	V _{CCA}	L23	NC	P21	I/O
D16	I/O	H21	I/O	M1	I/O	P22	I/O
D17	I/O	H22	I/O	M2	I/O	P23	I/O
D18	I/O	H23	I/O	M3	I/O	R1	I/O
D19	I/O	J1	NC	M4	V _{CCA}	R2	I/O
D20	I/O	J2	I/O	M10	GND	R3	I/O
D21	I/O	J3	I/O	M11	GND	R4	I/O
D22	I/O	J4	I/O	M12	GND	R20	I/O
D23	I/O	J20	I/O	M13	GND	R21	I/O
E1	V _{CCI}	J21	I/O	M14	GND	R22	I/O
E2	I/O	J22	I/O	M20	V _{CCA}	R23	I/O
E3	I/O	J23	I/O	M21	I/O	T1	I/O
E4	I/O	K1	I/O	M22	I/O	T2	I/O
E20	I/O	К2	I/O	M23	V _{CCI}	T3	I/O
E21	I/O	К3	I/O	N1	I/O	T4	I/O
E22	I/O	К4	I/O	N2	TRST, I/O	T20	I/O
E23	I/O	K10	GND	N3	I/O	T21	I/O
F1	I/O	K11	GND	N4	I/O	T22	I/O
F2	TMS	K12	GND	N10	GND	T23	I/O
F3	I/O	K13	GND	N11	GND	U1	I/O
F4	I/O	K14	GND	N12	GND	U2	I/O
F20	I/O	K20	I/O	N13	GND	U3	V _{CCA}
F21	I/O	K21	I/O	N14	GND	U4	I/O
F22	I/O	K22	I/O	N20	NC	U20	I/O
F23	I/O	K23	I/O	N21	I/O	U21	V _{CCA}
G1	I/O	L1	I/O	N22	I/O	U22	I/O
G2	I/O	L2	I/O	N23	I/O	U23	I/O
G3	I/O	L3	I/O	P1	I/O	V1	V _{CCI}
G4	I/O	L4	NC	P2	I/O	V2	I/O
G20	I/O	L10	GND	P3	I/O	V3	I/O
G21	I/O	L11	GND	P4	I/O	V4	I/O
G22	I/O	L12	GND	P10	GND	V20	I/O
G23	GND	L13	GND	P11	GND	V21	I/O

144-Pin FBGA



Figure 3-6 • 144-Pin FBGA (Top View)

Note

For Package Manufacturing and Environmental information, visit Resource center at http://www.actel.com/products/rescenter/package/index.html.

Previous Version	Changes in Current Version (v5.3)	Page
v4.0	Table 2-12 was updated.	2-11
(continued)	The was updated.	2-14
	The "Sample Path Calculations" were updated.	2-14
	Table 2-13 was updated.	2-17
	Table 2-13 was updated.	2-17
	All timing tables were updated.	2-18 to 2-52
v3.0	The "Actel Secure Programming Technology with FuseLock™ Prevents Reverse Engineering and Design Theft" section was updated.	1-i
	The "Ordering Information" section was updated.	1-ii
	The "Temperature Grade Offering" section was updated.	1-iii
	The Figure 1-1 • SX-A Family Interconnect Elements was updated.	1-1
	The ""Clock Resources" section" was updated	1-5
	The Table 1-1 • SX-A Clock Resources is new.	1-5
	The "User Security" section is new.	1-7
	The "I/O Modules" section was updated.	1-7
	The Table 1-2 • I/O Features was updated.	1-8
	The Table 1-3 • I/O Characteristics for All I/O Configurations is new.	1-8
	The Table 1-4 • Power-Up Time at which I/Os Become Active is new	1-8
	The Figure 1-12 • Device Selection Wizard is new.	1-9
	The "Boundary-Scan Pin Configurations and Functions" section is new.	1-9
	The Table 1-9 • Device Configuration Options for Probe Capability (TRST Pin Reserved) is new.	1-11
	The "SX-A Probe Circuit Control Pins" section was updated.	1-12
	The "Design Considerations" section was updated.	1-12
	The Figure 1-13 • Probe Setup was updated.	1-12
	The Design Environment was updated.	1-13
	The Figure 1-13 • Design Flow is new.	1-11
	The "Absolute Maximum Ratings*" section was updated.	1-12
	The "Recommended Operating Conditions" section was updated.	1-12
	The "Electrical Specifications" section was updated.	1-12
	The "2.5V LVCMOS2 Electrical Specifications" section was updated.	1-13
	The "SX-A Timing Model" and "Sample Path Calculations" equations were updated.	1-23
	The "Pin Description" section was updated.	1-15
v2.0.1	The "Design Environment" section has been updated.	1-13
	The "I/O Modules" section, and Table 1-2 • I/O Features have been updated.	1-8
	The "SX-A Timing Model" section and the "Timing Characteristics" section have new timing numbers.	1-23