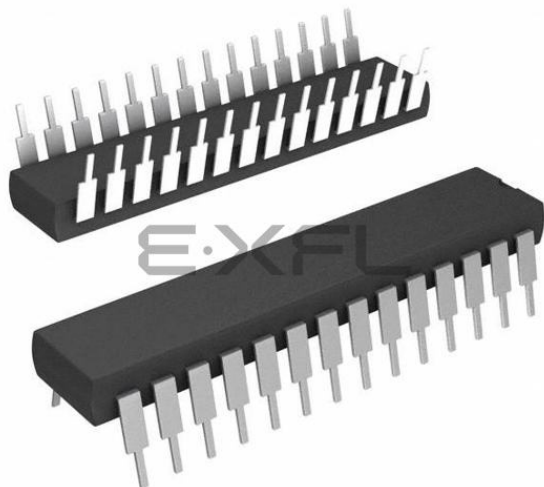




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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	25
Program Memory Size	3.5KB (2K x 14)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 11x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Through Hole
Package / Case	28-DIP (0.300", 7.62mm)
Supplier Device Package	28-SPDIP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf722-i-sp

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3.5 Brown-Out Reset (BOR)

Brown-out Reset is enabled by programming the BOREN<1:0> bits in the Configuration register. The brown-out trip point is selectable from two trip points via the BORV bit in the Configuration register.

Between the POR and BOR, complete voltage range coverage for execution protection can be implemented.

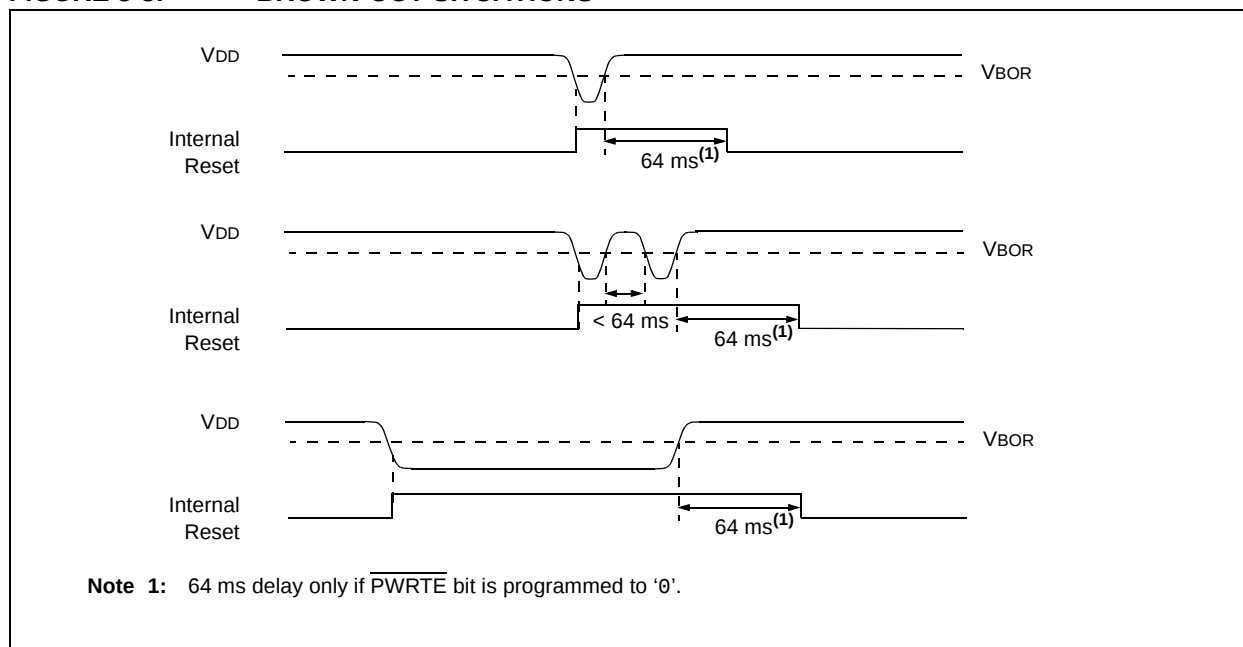
Two bits are used to enable the BOR. When BOREN = 11, the BOR is always enabled. When BOREN = 10, the BOR is enabled, but disabled during Sleep. When BOREN = 0X, the BOR is disabled.

If VDD falls below VBOR for greater than parameter (TBOR) (see **Section 23.0 “Electrical Specifications”**), the brown-out situation will reset the device. This will occur regardless of VDD slew rate. A Reset is not ensured to occur if VDD falls below VBOR for more than TBOR.

If VDD drops below VBOR while the Power-up Timer is running, the chip will go back into a Brown-out Reset and the Power-up Timer will be re-initialized. Once VDD rises above VBOR, the Power-up Timer will execute a 64 ms Reset.

Note: When erasing Flash program memory, the BOR is forced to enabled at the minimum BOR setting to ensure that any code protection circuitry is operating properly.

FIGURE 3-3: BROWN-OUT SITUATIONS



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TABLE 3-4: INITIALIZATION CONDITION FOR REGISTERS (CONTINUED)

Register	Address	Power-on Reset/ Brown-out Reset ⁽¹⁾	MCLR Reset/ WDT Reset	Wake-up from Sleep through Interrupt/Time out
PCON	8Eh	---- --qq	---- --uu ^(1,5)	---- --uu
T1GCON	8Fh	0000 0x00	uuuu uxuu	uuuu uxuu
OSCCON	90h	--10 qq--	--10 qq--	--uu qq--
OSCTUNE	91h	--00 0000	--uu uuuu	--uu uuuu
PR2	92h	1111 1111	1111 1111	uuuu uuuu
SSPAD	93h	0000 0000	0000 0000	uuuu uuuu
SSPMASK	93h	1111 1111	1111 1111	uuuu uuuu
SSPSTAT	94h	0000 0000	0000 0000	uuuu uuuu
WPUB	95h	1111 1111	1111 1111	uuuu uuuu
IOCB	96h	0000 0000	0000 0000	uuuu uuuu
TXSTA	98h	0000 -010	0000 -010	uuuu -uuu
SPBRG	99h	0000 0000	0000 0000	uuuu uuuu
APFCON	9Ch	---- --00	---- --00	---- --uu
FVRCON	9Dh	q000 --00	q000 --00	uuuu --uu
ADCON1	9Fh	-000 --00	-000 --00	-uuu --uu
CPSCON0	108h	0--- 0000	0--- 0000	u--- uuuu
CPSCON1	109h	---- 0000	---- 0000	---- uuuu
PMDATL	10Ch	xxxx xxxx	xxxx xxxx	uuuu uuuu
PMADRL	10Dh	xxxx xxxx	xxxx xxxx	uuuu uuuu
PMDATH	10Eh	--xx xxxx	--xx xxxx	--uu uuuu
PMADRH	10Fh	---x xxxx	---x xxxx	---u uuuu
ANSELA	185h	--11 1111	--11 1111	--uu uuuu
ANSELB	186h	--11 1111	--11 1111	--uu uuuu
ANSELD ⁽⁶⁾	188h	1111 1111	1111 1111	uuuu uuuu
ANSELE	189h	---- -111	---- -111	---- -uuu
PMCON1	18Ch	1--- ---0	1--- ---0	u--- ---u

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

- Note**
- 1: If V_{DD} goes too low, Power-on Reset will be activated and registers will be affected differently.
 - 2: One or more bits in INTCON and/or PIR1 and PIR2 will be affected (to cause wake-up).
 - 3: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).
 - 4: See Table 3-5 for Reset value for specific condition.
 - 5: If Reset was due to brown-out, then bit 0 = 0. All other Resets will cause bit 0 = u.
 - 6: PIC16F724/727/PIC16LF724/727 only.

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FIGURE 6-22: BLOCK DIAGRAM OF RE<2:0>

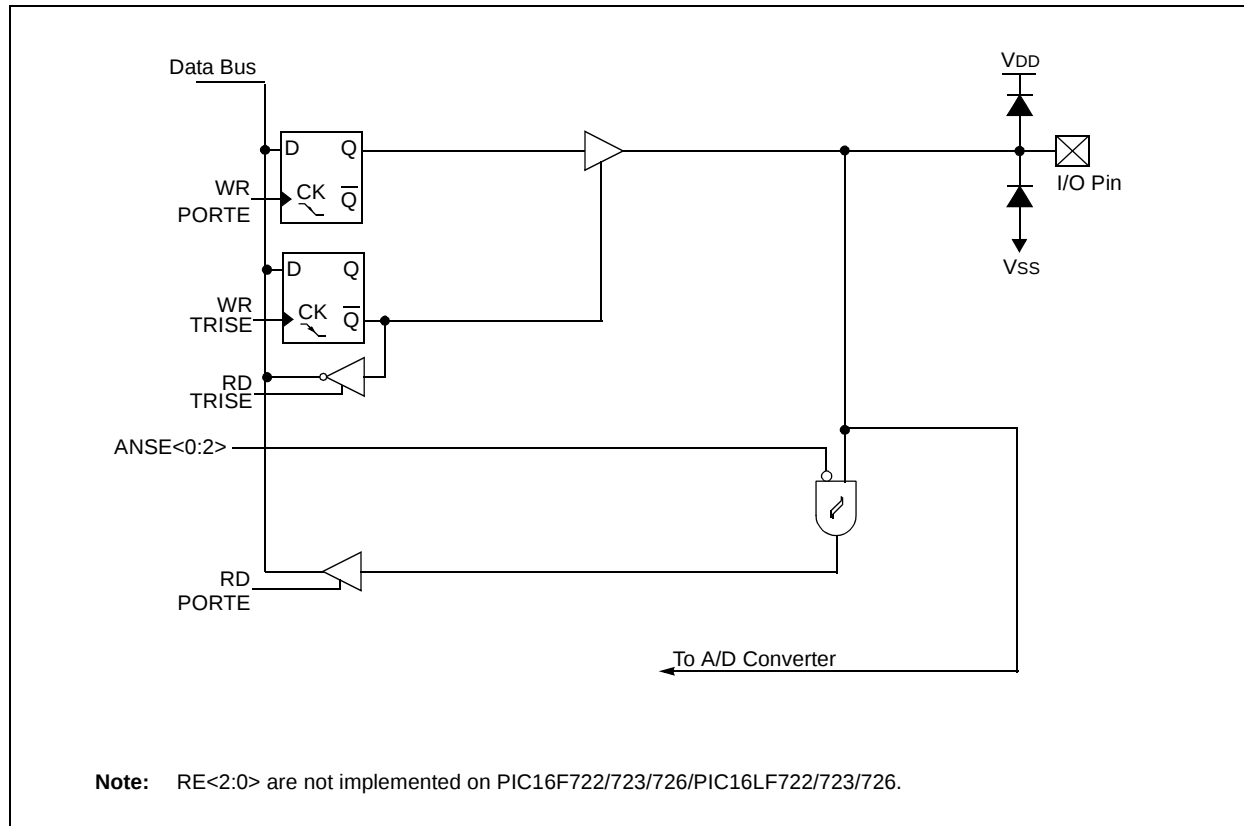
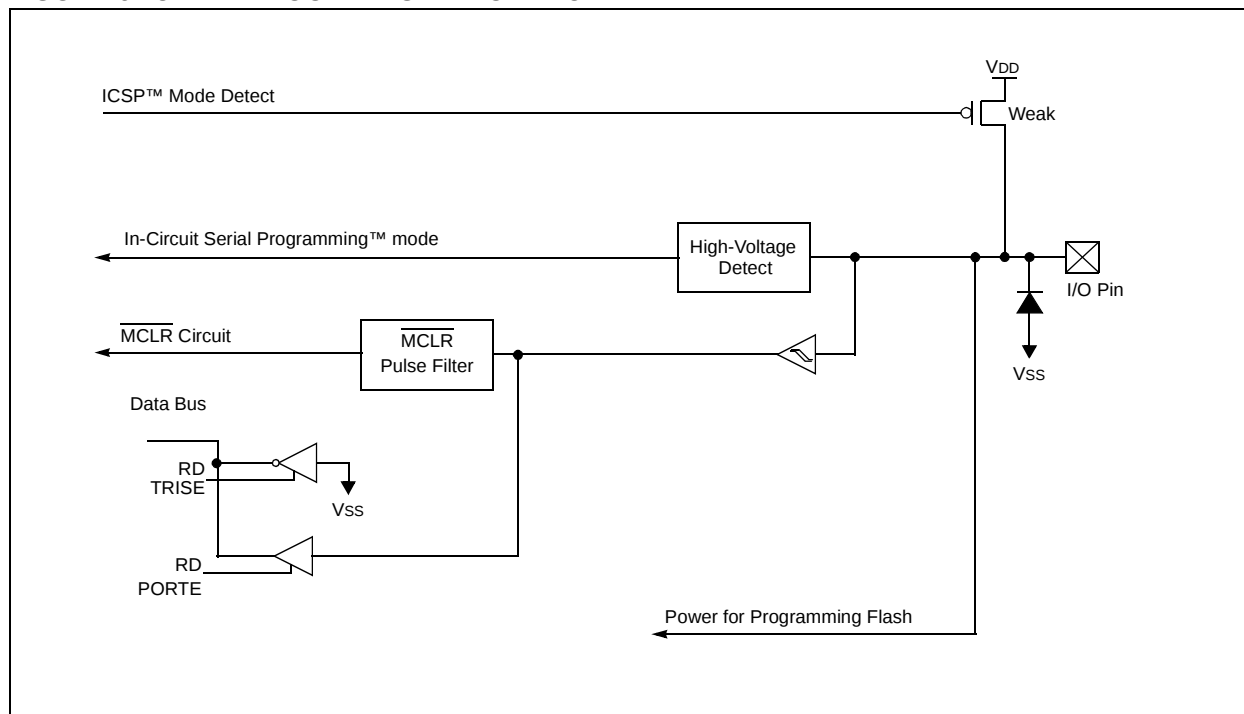


FIGURE 6-23: BLOCK DIAGRAM OF RE3



7.2 Clock Source Modes

Clock source modes can be classified as external or internal.

- Internal clock source (INTOSC) is contained within the oscillator module and derived from a 500 kHz high precision oscillator. The oscillator module has eight selectable output frequencies, with a maximum internal frequency of 16 MHz.
- External clock modes rely on external circuitry for the clock source. Examples are: oscillator modules (EC mode), quartz crystal resonators or ceramic resonators (LP, XT and HS modes) and Resistor-Capacitor (RC) mode circuits.

The system clock can be selected between external or internal clock sources via the FOSC bits of the Configuration Word 1.

7.3 Internal Clock Modes

The oscillator module has eight output frequencies derived from a 500 kHz high precision oscillator. The IRCF bits of the OSCCON register select the postscaler applied to the clock source dividing the frequency by 1, 2, 4 or 8. Setting the PLEN bit of the Configuration Word 1 locks the internal clock source to 16 MHz before the postscaler is selected by the IRCF bits. The PLEN bit must be set or cleared at the time of programming; therefore, only the upper or low four clock source frequencies are selectable in software.

7.3.1 INTOSC AND INTOSCIO MODES

The INTOSC and INTOSCIO modes configure the internal oscillators as the system clock source when the device is programmed using the oscillator selection or the FOSC<2:0> bits in the CONFIG1 register. See **Section 8.0 “Device Configuration”** for more information.

In INTOSC mode, OSC1/CLKIN is available for general purpose I/O. OSC2/CLKOUT outputs the selected internal oscillator frequency divided by 4. The CLKOUT signal may be used to provide a clock for external circuitry, synchronization, calibration, test or other application requirements.

In INTOSCIO mode, OSC1/CLKIN and OSC2/CLKOUT are available for general purpose I/O.

7.3.2 FREQUENCY SELECT BITS (IRCF)

The output of the 500 kHz INTOSC and 16 MHz INTOSC, with Phase-Locked Loop enabled, connect to a postscaler and multiplexer (see Figure 7-1). The Internal Oscillator Frequency Select bits (IRCF) of the OSCCON register select the frequency output of the internal oscillator. Depending upon the PLEN bit, one of four frequencies of two frequency sets can be selected via software:

If PLEN = 1, frequency selection is as follows:

- 16 MHz
- 8 MHz (default after Reset)
- 4 MHz
- 2 MHz

If PLEN = 0, frequency selection is as follows:

- 500 kHz
- 250 kHz (default after Reset)
- 125 kHz
- 62.5 kHz

Note: Following any Reset, the IRCF<1:0> bits of the OSCCON register are set to '10' and the frequency selection is set to 8 MHz or 250 kHz. The user can modify the IRCF bits to select a different frequency.

There is no start-up delay before a new frequency selected in the IRCF bits takes effect. This is because the old and new frequencies are derived from INTOSC via the postscaler and multiplexer.

Start-up delay specifications are located in the Table 23-2 in **Section 23.0 “Electrical Specifications”**.

14.1 Analog MUX

The capacitive sensing module can monitor up to 16 inputs. The capacitive sensing inputs are defined as CPS<15:0>. To determine if a frequency change has occurred the user must:

- Select the appropriate CPS pin by setting the CPSCH<3:0> bits of the CPSCON1 register
- Set the corresponding ANSEL bit
- Set the corresponding TRIS bit
- Run the software algorithm

Selection of the CPSx pin while the module is enabled will cause the capacitive sensing oscillator to be on the CPSx pin. Failure to set the corresponding ANSEL and TRIS bits can cause the capacitive sensing oscillator to stop, leading to false frequency readings.

14.2 Capacitive Sensing Oscillator

The capacitive sensing oscillator consists of a constant current source and a constant current sink, to produce a triangle waveform. The CPSOUT bit of the CPSCON0 register shows the status of the capacitive sensing oscillator, whether it is a sinking or sourcing current. The oscillator is designed to drive a capacitive load (single PCB pad) and at the same time, be a clock source to either Timer0 or Timer1. The oscillator has three different current settings as defined by CPSRNG<1:0> of the CPSCON0 register. The different current settings for the oscillator serve two purposes:

- Maximize the number of counts in a timer for a fixed time base
- Maximize the count differential in the timer during a change in frequency

14.3 Timer resources

To measure the change in frequency of the capacitive sensing oscillator, a fixed time base is required. For the period of the fixed time base, the capacitive sensing oscillator is used to clock either Timer0 or Timer1. The frequency of the capacitive sensing oscillator is equal to the number of counts in the timer divided by the period of the fixed time base.

14.4 Fixed Time Base

To measure the frequency of the capacitive sensing oscillator, a fixed time base is required. Any timer resource or software loop can be used to establish the fixed time base. It is up to the end user to determine the method in which the fixed time base is generated.

Note: The fixed time base can not be generated by timer resource the capacitive sensing oscillator is clocking.

14.4.1 TIMER0

To select Timer0 as the timer resource for the capacitive sensing module:

- Set the T0XCS bit of the CPSCON0 register
- Clear the T0CS bit of the OPTION register

When Timer0 is chosen as the timer resource, the capacitive sensing oscillator will be the clock source for Timer0. Refer to **Section 11.0 “Timer0 Module”** for additional information.

14.4.2 TIMER1

To select Timer1 as the timer resource for the capacitive sensing module, set the TMR1CS<1:0> of the T1CON register to '11'. When Timer1 is chosen as the timer resource, the capacitive sensing oscillator will be the clock source for Timer1. Because the Timer1 module has a gate control, developing a time base for the frequency measurement can be simplified using either:

- The Timer0 overflow flag
- The Timer2 overflow flag
- The WDT overflow flag

It is recommend that one of these flags, in conjunction with the toggle mode of the Timer1 Gate, is used to develop the fixed time base required by the software portion of the capacitive sensing module. Refer to **Section 12.0 “Timer1 Module with Gate Control”** for additional information.

TABLE 14-1: TIMER1 ENABLE FUNCTION

TMR1ON	TMR1GE	Timer1 Operation
0	0	Off
0	1	Off
1	0	On
1	1	Count Enabled by input

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15.1 Capture Mode

In Capture mode, CCPRxH:CCPRxL captures the 16-bit value of the TMR1 register when an event occurs on pin CCPx. An event is defined as one of the following and is configured by the CCPxM<3:0> bits of the CCPxCON register:

- Every falling edge
- Every rising edge
- Every 4th rising edge
- Every 16th rising edge

When a capture is made, the Interrupt Request Flag bit CCPxIF of the PIRx register is set. The interrupt flag must be cleared in software. If another capture occurs before the value in the CCPRxH, CCPRxL register pair is read, the old captured value is overwritten by the new captured value (refer to Figure 15-1).

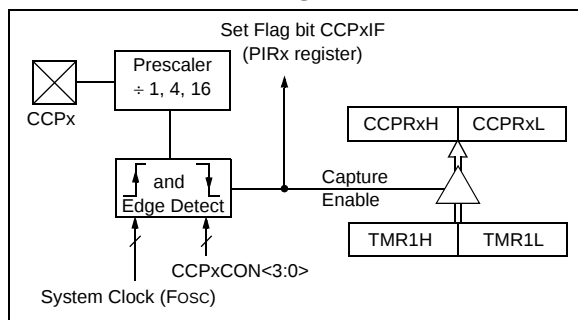
15.1.1 CCPx PIN CONFIGURATION

In Capture mode, the CCPx pin should be configured as an input by setting the associated TRIS control bit.

Either RC1 or RB3 can be selected as the CCP2 pin. Refer to **Section 6.1 “Alternate Pin Function”** for more information.

Note: If the CCPx pin is configured as an output, a write to the port can cause a capture condition.

FIGURE 15-1: CAPTURE MODE OPERATION BLOCK DIAGRAM



15.1.2 TIMER1 MODE SELECTION

Timer1 must be running in Timer mode or Synchronized Counter mode for the CCP module to use the capture feature. In Asynchronous Counter mode or when Timer1 is clocked at Fosc, the capture operation may not work.

15.1.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCPxIE interrupt enable bit of the PIRx register clear to avoid false interrupts. Additionally, the user should clear the CCPxIF interrupt flag bit of the PIRx register following any change in operating mode.

Note: Clocking Timer1 from the system clock (Fosc) should not be used in Capture Mode. In order for Capture Mode to recognize the trigger event on the CCPx pin, Timer1 must be clocked from the Instruction Clock (Fosc/4) or from an external clock source.

15.1.4 CCP PRESCALER

There are four prescaler settings specified by the CCPxM<3:0> bits of the CCPxCON register. Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. Any Reset will clear the prescaler counter.

Switching from one capture prescaler to another does not clear the prescaler and may generate a false interrupt. To avoid this unexpected operation, turn the module off by clearing the CCPxCON register before changing the prescaler (refer to Example 15-1).

EXAMPLE 15-1: CHANGING BETWEEN CAPTURE PRESCALERS

```
BANKSEL CCP1CON    ;Set Bank bits to point
                    ;to CCP1CON
CLRF    CCP1CON    ;Turn CCP module off
MOVLW   NEW_CAPT_PS;Load the W reg with
                    ; the new prescaler
MOVWF   CCP1CON    ; move value and CCP ON
                    ;Load CCP1CON with this
                    ; value
```

15.1.5 CAPTURE DURING SLEEP

Capture mode depends upon the Timer1 module for proper operation. There are two options for driving the Timer1 module in Capture mode. It can be driven by the instruction clock (Fosc/4), or by an external clock source.

If Timer1 is clocked by Fosc/4, then Timer1 will not increment during Sleep. When the device wakes from Sleep, Timer1 will continue from its previous state.

If Timer1 is clocked by an external clock source, then Capture mode will operate as defined in **Section 15.1 “Capture Mode”**.

TABLE 15-4: SUMMARY OF REGISTERS ASSOCIATED WITH COMPARE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ADCON0	—	—	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON	--00 0000	--00 0000
ANSELB	—	—	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	--11 1111	--11 1111
APFCON	—	—	—	—	—	—	SSSEL	CCP2SEL	---- --00	---- --00
CCP1CON	—	—	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	--00 0000	--00 0000
CCP2CON	—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	--00 0000	--00 0000
CCPRxL	Capture/Compare/PWM Register X Low Byte								xxxx xxxx	uuuu uuuu
CCPRxH	Capture/Compare/PWM Register X High Byte								xxxx xxxx	uuuu uuuu
INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIE2	—	—	—	—	—	—	—	CCP2IE	---- --0	---- --0
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIR2	—	—	—	—	—	—	—	CCP2IF	---- --0	---- --0
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	—	TMR1ON	0000 00-0	uuuu uu-u
T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GGO/DONE	T1GVAL	T1GSS1	T1GSS0	0000 0x00	0000 0x00
TMR1L	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register								xxxx xxxx	uuuu uuuu
TMR1H	Holding Register for the Most Significant Byte of the 16-bit TMR1 Register								xxxx xxxx	uuuu uuuu
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111

Legend: - = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the Compare.

15.3.2 PWM PERIOD

The PWM period is specified by the PR2 register of Timer2. The PWM period can be calculated using the formula of Equation 15-1.

EQUATION 15-1: PWM PERIOD

$$PWM\ Period = [(PR2) + 1] \cdot 4 \cdot T_{OSC} \cdot (TMR2\ Prescale\ Value)$$

Note: $T_{OSC} = 1/F_{OSC}$

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCPx pin is set. (Exception: If the PWM duty cycle = 0%, the pin will not be set.)
- The PWM duty cycle is latched from CCPRxL into CCPRxH.

Note: The Timer2 postscaler (refer to **Section 13.1 “Timer2 Operation”**) is not used in the determination of the PWM frequency.

15.3.3 PWM DUTY CYCLE

The PWM duty cycle is specified by writing a 10-bit value to multiple registers: CCPRxL register and DCxB<1:0> bits of the CCPxCON register. The CCPRxL contains the eight MSbs and the DCxB<1:0> bits of the CCPxCON register contain the two LSbs. CCPRxL and DCxB<1:0> bits of the CCPxCON register can be written to at any time. The duty cycle value is not latched into CCPRxH until after the period completes (i.e., a match between PR2 and TMR2 registers occurs). While using the PWM, the CCPRxH register is read-only.

Equation 15-2 is used to calculate the PWM pulse width.

Equation 15-3 is used to calculate the PWM duty cycle ratio.

EQUATION 15-2: PULSE WIDTH

$$Pulse\ Width = (CCPRxL:CCPxCON<5:4>) \cdot T_{OSC} \cdot (TMR2\ Prescale\ Value)$$

Note: $T_{OSC} = 1/F_{OSC}$

EQUATION 15-3: DUTY CYCLE RATIO

$$Duty\ Cycle\ Ratio = \frac{(CCPRxL:CCPxCON<5:4>)}{4(PR2 + 1)}$$

The CCPRxH register and a 2-bit internal latch are used to double buffer the PWM duty cycle. This double buffering is essential for glitchless PWM operation.

The 8-bit timer TMR2 register is concatenated with either the 2-bit internal system clock (F_{OSC}), or two bits of the prescaler, to create the 10-bit time base. The system clock is used if the Timer2 prescaler is set to 1:1.

When the 10-bit time base matches the CCPRxH and 2-bit latch, then the CCPx pin is cleared (refer to Figure 15-3).

16.1.2.4 Receive Framing Error

Each character in the receive FIFO buffer has a corresponding framing error Status bit. A framing error indicates that a Stop bit was not seen at the expected time. The framing error status is accessed via the FERR bit of the RCSTA register. The FERR bit represents the status of the top unread character in the receive FIFO. Therefore, the FERR bit must be read before reading the RCREG.

The FERR bit is read-only and only applies to the top unread character in the receive FIFO. A framing error (FERR = 1) does not preclude reception of additional characters. It is not necessary to clear the FERR bit. Reading the next character from the FIFO buffer will advance the FIFO to the next character and the next corresponding framing error.

The FERR bit can be forced clear by clearing the SPEN bit of the RCSTA register which resets the AUSART. Clearing the CREN bit of the RCSTA register does not affect the FERR bit. A framing error by itself does not generate an interrupt.

Note: If all receive characters in the receive FIFO have framing errors, repeated reads of the RCREG will not clear the FERR bit.
--

16.1.2.5 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before the FIFO is accessed. When this happens the OERR bit of the RCSTA register is set. The characters already in the FIFO buffer can be read but no additional characters will be received until the error is cleared. The error must be cleared by either clearing the CREN bit of the RCSTA register or by setting the AUSART by clearing the SPEN bit of the RCSTA register.

16.1.2.6 Receiving 9-bit Characters

The AUSART supports 9-bit character reception. When the RX9 bit of the RCSTA register is set the AUSART will shift nine bits into the RSR for each character received. The RX9D bit of the RCSTA register is the ninth and Most Significant data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RCREG.

16.1.2.7 Address Detection

A special Address Detection mode is available for use when multiple receivers share the same transmission line, such as in RS-485 systems. Address detection is enabled by setting the ADDEN bit of the RCSTA register.

Address detection requires 9-bit character reception. When address detection is enabled, only characters with the ninth data bit set will be transferred to the receive FIFO buffer, thereby setting the RCIF interrupt bit of the PIR1 register. All other characters will be ignored.

Upon receiving an address character, user software determines if the address matches its own. Upon address match, user software must disable address detection by clearing the ADDEN bit before the next Stop bit occurs. When user software detects the end of the message, determined by the message protocol used, software places the receiver back into the Address Detection mode by setting the ADDEN bit.

19.2 Wake-up Using Interrupts

When global interrupts are disabled (GIE cleared) and any interrupt source has both its interrupt enable bit and interrupt flag bit set, one of the following will occur:

- If the interrupt occurs **before** the execution of a SLEEP instruction, the SLEEP instruction will complete as a NOP. Therefore, the WDT and WDT prescaler and postscaler (if enabled) will not be cleared, the $\overline{TO}$ bit will not be set and the $\overline{PD}$ bit will not be cleared.
- If the interrupt occurs **during or after** the execution of a SLEEP instruction, the device will immediately wake-up from Sleep. The SLEEP instruction will be completely executed before the wake-up. Therefore, the WDT and WDT prescaler and postscaler (if enabled) will be cleared, the $\overline{TO}$ bit will be set and the $\overline{PD}$ bit will be cleared.

Even if the flag bits were checked before executing a SLEEP instruction, it may be possible for flag bits to become set before the SLEEP instruction completes. To determine whether a SLEEP instruction executed, test the $\overline{PD}$ bit. If the $\overline{PD}$ bit is set, the SLEEP instruction was executed as a NOP.

To ensure that the WDT is cleared, a CLRWD instruction should be executed before a SLEEP instruction.

FIGURE 19-1: WAKE-UP FROM SLEEP THROUGH INTERRUPT

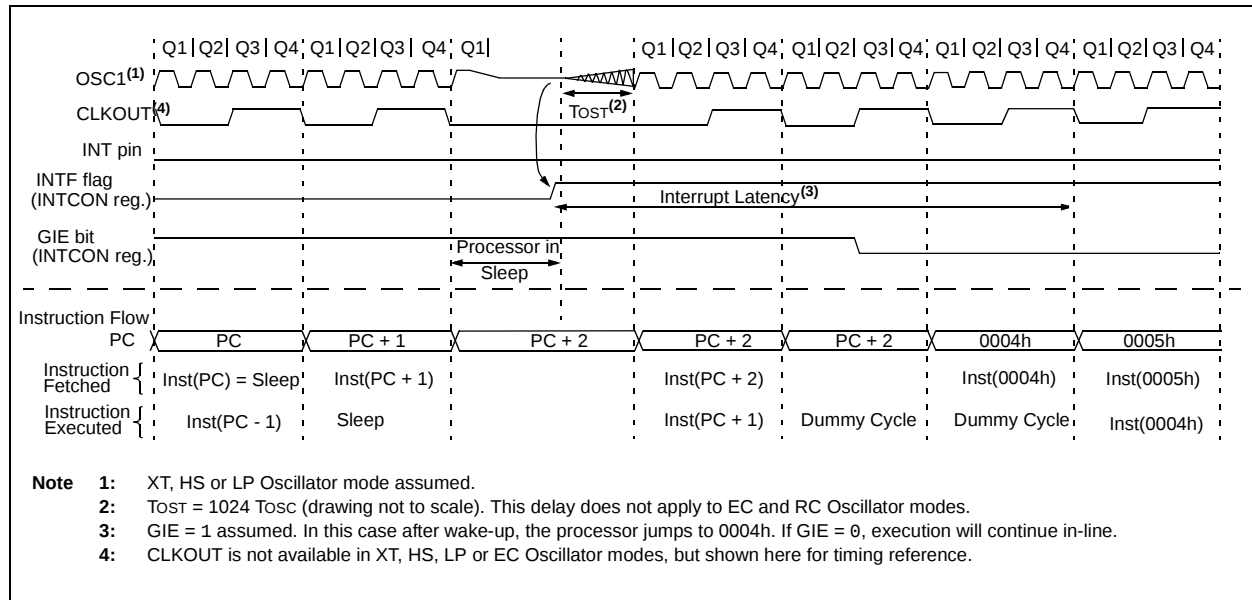


TABLE 19-1: SUMMARY OF REGISTERS ASSOCIATED WITH POWER-DOWN MODE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
IOCB	IOCB7	IOCB6	IOCB5	IOCB4	IOCB3	IOCB2	IOCB1	IOCB0	0000 0000	0000 0000
INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF	0000 0000	0000 0000
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIE2	—	—	—	—	—	—	—	CCP2IE	---- --0	---- --0
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
PIR2	—	—	—	—	—	—	—	CCP2IF	---- --0	---- --0

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used in Power-Down mode.

RETFIE	Return from Interrupt
Syntax:	[<i>label</i>] RETFIE
Operands:	None
Operation:	TOS → PC, 1 → GIE
Status Affected:	None
Description:	Return from Interrupt. Stack is POPed and Top-of-Stack (TOS) is loaded in the PC. Interrupts are enabled by setting Global Interrupt Enable bit, GIE (INTCON<7>). This is a 2-cycle instruction.
Words:	1
Cycles:	2
<u>Example:</u>	RETFIE After Interrupt PC = TOS GIE = 1

RETLW	Return with literal in W
Syntax:	[<i>label</i>] RETLW k
Operands:	0 ≤ k ≤ 255
Operation:	k → (W); TOS → PC
Status Affected:	None
Description:	The W register is loaded with the 8-bit literal 'k'. The program counter is loaded from the top of the stack (the return address). This is a 2-cycle instruction.
Words:	1
Cycles:	2
<u>Example:</u>	CALL TABLE;W contains table ;offset value • ;W now has table value • • ADDWF PC ;W = offset RETLW k1 ;Begin table RETLW k2 ; • • • RETLW kn ; End of table
	Before Instruction W = 0x07 After Instruction W = value of k8

RETURN	Return from Subroutine
Syntax:	[<i>label</i>] RETURN
Operands:	None
Operation:	TOS → PC
Status Affected:	None
Description:	Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a 2-cycle instruction.

PIC16(L)F722/3/4/6/7

FIGURE 23-9: BROWN-OUT RESET TIMING AND CHARACTERISTICS

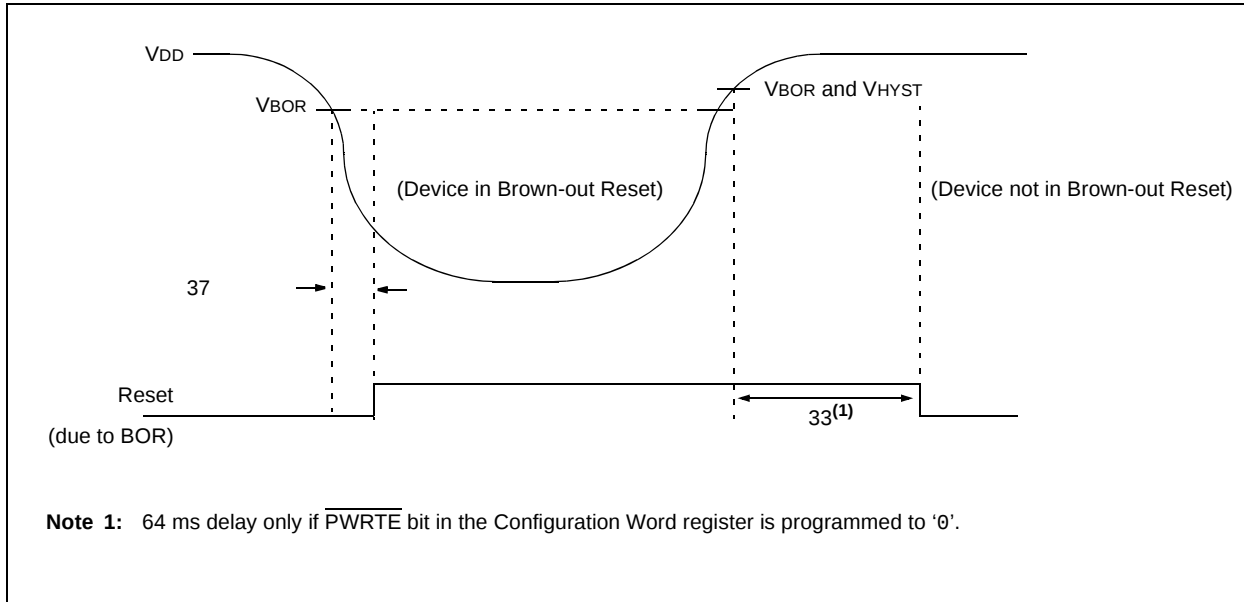


TABLE 23-4: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER, AND BROWN-OUT RESET PARAMETERS

Standard Operating Conditions (unless otherwise stated) Operating Temperature $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$							
Param No.	Sym.	Characteristic	Min.	Typ†	Max.	Units	Conditions
30	TMCL	MCLR Pulse Width (low)	2 5	— —	— —	μs μs	$V_{DD} = 3.3\text{-}5\text{V}$, -40°C to $+85^{\circ}\text{C}$ $V_{DD} = 3.3\text{-}5\text{V}$
31	TWDTLP	Low Power Watchdog Timer Time-out Period (No Prescaler)	10	18	27	ms	$V_{DD} = 3.3\text{V}\text{-}5\text{V}$
32	TOST	Oscillator Start-up Timer Period ^{(1), (2)}	—	1024	—	Tosc	(Note 3)
33*	TPWRT	Power-up Timer Period, $PWRT = 0$	40	65	140	ms	
34*	TIOZ	I/O high-impedance from MCLR Low or Watchdog Timer Reset	—	—	2.0	μs	
35	VBOR	Brown-out Reset Voltage	2.38 1.80	2.5 1.9	2.73 2.11	V	BORV=2.5V BORV=1.9V
36*	VHYS	Brown-out Reset Hysteresis	0	25	50	mV	-40°C to $+85^{\circ}\text{C}$
37*	TBORDC	Brown-out Reset DC Response Time	1	3	5 10	μs	$V_{DD} \leq V_{BOR}$, -40°C to $+85^{\circ}\text{C}$ $V_{DD} \leq V_{BOR}$

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (Tcy) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to the OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

2: By design.

3: Period of the slower clock.

4: To ensure these voltage tolerances, V_{DD} and V_{SS} must be capacitively decoupled as close to the device as possible. 0.1 μF and 0.01 μF values in parallel are recommended.

FIGURE 23-10: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS

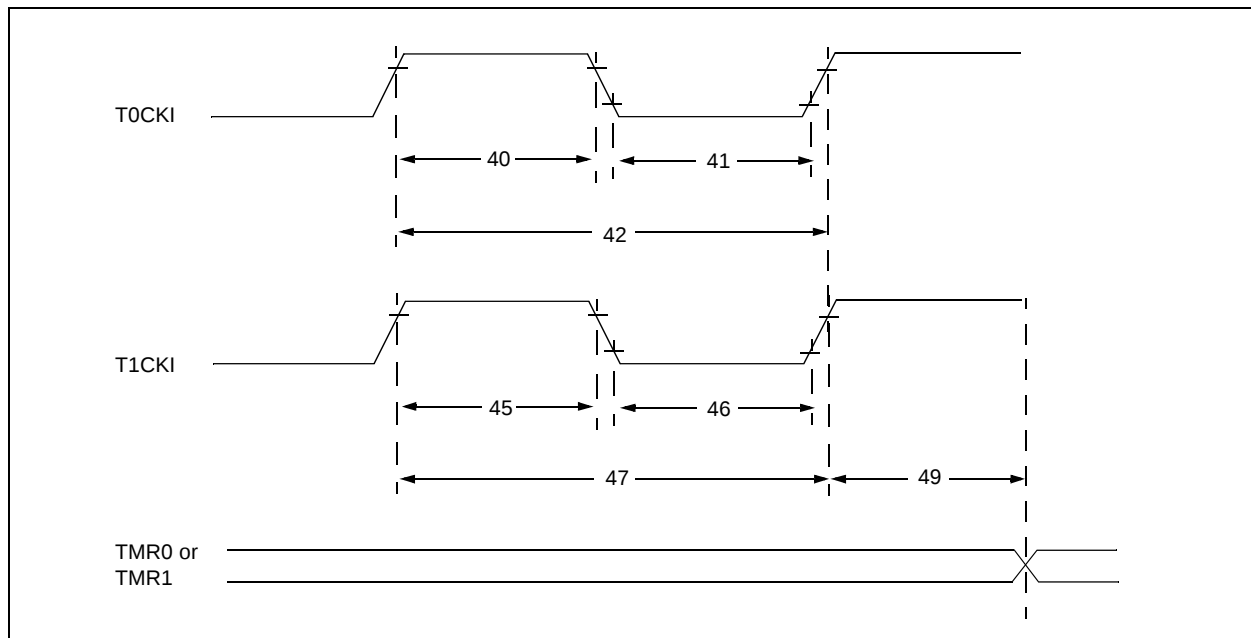


FIGURE 24-47: PIC16F722/3/4/6/7 WDT IPD vs. VDD, VCAP = 0.1 μ F

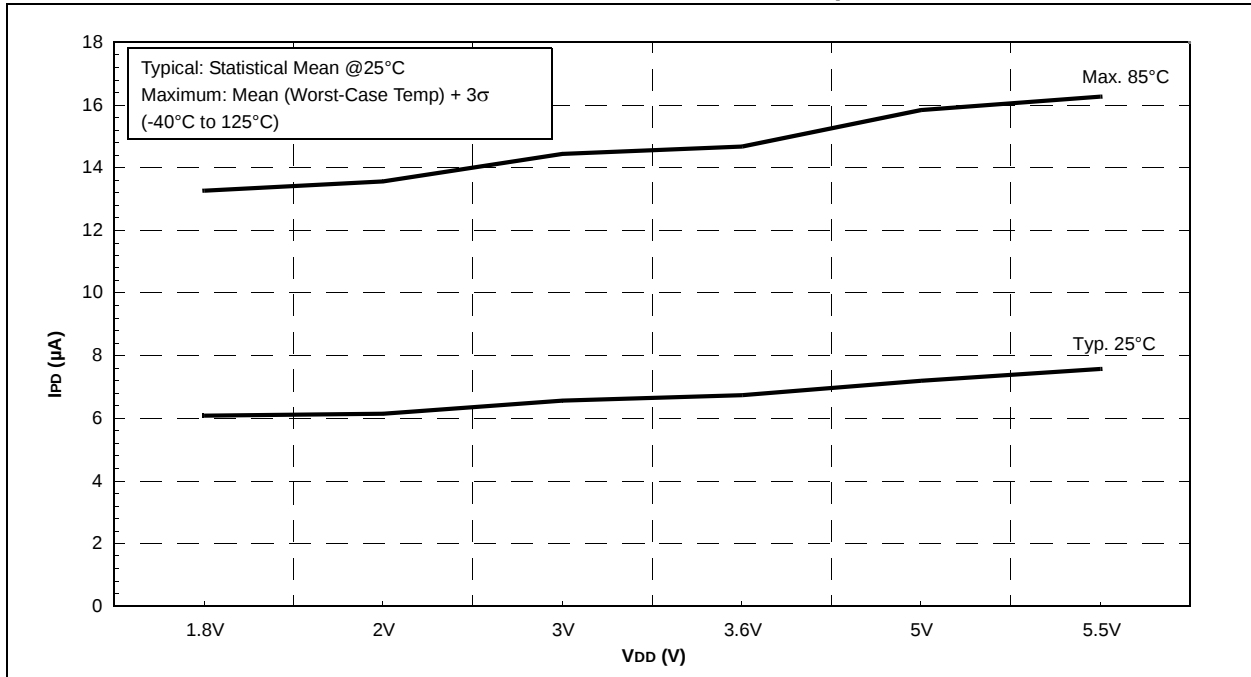


FIGURE 24-48: PIC16LF722/3/4/6/7 WDT IPD vs. VDD

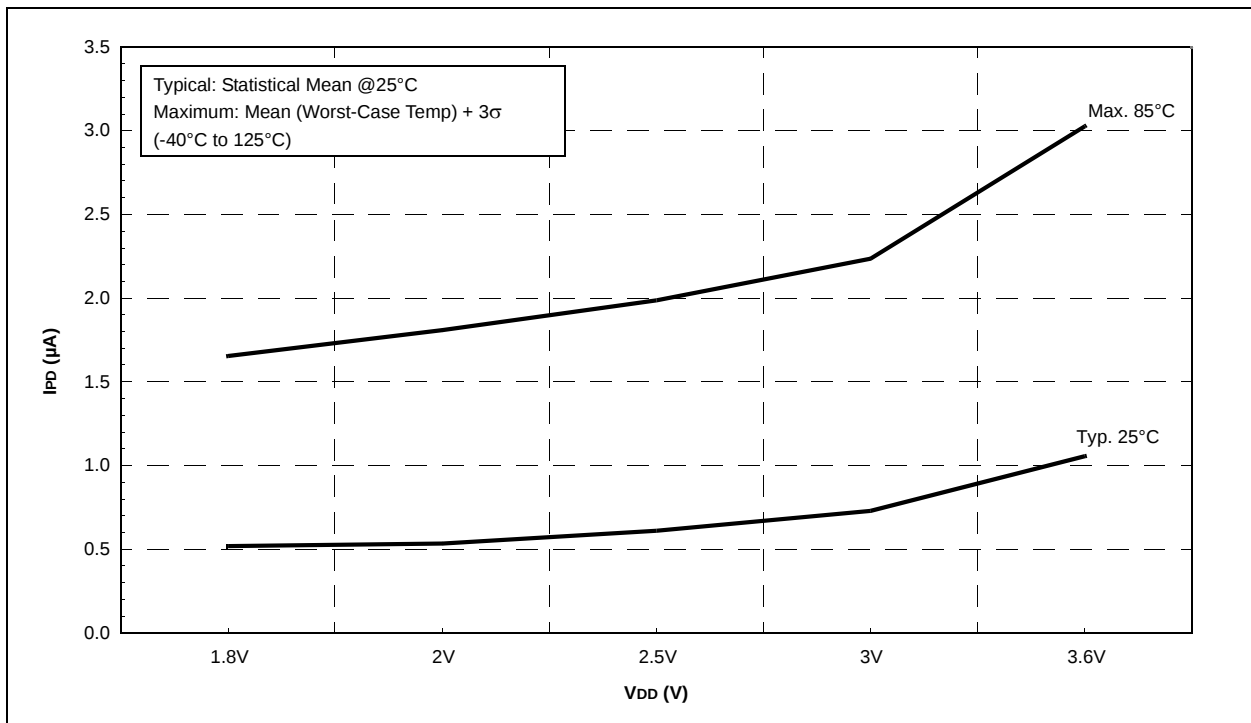


FIGURE 24-55: Vol vs. IOL OVER TEMPERATURE, VDD = 5.5V

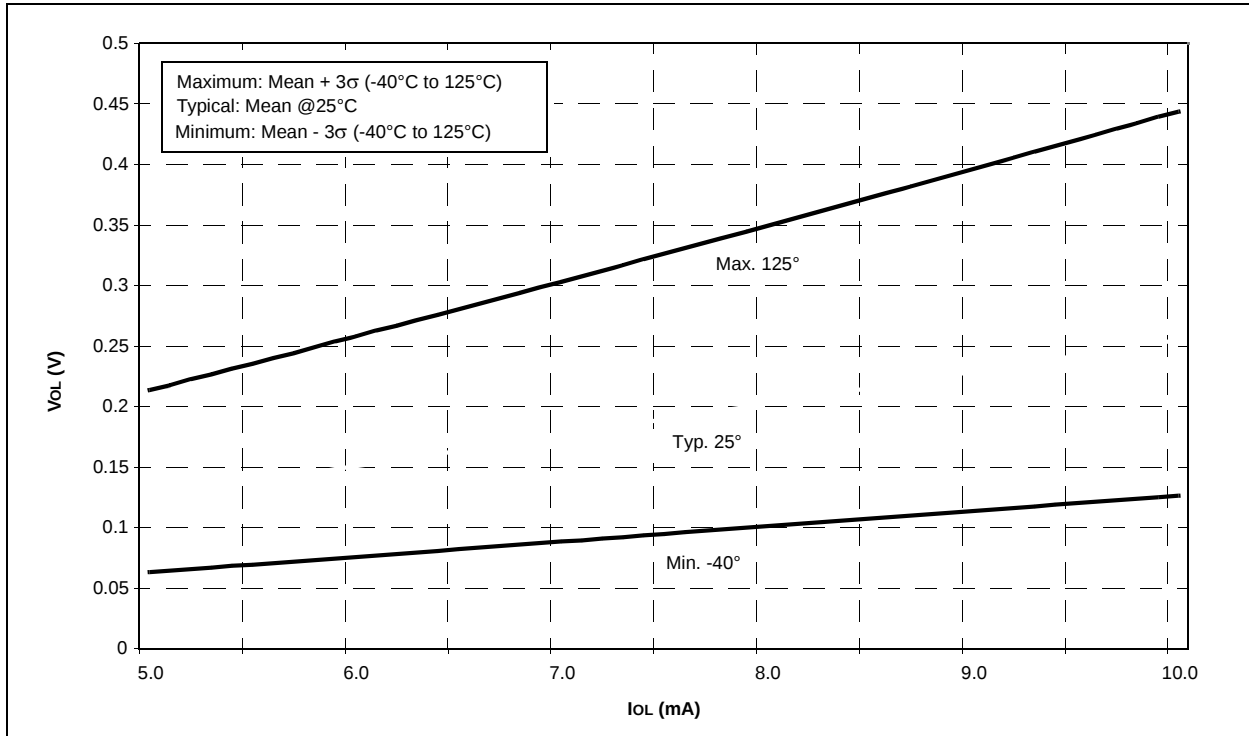


FIGURE 24-56: Vol vs. IOL OVER TEMPERATURE, VDD = 3.6V

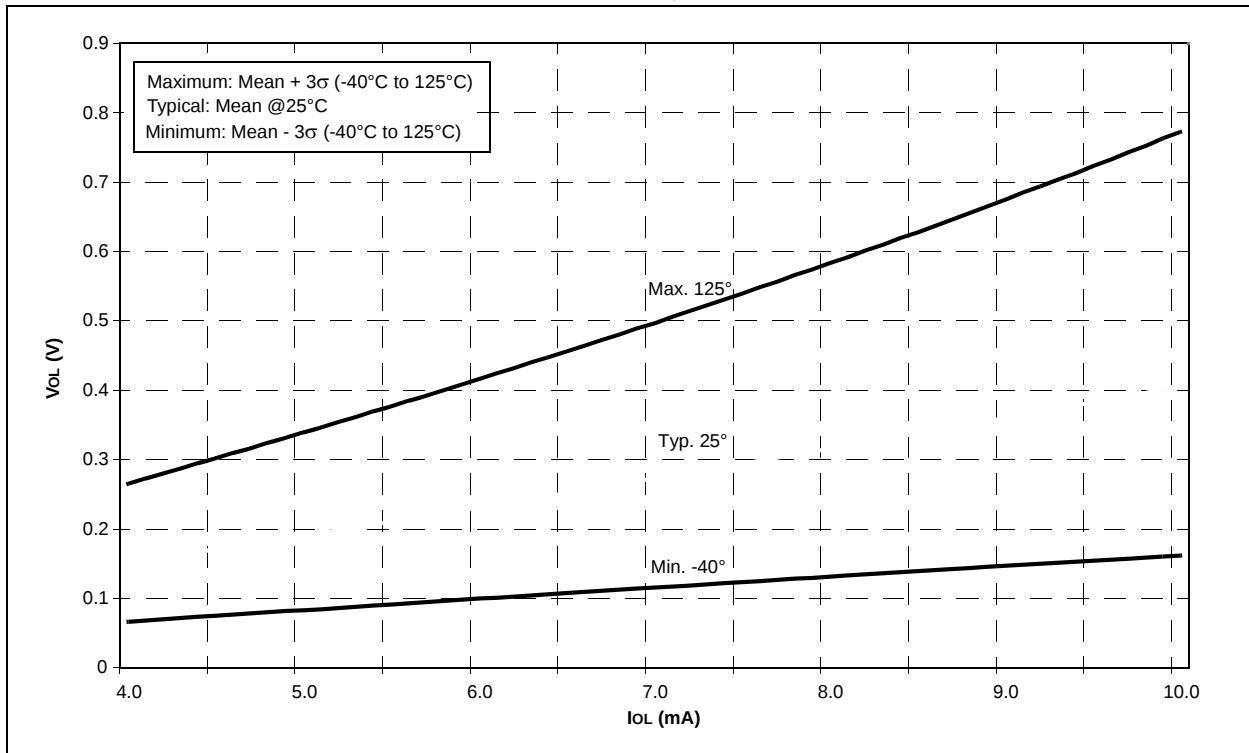


FIGURE 24-59: PIC16F722/3/4/6/7 WDT TIME-OUT PERIOD

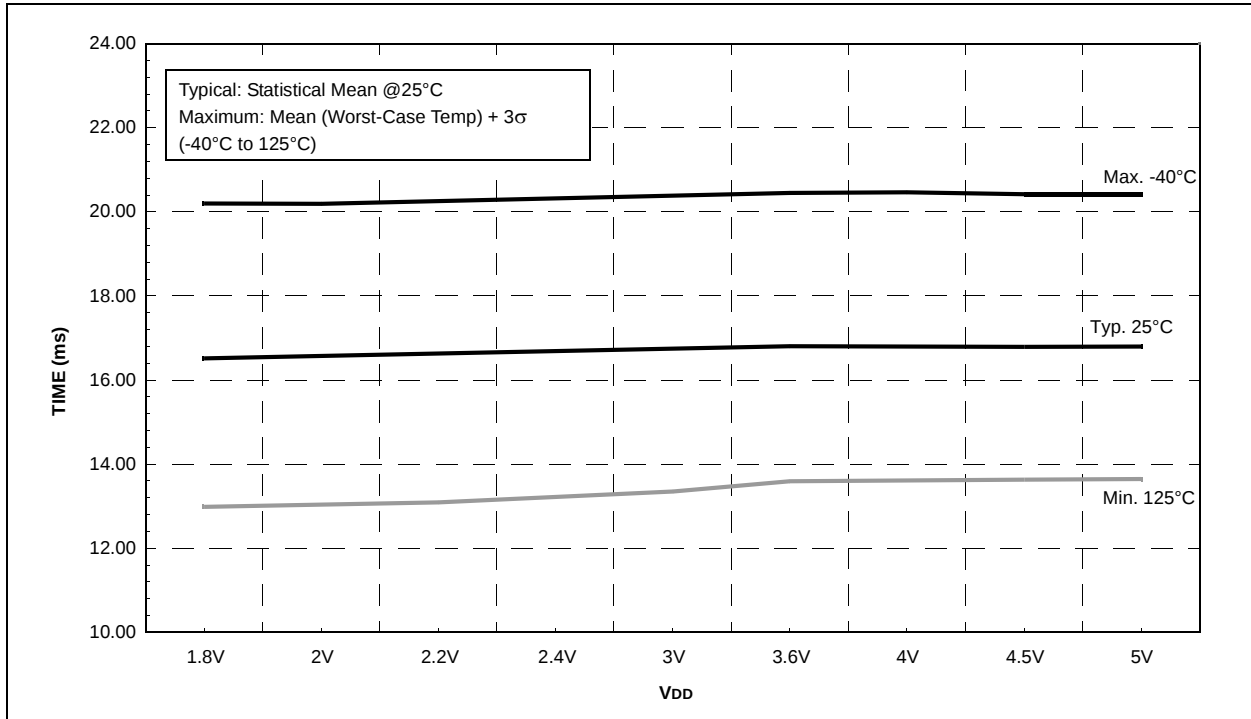
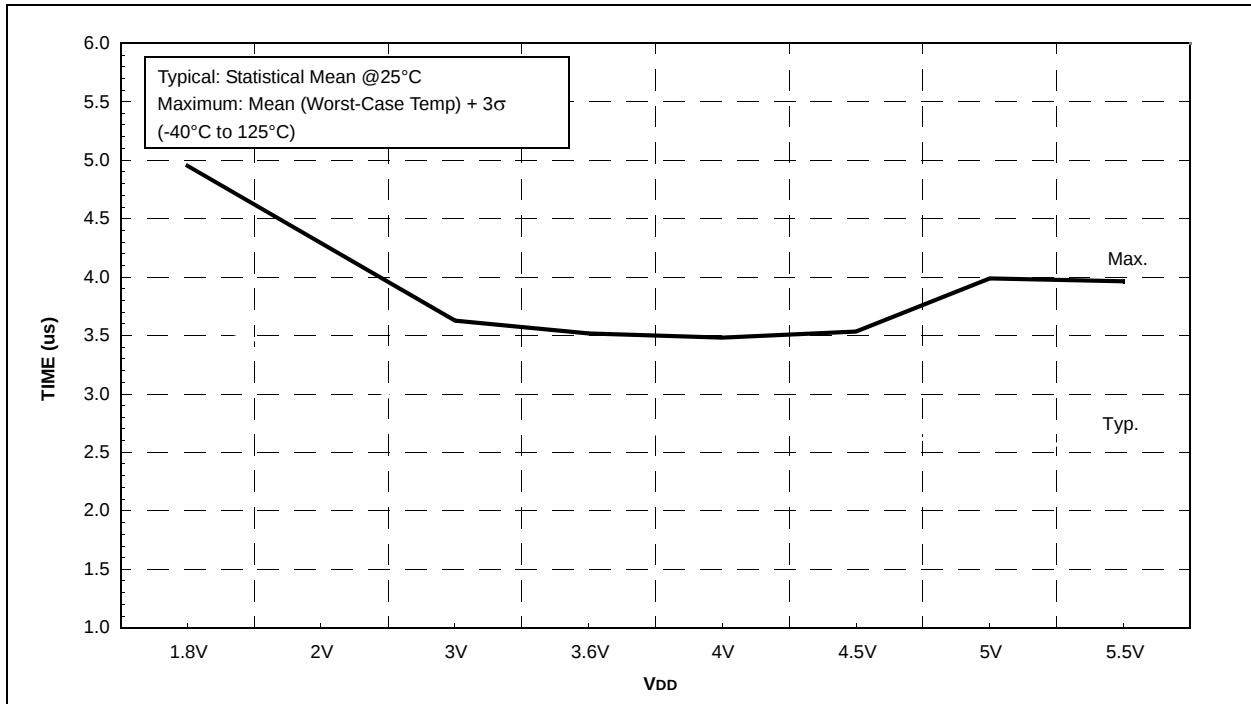


FIGURE 24-60: PIC16F722/3/4/6/7 HFINTOSC WAKE-UP FROM SLEEP START-UP TIME

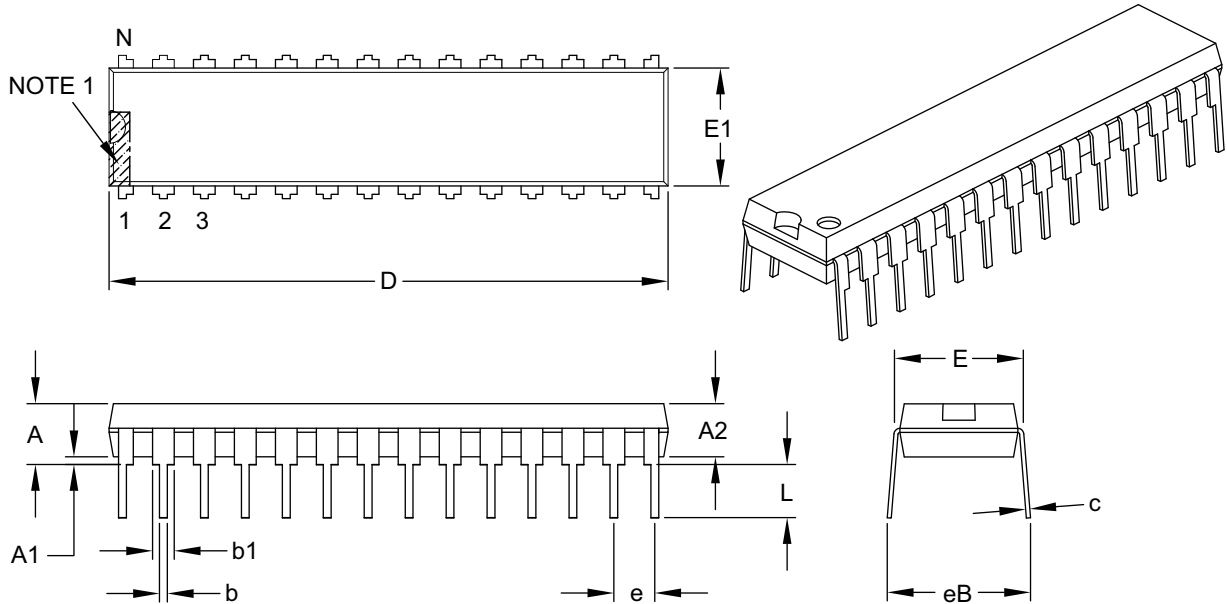


25.2 Package Details

The following sections give the technical details of the packages.

28-Lead Skinny Plastic Dual In-Line (SP) – 300 mil Body [SPDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Units		INCHES		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N	28		
Pitch	e	.100 BSC		
Top to Seating Plane	A	–	–	.200
Molded Package Thickness	A2	.120	.135	.150
Base to Seating Plane	A1	.015	–	–
Shoulder to Shoulder Width	E	.290	.310	.335
Molded Package Width	E1	.240	.285	.295
Overall Length	D	1.345	1.365	1.400
Tip to Seating Plane	L	.110	.130	.150
Lead Thickness	c	.008	.010	.015
Upper Lead Width	b1	.040	.050	.070
Lower Lead Width	b	.014	.018	.022
Overall Row Spacing §	eB	–	–	.430

Notes:

- Pin 1 visual index feature may vary, but must be located within the hatched area.
- § Significant Characteristic.
- Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
- Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-070B

Note the following details of the code protection feature on Microchip devices:

- Microchip products meet the specification contained in their particular Microchip Data Sheet.
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ISBN: 978-1-5224-0042-4

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