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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	6036
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	171
Number of Gates	108000
Voltage - Supply	2.25V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/a54sx72a-1pqg208i

Routing Resources

The routing and interconnect resources of SX-A devices are in the top two metal layers above the logic modules (Figure 1-1 on page 1-1), providing optimal use of silicon, thus enabling the entire floor of the device to be spanned with an uninterrupted grid of logic modules. Interconnection between these logic modules is achieved using the Actel patented metal-to-metal programmable antifuse interconnect elements. The antifuses are normally open circuits and, when programmed, form a permanent low-impedance connection.

Clusters and SuperClusters can be connected through the use of two innovative local routing resources called FastConnect and DirectConnect, which enable extremely fast and predictable interconnection of modules within Clusters and SuperClusters (Figure 1-5 on page 1-4 and Figure 1-6 on page 1-4). This routing architecture also dramatically reduces the number of antifuses required to complete a circuit, ensuring the highest possible performance, which is often required in applications such as fast counters, state machines, and data path logic. The interconnect elements (i.e., the antifuses and metal tracks) have lower capacitance and lower resistance than any other device of similar capacity, leading to the fastest signal propagation in the industry.

DirectConnect is a horizontal routing resource that provides connections from a C-cell to its neighboring R-Cell in a given SuperCluster. DirectConnect uses a hardwired signal path requiring no programmable interconnection to achieve its fast signal propagation time of less than 0.1 ns.

FastConnect enables horizontal routing between any two logic modules within a given SuperCluster, and vertical routing with the SuperCluster immediately below it. Only one programmable connection is used in a FastConnect path, delivering a maximum pin-to-pin propagation time of 0.3 ns.

In addition to DirectConnect and FastConnect, the architecture makes use of two globally oriented routing resources known as segmented routing and high-drive routing. The Actel segmented routing structure provides a variety of track lengths for extremely fast routing between SuperClusters. The exact combination of track lengths and antifuses within each path is chosen by the 100% automatic place-and-route software to minimize signal propagation delays.

The general system of routing tracks allows any logic module in the array to be connected to any other logic or I/O module. Within this system, most connections typically require three or fewer antifuses, resulting in fast and predictable performance.

The unique local and general routing structure featured in SX-A devices allows 100% pin-locking with full logic utilization, enables concurrent printed circuit board (PCB) development, reduces design time, and allows designers to achieve performance goals with minimum effort.

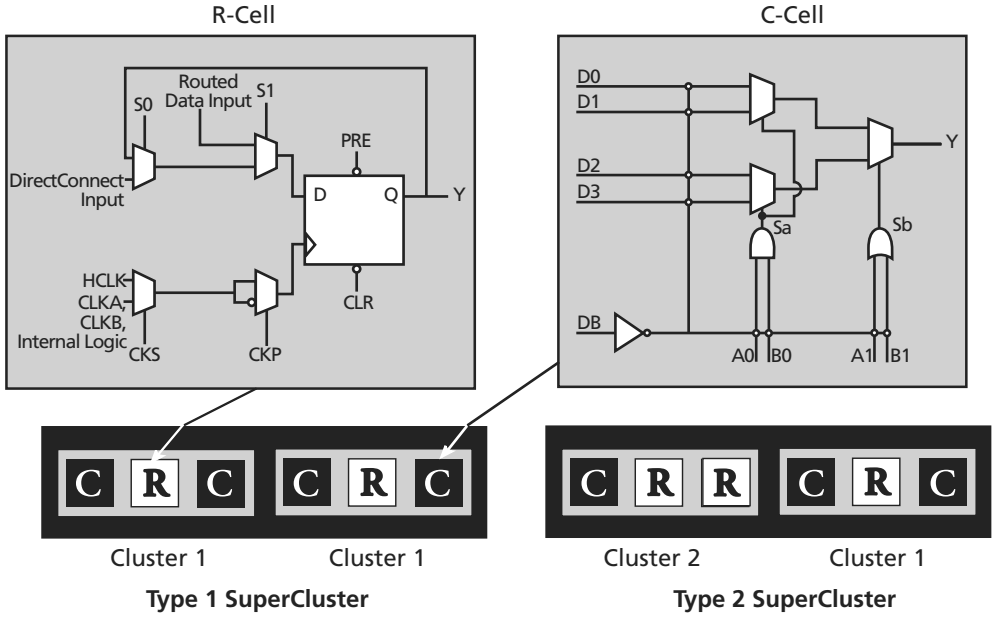


Figure 1-4 • **Cluster Organization**

Other Architectural Features

Technology

The Actel SX-A family is implemented on a high-voltage, twin-well CMOS process using 0.22 µ / 0.25 µ design rules. The metal-to-metal antifuse is comprised of a combination of amorphous silicon and dielectric material with barrier metals and has a programmed ('on' state) resistance of 25 Ω with capacitance of 1.0 fF for low signal impedance.

Performance

The unique architectural features of the SX-A family enable the devices to operate with internal clock frequencies of 350 MHz, causing very fast execution of even complex logic functions. The SX-A family is an optimal platform upon which to integrate the functionality previously contained in multiple complex programmable logic devices (CPLDs). In addition, designs that previously would have required a gate array to meet performance goals can be integrated into an SX-A device with dramatic improvements in cost and time-to-market. Using timing-driven place-and-route tools, designers can achieve highly deterministic device performance.

User Security

Reverse engineering is virtually impossible in SX-A devices because it is extremely difficult to distinguish between programmed and unprogrammed antifuses. In addition, since SX-A is a nonvolatile, single-chip solution, there is no configuration bitstream to intercept at device power-up.

The Actel FuseLock advantage ensures that unauthorized users will not be able to read back the contents of an Actel antifuse FPGA. In addition to the inherent strengths of the architecture, special security fuses that prevent internal probing and overwriting are hidden throughout the fabric of the device. They are located where they cannot be accessed or bypassed without destroying access to the rest of the device, making both invasive and more-subtle noninvasive attacks ineffective against Actel antifuse FPGAs.

Look for this symbol to ensure your valuable IP is secure (Figure 1-11).

Figure 1-11 • **FuseLock**

For more information, refer to Actel's *Implementation of Security in Actel Antifuse FPGAs* application note.

I/O Modules

For a simplified I/O schematic, refer to Figure 1 in the application note, *Actel eX, SX-A, and RTSX-S I/Os*.

Each user I/O on an SX-A device can be configured as an input, an output, a tristate output, or a bidirectional pin. Mixed I/O standards can be set for individual pins, though this is only allowed with the same voltage as the input. These I/Os, combined with array registers, can achieve clock-to-output-pad timing as fast as 3.8 ns, even without the dedicated I/O registers. In most FPGAs, I/O cells that have embedded latches and flip-flops, requiring instantiation in HDL code; this is a design complication not encountered in SX-A FPGAs. Fast pin-to-pin timing ensures that the device is able to interface with any other device in the system, which in turn enables parallel design of system components and reduces overall design time. All unused I/Os are configured as tristate outputs by the Actel Designer software, for maximum flexibility when designing new boards or migrating existing designs.

SX-A I/Os should be driven by high-speed push-pull devices with a low-resistance pull-up device when being configured as tristate output buffers. If the I/O is driven by a voltage level greater than V_{CCI} and a fast push-pull device is NOT used, the high-resistance pull-up of the driver and the internal circuitry of the SX-A I/O may create a voltage divider. This voltage divider could pull the input voltage below specification for some devices connected to the driver. A logic '1' may not be correctly presented in this case. For example, if an open drain driver is used with a pull-up resistor to 5 V to provide the logic '1' input, and V_{CCI} is set to 3.3 V on the SX-A device, the input signal may be pulled down by the SX-A input.

Each I/O module has an available power-up resistor of approximately 50 kΩ that can configure the I/O in a known state during power-up. For nominal pull-up and pull-down resistor values, refer to Table 1-4 on page 1-8 of the application note *Actel eX, SX-A, and RTSX-S I/Os*. Just slightly before V_{CCA} reaches 2.5 V, the resistors are disabled, so the I/Os will be controlled by user logic. See Table 1-2 on page 1-8 and Table 1-3 on page 1-8 for more information concerning available I/O features.

Boundary-Scan Testing (BST)

All SX-A devices are IEEE 1149.1 compliant and offer superior diagnostic and testing capabilities by providing Boundary Scan Testing (BST) and probing capabilities. The BST function is controlled through the special JTAG pins (TMS, TDI, TCK, TDO, and TRST). The functionality of the JTAG pins is defined by two available modes: Dedicated and Flexible. TMS cannot be employed as a user I/O in either mode.

Dedicated Mode

In Dedicated mode, all JTAG pins are reserved for BST; designers cannot use them as regular I/Os. An internal pull-up resistor is automatically enabled on both TMS and TDI pins, and the TMS pin will function as defined in the IEEE 1149.1 (JTAG) specification.

To select Dedicated mode, the user must reserve the JTAG pins in Actel's Designer software. Reserve the JTAG pins by checking the **Reserve JTAG** box in the Device Selection Wizard (Figure 1-12).

The default for the software is Flexible mode; all boxes are unchecked. Table 1-5 lists the definitions of the options in the Device Selection Wizard.

Figure 1-12 • **Device Selection Wizard**

Table 1-5 • **Reserve Pin Definitions**

Pin	Function
Reserve JTAG	Keeps pins from being used and changes the behavior of JTAG pins (no pull-up on TMS)
Reserve JTAG Test Reset	Regular I/O or JTAG reset with an internal pull-up
Reserve Probe	Keeps pins from being used or regular I/O

Flexible Mode

In Flexible mode, TDI, TCK, and TDO may be employed as either user I/Os or as JTAG input pins. The internal resistors on the TMS and TDI pins are not present in flexible JTAG mode.

To select the Flexible mode, uncheck the **Reserve JTAG** box in the Device Selection Wizard dialog in the Actel Designer software. In Flexible mode, TDI, TCK, and TDO pins may function as user I/Os or BST pins. The functionality is controlled by the BST Test Access Port (TAP) controller. The TAP controller receives two control inputs, TMS and TCK. Upon power-up, the TAP controller enters the Test-Logic-Reset state. In this state, TDI, TCK, and TDO function as user I/Os. The TDI, TCK, and TDO are transformed from user I/Os into BST pins when a rising edge on TCK is detected while TMS is at logic low. To return to Test-Logic Reset state, TMS must be high for at least five TCK cycles. **An external 10 k pull-up resistor to V_{CCI} should be placed on the TMS pin to pull it High by default.**

Table 1-6 describes the different configuration requirements of BST pins and their functionality in different modes.

Table 1-6 • **Boundary-Scan Pin Configurations and Functions**

Mode	Designer "Reserve JTAG" Selection	TAP Controller State
Dedicated (JTAG)	Checked	Any
Flexible (User I/O)	Unchecked	Test-Logic-Reset
Flexible (JTAG)	Unchecked	Any EXCEPT Test-Logic-Reset

TRST Pin

The TRST pin functions as a dedicated Boundary-Scan Reset pin when the **Reserve JTAG Test Reset** option is selected as shown in Figure 1-12. An internal pull-up resistor is permanently enabled on the TRST pin in this mode. Actel recommends connecting this pin to ground in normal operation to keep the JTAG state controller in the Test-Logic-Reset state. When JTAG is being used, it can be left floating or can be driven high.

When the **Reserve JTAG Test Reset** option is not selected, this pin will function as a regular I/O. If unused as an I/O in the design, it will be configured as a tristated output.

Related Documents

Application Notes

*Global Clock Networks in Actel's Antifuse Device*s
http://www.actel.com/documents/GlobalClk_AN.pdf
Using A54SX72A and RT54SX72S Quadrant Clocks
http://www.actel.com/documents/QCLK_AN.pdf
Implementation of Security in Actel Antifuse FPGAs
http://www.actel.com/documents/Antifuse_Security_AN.pdf
Actel eX, SX-A, and RTSX-S I/Os
http://www.actel.com/documents/AntifuseIO_AN.pdf
Actel SX-A and RT54SX-S Devices in Hot-Swap and Cold-Sparing Applications
http://www.actel.com/documents/HotSwapColdSparing_AN.pdf
Programming Antifuse Devices
http://www.actel.com/documents/AntifuseProgram_AN.pdf

Datasheets

HiRel SX-A Family FPGAs
http://www.actel.com/documents/HRSXA_DS.pdf
SX-A Automotive Family FPGAs
http://www.actel.com/documents/SXA_Auto_DS.pdf

User's Guides

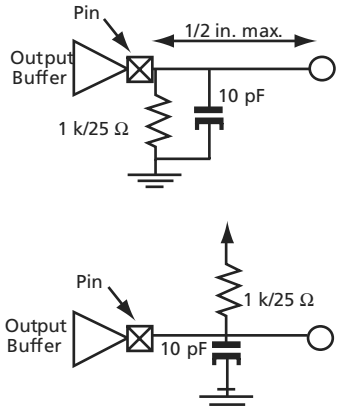
Silicon Sculptor User's Guide
http://www.actel.com/documents/SiliSculptII_Sculpt3_ug.pdf

Table 2-10 • **AC Specifications (3.3 V PCI Operation)**

Symbol	Parameter	Condition	Min.	Max.	Units
$I_{OH(AC)}$	Switching Current High	$0 < V_{OUT} \le 0.3V_{CCI}$ [1]	$-12V_{CCI}$	–	mA
		$0.3V_{CCI} \le V_{OUT} < 0.9V_{CCI}$ [1]	$(-17.1(V_{CCI} - V_{OUT}))$	–	mA
		$0.7V_{CCI} < V_{OUT} < V_{CCI}$ [1, 2]	–	EQ 2-3 on page 2-7	–
	(Test Point)	$V_{OUT} = 0.7V_{CC}$ [2]	–	$-32V_{CCI}$	mA
$I_{OL(AC)}$	Switching Current Low	$V_{CCI} > V_{OUT} \ge 0.6V_{CCI}$ [1]	$16V_{CCI}$	–	mA
		$0.6V_{CCI} > V_{OUT} > 0.1V_{CCI}$ [1]	$(26.7V_{OUT})$	–	mA
		$0.18V_{CCI} > V_{OUT} > 0$ [1, 2]	–	EQ 2-4 on page 2-7	–
	(Test Point)	$V_{OUT} = 0.18V_{CC}$ [2]	–	$38V_{CCI}$	mA
I_{CL}	Low Clamp Current	$-3 < V_{IN} \le -1$	$-25 + (V_{IN} + 1)/0.015$	–	mA
I_{CH}	High Clamp Current	$V_{CCI} + 4 > V_{IN} \ge V_{CCI} + 1$	$25 + (V_{IN} - V_{CCI} - 1)/0.015$	–	mA
$slew_R$	Output Rise Slew Rate	$0.2V_{CCI}$ - $0.6V_{CCI}$ load [3]	1	4	V/ns
$slew_F$	Output Fall Slew Rate	$0.6V_{CCI}$ - $0.2V_{CCI}$ load [3]	1	4	V/ns

Notes:

1. *Refer to the V/I curves in Figure 2-2 on page 2-7. Switching current characteristics for REQ# and GNT# are permitted to be one half of that specified here; i.e., half size output drivers may be used on these signals. This specification does not apply to CLK and RST#, which are system outputs. "Switching Current High" specifications are not relevant to SERR#, INTA#, INTB#, INTC#, and INTD#, which are open drain outputs.*
2. *Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (C and D) are provided with the respective diagrams in Figure 2-2 on page 2-7. The equation defined maximum should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.*
3. *This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per the latest revision of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is required (the maximum is no longer simply a guideline). Rise slew rate does not apply to open drain outputs.*



Guidelines for Estimating Power

The following guidelines are meant to represent worst-case scenarios; they can be generally used to predict the upper limits of power dissipation:

Logic Modules (m) = 20% of modules

Inputs Switching (n) = Number inputs/4

Outputs Switching (p) = Number of outputs/4

CLKA Loads (q1) = 20% of R-cells

CLKB Loads (q2) = 20% of R-cells

Load Capacitance (CL) = 35 pF

Average Logic Module Switching Rate (fm) = f/10

Average Input Switching Rate (fn) =f/5

Average Output Switching Rate (fp) = f/10

Average CLKA Rate (fq1) = f/2

Average CLKB Rate (fq2) = f/2

Average HCLK Rate (fs1) = f

HCLK loads (s1) = 20% of R-cells

To assist customers in estimating the power dissipations of their designs, Actel has published the *eX, SX-A and RT54SX-S Power Calculator* worksheet.

Thermal Characteristics

Introduction

The temperature variable in Actel Designer software refers to the junction temperature, not the ambient, case, or board temperatures. This is an important distinction because dynamic and static power consumption will cause the chip's junction to be higher than the ambient, case, or board temperatures. EQ 2-9 and EQ 2-10 give the relationship between thermal resistance, temperature gradient and power.

$$\theta_{JA} = \frac{T_J - T_A}{P}$$

EQ 2-9

$$\theta_{JA} = \frac{T_C - T_A}{P}$$

EQ 2-10

Where:

θ_{JA} = Junction-to-air thermal resistance

θ_{JC} = Junction-to-case thermal resistance

T_J = Junction temperature

T_A = Ambient temperature

T_C = Ambient temperature

P = total power dissipated by the device

Table 2-12 • **Package Thermal Characteristics**

Package Type	Pin Count	θ_{JC}	θ_{JA}			Units
			Still Air	1.0 m/s 200 ft./min.	2.5 m/s 500 ft./min.	
Thin Quad Flat Pack (TQFP)	100	14	33.5	27.4	25	°C/W
Thin Quad Flat Pack (TQFP)	144	11	33.5	28	25.7	°C/W
Thin Quad Flat Pack (TQFP)	176	11	24.7	19.9	18	°C/W
Plastic Quad Flat Pack (PQFP)[1]	208	8	26.1	22.5	20.8	°C/W
Plastic Quad Flat Pack (PQFP) with Heat Spreader[2]	208	3.8	16.2	13.3	11.9	°C/W
Plastic Ball Grid Array (PBGA)	329	3	17.1	13.8	12.8	°C/W
Fine Pitch Ball Grid Array (FBGA)	144	3.8	26.9	22.9	21.5	°C/W
Fine Pitch Ball Grid Array (FBGA)	256	3.8	26.6	22.8	21.5	°C/W
Fine Pitch Ball Grid Array (FBGA)	484	3.2	18	14.7	13.6	°C/W

Notes:

1. *The A54SX08A PQ208 has no heat spreader.*
2. *The SX-A PQ208 package has a heat spreader for A54SX16A, A54SX32A, and A54SX72A.*

To determine the heat sink's thermal performance, use the following equation:

$$\theta_{JA(TOTAL)} = \theta_{JC} + \theta_{CS} + \theta_{SA}$$

EQ 2-14

where:

θ_{CS} = 0.37°C/W
= thermal resistance of the interface material between the case and the heat sink, usually provided by the thermal interface manufacturer

θ_{SA} = thermal resistance of the heat sink in °C/W

$$\theta_{SA} = \theta_{JA(TOTAL)} - \theta_{JC} - \theta_{CS}$$

EQ 2-15

$$\theta_{SA} = 13.33°C/W - 3.20°C/W - 0.37°C/W$$

$$\theta_{SA} = 9.76°C/W$$

A heat sink with a thermal resistance of 9.76°C/W or better should be used. Thermal resistance of heat sinks is a function of airflow. The heat sink performance can be significantly improved with the presence of airflow.

Carefully estimating thermal resistance is important in the long-term reliability of an Actel FPGA. Design engineers should always correlate the power consumption of the device with the maximum allowable power dissipation of the package selected for that device, using the provided thermal resistance data.

Note: The values may vary depending on the application.

Table 2-14 • **A54SX08A Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCA} = 2.25 V, V_{CCI} = 3.0 V, T_J = 70°C)

Parameter	Description	–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t_{INYH}	Input Data Pad to Y High 5 V PCI		0.5		0.6		0.7		0.9	ns
t_{INYL}	Input Data Pad to Y Low 5 V PCI		0.8		0.9		1.1		1.5	ns
t_{INYH}	Input Data Pad to Y High 5 V TTL		0.5		0.6		0.7		0.9	ns
t_{INYL}	Input Data Pad to Y Low 5 V TTL		0.8		0.9		1.1		1.5	ns
Input Module Predicted Routing Delays[2]										
t_{IRD1}	FO = 1 Routing Delay		0.3		0.3		0.4		0.6	ns
t_{IRD2}	FO = 2 Routing Delay		0.5		0.5		0.6		0.8	ns
t_{IRD3}	FO = 3 Routing Delay		0.6		0.7		0.8		1.1	ns
t_{IRD4}	FO = 4 Routing Delay		0.8		0.9		1		1.4	ns
t_{IRD8}	FO = 8 Routing Delay		1.4		1.5		1.8		2.5	ns
t_{IRD12}	FO = 12 Routing Delay		2		2.2		2.6		3.6	ns

Notes:

1. *For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual performance.*

Table 2-18 • **A54SX08A Timing Characteristics (Worst-Case Commercial Conditions V_{CCA} = 2.25 V, V_{CCI} = 2.3 V, T_J = 70°C)**

Parameter	Description	–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
2.5 V LVCMOS Output Module Timing[1,2]										
t_{DLH}	Data-to-Pad Low to High		3.9		4.4		5.2		7.2	ns
t_{DHL}	Data-to-Pad High to Low		3.0		3.4		3.9		5.5	ns
t_{DHLS}	Data-to-Pad High to Low—low slew		13.3		15.1		17.7		24.8	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.8		3.2		3.7		5.2	ns
t_{ENZLS}	Data-to-Pad, Z to L—low slew		13.7		15.5		18.2		25.5	ns
t_{ENZH}	Enable-to-Pad, Z to H		3.9		4.4		5.2		7.2	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.5		2.8		3.3		4.7	ns
t_{ENHZ}	Enable-to-Pad, H to Z		3.0		3.4		3.9		5.5	ns
d_{TLH}[3]	Delta Low to High		0.037		0.043		0.051		0.071	ns/pF
d_{THL}[3]	Delta High to Low		0.017		0.023		0.023		0.037	ns/pF
d_{THLS}[3]	Delta High to Low—low slew		0.06		0.071		0.086		0.117	ns/pF

Note:

1. *Delays based on 35 pF loading.*
2. *The equivalent I/O Attribute Editor settings for 2.5 V LVCMOS is 2.5 V LVTTL in the software.*
3. *To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation:*
   Slew Rate [V/ns] = $(0.1*V_{CCI} – 0.9*V_{CCI})/ (C_{load} * d_{T[LH|HL|HLS]})$
   where C_{load} is the load capacitance driven by the I/O in pF
   $d_{T[LH|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.

Table 2-20 • **A54SX08A Timing Characteristics (Worst-Case Commercial Conditions V_{CCA} = 2.25 V, V_{CCI} = 4.75 V, T_J = 70°C)**

Parameter	Description	–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
5 V PCI Output Module Timing[1]										
t_{DLH}	Data-to-Pad Low to High		2.4		2.8		3.2		4.5	ns
t_{DHL}	Data-to-Pad High to Low		3.2		3.6		4.2		5.9	ns
t_{ENZL}	Enable-to-Pad, Z to L		1.5		1.7		2.0		2.8	ns
t_{ENZH}	Enable-to-Pad, Z to H		2.4		2.8		3.2		4.5	ns
t_{ENLZ}	Enable-to-Pad, L to Z		3.5		3.9		4.6		6.4	ns
t_{ENHZ}	Enable-to-Pad, H to Z		3.2		3.6		4.2		5.9	ns
d_{TLH}[2]	Delta Low to High		0.016		0.02		0.022		0.032	ns/pF
d_{THL}[2]	Delta High to Low		0.03		0.032		0.04		0.052	ns/pF
5 V TTL Output Module Timing[3]										
t_{DLH}	Data-to-Pad Low to High		2.4		2.8		3.2		4.5	ns
t_{DHL}	Data-to-Pad High to Low		3.2		3.6		4.2		5.9	ns
t_{DHLS}	Data-to-Pad High to Low—low slew		7.6		8.6		10.1		14.2	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.4		2.7		3.2		4.5	ns
t_{ENZLS}	Enable-to-Pad, Z to L—low slew		8.4		9.5		11.0		15.4	ns
t_{ENZH}	Enable-to-Pad, Z to H		2.4		2.8		3.2		4.5	ns
t_{ENLZ}	Enable-to-Pad, L to Z		4.2		4.7		5.6		7.8	ns
t_{ENHZ}	Enable-to-Pad, H to Z		3.2		3.6		4.2		5.9	ns
d_{TLH}	Delta Low to High		0.017		0.017		0.023		0.031	ns/pF
d_{THL}	Delta High to Low		0.029		0.031		0.037		0.051	ns/pF
d_{THLS}	Delta High to Low—low slew		0.046		0.057		0.066		0.089	ns/pF

Notes:

1. *Delays based on 50 pF loading.*
2. *To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation:*
   Slew Rate [V/ns] = $(0.1*V_{CCI} – 0.9*V_{CCI})$ / $(C_{load} * d_{T[LH|HL|HLS]})$
   where C_{load} is the load capacitance driven by the I/O in pF
   $d_{T[LH|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.
3. *Delays based on 35 pF loading.*

Table 2-22 • **A54SX16A Timing Characteristics (Worst-Case Commercial Conditions V_{CCA} = 2.25 V, V_{CCI} = 2.25 V, T_J = 70°C)**

Parameter	Description	–3 Speed*		–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Dedicated (Hardwired) Array Clock Networks												
t_{HCKH}	Input Low to High (Pad to R-cell Input)		1.2		1.4		1.6		1.8		2.8	ns
t_{HCKL}	Input High to Low (Pad to R-cell Input)		1.0		1.1		1.2		1.5		2.2	ns
t_{HPWH}	Minimum Pulse Width High	1.4		1.7		1.9		2.2		3.0		ns
t_{HPWL}	Minimum Pulse Width Low	1.4		1.7		1.9		2.2		3.0		ns
t_{HCKSW}	Maximum Skew		0.3		0.3		0.4		0.4		0.7	ns
t_{HP}	Minimum Period	2.8		3.4		3.8		4.4		6.0		ns
f_{HMAX}	Maximum Frequency		357		294		263		227		167	MHz
Routed Array Clock Networks												
t_{RCKH}	Input Low to High (Light Load) (Pad to R-cell Input)		1.0		1.2		1.3		1.6		2.2	ns
t_{RCKL}	Input High to Low (Light Load) (Pad to R-cell Input)		1.1		1.3		1.5		1.7		2.4	ns
t_{RCKH}	Input Low to High (50% Load) (Pad to R-cell Input)		1.1		1.3		1.5		1.7		2.4	ns
t_{RCKL}	Input High to Low (50% Load) (Pad to R-cell Input)		1.1		1.3		1.5		1.7		2.4	ns
t_{RCKH}	Input Low to High (100% Load) (Pad to R-cell Input)		1.3		1.5		1.7		2.0		2.8	ns
t_{RCKL}	Input High to Low (100% Load) (Pad to R-cell Input)		1.3		1.5		1.7		2.0		2.8	ns
t_{RPWH}	Minimum Pulse Width High	1.4		1.7		1.9		2.2		3.0		ns
t_{RPWL}	Minimum Pulse Width Low	1.4		1.7		1.9		2.2		3.0		ns
t_{RCKSW}	Maximum Skew (Light Load)		0.8		0.9		1.0		1.2		1.7	ns
t_{RCKSW}	Maximum Skew (50% Load)		0.8		0.9		1.0		1.2		1.7	ns
t_{RCKSW}	Maximum Skew (100% Load)		1.0		1.1		1.3		1.5		2.1	ns

Note: **All –3 speed grades have been discontinued.*

Table 2-25 • **A54SX16A Timing Characteristics (Worst-Case Commercial Conditions V_{CCA} = 2.25 V, V_{CCI} = 2.25 V, T_J = 70°C)**

Parameter	Description	–3 Speed[1]		–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
2.5 V LVCMOS Output Module Timing [2, 3]												
t_{DLH}	Data-to-Pad Low to High		3.4		3.9		4.5		5.2		7.3	ns
t_{DHL}	Data-to-Pad High to Low		2.6		3.0		3.3		3.9		5.5	ns
t_{DHLS}	Data-to-Pad High to Low—low slew		11.6		13.4		15.2		17.9		25.0	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.4		2.8		3.2		3.7		5.2	ns
t_{ENZLS}	Data-to-Pad, Z to L—low slew		11.8		13.7		15.5		18.2		25.5	ns
t_{ENZH}	Enable-to-Pad, Z to H		3.4		3.9		4.5		5.2		7.3	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.1		2.5		2.8		3.3		4.7	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.6		3.0		3.3		3.9		5.5	ns
d_{TLH}[4]	Delta Low to High		0.031		0.037		0.043		0.051		0.071	ns/pF
d_{THL}[4]	Delta High to Low		0.017		0.017		0.023		0.023		0.037	ns/pF
d_{THLS}[4]	Delta High to Low—low slew		0.057		0.06		0.071		0.086		0.117	ns/pF

Note:

1. *All –3 speed grades have been discontinued.*
2. *Delays based on 35 pF loading.*
3. *The equivalent IO Attribute settings for 2.5 V LVCMOS is 2.5 V LVTTL in the software.*
4. *To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation:*
   Slew Rate [V/ns] = $(0.1*V_{CCI} – 0.9*V_{CCI})/ (C_{load} * d_{T[LH|HL|HLS]})$
   where C_{load} is the load capacitance driven by the I/O in pF
   $d_{T[LH|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.

Table 2-28 • **A54SX32A Timing Characteristics (Worst-Case Commercial Conditions, V_{CCA} = 2.25 V, V_{CCI} = 3.0 V, T_J = 70°C)**

Parameter	Description	–3 Speed[1]		–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
C-Cell Propagation Delays[2]												
t_{PD}	Internal Array Module		0.8		0.9		1.1		1.2		1.7	ns
Predicted Routing Delays[3]												
t_{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1		0.1	ns
t_{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.3		0.3		0.4		0.6	ns
t_{RD1}	FO = 1 Routing Delay		0.3		0.3		0.4		0.5		0.6	ns
t_{RD2}	FO = 2 Routing Delay		0.4		0.5		0.5		0.6		0.8	ns
t_{RD3}	FO = 3 Routing Delay		0.5		0.6		0.7		0.8		1.1	ns
t_{RD4}	FO = 4 Routing Delay		0.7		0.8		0.9		1.0		1.4	ns
t_{RD8}	FO = 8 Routing Delay		1.2		1.4		1.5		1.8		2.5	ns
t_{RD12}	FO = 12 Routing Delay		1.7		2.0		2.2		2.6		3.6	ns
R-Cell Timing												
t_{RCO}	Sequential Clock-to-Q		0.6		0.7		0.8		0.9		1.3	ns
t_{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.6		0.8		1.0	ns
t_{PRESET}	Asynchronous Preset-to-Q		0.6		0.7		0.7		0.9		1.2	ns
t_{SUD}	Flip-Flop Data Input Set-Up	0.6		0.7		0.8		0.9		1.2		ns
t_{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		0.0		ns
t_{WASYN}	Asynchronous Pulse Width	1.2		1.4		1.5		1.8		2.5		ns
$t_{RECASYN}$	Asynchronous Recovery Time	0.3		0.4		0.4		0.5		0.7		ns
t_{HASYN}	Asynchronous Removal Time	0.3		0.3		0.3		0.4		0.6		ns
t_{MPW}	Clock Pulse Width	1.4		1.6		1.8		2.1		2.9		ns
Input Module Propagation Delays												
t_{INYH}	Input Data Pad to Y High 2.5 V LVCMOS		0.6		0.7		0.8		0.9		1.2	ns
t_{INYL}	Input Data Pad to Y Low 2.5 V LVCMOS		1.2		1.3		1.5		1.8		2.5	ns
t_{INYH}	Input Data Pad to Y High 3.3 V PCI		0.5		0.6		0.6		0.7		1.0	ns
t_{INYL}	Input Data Pad to Y Low 3.3 V PCI		0.6		0.7		0.8		0.9		1.3	ns
t_{INYH}	Input Data Pad to Y High 3.3 V LVTTL		0.8		0.9		1.0		1.2		1.6	ns
t_{INYL}	Input Data Pad to Y Low 3.3 V LVTTL		1.4		1.6		1.8		2.2		3.0	ns

Notes:

1. *All –3 speed grades have been discontinued.*
2. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn} , t_{RCO} + t_{RD1} + t_{PDn} , or t_{PD1} + t_{RD1} + t_{SUD} , whichever is appropriate.*
3. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual performance.*

Table 2-32 • **A54SX32A Timing Characteristics (Worst-Case Commercial Conditions V_{CCA} = 2.25 V, V_{CCI} = 2.3 V, T_J = 70°C)**

Parameter	Description	–3 Speed[1]		–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
2.5 V LVCMOS Output Module Timing [2,3]												
t_{DLH}	Data-to-Pad Low to High		3.3		3.8		4.2		5.0		7.0	ns
t_{DHL}	Data-to-Pad High to Low		2.5		2.9		3.2		3.8		5.3	ns
t_{DHLS}	Data-to-Pad High to Low—low slew		11.1		12.8		14.5		17.0		23.8	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.4		2.8		3.2		3.7		5.2	ns
t_{ENZLS}	Data-to-Pad, Z to L—low slew		11.8		13.7		15.5		18.2		25.5	ns
t_{ENZH}	Enable-to-Pad, Z to H		3.3		3.8		4.2		5.0		7.0	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.1		2.5		2.8		3.3		4.7	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.5		2.9		3.2		3.8		5.3	ns
d_{TLH}[4]	Delta Low to High		0.031		0.037		0.043		0.051		0.071	ns/pF
d_{THL}[4]	Delta High to Low		0.017		0.017		0.023		0.023		0.037	ns/pF
d_{THLS}[4]	Delta High to Low—low slew		0.057		0.06		0.071		0.086		0.117	ns/pF

Note:

1. *All –3 speed grades have been discontinued.*
2. *Delays based on 35 pF loading.*
3. *The equivalent IO Attribute settings for 2.5 V LVCMOS is 2.5 V LVTTL in the software.*
4. *To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation:*
   Slew Rate [V/ns] = $(0.1*V_{CCI} – 0.9*V_{CCI})/ (C_{load} * d_{T[LH|HL|HLS]})$
   where C_{load} is the load capacitance driven by the I/O in pF
   $d_{T[LH|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.

Table 2-34 • **A54SX32A Timing Characteristics (Worst-Case Commercial Conditions V_{CCA} = 2.25 V, V_{CCI} = 4.75 V, T_J = 70°C)**

Parameter	Description	–3 Speed[1]		–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
5 V PCI Output Module Timing[2]												
t_{DLH}	Data-to-Pad Low to High		2.1		2.4		2.8		3.2		4.5	ns
t_{DHL}	Data-to-Pad High to Low		2.8		3.2		3.6		4.2		5.9	ns
t_{ENZL}	Enable-to-Pad, Z to L		1.3		1.5		1.7		2.0		2.8	ns
t_{ENZH}	Enable-to-Pad, Z to H		2.1		2.4		2.8		3.2		4.5	ns
t_{ENLZ}	Enable-to-Pad, L to Z		3.0		3.5		3.9		4.6		6.4	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.8		3.2		3.6		4.2		5.9	ns
d_{TLH}[3]	Delta Low to High		0.016		0.016		0.02		0.022		0.032	ns/pF
d_{THL}[3]	Delta High to Low		0.026		0.03		0.032		0.04		0.052	ns/pF
5 V TTL Output Module Timing[4]												
t_{DLH}	Data-to-Pad Low to High		1.9		2.2		2.5		2.9		4.1	ns
t_{DHL}	Data-to-Pad High to Low		2.5		2.9		3.3		3.9		5.4	ns
t_{DHLS}	Data-to-Pad High to Low—low slew		6.6		7.6		8.6		10.1		14.2	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.1		2.4		2.7		3.2		4.5	ns
t_{ENZLS}	Enable-to-Pad, Z to L—low slew		7.4		8.4		9.5		11.0		15.4	ns
t_{ENZH}	Enable-to-Pad, Z to H		1.9		2.2		2.5		2.9		4.1	ns
t_{ENLZ}	Enable-to-Pad, L to Z		3.6		4.2		4.7		5.6		7.8	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.5		2.9		3.3		3.9		5.4	ns
d_{TLH}[3]	Delta Low to High		0.014		0.017		0.017		0.023		0.031	ns/pF
d_{THL}[3]	Delta High to Low		0.023		0.029		0.031		0.037		0.051	ns/pF
d_{THLS}[3]	Delta High to Low—low slew		0.043		0.046		0.057		0.066		0.089	ns/pF

Notes:

1. *All –3 speed grades have been discontinued.*
2. *Delays based on 50 pF loading.*
3. *To obtain the slew rate, substitute the appropriate Delta value, load capacitance, and the V_{CCI} value into the following equation:*
   Slew Rate [V/ns] = $(0.1*V_{CCI} – 0.9*V_{CCI})/ (C_{load} * d_{T[LH|HL|HLS]})$
   where C_{load} is the load capacitance driven by the I/O in pF
   $d_{T[LH|HL|HLS]}$ is the worst case delta value from the datasheet in ns/pF.
4. *Delays based on 35 pF loading.*

Table 2-35 • **A54SX72A Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCA} = 2.25 V, V_{CCI} = 3.0 V, T_J = 70°C)

Parameter	Description	–3 Speed[1]		–2 Speed		–1 Speed		Std. Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t_{INYH}	Input Data Pad to Y High 5 V PCI		0.5		0.6		0.7		0.8		1.1	ns
t_{INYL}	Input Data Pad to Y Low 5 V PCI		0.8		0.9		1.0		1.2		1.6	ns
t_{INYH}	Input Data Pad to Y High 5 V TTL		0.7		0.8		0.9		1.0		1.4	ns
t_{INYL}	Input Data Pad to Y Low 5 V TTL		0.9		1.1		1.2		1.4		1.9	ns
Input Module Predicted Routing Delays[3]												
t_{IRD1}	FO = 1 Routing Delay		0.3		0.3		0.4		0.5		0.7	ns
t_{IRD2}	FO = 2 Routing Delay		0.4		0.5		0.6		0.7		1	ns
t_{IRD3}	FO = 3 Routing Delay		0.5		0.7		0.8		0.9		1.3	ns
t_{IRD4}	FO = 4 Routing Delay		0.7		0.9		1		1.1		1.5	ns
t_{IRD8}	FO = 8 Routing Delay		1.2		1.5		1.7		2.1		2.9	ns
t_{IRD12}	FO = 12 Routing Delay		1.7		2.2		2.5		3		4.2	ns

Notes:

1. *All –3 speed grades have been discontinued.*
2. *For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.*
3. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual performance.*

329-Pin PBGA	
Pin Number	**A54SX32A Function**
D11	V_{CCA}
D12	NC
D13	I/O
D14	I/O
D15	I/O
D16	I/O
D17	I/O
D18	I/O
D19	I/O
D20	I/O
D21	I/O
D22	I/O
D23	I/O
E1	V_{CCI}
E2	I/O
E3	I/O
E4	I/O
E20	I/O
E21	I/O
E22	I/O
E23	I/O
F1	I/O
F2	TMS
F3	I/O
F4	I/O
F20	I/O
F21	I/O
F22	I/O
F23	I/O
G1	I/O
G2	I/O
G3	I/O
G4	I/O
G20	I/O
G21	I/O
G22	I/O
G23	GND

329-Pin PBGA	
Pin Number	**A54SX32A Function**
H1	I/O
H2	I/O
H3	I/O
H4	I/O
H20	V_{CCA}
H21	I/O
H22	I/O
H23	I/O
J1	NC
J2	I/O
J3	I/O
J4	I/O
J20	I/O
J21	I/O
J22	I/O
J23	I/O
K1	I/O
K2	I/O
K3	I/O
K4	I/O
K10	GND
K11	GND
K12	GND
K13	GND
K14	GND
K20	I/O
K21	I/O
K22	I/O
K23	I/O
L1	I/O
L2	I/O
L3	I/O
L4	NC
L10	GND
L11	GND
L12	GND
L13	GND

329-Pin PBGA	
Pin Number	**A54SX32A Function**
L14	GND
L20	NC
L21	I/O
L22	I/O
L23	NC
M1	I/O
M2	I/O
M3	I/O
M4	V_{CCA}
M10	GND
M11	GND
M12	GND
M13	GND
M14	GND
M20	V_{CCA}
M21	I/O
M22	I/O
M23	V_{CCI}
N1	I/O
N2	TRST, I/O
N3	I/O
N4	I/O
N10	GND
N11	GND
N12	GND
N13	GND
N14	GND
N20	NC
N21	I/O
N22	I/O
N23	I/O
P1	I/O
P2	I/O
P3	I/O
P4	I/O
P10	GND
P11	GND

329-Pin PBGA	
Pin Number	**A54SX32A Function**
P12	GND
P13	GND
P14	GND
P20	I/O
P21	I/O
P22	I/O
P23	I/O
R1	I/O
R2	I/O
R3	I/O
R4	I/O
R20	I/O
R21	I/O
R22	I/O
R23	I/O
T1	I/O
T2	I/O
T3	I/O
T4	I/O
T20	I/O
T21	I/O
T22	I/O
T23	I/O
U1	I/O
U2	I/O
U3	V_{CCA}
U4	I/O
U20	I/O
U21	V_{CCA}
U22	I/O
U23	I/O
V1	V_{CCI}
V2	I/O
V3	I/O
V4	I/O
V20	I/O
V21	I/O

484-Pin FBGA

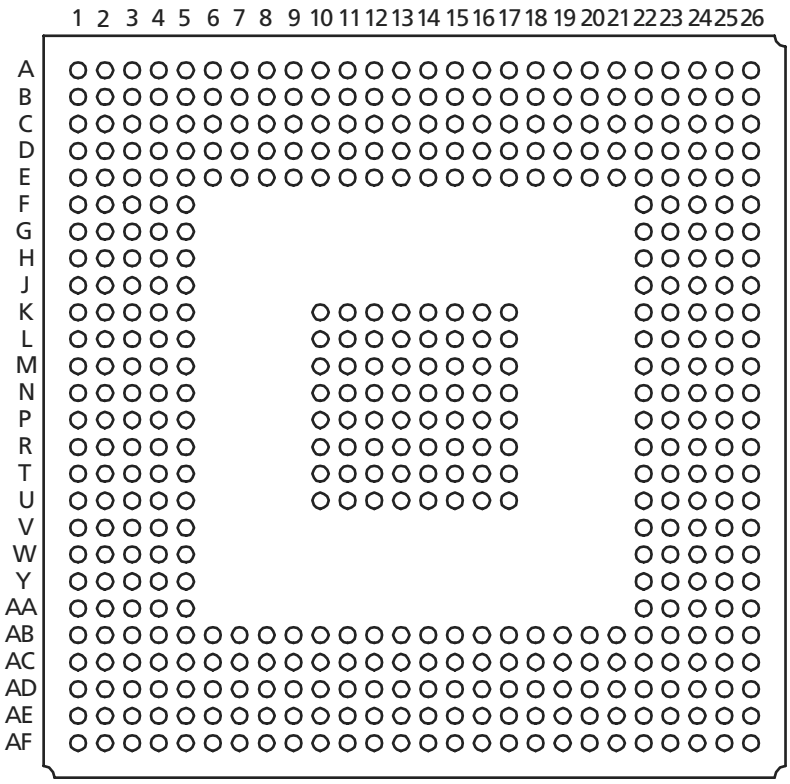


Figure 3-8 • **484-Pin FBGA (Top View)**

Note

For Package Manufacturing and Environmental information, visit Resource center at http://www.actel.com/products/rescenter/package/index.html.