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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I ² C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	5
Program Memory Size	7KB (4K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	2.3V ~ 5.5V
Data Converters	A/D 4x10b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	8-SOIC (0.154", 3.90mm Width)
Supplier Device Package	8-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic12f1840-e-sn

TABLE 3-3: PIC12(L)F1840 MEMORY MAP (CONTINUED)

BANK 8		BANK 9		BANK 10		BANK 11		BANK 12		BANK 13		BANK 14		BANK 15	
400h	Core Registers (Table 3-2)	480h	Core Registers (Table 3-2)	500h	Core Registers (Table 3-2)	580h	Core Registers (Table 3-2)	600h	Core Registers (Table 3-2)	680h	Core Registers (Table 3-2)	700h	Core Registers (Table 3-2)	780h	Core Registers (Table 3-2)
40Bh	Unimplemented Read as '0'	48Bh	Unimplemented Read as '0'	50Bh	Unimplemented Read as '0'	58Bh	Unimplemented Read as '0'	60Bh	Unimplemented Read as '0'	68Bh	Unimplemented Read as '0'	70Bh	Unimplemented Read as '0'	78Bh	Unimplemented Read as '0'
40Ch		48Ch		50Ch		58Ch		60Ch		68Ch		70Ch		78Ch	
46Fh	Common RAM (Accesses 70h – 7Fh)	4EFh	Common RAM (Accesses 70h – 7Fh)	56Fh	Common RAM (Accesses 70h – 7Fh)	5EFh	Common RAM (Accesses 70h – 7Fh)	66Fh	Common RAM (Accesses 70h – 7Fh)	6EFh	Common RAM (Accesses 70h – 7Fh)	76Fh	Common RAM (Accesses 70h – 7Fh)	7EFh	Common RAM (Accesses 70h – 7Fh)
470h		4F0h		570h		5F0h		670h		6F0h		770h		7F0h	
47Fh		4FFh		57Fh		5FFh		67Fh		6FFh		77Fh		7FFh	
BANK 16		BANK 17		BANK 18		BANK 19		BANK 20		BANK 21		BANK 22		BANK 23	
800h	Core Registers (Table 3-2)	880h	Core Registers (Table 3-2)	900h	Core Registers (Table 3-2)	980h	Core Registers (Table 3-2)	A00h	Core Registers (Table 3-2)	A80h	Core Registers (Table 3-2)	B00h	Core Registers (Table 3-2)	B80h	Core Registers (Table 3-2)
80Bh	Unimplemented Read as '0'	88Bh	Unimplemented Read as '0'	90Bh	Unimplemented Read as '0'	98Bh	Unimplemented Read as '0'	A0Bh	Unimplemented Read as '0'	A8Bh	Unimplemented Read as '0'	B0Bh	Unimplemented Read as '0'	B8Bh	Unimplemented Read as '0'
80Ch		88Ch		90Ch		98Ch		A0Ch		A8Ch		B0Ch		B8Ch	
86Fh	Common RAM (Accesses 70h – 7Fh)	8EFh	Common RAM (Accesses 70h – 7Fh)	96Fh	Common RAM (Accesses 70h – 7Fh)	9EFh	Common RAM (Accesses 70h – 7Fh)	A6Fh	Common RAM (Accesses 70h – 7Fh)	AEFh	Common RAM (Accesses 70h – 7Fh)	B6Fh	Common RAM (Accesses 70h – 7Fh)	BEFh	Common RAM (Accesses 70h – 7Fh)
870h		8F0h		970h		9F0h		A70h		AF0h		B70h		BF0h	
87Fh		8FFh		97Fh		9FFh		A7Fh		AFFh		B7Fh		BFFh	
BANK 24		BANK 25		BANK 26		BANK 27		BANK 28		BANK 29		BANK 30			
C00h	Core Registers (Table 3-2)	C80h	Core Registers (Table 3-2)	D00h	Core Registers (Table 3-2)	D80h	Core Registers (Table 3-2)	E00h	Core Registers (Table 3-2)	E80h	Core Registers (Table 3-2)	F00h	Core Registers (Table 3-2)		
C0Bh	Unimplemented Read as '0'	C8Bh	Unimplemented Read as '0'	D0Bh	Unimplemented Read as '0'	D8Bh	Unimplemented Read as '0'	E0Bh	Unimplemented Read as '0'	E8Bh	Unimplemented Read as '0'	F0Bh	Unimplemented Read as '0'		
C0Ch		C8Ch		D0Ch		D8Ch		E0Ch		E8Ch		F0Ch			
C6Fh	Common RAM (Accesses 70h – 7Fh)	CEFh	Common RAM (Accesses 70h – 7Fh)	D6Fh	Common RAM (Accesses 70h – 7Fh)	DEFh	Common RAM (Accesses 70h – 7Fh)	E6Fh	Common RAM (Accesses 70h – 7Fh)	EEFh	Common RAM (Accesses 70h – 7Fh)	F6Fh	Common RAM (Accesses 70h – 7Fh)		
C70h		CF0h		D70h		DF0h		E70h		EF0h		F70h			
C7Fh		CFh		D7Fh		DFh		E7Fh		EFh		F7Fh			

Legend: = Unimplemented data memory locations, read as '0'

5.6 Register Definitions: Oscillator Control

REGISTER 5-1: OSCCON: OSCILLATOR CONTROL REGISTER

R/W-0/0	R/W-0/0	R/W-1/1	R/W-1/1	R/W-1/1	U-0	R/W-0/0	R/W-0/0
SPLLEN	IRCF<3:0>			—	SCS<1:0>		
bit 7						bit 0	

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

- bit 7 **SPLLEN:** Software PLL Enable bit
If PLEN in Configuration Words = 1:
 SPLLEN bit is ignored. 4x PLL is always enabled (subject to oscillator requirements)
If PLEN in Configuration Words = 0:
 1 = 4x PLL is enabled
 0 = 4x PLL is disabled
- bit 6-3 **IRCF<3:0>:** Internal Oscillator Frequency Select bits
 1111 = 16 MHz HF
 1110 = 8 MHz or 32 MHz HF (see **Section 5.2.2.1 "HFINTOSC"**)
 1101 = 4 MHz HF
 1100 = 2 MHz HF
 1011 = 1 MHz HF
 1010 = 500 kHz HF⁽¹⁾
 1001 = 250 kHz HF⁽¹⁾
 1000 = 125 kHz HF⁽¹⁾
 0111 = 500 kHz MF (default upon Reset)
 0110 = 250 kHz MF
 0101 = 125 kHz MF
 0100 = 62.5 kHz MF
 0011 = 31.25 kHz HF⁽¹⁾
 0010 = 31.25 kHz MF
 000x = 31 kHz LF
- bit 2 **Unimplemented:** Read as '0'
- bit 1-0 **SCS<1:0>:** System Clock Select bits
 1x = Internal oscillator block
 01 = Timer1 oscillator
 00 = Clock determined by FOSC<2:0> in Configuration Words.

Note 1: Duplicate frequency derived from HFINTOSC.

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6.0 REFERENCE CLOCK MODULE

The reference clock module provides the ability to send a divided clock to the clock output pin of the device (CLKR) and provide a secondary internal clock source to the modulator module. This module is available in all oscillator configurations and allows the user to select a greater range of clock sub-multiples to drive external devices in the application. The reference clock module includes the following features:

- System clock is the source
- Available in all oscillator configurations
- Programmable clock divider
- Output enable to a port pin
- Selectable duty cycle
- Slew rate control

The reference clock module is controlled by the CLKRCON register (Register 6-1) and is enabled when setting the CLKREN bit. To output the divided clock signal to the CLKR port pin, the CLKROE bit must be set. The CLKRDIV<2:0> bits enable the selection of eight different clock divider options. The CLKRDC<1:0> bits can be used to modify the duty cycle of the output clock⁽¹⁾. The CLKRSLR bit controls slew rate limiting.

Note 1: If the base clock rate is selected without a divider, the output clock will always have a duty cycle equal to that of the source clock, unless a 0% duty cycle is selected. If the clock divider is set to base clock/2, then 25% and 75% duty cycle accuracy will be dependent upon the source clock.

For information on using the reference clock output with the modulator module, see **Section 23.0 “Data Signal Modulator”**.

6.1 Slew Rate

The slew rate limitation on the output port pin can be disabled. The slew rate limitation is removed by clearing the CLKRSLR bit in the CLKRCON register.

6.2 Effects of a Reset

Upon any device Reset, the reference clock module is disabled. The user’s firmware is responsible for initializing the module before enabling the output. The registers are reset to their default values.

6.3 Conflicts with the CLKR Pin

There are two cases when the reference clock output signal cannot be output to the CLKR pin, if:

- LP, XT or HS Oscillator mode is selected.
- CLKOUT function is enabled.

Even if either of these cases are true, the module can still be enabled and the reference clock signal may be used in conjunction with the modulator module.

6.3.1 OSCILLATOR MODES

If LP, XT or HS oscillator modes are selected, the OSC2/CLKR pin must be used as an oscillator input pin and the CLKR output cannot be enabled. See **Section 5.2 “Clock Source Types”** for more information on different oscillator modes.

6.3.2 CLKOUT FUNCTION

The CLKOUT function has a higher priority than the reference clock module. Therefore, if the CLKOUT function is enabled by the CLKOUTEN bit in Configuration Words, Fosc/4 will always be output on the port pin. Reference **Section 4.0 “Device Configuration”** for more information.

6.4 Operation During Sleep

As the reference clock module relies on the system clock as its source, and the system clock is disabled in Sleep, the module does not function in Sleep, even if an external clock source or the Timer1 clock source is configured as the system clock. The module outputs will remain in their current state until the device exits Sleep.

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TABLE 6-1: SUMMARY OF REGISTERS ASSOCIATED WITH REFERENCE CLOCK SOURCES

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
CLKRCON	CLKREN	CLKROE	CLKRSLR	CLKRDC<1:0>		CLKRDIV<2:0>			57

Legend: — = unimplemented locations read as '0'. Shaded cells are not used by reference clock sources.

TABLE 6-2: SUMMARY OF CONFIGURATION WORD WITH REFERENCE CLOCK SOURCES

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page
CONFIG1	13:8	—	—	FCMEN	IESO	CLKOUTEN	BOREN<1:0>	CPD	33	
	7:0	CP	MCLRE	PWRTE	WDTE<1:0>	FOSC<2:0>				

Legend: — = unimplemented locations read as '0'. Shaded cells are not used by reference clock sources.

19.2 Comparator Control

The comparator has two control registers: CM1CON0 and CM1CON1.

The CM1CON0 register (see Register 19-1) contains Control and Status bits for the following:

- Enable
- Output selection
- Output polarity
- Speed/Power selection
- Hysteresis enable
- Output synchronization

The CM1CON1 register (see Register 19-2) contains Control bits for the following:

- Interrupt enable
- Interrupt edge polarity
- Positive input channel selection
- Negative input channel selection

19.2.1 COMPARATOR ENABLE

Setting the C1ON bit of the CM1CON0 register enables the comparator for operation. Clearing the C1ON bit disables the comparator resulting in minimum current consumption.

19.2.2 COMPARATOR OUTPUT SELECTION

The output of the comparator can be monitored by reading either the C1OUT bit of the CM1CON0 register or the MC1OUT bit of the CMOUT register. In order to make the output available for an external connection, the following conditions must be true:

- C1OE bit of the CM1CON0 register must be set
- Corresponding TRIS bit must be cleared
- C1ON bit of the CM1CON0 register must be set

Note 1: The C1OE bit of the CM1CON0 register overrides the PORT data latch. Setting the C1ON bit of the CM1CON0 register has no impact on the port override.

2: The internal output of the comparator is latched with each instruction cycle. Unless otherwise specified, external outputs are not latched.

19.2.3 COMPARATOR OUTPUT POLARITY

Inverting the output of the comparator is functionally equivalent to swapping the comparator inputs. The polarity of the comparator output can be inverted by setting the C1POL bit of the CM1CON0 register. Clearing the C1POL bit results in a non-inverted output.

Table 19-2 shows the output state versus input conditions, including polarity control.

TABLE 19-2: COMPARATOR OUTPUT STATE VS. INPUT CONDITIONS

Input Condition	C1POL	C1OUT
$C1V_N > C1V_P$	0	0
$C1V_N < C1V_P$	0	1
$C1V_N > C1V_P$	1	1
$C1V_N < C1V_P$	1	0

19.2.4 COMPARATOR SPEED/POWER SELECTION

The trade-off between speed or power can be optimized during program execution with the C1SP control bit. The default state for this bit is '1' which selects the normal speed mode. Device power consumption can be optimized at the cost of slower comparator propagation delay by clearing the C1SP bit to '0'.

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22.1 Timer2 Operation

The clock input to the Timer2 modules is the system instruction clock ($F_{osc}/4$).

TMR2 increments from 00h on each clock edge.

A 4-bit counter/prescaler on the clock input allows direct input, divide-by-4 and divide-by-16 prescale options. These options are selected by the prescaler control bits, T2CKPS<1:0> of the T2CON register. The value of TMR2 is compared to that of the Period register, PR2, on each clock cycle. When the two values match, the comparator generates a match signal as the timer output. This signal also resets the value of TMR2 to 00h on the next cycle and drives the output counter/postscaler (see **Section 22.2 “Timer2 Interrupt”**).

The TMR2 and PR2 registers are both directly readable and writable. The TMR2 register is cleared on any device Reset, whereas the PR2 register initializes to FFh. Both the prescaler and postscaler counters are cleared on the following events:

- a write to the TMR2 register
- a write to the T2CON register
- Power-on Reset (POR)
- Brown-out Reset (BOR)
- MCLR Reset
- Watchdog Timer (WDT) Reset
- Stack Overflow Reset
- Stack Underflow Reset
- RESET Instruction

Note: TMR2 is not cleared when T2CON is written.

22.2 Timer2 Interrupt

Timer2 can also generate an optional device interrupt. The Timer2 output signal (TMR2-to-PR2 match) provides the input for the 4-bit counter/postscaler. This counter generates the TMR2 match interrupt flag which is latched in TMR2IF of the PIR1 register. The interrupt is enabled by setting the TMR2 Match Interrupt Enable bit, TMR2IE of the PIE1 register.

A range of 16 postscale options (from 1:1 through 1:16 inclusive) can be selected with the postscaler control bits, T2OUTPS<3:0>, of the T2CON register.

22.3 Timer2 Output

The unscaled output of TMR2 is available primarily to the CCP1 module, where it is used as a time base for operations in PWM mode.

Timer2 can be optionally used as the shift clock source for the MSSP1 module operating in SPI mode. Additional information is provided in **Section 25.1 “Master SSP (MSSP1) Module Overview”**

22.4 Timer2 Operation During Sleep

Timer2 cannot be operated while the processor is in Sleep mode. The contents of the TMR2 and PR2 registers will remain unchanged while the processor is in Sleep mode.

22.5 Register Definitions: Timer2 Control

REGISTER 22-1: T2CON: TIMER2 CONTROL REGISTER

U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	T2OUTPS<3:0>			TMR2ON	T2CKPS<1:0>		
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

bit 7 **Unimplemented:** Read as '0'

bit 6-3 **T2OUTPS<3:0>:** Timer Output Postscaler Select bits

1111 = 1:16 Postscaler
 1110 = 1:15 Postscaler
 1101 = 1:14 Postscaler
 1100 = 1:13 Postscaler
 1011 = 1:12 Postscaler
 1010 = 1:11 Postscaler
 1001 = 1:10 Postscaler
 1000 = 1:9 Postscaler
 0111 = 1:8 Postscaler
 0110 = 1:7 Postscaler
 0101 = 1:6 Postscaler
 0100 = 1:5 Postscaler
 0011 = 1:4 Postscaler
 0010 = 1:3 Postscaler
 0001 = 1:2 Postscaler
 0000 = 1:1 Postscaler

bit 2 **TMR2ON:** Timer2 On bit

1 = Timer2 is on
 0 = Timer2 is off

bit 1-0 **T2CKPS<1:0>:** Timer2 Clock Prescale Select bits

11 = Prescaler is 64
 10 = Prescaler is 16
 01 = Prescaler is 4
 00 = Prescaler is 1

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24.4 PWM (Enhanced Mode)

The enhanced PWM mode generates a Pulse-Width Modulation (PWM) signal on up to two different output pins with up to ten bits of resolution. The period, duty cycle, and resolution are controlled by the following registers:

- PR2 registers
- T2CON registers
- CCPR1L registers
- CCP1CON registers

The ECCP modules have the following additional PWM registers which control Auto-shutdown, Auto-restart, Dead-band Delay and PWM Steering modes:

- CCP1AS registers
- PSTR1CON registers
- PWM1CON registers

The enhanced PWM module can generate the following three PWM Output modes:

- Single PWM
- Half-Bridge PWM
- Single PWM with PWM Steering Mode

To select an Enhanced PWM Output mode, the P1M bits of the CCP1CON register must be configured appropriately.

The PWM outputs are multiplexed with I/O pins and are designated P1A and P1B. The polarity of the PWM pins is configurable and is selected by setting the bits CCP1M<3:0> in the CCP1CON register appropriately.

Figure 24-5 shows an example of a simplified block diagram of the Enhanced PWM module.

Table 24-8 shows the pin assignments for various Enhanced PWM modes.

Note 1: The corresponding TRIS bit must be cleared to enable the PWM output on the CCP1 pin.

2: Clearing the CCP1CON register will relinquish control of the CCP1 pin.

3: Any pin not used in the enhanced PWM mode is available for alternate pin functions, if applicable.

4: To prevent the generation of an incomplete waveform when the PWM is first enabled, the ECCP module waits until the start of a new PWM period before generating a PWM signal.

FIGURE 24-5: EXAMPLE SIMPLIFIED BLOCK DIAGRAM OF THE ENHANCED PWM MODE

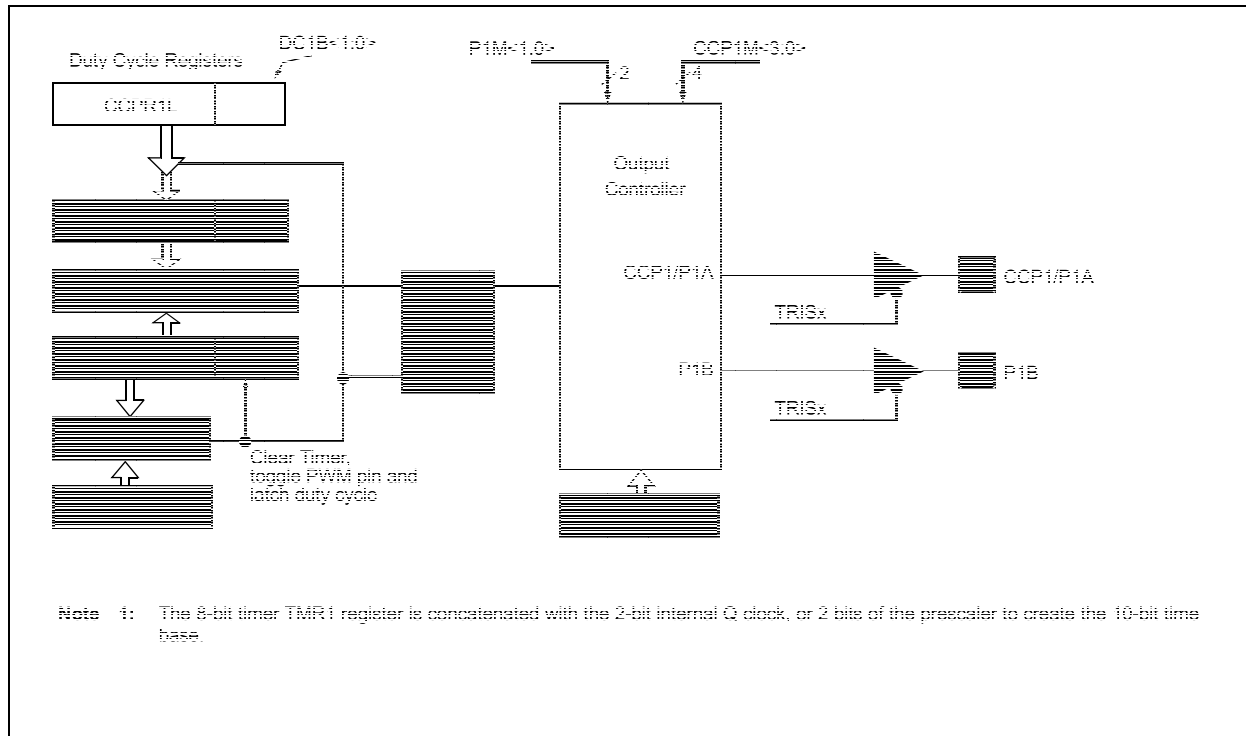
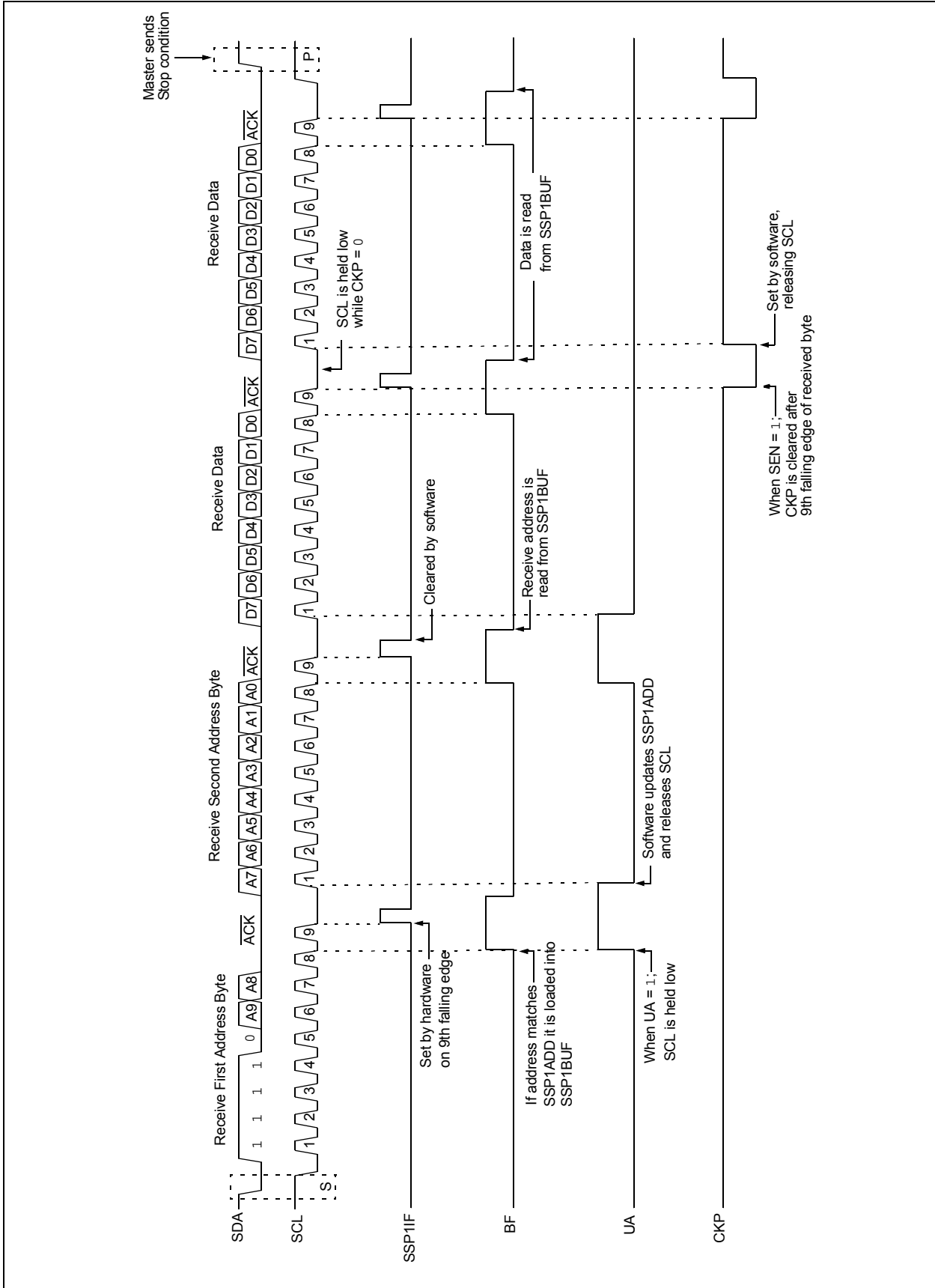


FIGURE 25-20: I²C SLAVE, 10-BIT ADDRESS, RECEPTION (SEN = 1, AHEN = 0, DHEN = 0)



26.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

The Enhanced Universal Synchronous Asynchronous Receiver Transmitter (EUSART) module is a serial I/O communications peripheral. It contains all the clock generators, shift registers and data buffers necessary to perform an input or output serial data transfer independent of device program execution. The EUSART, also known as a Serial Communications Interface (SCI), can be configured as a full-duplex asynchronous system or half-duplex synchronous system. Full-Duplex mode is useful for communications with peripheral systems, such as CRT terminals and personal computers. Half-Duplex Synchronous mode is intended for communications with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs or other microcontrollers. These devices typically do not have internal clocks for baud rate generation and require the external clock signal provided by a master synchronous device.

The EUSART module includes the following capabilities:

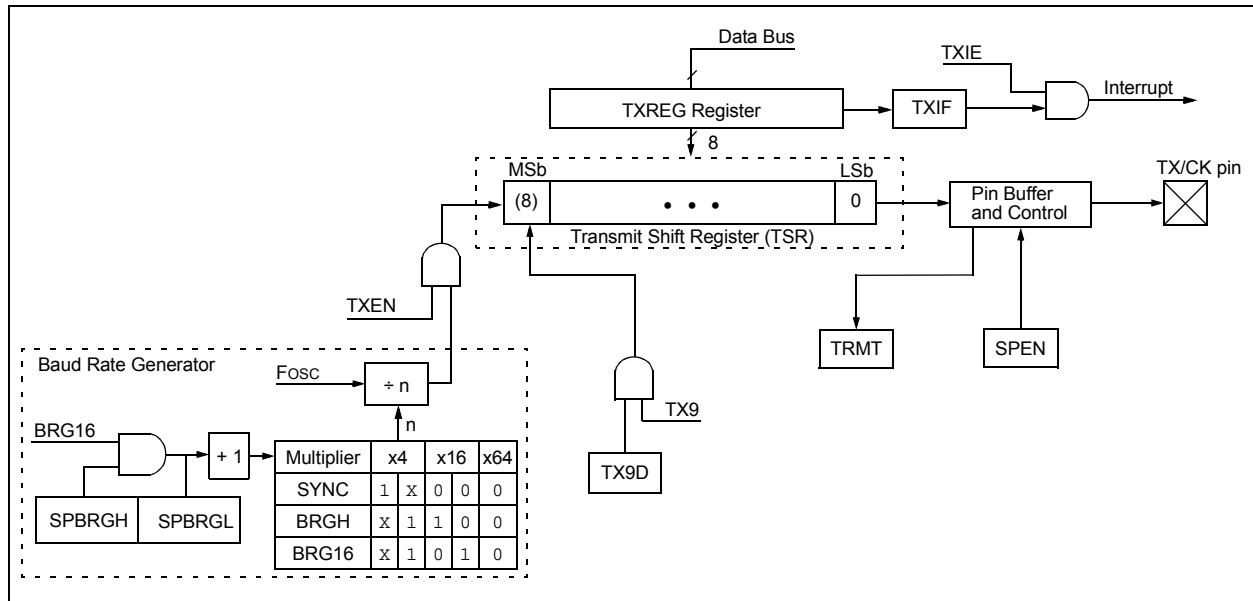
- Full-duplex asynchronous transmit and receive
- Two-character input buffer
- One-character output buffer
- Programmable 8-bit or 9-bit character length
- Address detection in 9-bit mode
- Input buffer overrun error detection
- Received character framing error detection
- Half-duplex synchronous master
- Half-duplex synchronous slave
- Programmable clock polarity in synchronous modes
- Sleep operation

The EUSART module implements the following additional features, making it ideally suited for use in Local Interconnect Network (LIN) bus systems:

- Automatic detection and calibration of the baud rate
- Wake-up on Break reception
- 13-bit Break character transmit

Block diagrams of the EUSART transmitter and receiver are shown in Figure 26-1 and Figure 26-2.

FIGURE 26-1: EUSART TRANSMIT BLOCK DIAGRAM



26.1.2.4 Receive Framing Error

Each character in the receive FIFO buffer has a corresponding framing error Status bit. A framing error indicates that a Stop bit was not seen at the expected time. The framing error status is accessed via the FERR bit of the RCSTA register. The FERR bit represents the status of the top unread character in the receive FIFO. Therefore, the FERR bit must be read before reading the RCREG.

The FERR bit is read-only and only applies to the top unread character in the receive FIFO. A framing error (FERR = 1) does not preclude reception of additional characters. It is not necessary to clear the FERR bit. Reading the next character from the FIFO buffer will advance the FIFO to the next character and the next corresponding framing error.

The FERR bit can be forced clear by clearing the SPEN bit of the RCSTA register which resets the EUSART. Clearing the CREN bit of the RCSTA register does not affect the FERR bit. A framing error by itself does not generate an interrupt.

Note: If all receive characters in the receive FIFO have framing errors, repeated reads of the RCREG will not clear the FERR bit.

26.1.2.5 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before the FIFO is accessed. When this happens the OERR bit of the RCSTA register is set. The characters already in the FIFO buffer can be read but no additional characters will be received until the error is cleared. The error must be cleared by either clearing the CREN bit of the RCSTA register or by resetting the EUSART by clearing the SPEN bit of the RCSTA register.

26.1.2.6 Receiving 9-bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RCSTA register is set the EUSART will shift nine bits into the RSR for each character received. The RX9D bit of the RCSTA register is the ninth and Most Significant data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RCREG.

26.1.2.7 Address Detection

A special Address Detection mode is available for use when multiple receivers share the same transmission line, such as in RS-485 systems. Address detection is enabled by setting the ADDEN bit of the RCSTA register.

Address detection requires 9-bit character reception. When address detection is enabled, only characters with the ninth data bit set will be transferred to the receive FIFO buffer, thereby setting the RCIF interrupt bit. All other characters will be ignored.

Upon receiving an address character, user software determines if the address matches its own. Upon address match, user software must disable address detection by clearing the ADDEN bit before the next Stop bit occurs. When user software detects the end of the message, determined by the message protocol used, software places the receiver back into the Address Detection mode by setting the ADDEN bit.

26.6 EUSART Operation During Sleep

The EUSART will remain active during Sleep only in the Synchronous Slave mode. All other modes require the system clock and therefore cannot generate the necessary signals to run the Transmit or Receive Shift registers during Sleep.

Synchronous Slave mode uses an externally generated clock to run the Transmit and Receive Shift registers.

26.6.1 SYNCHRONOUS RECEIVE DURING SLEEP

To receive during Sleep, all the following conditions must be met before entering Sleep mode:

- RCSTA and TXSTA Control registers must be configured for Synchronous Slave Reception (see **Section 26.5.2.4 “Synchronous Slave Reception Set-up:”**).
- If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
- The RCIF interrupt flag must be cleared by reading RCREG to unload any pending characters in the receive buffer.

Upon entering Sleep mode, the device will be ready to accept data and clocks on the RX/DT and TX/CK pins, respectively. When the data word has been completely clocked in by the external device, the RCIF interrupt flag bit of the PIR1 register will be set. Thereby, waking the processor from Sleep.

Upon waking from Sleep, the instruction following the SLEEP instruction will be executed. If the Global Interrupt Enable (GIE) bit of the INTCON register is also set, then the Interrupt Service Routine at address 004h will be called.

26.6.2 SYNCHRONOUS TRANSMIT DURING SLEEP

To transmit during Sleep, all the following conditions must be met before entering Sleep mode:

- RCSTA and TXSTA Control registers must be configured for synchronous slave transmission (see **Section 26.5.2.2 “Synchronous Slave Transmission Set-up:”**).
- The TXIF interrupt flag must be cleared by writing the output data to the TXREG, thereby filling the TSR and transmit buffer.
- If interrupts are desired, set the TXIE bit of the PIE1 register and the PEIE bit of the INTCON register.
- Interrupt enable bits TXIE of the PIE1 register and PEIE of the INTCON register must set.

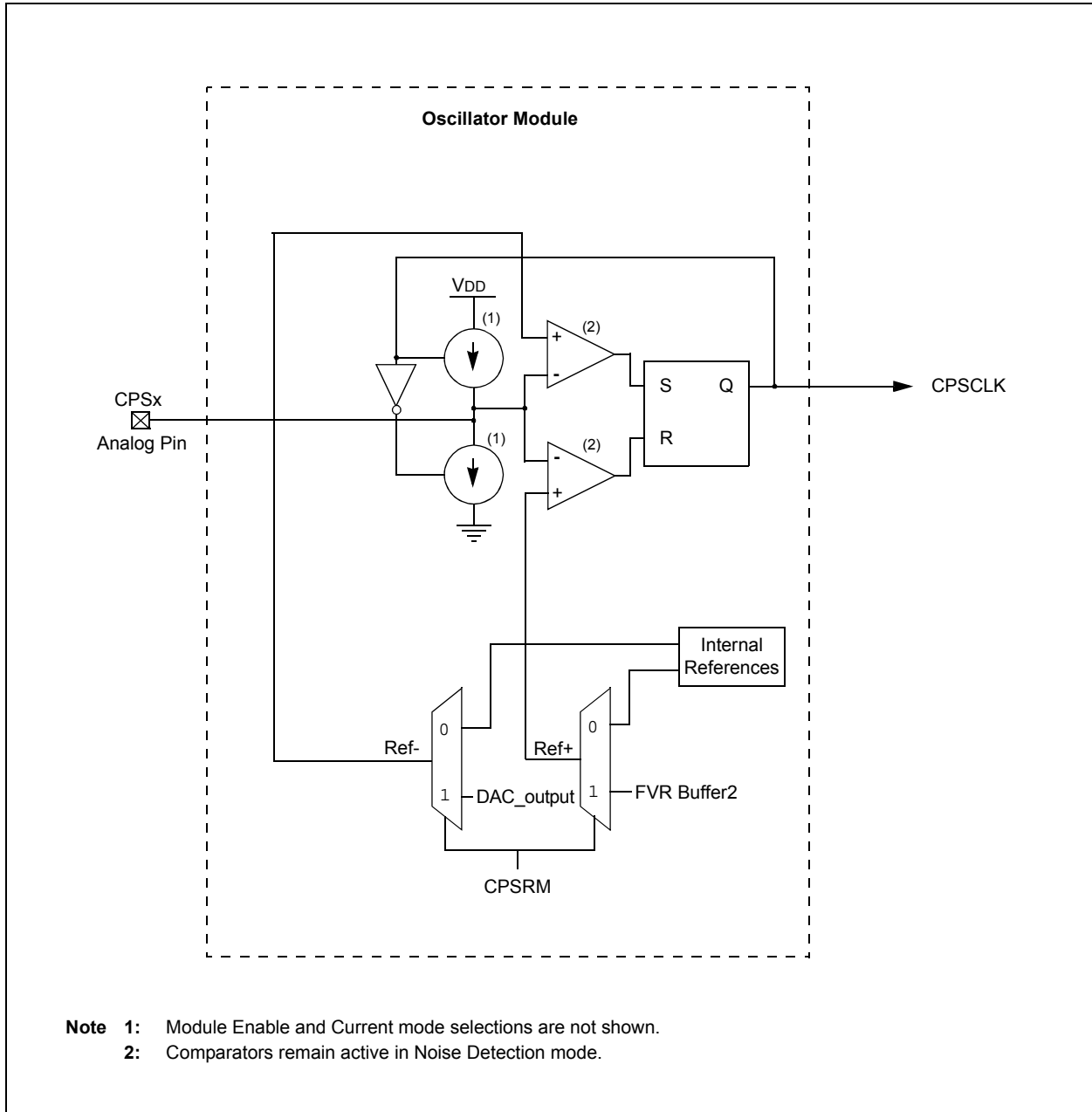
Upon entering Sleep mode, the device will be ready to accept clocks on TX/CK pin and transmit data on the RX/DT pin. When the data word in the TSR has been completely clocked out by the external device, the pending byte in the TXREG will transfer to the TSR and the TXIF flag will be set. Thereby, waking the processor from Sleep. At this point, the TXREG is available to accept another character for transmission, which will clear the TXIF flag.

Upon waking from Sleep, the instruction following the SLEEP instruction will be executed. If the Global Interrupt Enable (GIE) bit is also set then the Interrupt Service Routine at address 0004h will be called.

26.6.3 ALTERNATE PIN LOCATIONS

This module incorporates I/O pins that can be moved to other locations with the use of the alternate pin function register, APFCON. To determine which pins can be moved and what their default locations are upon a Reset, see **Section 12.1 “Alternate Pin Function”** for more information.

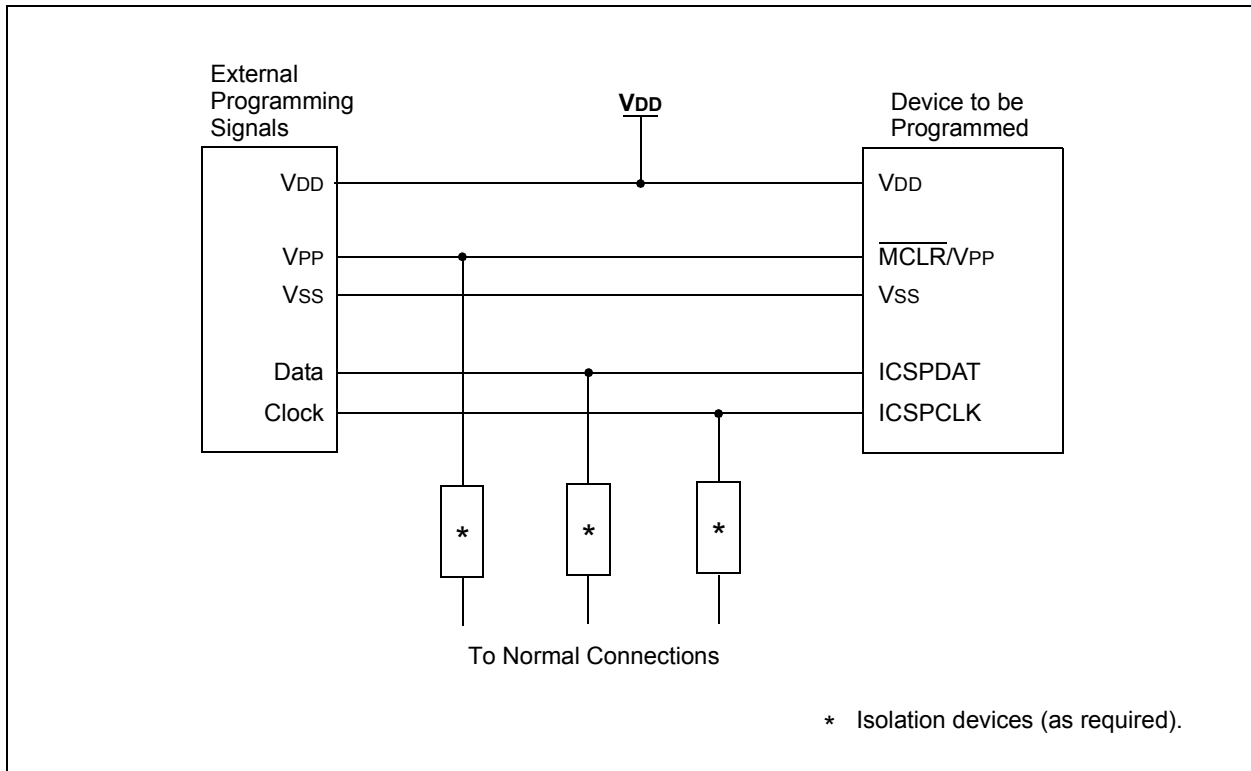
FIGURE 27-2: CAPACITIVE SENSING OSCILLATOR BLOCK DIAGRAM



For additional interface recommendations, refer to your specific device programmer manual prior to PCB design.

It is recommended that isolation devices be used to separate the programming pins from other circuitry. The type of isolation is highly dependent on the specific application and may include devices such as resistors, diodes, or even jumpers. See Figure 28-4 for more information.

FIGURE 28-4: TYPICAL CONNECTION FOR ICSP™ PROGRAMMING



30.8 AC Characteristics: PIC12(L)F1840-I/E

FIGURE 30-6: CLOCK TIMING

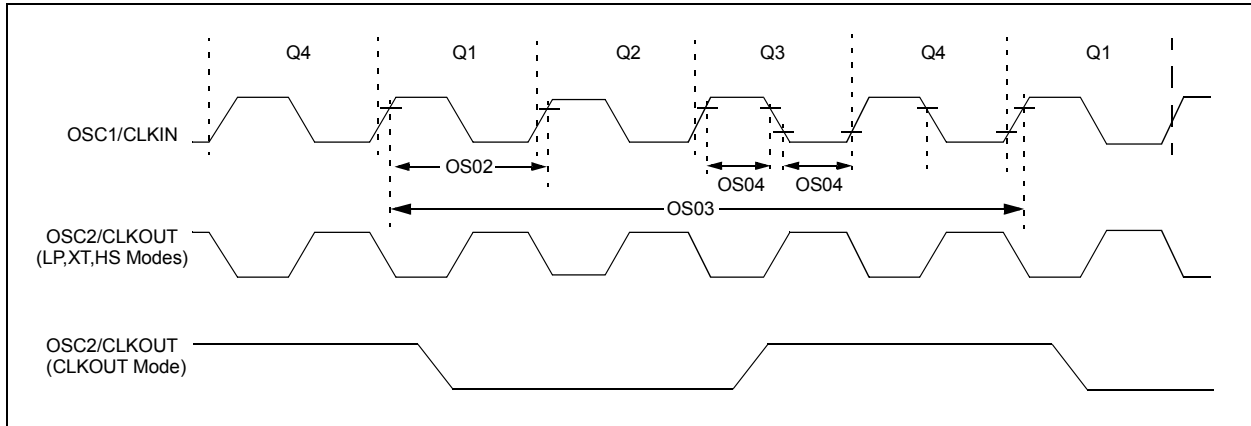


TABLE 30-1: CLOCK OSCILLATOR TIMING REQUIREMENTS

Standard Operating Conditions (unless otherwise stated)							
Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$							
Param No.	Sym.	Characteristic	Min.	Typ†	Max.	Units	Conditions
OS01	Fosc	External CLKIN Frequency ⁽¹⁾	DC	—	0.5	MHz	External Clock (ECL)
			DC	—	4	MHz	External Clock (ECM)
			DC	—	32	MHz	External Clock (ECH)
		Oscillator Frequency ⁽¹⁾	—	32.768	—	kHz	LP Oscillator
			0.1	—	4	MHz	XT Oscillator
1	—		4	MHz	HS Oscillator		
1	—		20	MHz	HS Oscillator, $V_{DD} > 2.7\text{V}$		
DC	—	4	MHz	RC Oscillator, $V_{DD} > 2.0\text{V}$			
OS02	Tosc	External CLKIN Period ⁽¹⁾	27	—	∞	μs	LP Oscillator
			250	—	∞	ns	XT Oscillator
			50	—	∞	ns	HS Oscillator
			50	—	∞	ns	External Clock (EC)
		Oscillator Period ⁽¹⁾	—	30.5	—	μs	LP Oscillator
250	—	10,000	ns	XT Oscillator			
50	—	1,000	ns	HS Oscillator			
250	—	—	ns	RC Oscillator			
OS03	Tcy	Instruction Cycle Time ⁽¹⁾	200	Tcy	DC	ns	$T_{CY} = 4/F_{OSC}$
OS04*	TosH, TosL	External CLKIN High, External CLKIN Low	2	—	—	μs	LP oscillator
			100	—	—	ns	XT oscillator
			20	—	—	ns	HS oscillator
OS05*	TosR, TosF	External CLKIN Rise, External CLKIN Fall	0	—	—	ns	LP oscillator
			0	—	—	ns	XT oscillator
			0	—	—	ns	HS oscillator

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (Tcy) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

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TABLE 30-9: ADC CONVERSION REQUIREMENTS

Operating Conditions (unless otherwise stated) VDD = 3.0V, TA = 25°C							
Param No.	Sym.	Characteristic	Min.	Typ†	Max.	Units	Conditions
AD130*	TAD	ADC Clock Period	1.0	—	9.0	μs	Fosc-based
		ADC Internal RC Oscillator Period	1.0	2.0	6.0	μs	ADCS<2:0> = x11 (ADC FRC mode)
AD131	TcNV	Conversion Time (not including Acquisition Time) (Note 1)	—	11	—	TAD	Set GO/DONE bit to conversion complete
AD132*	TACQ	Acquisition Time	—	5.0	—	μs	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The ADRES register may be read on the following TcY cycle.

FIGURE 30-12: ADC CONVERSION TIMING (NORMAL MODE)

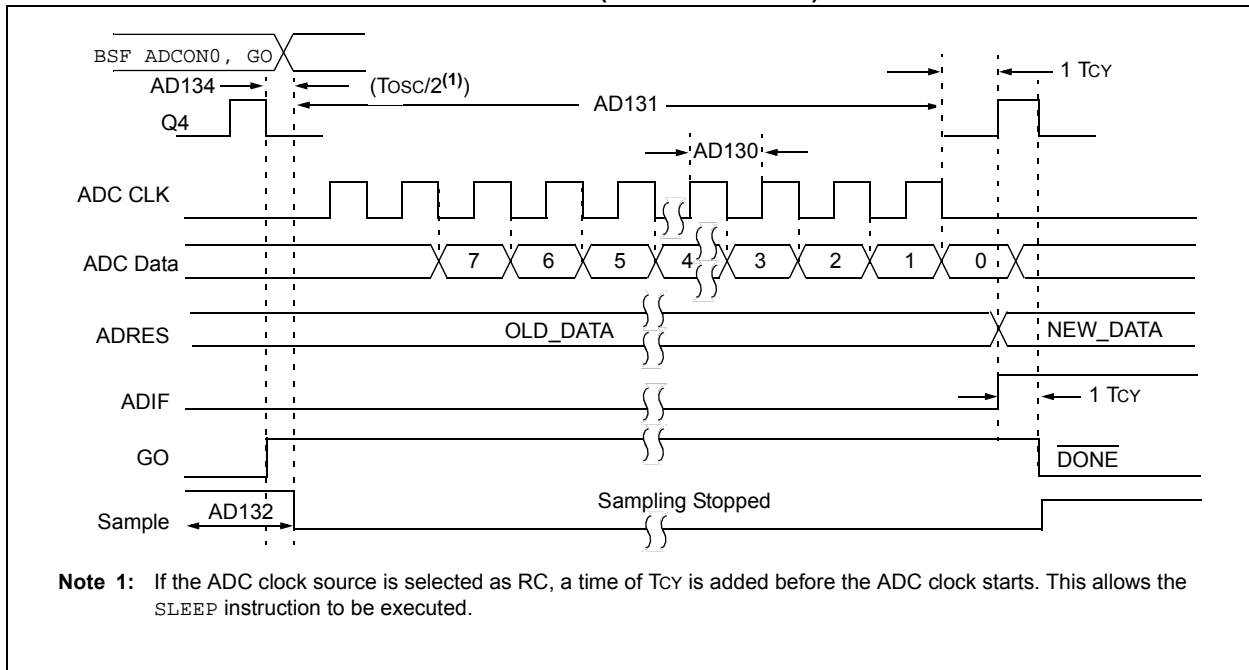


FIGURE 30-14: USART SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING

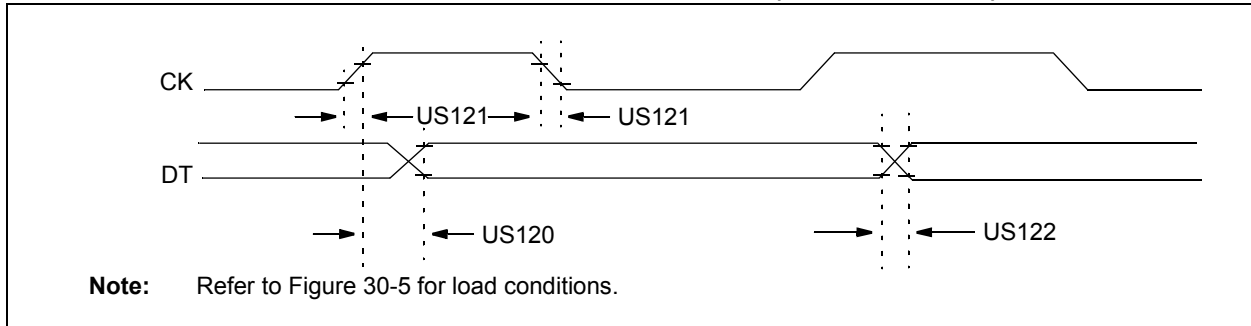


TABLE 30-12: USART SYNCHRONOUS TRANSMISSION REQUIREMENTS

Standard Operating Conditions (unless otherwise stated)							
Operating Temperature $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$							
Param. No.	Symbol	Characteristic		Min.	Max.	Units	Conditions
US120	TCKH2DTV	SYNC XMIT (Master and Slave) Clock high to data-out valid	3.0-5.5V	—	80	ns	
			1.8-5.5V	—	100	ns	
US121	TCKRF	Clock out rise time and fall time (Master mode)	3.0-5.5V	—	45	ns	
			1.8-5.5V	—	50	ns	
US122	TDTRF	Data-out rise time and fall time	3.0-5.5V	—	45	ns	
			1.8-5.5V	—	50	ns	

FIGURE 30-15: USART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING

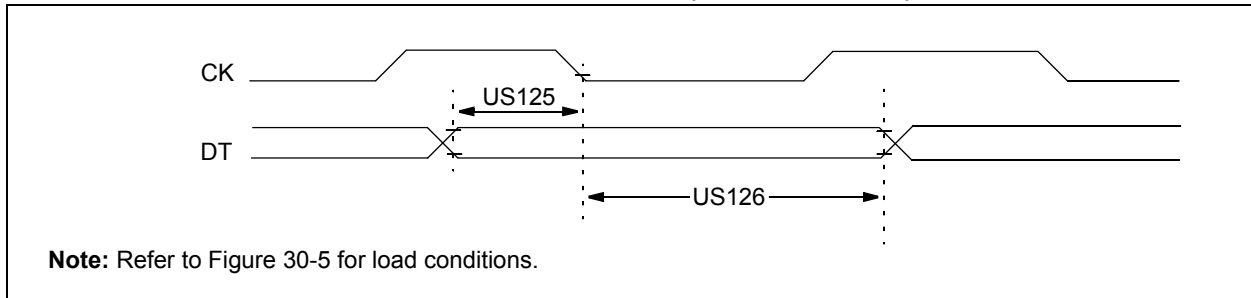


TABLE 30-13: USART SYNCHRONOUS RECEIVE REQUIREMENTS

Standard Operating Conditions (unless otherwise stated)							
Operating Temperature $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$							
Param. No.	Symbol	Characteristic	Min.	Max.	Units	Conditions	
US125	TDTV2CKL	SYNC RCV (Master and Slave) Data-hold before CK ↓ (DT hold time)	10	—	ns		
US126	TCKL2DTL	Data-hold after CK ↓ (DT hold time)	15	—	ns		

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FIGURE 30-21: I²C™ BUS DATA TIMING

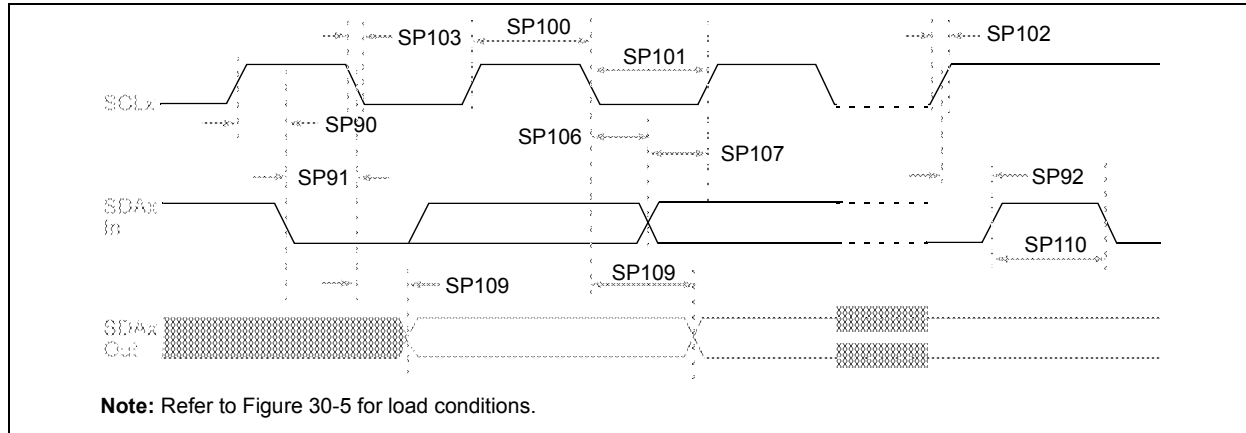


TABLE 30-16: I²C™ BUS DATA REQUIREMENTS

Param. No.	Symbol	Characteristic		Min.	Max.	Units	Conditions
SP100*	THIGH	Clock high time	100 kHz mode	4.0	—	μs	Device must operate at a minimum of 1.5 MHz
			400 kHz mode	0.6	—	μs	Device must operate at a minimum of 10 MHz
		SSP module	1.5Tcy	—	—		
SP101*	TLOW	Clock low time	100 kHz mode	4.7	—	μs	Device must operate at a minimum of 1.5 MHz
			400 kHz mode	1.3	—	μs	Device must operate at a minimum of 10 MHz
		SSP module	1.5Tcy	—	—		
SP102*	TR	SDA and SCL rise time	100 kHz mode	—	1000	ns	
			400 kHz mode	20 + 0.1Cb	300	ns	Cb is specified to be from 10-400 pF
SP103*	TF	SDA and SCL fall time	100 kHz mode	—	250	ns	
			400 kHz mode	20 + 0.1Cb	250	ns	Cb is specified to be from 10-400 pF
SP106*	THD:DAT	Data input hold time	100 kHz mode	0	—	ns	
			400 kHz mode	0	0.9	μs	
SP107*	TSU:DAT	Data input setup time	100 kHz mode	250	—	ns	(Note 2)
			400 kHz mode	100	—	ns	
SP109*	TAA	Output valid from clock	100 kHz mode	—	3500	ns	(Note 1)
			400 kHz mode	—	—	ns	
SP110*	TBUF	Bus free time	100 kHz mode	4.7	—	μs	Time the bus must be free before a new transmission can start
			400 kHz mode	1.3	—	μs	
SP111	CB	Bus capacitive loading		—	400	pF	

* These parameters are characterized but not tested.

Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCL to avoid unintended generation of Start or Stop conditions.

2: A Fast mode (400 kHz) I²C™ bus device can be used in a Standard mode (100 kHz) I²C bus system, but the requirement TSU:DAT ≥ 250 ns must then be met. This will automatically be the case if the device does not stretch the low period of the SCL signal. If such a device does stretch the low period of the SCL signal, it must output the next data bit to the SDA line TR max. + TSU:DAT = 1000 + 250 = 1250 ns (according to the Standard mode I²C bus specification), before the SCL line is released.

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TABLE 30-23: A/D CONVERTER (ADC) CHARACTERISTICS FOR PIC12F1840-H (High Temp.)

PIC12F1840			Standard Operating Conditions: (unless otherwise stated) Operating Temperature: $-40^{\circ}\text{C} \leq T_A \leq +150^{\circ}\text{C}$ for High Temperature				
Param No.	Sym.	Characteristic	Min.	Typ.	Max.	Units	Conditions
AD04	E _{OFF}	Offset Error	—	—	3.5	LSB	No missing codes V _{REF} = 3.0V

† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

- Note 1:** Total Absolute Error includes integral, differential, offset and gain errors.
2: The A/D conversion result never decreases with an increase in the input voltage and has no missing codes.
3: ADC V_{REF} is from external V_{REF}, V_{DD} pin or F_VR, whichever is selected as reference input.

TABLE 30-24: COMPARATOR SPECIFICATIONS FOR PIC12F1840-H (High Temp.)

PIC12F1840			Standard Operating Conditions: (unless otherwise stated) Operating Temperature: $-40^{\circ}\text{C} \leq T_A \leq +150^{\circ}\text{C}$ for High Temperature				
Param No.	Sym.	Characteristic	Min.	Typ.	Max.	Units	Conditions
CM01	V _{IOFF}	Input Offset Voltage	—	—	±70	mV	High-Power mode, V _{ICM} = V _{DD} /2

TABLE 30-25: CAP SENSE OSCILLATOR SPECIFICATIONS FOR PIC12F1840-H (High Temp.)

PIC12F1840			Standard Operating Conditions: (unless otherwise stated) Operating Temperature: $-40^{\circ}\text{C} \leq T_A \leq +150^{\circ}\text{C}$ for High Temperature				
Param No.	Sym.	Characteristic	Min.	Typ.	Max.	Units	Conditions
All	All	All	—	—	—	—	This module is not intended for use in high temperature devices.

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FIGURE 31-5: I_{DD} TYPICAL, XT AND EXTRC OSCILLATOR, PIC12F1840 ONLY

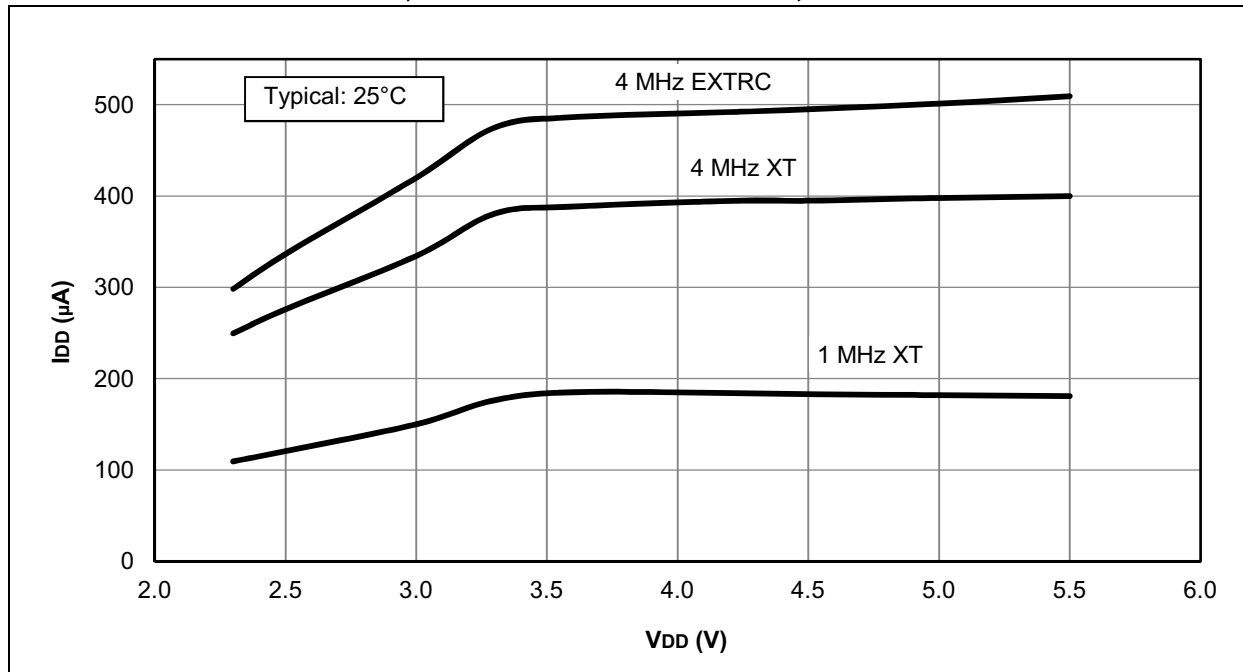


FIGURE 31-6: I_{DD} MAXIMUM, XT AND EXTRC OSCILLATOR, PIC12F1840 ONLY

