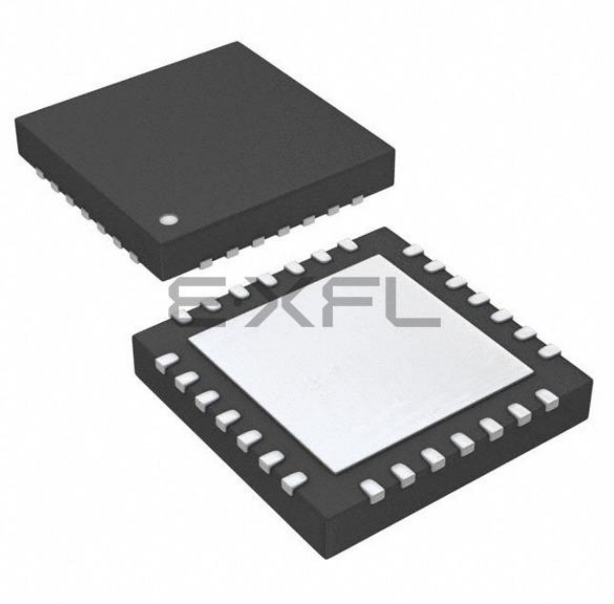


Welcome to E-XFL.COM

What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	25
Program Memory Size	7KB (4K x 14)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	192 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 5.5V
Data Converters	A/D 11x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	28-UFQFN Exposed Pad
Supplier Device Package	28-UQFN (4x4)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16f723-e-mv

FIGURE 2-4: PIC16F722/LF722 SPECIAL FUNCTION REGISTERS

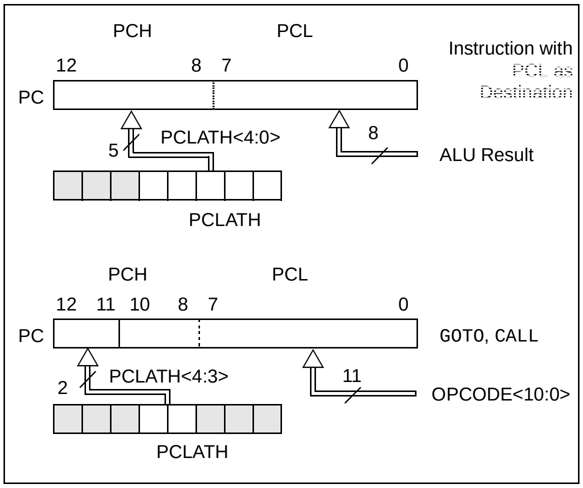
Bank 0	File Address	Bank 1	File Address	Bank 2	File Address	Bank 3	File Address
Indirect addr.(*)	00h	Indirect addr.(*)	80h	Indirect addr.(*)	100h	Indirect addr.(*)	180h
TMR0	01h	OPTION	81h	TMR0	101h	OPTION	181h
PCL	02h	PCL	82h	PCL	102h	PCL	182h
STATUS	03h	STATUS	83h	STATUS	103h	STATUS	183h
FSR	04h	FSR	84h	FSR	104h	FSR	184h
PORTA	05h	TRISA	85h		105h	ANSELA	185h
PORTB	06h	TRISB	86h		106h	ANSELB	186h
PORTC	07h	TRISC	87h		107h		187h
	08h		88h	CPSCON0	108h		188h
PORTE	09h	TRISE	89h	CPSCON1	109h		189h
PCLATH	0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH	18Ah
INTCON	0Bh	INTCON	8Bh	INTCON	10Bh	INTCON	18Bh
PIR1	0Ch	PIE1	8Ch	PMDATL	10Ch	PMCON1	18Ch
PIR2	0Dh	PIE2	8Dh	PMADRL	10Dh	Reserved	18Dh
TMR1L	0Eh	PCON	8Eh	PMDATH	10Eh	Reserved	18Eh
TMR1H	0Fh	T1GCON	8Fh	PMADRH	10Fh	Reserved	18Fh
T1CON	10h	OSCCON	90h		110h		190h
TMR2	11h	OSCTUNE	91h		111h		191h
T2CON	12h	PR2	92h		112h		192h
SSPBUF	13h	SSPADD/SSPMSK	93h		113h		193h
SSPCON	14h	SSPSTAT	94h		114h		194h
CCPR1L	15h	WPUB	95h		115h		195h
CCPR1H	16h	IOCB	96h		116h		196h
CCP1CON	17h		97h		117h		197h
RCSTA	18h	TXSTA	98h		118h		198h
TXREG	19h	SPBRG	99h		119h		199h
RCREG	1Ah		9Ah		11Ah		19Ah
CCPR2L	1Bh		9Bh		11Bh		19Bh
CCPR2H	1Ch	APFCON	9Ch		11Ch		19Ch
CCP2CON	1Dh	FVRCON	9Dh		11Dh		19Dh
ADRES	1Eh		9Eh		11Eh		19Eh
ADCON0	1Fh	ADCON1	9Fh		11Fh		19Fh
General Purpose Register 96 Bytes	20h–7Fh	General Purpose Register 32 Bytes	A0h–BFh		120h–16Fh		1A0h–1EFh
			C0h–EFh				
		Accesses 70h-7Fh	F0h–FFh	Accesses 70h-7Fh	170h–17Fh	Accesses 70h-7Fh	1F0h–1FFh

Legend: (shaded) = Unimplemented data memory locations, read as '0'.
* = Not a physical register.

2.3 PCL and PCLATH

The Program Counter (PC) is 13 bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<12:8>) is not directly readable or writable and comes from PCLATH. On any Reset, the PC is cleared. Figure 2-7 shows the two situations for the loading of the PC. The upper example in Figure 2-7 shows how the PC is loaded on a write to PCL (PCLATH<4:0> $\rightarrow$ PCH). The lower example in Figure 2-7 shows how the PC is loaded during a `CALL` or `GOTO` instruction (PCLATH<4:3> $\rightarrow$ PCH).

FIGURE 2-7: LOADING OF PC IN DIFFERENT SITUATIONS



2.3.1 COMPUTED `GOTO`

A computed `GOTO` is accomplished by adding an offset to the program counter (`ADDWF PCL`). When performing a table read using a computed `GOTO` method, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block). Refer to the Application Note *AN556, Implementing a Table Read* (DS00556).

2.3.2 STACK

All devices have an 8-level x 13-bit wide hardware stack (refer to Figures 2-1 and 2-3). The stack space is not part of either program or data space and the Stack Pointer is not readable or writable. The PC is PUSHed onto the stack when a `CALL` instruction is executed or an interrupt causes a branch. The stack is POPed in the event of a `RETURN`, `RETLW` or a `RETFIE` instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer. This means that after the stack has been PUSHed eight times, the ninth PUSH overwrites the value that was stored from the first PUSH. The tenth PUSH overwrites the second PUSH (and so on).

> **Note 1:** There are no Status bits to indicate Stack Overflow or Stack Underflow conditions.
>
> **2:** There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the `CALL`, `RETURN`, `RETLW` and `RETFIE` instructions or the vectoring to an interrupt address.

2.4 Program Memory Paging

All devices are capable of addressing a continuous 8K word block of program memory. The `CALL` and `GOTO` instructions provide only 11 bits of address to allow branching within any 2K program memory page. When doing a `CALL` or `GOTO` instruction, the upper two bits of the address are provided by PCLATH<4:3>. When doing a `CALL` or `GOTO` instruction, the user must ensure that the page select bits are programmed so that the desired program memory page is addressed. If a return from a `CALL` instruction (or interrupt) is executed, the entire 13-bit PC is POPed off the stack. Therefore, manipulation of the PCLATH<4:3> bits is not required for the `RETURN` instructions (which POPs the address from the stack).

> **Note:** The contents of the PCLATH register are unchanged after a `RETURN` or `RETFIE` instruction is executed. The user must rewrite the contents of the PCLATH register for any subsequent subroutine calls or `GOTO` instructions.

Example 2-1 shows the calling of a subroutine in page 1 of the program memory. This example assumes that PCLATH is saved and restored by the Interrupt Service Routine (if interrupts are used).

EXAMPLE 2-1: CALL OF A SUBROUTINE IN PAGE 1 FROM PAGE 0

```
        ORG 500h
        PAGESEL SUB_P1  ;Select page 1
                        ;(800h-FFFh)
        CALL    SUB1_P1 ;Call subroutine in
        :               ;page 1 (800h-FFFh)
        :
        ORG     900h    ;page 1 (800h-FFFh)
SUB1_P1
        :               ;called subroutine
                        ;page 1 (800h-FFFh)
        :
        RETURN          ;return to
                        ;Call subroutine
                        ;in page 0
                        ;(000h-7FFh)
```

EXAMPLE 4-1: SAVING W, STATUS AND PCLATH REGISTERS IN RAM

```
MOVWF   W_TEMP            ;Copy W to W_TEMP register
SWAPF   STATUS,W          ;Swap status to be saved into W
                          ;Swaps are used because they do not affect the status bits
BANKSEL STATUS_TEMP       ;Select regardless of current bank
MOVWF   STATUS_TEMP       ;Copy status to bank zero STATUS_TEMP register
MOVF    PCLATH,W          ;Copy PCLATH to W register
MOVWF   PCLATH_TEMP       ;Copy W register to PCLATH_TEMP
:
:(ISR)                    ;Insert user code here
:
BANKSEL STATUS_TEMP       ;Select regardless of current bank
MOVF    PCLATH_TEMP,W     ;
MOVWF   PCLATH            ;Restore PCLATH
SWAPF   STATUS_TEMP,W     ;Swap STATUS_TEMP register into W
                          ;(sets bank to original state)
MOVWF   STATUS            ;Move W into STATUS register
SWAPF   W_TEMP,F          ;Swap W_TEMP
SWAPF   W_TEMP,W          ;Swap W_TEMP into W
```

4.5.1 INTCON REGISTER

The INTCON register is a readable and writable register, which contains the various enable and flag bits for TMR0 register overflow, PORTB change and external RB0/INT/SEG0 pin interrupts.

Note:	Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit, GIE of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 6-7: WPUB: WEAK PULL-UP PORTB REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
WPUB7	WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **WPUB<7:0>**: Weak Pull-up Register bits
`1` = Pull-up enabled
`0` = Pull-up disabled

Note 1: Global $\overline{\text{RBPU}}$ bit of the OPTION register must be cleared for individual pull-ups to be enabled.
2: The weak pull-up device is automatically disabled if the pin is in configured as an output.

REGISTER 6-8: IOCB: INTERRUPT-ON-CHANGE PORTB REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
IOCB7	IOCB6	IOCB5	IOCB4	IOCB3	IOCB2	IOCB1	IOCB0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **IOCB<7:0>:** Interrupt-on-Change PORTB Control bits
`1` = Interrupt-on-change enabled
`0` = Interrupt-on-change disabled

REGISTER 6-9: ANSELB: PORTB ANALOG SELECT REGISTER

U-0	U-0	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
—	—	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 **Unimplemented:** Read as '`0`'

bit 5-0 **ANSB<5:0>**: Analog Select between Analog or Digital Function on Pins RB<5:0>, respectively
`0` = Digital I/O. Pin is assigned to port or Digital special function.
`1` = Analog input. Pin is assigned as analog input[(1)]. Digital Input buffer disabled.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

REGISTER 6-13: TRISD: PORTD TRI-STATE REGISTER(1)

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-0 **TRISD<7:0>:** PORTD Tri-State Control bits
`1` = PORTD pin configured as an input (tri-stated)
`0` = PORTD pin configured as an output

Note 1: TRISD is not implemented on PIC16F722/723/726/PIC16LF722/723/726 devices, read as ‘`0`’.

REGISTER 6-14: ANSELD: PORTD ANALOG SELECT REGISTER(2)

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
ANSD7	ANSD6	ANSD5	ANSD4	ANSD3	ANSD2	ANSD1	ANSD0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-0 **ANSD<7:0>**: Analog Select between Analog or Digital Function on Pins RD<7:0>, respectively
`0` = Digital I/O. Pin is assigned to port or Digital special function.
`1` = Analog input. Pin is assigned as analog input(1). Digital Input buffer disabled.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.
2: ANSELD register is not implemented on the PIC16F722/723/726/PIC16LF722/723/726. Read as ‘`0`’.

Note:	PORTD is available on PIC16F724/LF724 and PIC16F727/LF727 only.

6.5.2 RD0/CPS8

Figure 6-21 shows the diagram for these pins. They are configurable to function as one of the following:

- a general purpose I/O
- a capacitive sensing input

6.5.3 RD1/CPS9

Figure 6-21 shows the diagram for these pins. They are configurable to function as one of the following:

- a general purpose I/O
- a capacitive sensing input

6.5.4 RD2/CPS10

Figure 6-21 shows the diagram for these pins. They are configurable to function as one of the following:

- a general purpose I/O
- a capacitive sensing input

6.5.5 RD3/CPS11

Figure 6-21 shows the diagram for these pins. They are configurable to function as one of the following:

- a general purpose I/O
- a capacitive sensing input

REGISTER 8-1: CONFIG1: CONFIGURATION WORD REGISTER 1 (CONTINUED)

bit 4	**$\overline{\text{PWRTE}}$:** Power-up Timer Enable bit 1 = PWRT disabled 0 = PWRT enabled
bit 3	**WDTE:** Watchdog Timer Enable bit 1 = WDT enabled 0 = WDT disabled
bit 2-0	**FOSC<2:0>:** Oscillator Selection bits 111 = RC oscillator: CLKOUT function on RA6/OSC2/CLKOUT pin, RC on RA7/OSC1/CLKIN 110 = RCIO oscillator: I/O function on RA6/OSC2/CLKOUT pin, RC on RA7/OSC1/CLKIN 101 = INTOSC oscillator: CLKOUT function on RA6/OSC2/CLKOUT pin, I/O function on RA7/OSC1/CLKIN 100 = INTOSCIO oscillator: I/O function on RA6/OSC2/CLKOUT pin, I/O function on RA7/OSC1/CLKIN 011 = EC: I/O function on RA6/OSC2/CLKOUT pin, CLKIN on RA7/OSC1/CLKIN 010 = HS oscillator: High-speed crystal/resonator on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN 001 = XT oscillator: Crystal/resonator on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN 000 = LP oscillator: Low-power crystal on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN

Note 1: Enabling Brown-out Reset does not automatically enable Power-up Timer.
2: The entire program memory will be erased when the code protection is turned off.
3: When $\overline{\text{MCLR}}$ is asserted in INTOSC or RC mode, the internal clock oscillator is disabled.
4: MPLAB® X IDE masks unimplemented Configuration bits to ‘0’.

REGISTER 8-2: CONFIG2: CONFIGURATION WORD REGISTER 2

		U-1(1)	U-1(1)	U-1(1)	U-1(1)	U-1(1)	U-1(1)
—	—	—	—	—	—	—	—
bit 15							bit 8

U-1(1)	U-1(1)	R/P-1	R/P-1	U-1(1)	U-1(1)	U-1(1)	U-1(1)
—	—	VCAPEN1	VCAPEN0	—	—	—	—
bit 7							bit 0

Legend:	P = Programmable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 13-6	**Unimplemented**: Read as ‘1’
bit 5-4	**VCAPEN<1:0>:** Voltage Regulator Capacitor Enable bits <u>For the PIC16LF72X</u>: These bits are ignored. All VCAP pin functions are disabled. <u>For the PIC16F72X</u>: 00 = VCAP functionality is enabled on RA0 01 = VCAP functionality is enabled on RA5 10 = VCAP functionality is enabled on RA6 11 = All VCAP functions are disabled (not recommended)
bit 3-0	**Unimplemented**: Read as ‘1’

Note 1: MPLAB® X IDE masks unimplemented Configuration bits to ‘0’.

9.0 ANALOG-TO-DIGITAL CONVERTER (ADC) MODULE

The Analog-to-Digital Converter (ADC) allows conversion of an analog input signal to a 8-bit binary representation of that signal. This device uses analog inputs, which are multiplexed into a single sample and hold circuit. The output of the sample and hold is connected to the input of the converter. The converter generates a 8-bit binary result via successive approximation and stores the conversion result into the ADC result register (ADRES). Figure 9-1 shows the block diagram of the ADC.

The ADC voltage reference is software selectable to be either internally generated or externally supplied.

The ADC can generate an interrupt upon completion of a conversion. This interrupt can be used to wake-up the device from Sleep.

FIGURE 9-1: ADC BLOCK DIAGRAM

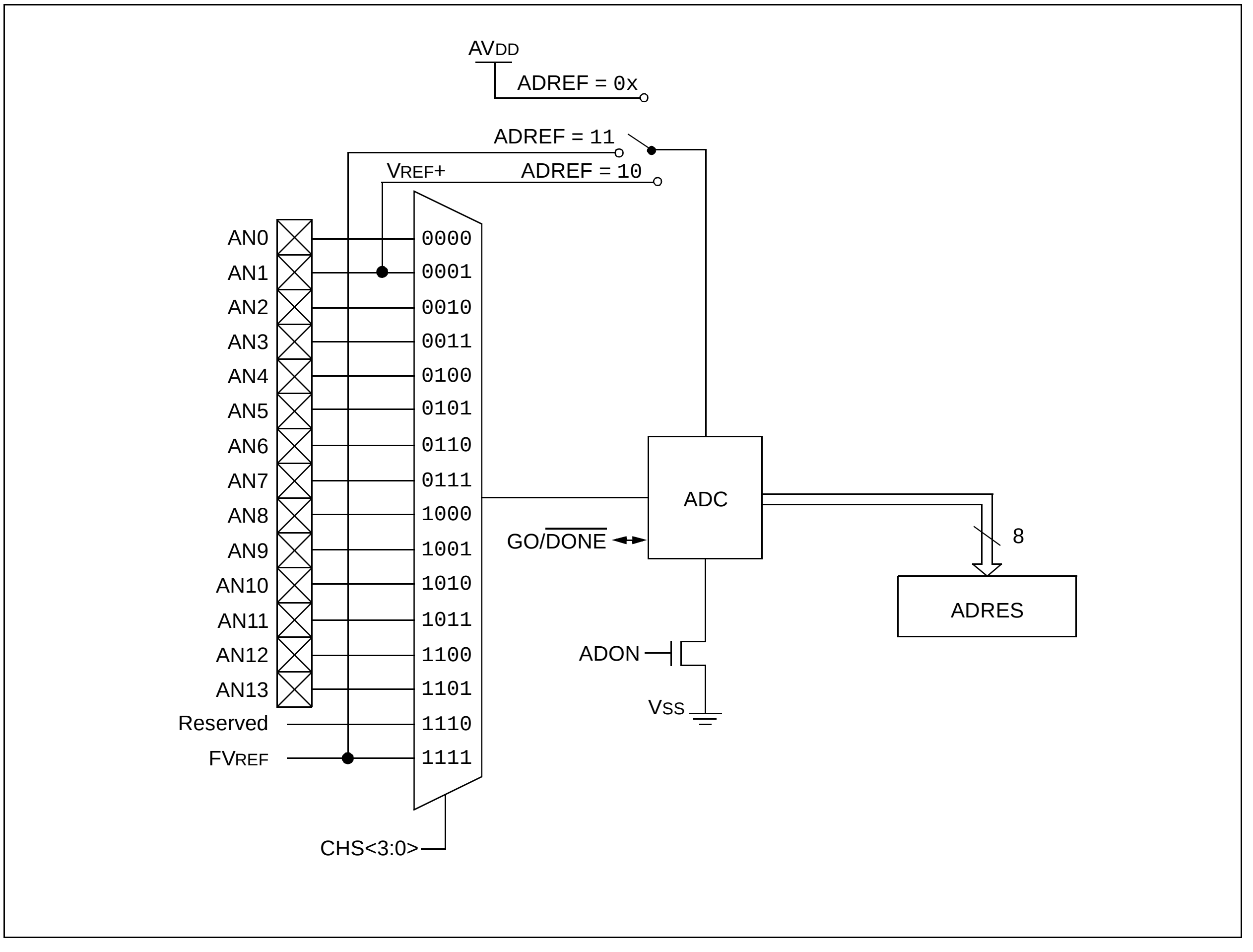
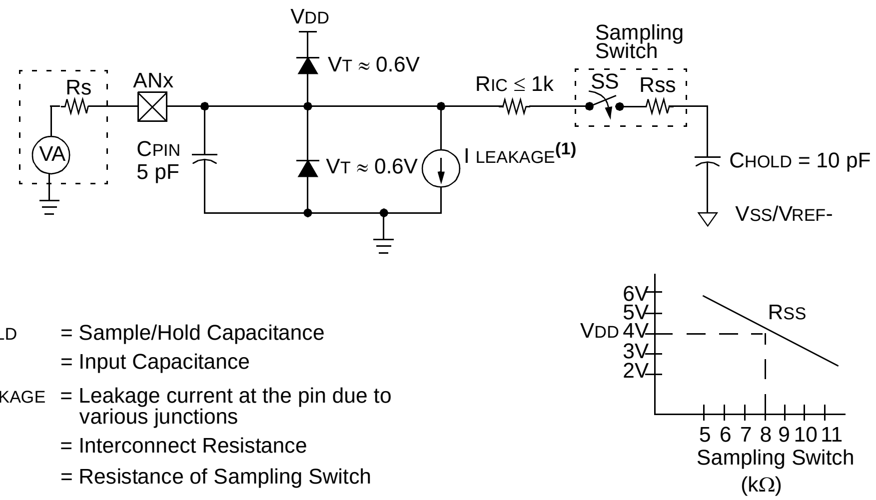


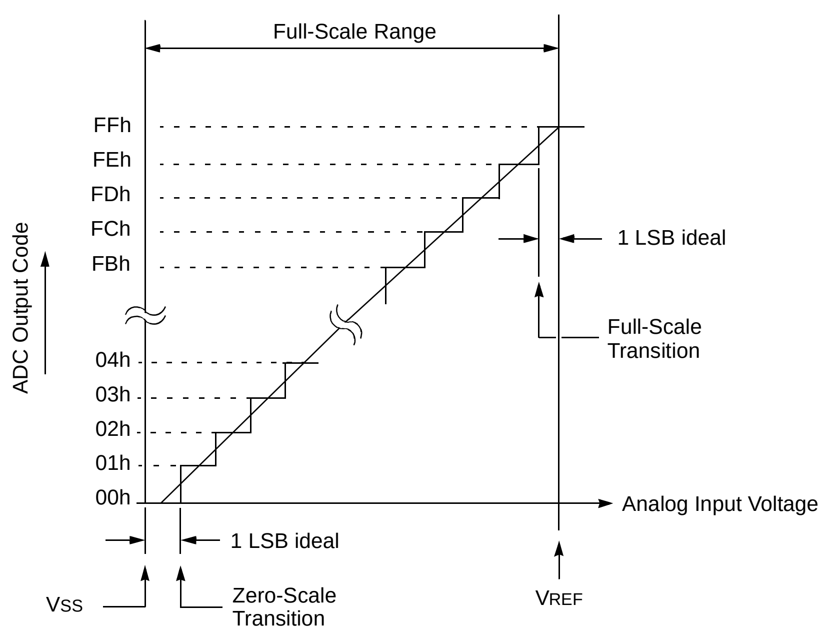
FIGURE 9-3: ANALOG INPUT MODEL



Legend:
- CHOLD = Sample/Hold Capacitance
- CPIN = Input Capacitance
- I LEAKAGE = Leakage current at the pin due to various junctions
- RIC = Interconnect Resistance
- RSS = Resistance of Sampling Switch
- SS = Sampling Switch
- VT = Threshold Voltage

Note 1: Refer to **Section 23.0 "Electrical Specifications"**.

FIGURE 9-4: ADC TRANSFER FUNCTION



11.1.3 SOFTWARE PROGRAMMABLE PRESCALER

A single software programmable prescaler is available for use with either Timer0 or the Watchdog Timer (WDT), but not both simultaneously. The prescaler assignment is controlled by the PSA bit of the OPTION register. To assign the prescaler to Timer0, the PSA bit must be cleared to a '0'.

There are eight prescaler options for the Timer0 module ranging from 1:2 to 1:256. The prescale values are selectable via the PS<2:0> bits of the OPTION register. In order to have a 1:1 prescaler value for the Timer0 module, the prescaler must be assigned to the WDT module.

The prescaler is not readable or writable. When assigned to the Timer0 module, all instructions writing to the TMR0 register will clear the prescaler.

Note:	When the prescaler is assigned to WDT, a `CLRWDT` instruction will clear the prescaler along with the WDT.

11.1.4 TIMER0 INTERRUPT

Timer0 will generate an interrupt when the TMR0 register overflows from FFh to 00h. The T0IF interrupt flag bit of the INTCON register is set every time the TMR0 register overflows, regardless of whether or not the Timer0 interrupt is enabled. The T0IF bit can only be cleared in software. The Timer0 interrupt enable is the T0IE bit of the INTCON register.

Note:	The Timer0 interrupt cannot wake the processor from Sleep since the timer is frozen during Sleep.

11.1.5 8-BIT COUNTER MODE SYNCHRONIZATION

When in 8-Bit Counter mode, the incrementing edge on the T0CKI pin must be synchronized to the instruction clock. Synchronization can be accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the instruction clock. The high and low periods of the external clocking source must meet the timing requirements as shown in **Section 23.0 "Electrical Specifications"**.

12.3 Timer1 Prescaler

Timer1 has four prescaler options allowing 1, 2, 4 or 8 divisions of the clock input. The T1CKPS bits of the T1CON register control the prescale counter. The prescale counter is not directly readable or writable; however, the prescaler counter is cleared upon a write to TMR1H or TMR1L.

12.4 Timer1 Oscillator

A dedicated low-power 32.768 kHz oscillator circuit is built-in between pins T1OSI (input) and T1OSO (amplifier output). This internal circuit is to be used in conjunction with an external 32.768 kHz crystal.

The oscillator circuit is enabled by setting the T1OSCEN bit of the T1CON register. The oscillator will continue to run during Sleep.

Note:	The oscillator requires a start-up and stabilization time before use. Thus, T1OSCEN should be set and a suitable delay observed prior to enabling Timer1.

12.5 Timer1 Operation in Asynchronous Counter Mode

If control bit $\overline{\text{T1SYNC}}$ of the T1CON register is set, the external clock input is not synchronized. The timer increments asynchronously to the internal phase clocks. If external clock source is selected then the timer will continue to run during Sleep and can generate an interrupt on overflow, which will wake-up the processor. However, special precautions in software are needed to read/write the timer (see **Section 12.5.1 "Reading and Writing Timer1 in Asynchronous Counter Mode"**).

Note:	When switching from synchronous to asynchronous operation, it is possible to skip an increment. When switching from asynchronous to synchronous operation, it is possible to produce an additional increment.

12.5.1 READING AND WRITING TIMER1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L while the timer is running from an external asynchronous clock will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself, poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers, while the register is incrementing. This may produce an unpredictable value in the TMR1H:TMR1L register pair.

12.11 Timer1 Control Register

The Timer1 Control register (T1CON), shown in Register 12-1, is used to control Timer1 and select the various features of the Timer1 module.

REGISTER 12-1: T1CON: TIMER1 CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0
TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	—	TMR1ON
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-6 **TMR1CS<1:0>:** Timer1 Clock Source Select bits
11 = Timer1 clock source is Capacitive Sensing Oscillator (CAPOSC)
10 = Timer1 clock source is pin or oscillator:
If T1OSCEN = 0:
External clock from T1CKI pin (on the rising edge)
If T1OSCEN = 1:
Crystal oscillator on T1OSI/T1OSO pins
01 = Timer1 clock source is system clock (FOSC)
00 = Timer1 clock source is instruction clock (FOSC/4)

bit 5-4 **T1CKPS<1:0>:** Timer1 Input Clock Prescale Select bits
11 = 1:8 Prescale value
10 = 1:4 Prescale value
01 = 1:2 Prescale value
00 = 1:1 Prescale value

bit 3 **T1OSCEN:** LP Oscillator Enable Control bit
1 = Dedicated Timer1 oscillator circuit enabled
0 = Dedicated Timer1 oscillator circuit disabled

bit 2 **T1SYNC:** Timer1 External Clock Input Synchronization Control bit
TMR1CS<1:0> = 1X
1 = Do not synchronize external clock input
0 = Synchronize external clock input with system clock (FOSC)

TMR1CS<1:0> = 0X
This bit is ignored. Timer1 uses the internal clock when TMR1CS<1:0> = 1X.

bit 1 **Unimplemented:** Read as ‘0’

bit 0 **TMR1ON:** Timer1 On bit
1 = Enables Timer1
0 = Stops Timer1
Clears Timer1 Gate flip-flop

14.5 Software Control

The software portion of the capacitive sensing module is required to determine the change in frequency of the capacitive sensing oscillator. This is accomplished by the following:

- Setting a fixed time base to acquire counts on Timer0 or Timer1
- Establishing the nominal frequency for the capacitive sensing oscillator
- Establishing the reduced frequency for the capacitive sensing oscillator due to an additional capacitive load
- Set the frequency threshold

14.5.1 NOMINAL FREQUENCY (NO CAPACITIVE LOAD)

To determine the nominal frequency of the capacitive sensing oscillator:

- Remove any extra capacitive load on the selected CPSx pin
- At the start of the fixed time base, clear the timer resource
- At the end of the fixed time base save the value in the timer resource

The value of the timer resource is the number of oscillations of the capacitive sensing oscillator for the given time base. The frequency of the capacitive sensing oscillator is equal to the number of counts on in the timer divided by the period of the fixed time base.

14.5.2 REDUCED FREQUENCY (ADDITIONAL CAPACITIVE LOAD)

The extra capacitive load will cause the frequency of the capacitive sensing oscillator to decrease. To determine the reduced frequency of the capacitive sensing oscillator:

- Add a typical capacitive load on the selected CPSx pin
- Use the same fixed-time base as the nominal frequency measurement
- At the start of the fixed-time base, clear the timer resource
- At the end of the fixed-time base save the value in the timer resource

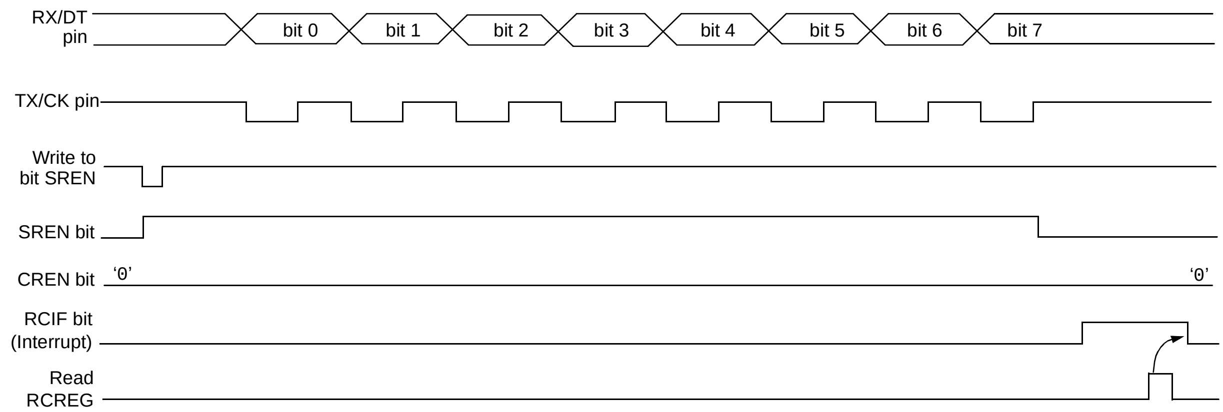
The value of the timer resource is the number of oscillations of the capacitive sensing oscillator with an additional capacitive load. The frequency of the capacitive sensing oscillator is equal to the number of counts on in the timer divided by the period of the fixed time base. This frequency should be less than the value obtained during the nominal frequency measurement.

14.5.3 FREQUENCY THRESHOLD

The frequency threshold should be placed midway between the value of nominal frequency and the reduced frequency of the capacitive sensing oscillator. Refer to Application Note AN1103, *Software Handling for Capacitive Sensing* (DS01103) for more detailed information the software required for capacitive sensing module.

Note:	For more information on general Capacitive Sensing refer to Application Notes: • AN1101, *Introduction to Capacitive Sensing* (DS01101) • *AN1102, Layout and Physical Design Guidelines for Capacitive Sensing* (DS01102).

FIGURE 16-8: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)



Note: Timing diagram demonstrates Synchronous Master mode with bit SREN = 1 and bit BRGH = 0.

TABLE 16-7: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
RCREG	AUSART Receive Data Register								0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000X	0000 000X
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for Synchronous Master Reception.

17.2.2 START AND STOP CONDITIONS

During times of no data transfer (Idle time), both the clock line (SCL) and the data line (SDA) are pulled high through external pull-up resistors. The Start and Stop conditions determine the start and stop of data transmission. The Start condition is defined as a high-to-low transition of the SDA line while SCL is high. The Stop condition is defined as a low-to-high transition of the SDA line while SCL is high.

Figure 17-9 shows the Start and Stop conditions. A master device generates these conditions for starting and terminating data transfer. Due to the definition of the Start and Stop conditions, when data is being transmitted, the SDA line can only change state when the SCL line is low.

17.2.3 ACKNOWLEDGE

After the valid reception of an address or data byte, the hardware automatically will generate the Acknowledge ($\overline{\text{ACK}}$) pulse and load the SSPBUF register with the received value currently in the SSPSR register. There are certain conditions that will cause the SSP module not to generate this $\overline{\text{ACK}}$ pulse. They include any or all of the following:

- The Buffer Full bit, BF of the SSPSTAT register, was set before the transfer was received.
- The SSP Overflow bit, SSPOV of the SSPCON register, was set before the transfer was received.
- The SSP Module is being operated in Firmware Master mode.

In such a case, the SSPSR register value is not loaded into the SSPBUF, but bit SSPIF of the PIR1 register is set. Table 17-2 shows the results of when a data transfer byte is received, given the status of bits BF and SSPOV. Flag bit BF is cleared by reading the SSPBUF register, while bit SSPOV is cleared through software.

FIGURE 17-9: START AND STOP CONDITIONS

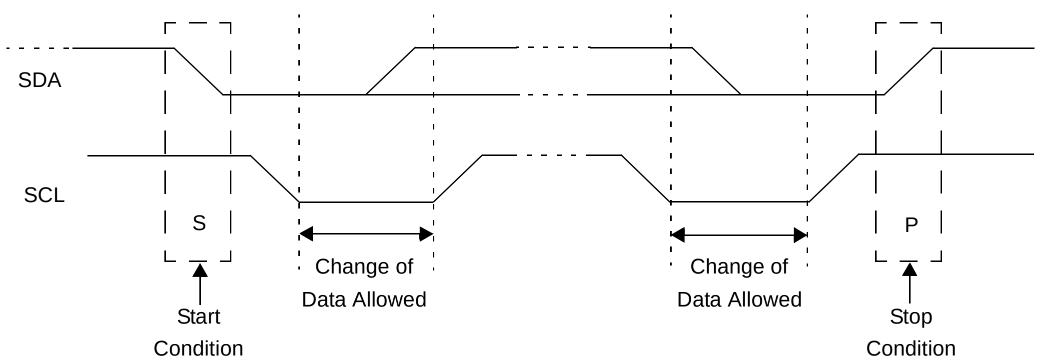


TABLE 17-2: DATA TRANSFER RECEIVED BYTE ACTIONS

Status Bits as Data Transfer is Received		SSPSR → SSPBUF	Generate $\overline{\text{ACK}}$ Pulse	Set bit SSPIF (SSP Interrupt occurs if enabled)
BF	SSPOV			
0	0	Yes	Yes	Yes
1	0	No	No	Yes
1	1	No	No	Yes
0	1	No	No	Yes

Note 1: Shaded cells show the conditions where the user software did not properly clear the overflow condition.

TABLE 21-2: PIC16(L)F722/3/4/6/7 INSTRUCTION SET

Mnemonic, Operands		Description	Cycles	14-Bit Opcode MSb			LSb	Status Affected	Notes
BYTE-ORIENTED FILE REGISTER OPERATIONS									
ADDWF	f, d	Add W and f	1	00	0111	dfff	ffff	C, DC, Z	**1, 2**
ANDWF	f, d	AND W with f	1	00	0101	dfff	ffff	Z	**1, 2**
CLRF	f	Clear f	1	00	0001	lfff	ffff	Z	**2**
CLRW	–	Clear W	1	00	0001	0xxx	xxxx	Z	
COMF	f, d	Complement f	1	00	1001	dfff	ffff	Z	**1, 2**
DECF	f, d	Decrement f	1	00	0011	dfff	ffff	Z	**1, 2**
DECFSZ	f, d	Decrement f, Skip if 0	1**(2)**	00	1011	dfff	ffff		**1, 2, 3**
INCF	f, d	Increment f	1	00	1010	dfff	ffff	Z	**1, 2**
INCFSZ	f, d	Increment f, Skip if 0	1**(2)**	00	1111	dfff	ffff		**1, 2, 3**
IORWF	f, d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	**1, 2**
MOVF	f, d	Move f	1	00	1000	dfff	ffff	Z	**1, 2**
MOVWF	f	Move W to f	1	00	0000	lfff	ffff		
NOP	–	No Operation	1	00	0000	0xx0	0000		
RLF	f, d	Rotate Left f through Carry	1	00	1101	dfff	ffff	C	**1, 2**
RRF	f, d	Rotate Right f through Carry	1	00	1100	dfff	ffff	C	**1, 2**
SUBWF	f, d	Subtract W from f	1	00	0010	dfff	ffff	C, DC, Z	**1, 2**
SWAPF	f, d	Swap nibbles in f	1	00	1110	dfff	ffff		**1, 2**
XORWF	f, d	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	**1, 2**
BIT-ORIENTED FILE REGISTER OPERATIONS									
BCF	f, b	Bit Clear f	1	01	00bb	bfff	ffff		**1, 2**
BSF	f, b	Bit Set f	1	01	01bb	bfff	ffff		**1, 2**
BTFSC	f, b	Bit Test f, Skip if Clear	1 **(2)**	01	10bb	bfff	ffff		**3**
BTFSS	f, b	Bit Test f, Skip if Set	1 **(2)**	01	11bb	bfff	ffff		**3**
LITERAL AND CONTROL OPERATIONS									
ADDLW	k	Add literal and W	1	11	111x	kkkk	kkkk	C, DC, Z	
ANDLW	k	AND literal with W	1	11	1001	kkkk	kkkk	Z	
CALL	k	Call Subroutine	2	10	0kkk	kkkk	kkkk		
CLRWDT	–	Clear Watchdog Timer	1	00	0000	0110	0100	$\overline{TO}$, $\overline{PD}$	
GOTO	k	Go to address	2	10	1kkk	kkkk	kkkk		
IORLW	k	Inclusive OR literal with W	1	11	1000	kkkk	kkkk	Z	
MOVLW	k	Move literal to W	1	11	00xx	kkkk	kkkk		
RETFIE	–	Return from interrupt	2	00	0000	0000	1001		
RETLW	k	Return with literal in W	2	11	01xx	kkkk	kkkk		
RETURN	–	Return from Subroutine	2	00	0000	0000	1000		
SLEEP	–	Go into Standby mode	1	00	0000	0110	0011	$\overline{TO}$, $\overline{PD}$	
SUBLW	k	Subtract W from literal	1	11	110x	kkkk	kkkk	C, DC, Z	
XORLW	k	Exclusive OR literal with W	1	11	1010	kkkk	kkkk	Z	

Note 1: When an I/O register is modified as a function of itself (e.g., `MOVF PORTA, 1`), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and where applicable, d = 1), the prescaler will be cleared if assigned to the Timer0 module.

3: If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a `NOP`.

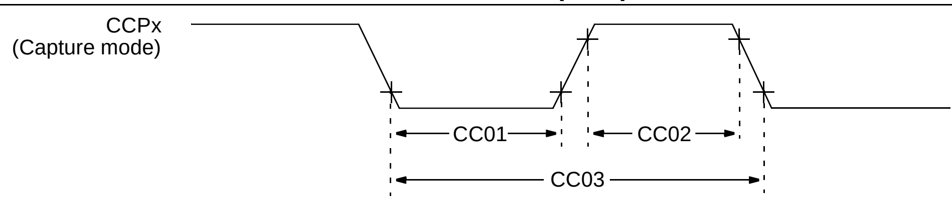
TABLE 23-5: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

Standard Operating Conditions (unless otherwise stated)
Operating Temperature -40°C ≤ TA ≤ +125°C

Param No.	Sym.	Characteristic		Min.	Typ†	Max.	Units	Conditions
40*	TT0H	T0CKI High Pulse Width	No Prescaler	0.5 TCY + 20	—	—	ns	
			With Prescaler	10	—	—	ns	
41*	TT0L	T0CKI Low Pulse Width	No Prescaler	0.5 TCY + 20	—	—	ns	
			With Prescaler	10	—	—	ns	
42*	TT0P	T0CKI Period		Greater of: 20 or $\frac{TCY + 40}{N}$	—	—	ns	N = prescale value (2, 4, ..., 256)
45*	TT1H	T1CKI High Time	Synchronous, No Prescaler	0.5 TCY + 20	—	—	ns	
			Synchronous, with Prescaler	15	—	—	ns	
			Asynchronous	30	—	—	ns	
46*	TT1L	T1CKI Low Time	Synchronous, No Prescaler	0.5 TCY + 20	—	—	ns	
			Synchronous, with Prescaler	15	—	—	ns	
			Asynchronous	30	—	—	ns	
47*	TT1P	T1CKI Input Period	Synchronous	Greater of: 30 or $\frac{TCY + 40}{N}$	—	—	ns	N = prescale value (1, 2, 4, 8)
			Asynchronous	60	—	—	ns	
48	FT1	Timer1 Oscillator Input Frequency Range (oscillator enabled by setting bit T1OSCEN)		32.4	32.768	33.1	kHz	
49*	TCKEZTMR1	Delay from External Clock Edge to Timer Increment		2 TOSC	—	7 TOSC	—	Timers in Sync mode

* These parameters are characterized but not tested.
† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 23-11: CAPTURE/COMPARE/PWM TIMINGS (CCP)



Note: Refer to Figure 23-2 for load conditions.

TABLE 23-6: CAPTURE/COMPARE/PWM REQUIREMENTS (CCP)

Standard Operating Conditions (unless otherwise stated)
Operating Temperature -40°C ≤ TA ≤ +125°C

Param No.	Sym.	Characteristic		Min.	Typ†	Max.	Units	Conditions
CC01*	TccL	CCPx Input Low Time	No Prescaler	0.5TCY + 20	—	—	ns	
			With Prescaler	20	—	—	ns	
CC02*	TccH	CCPx Input High Time	No Prescaler	0.5TCY + 20	—	—	ns	
			With Prescaler	20	—	—	ns	
CC03*	TccP	CCPx Input Period		$\frac{3TCY + 40}{N}$	—	—	ns	N = prescale value (1, 4 or 16)

* These parameters are characterized but not tested.
† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

TABLE 23-13: I^2C BUS DATA REQUIREMENTS

Param. No.	Symbol	Characteristic		Min.	Max.	Units	Conditions
SP100*	THIGH	Clock high time	100 kHz mode	4.0	—	μs	Device must operate at a minimum of 1.5 MHz
			400 kHz mode	0.6	—	μs	Device must operate at a minimum of 10 MHz
			SSP Module	1.5TCY	—		
SP101*	TLOW	Clock low time	100 kHz mode	4.7	—	μs	Device must operate at a minimum of 1.5 MHz
			400 kHz mode	1.3	—	μs	Device must operate at a minimum of 10 MHz
			SSP Module	1.5TCY	—		
SP102*	TR	SDA and SCL rise time	100 kHz mode	—	1000	ns	
			400 kHz mode	20 + 0.1CB	300	ns	CB is specified to be from 10-400 pF
SP103*	TF	SDA and SCL fall time	100 kHz mode	—	250	ns	
			400 kHz mode	20 + 0.1CB	250	ns	CB is specified to be from 10-400 pF
SP106*	THD:DAT	Data input hold time	100 kHz mode	0	—	ns	
			400 kHz mode	0	0.9	μs	
SP107*	TSU:DAT	Data input setup time	100 kHz mode	250	—	ns	**(Note 2)**
			400 kHz mode	100	—	ns	
SP109*	TAA	Output valid from clock	100 kHz mode	—	3500	ns	**(Note 1)**
			400 kHz mode	—	—	ns	
SP110*	TBUF	Bus free time	100 kHz mode	4.7	—	μs	Time the bus must be free before a new transmission can start
			400 kHz mode	1.3	—	μs	
SP111	CB	Bus capacitive loading		—	400	pF	

* These parameters are characterized but not tested.

Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCL to avoid unintended generation of Start or Stop conditions.

2: A Fast mode (400 kHz) I^2C bus device can be used in a Standard mode (100 kHz) I^2C bus system, but the requirement TSU:DAT ≥ 250 ns must then be met. This will automatically be the case if the device does not stretch the low period of the SCL signal. If such a device does stretch the low period of the SCL signal, it must output the next data bit to the SDA line TR max. + TSU:DAT = 1000 + 250 = 1250 ns (according to the Standard mode I^2C bus specification), before the SCL line is released.

FIGURE 24-5: PIC16F722/3/4/6/7 MAXIMUM IDD vs. VDD OVER FOSC, EXTRC MODE, VCAP = 0.1 µF

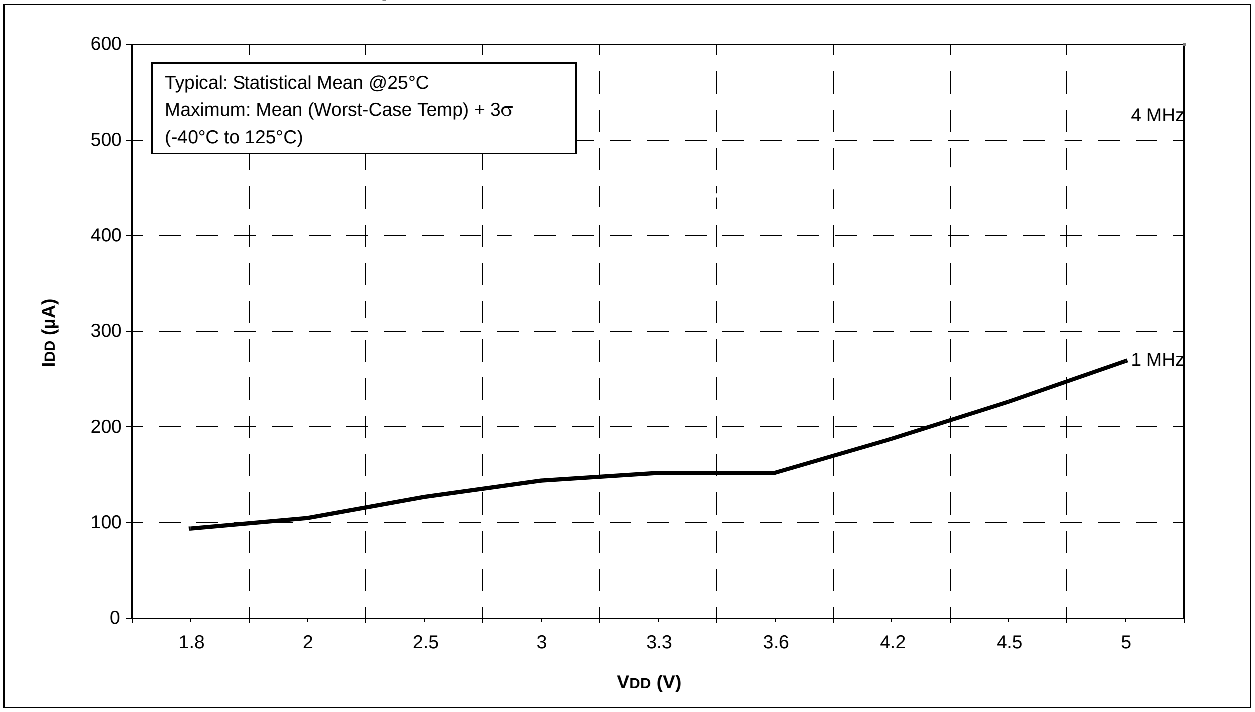


FIGURE 24-6: PIC16LF722/3/4/6/7 MAXIMUM IDD vs. VDD OVER FOSC, EXTRC MODE

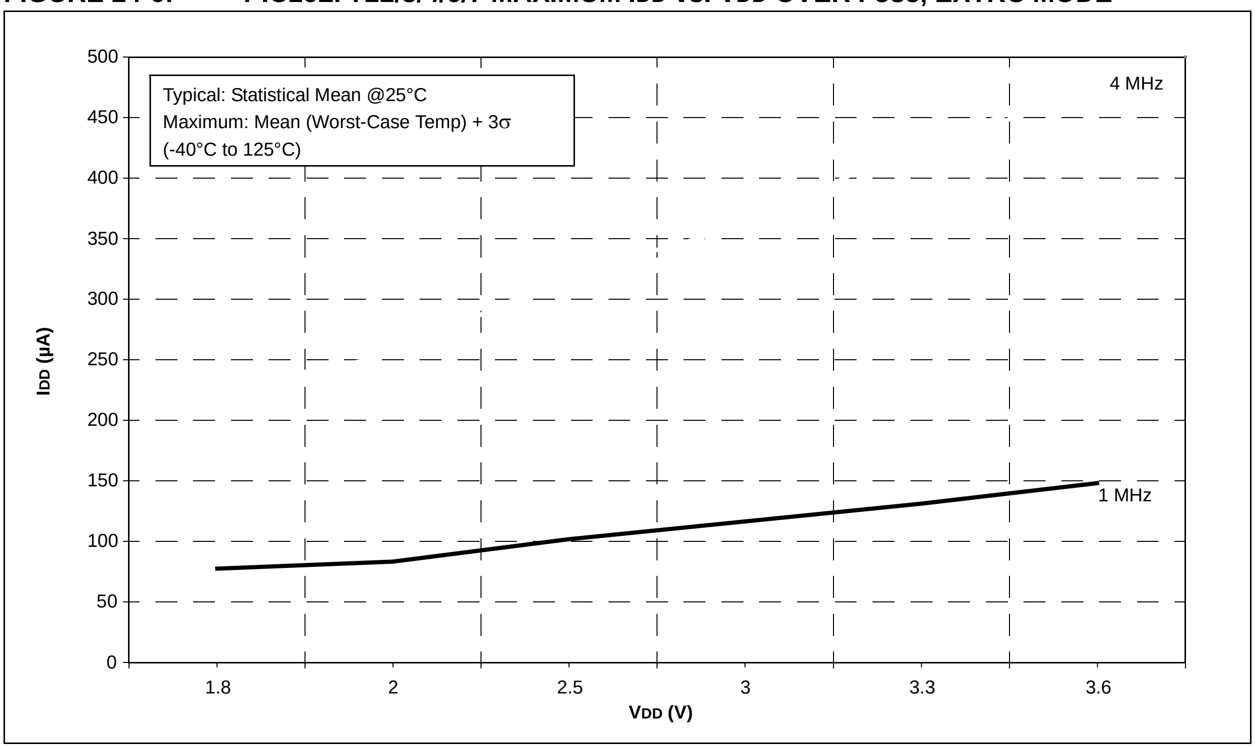


FIGURE 24-37: PIC16F722/3/4/6/7 CAP SENSE MEDIUM POWER IPD vs. VDD, VCAP = 0.1 µF

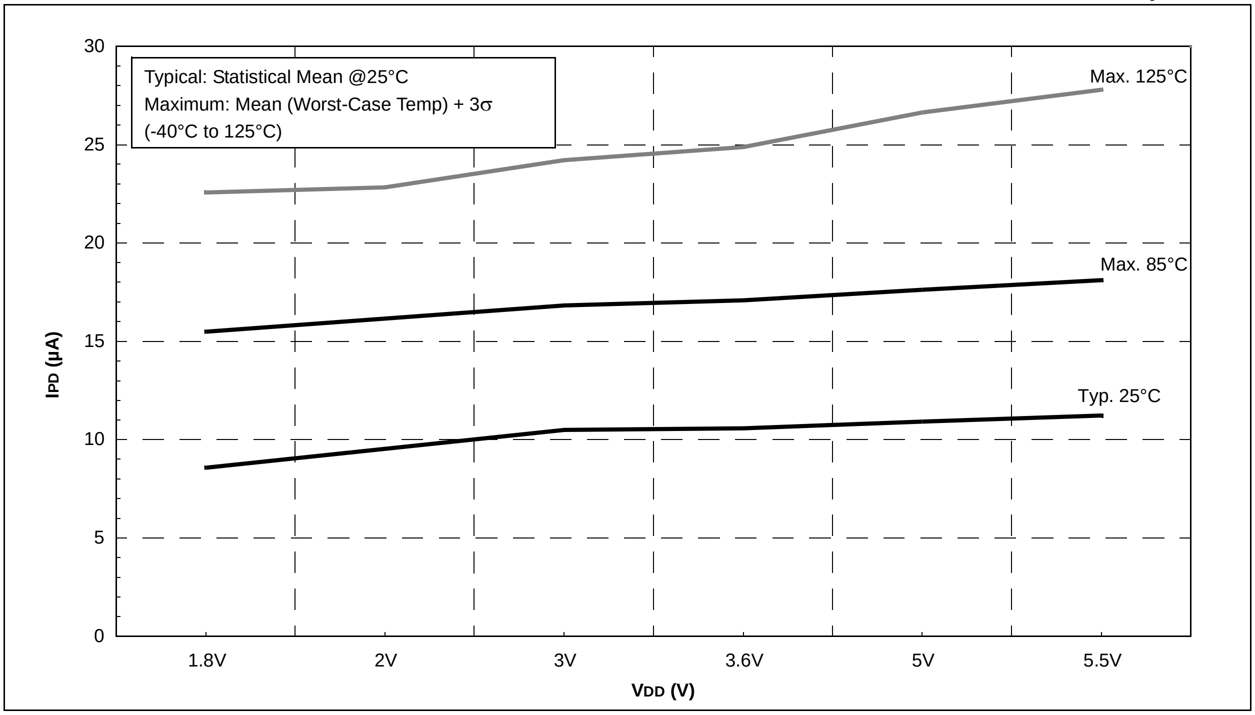
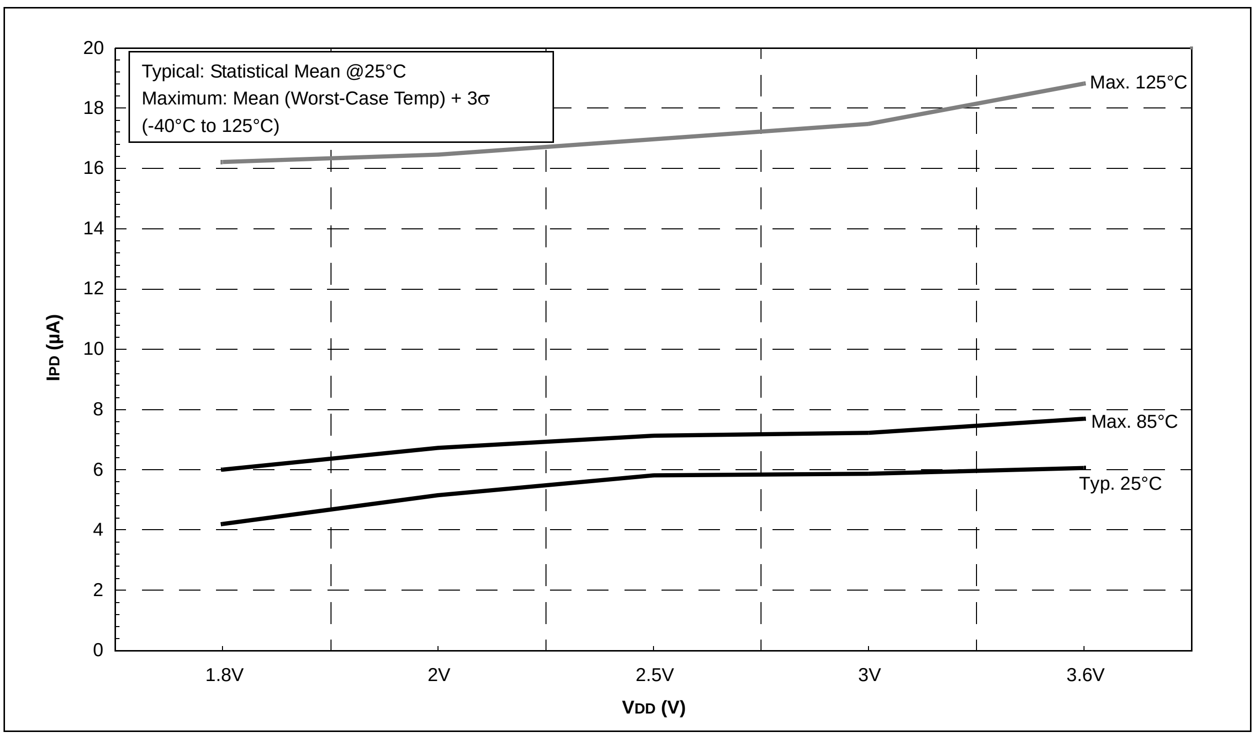


FIGURE 24-38: PIC16LF722/3/4/6/7 CAP SENSE MEDIUM POWER IPD vs. VDD



Note the following details of the code protection feature on Microchip devices:

- Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as "unbreakable."

Code protection is constantly evolving. We at Microchip are committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or other copyrighted work, you may have a right to sue for relief under that Act.

Trademarks

ISBN: 978-1-5224-0042-4

QUALITY MANAGEMENT SYSTEM CERTIFIED BY DNV

═ ISO/TS 16949 ═

Microchip received ISO/TS-16949:2009 certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona; Gresham, Oregon and design centers in California and India. The Company's quality system processes and procedures are for its PIC® MCUs and dsPIC® DSCs, KEELOQ® code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.