



Welcome to E-XFL.COM

What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

Details

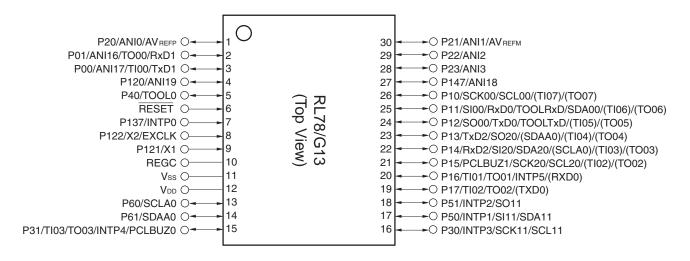
Product Status	Discontinued at Digi-Key
Core Processor	RL78
Core Size	16-Bit
Speed	32MHz
Connectivity	CSI, I ² C, LINbus, UART/USART
Peripherals	DMA, LVD, POR, PWM, WDT
Number of I/O	38
Program Memory Size	128KB (128K x 8)
Program Memory Type	FLASH
EEPROM Size	8K x 8
RAM Size	12K x 8
Voltage - Supply (Vcc/Vdd)	1.6V ~ 5.5V
Data Converters	A/D 12x8/10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	52-LQFP
Supplier Device Package	52-LQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/renesas-electronics-america/r5f100jggfa-x0

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

1.3.4 30-pin products

• 30-pin plastic LSSOP (7.62 mm (300), 0.65 mm pitch)



Caution Connect the REGC pin to Vss via a capacitor (0.47 to 1 μ F).

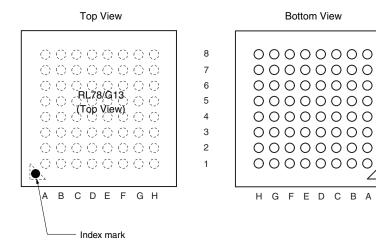
Remarks 1. For pin identification, see 1.4 Pin Identification.

Functions in parentheses in the above figure can be assigned via settings in the peripheral I/O redirection register (PIOR). Refer to Figure 4-8 Format of Peripheral I/O Redirection Register (PIOR) in the RL78/G13 User's Manual.



Bottom View

• 64-pin plastic VFBGA (4 × 4 mm, 0.4 mm pitch)



Pin No.	Name	Pin No.	Name	Pin No.	Name	Pin No.	Name
A1	P05/TI05/TO05	C1	P51/INTP2/SO11	E1	P13/TxD2/SO20/ (SDAA0)/(TI04)/(TO04)	G1	P146
A2	P30/INTP3/RTC1HZ /SCK11/SCL11	C2	P71/KR1/SI21/SDA21	E2	P14/RxD2/SI20/SDA20 /(SCLA0)/(TI03)/(TO03)	G2	P25/ANI5
A3	P70/KR0/SCK21 /SCL21	СЗ	P74/KR4/INTP8/SI01 /SDA01	E3	P15/SCK20/SCL20/ (TI02)/(TO02)	G3	P24/ANI4
A4	P75/KR5/INTP9 /SCK01/SCL01	C4	P52/(INTP10)	E4	P16/TI01/TO01/INTP5 /(SI00)/(RxD0)	G4	P22/ANI2
A5	P77/KR7/INTP11/ (TxD2)	C5	P53/(INTP11)	E5	P03/ANI16/SI10/RxD1 /SDA10	G5	P130
A6	P61/SDAA0	C6	P63	E6	P41/TI07/TO07	G6	P02/ANI17/SO10/TxD1
A7	P60/SCLA0	C7	Vss	E7	RESET	G7	P00/TI00
A8	EVDD0	C8	P121/X1	E8	P137/INTP0	G8	P124/XT2/EXCLKS
B1	P50/INTP1/SI11 /SDA11	D1	P55/(PCLBUZ1)/ (SCK00)	F1	P10/SCK00/SCL00/ (TI07)/(TO07)	H1	P147/ANI18
B2	P72/KR2/SO21	D2	P06/TI06/TO06	F2	P11/SI00/RxD0 /TOOLRxD/SDA00/ (TI06)/(TO06)	H2	P27/ANI7
В3	P73/KR3/SO01	D3	P17/TI02/TO02/ (SO00)/(TxD0)	F3	P12/SO00/TxD0 /TOOLTxD/(INTP5)/ (TI05)/(TO05)	H3	P26/ANI6
B4	P76/KR6/INTP10/ (RxD2)	D4	P54	F4	P21/ANI1/AVREFM	H4	P23/ANI3
B5	P31/TI03/TO03 /INTP4/(PCLBUZ0)	D5	P42/TI04/TO04	F5	P04/SCK10/SCL10	H5	P20/ANI0/AVREFP
B6	P62	D6	P40/TOOL0	F6	P43	H6	P141/PCLBUZ1/INTP7
B7	Vdd	D7	REGC	F7	P01/TO00	H7	P140/PCLBUZ0/INTP6
B8	EVsso	D8	P122/X2/EXCLK	F8	P123/XT1	H8	P120/ANI19

Cautions 1. Make EVsso pin the same potential as Vss pin.

- 2. Make VDD pin the potential that is higher than EVDD0 pin.
- 3. Connect the REGC pin to Vss via a capacitor (0.47 to 1 μ F).

Remarks 1. For pin identification, see 1.4 Pin Identification.

- 2. When using the microcontroller for an application where the noise generated inside the microcontroller must be reduced, it is recommended to supply separate powers to the VDD and EVDD0 pins and connect the Vss and EVss0 pins to separate ground lines.
- 3. Functions in parentheses in the above figure can be assigned via settings in the peripheral I/O redirection register (PIOR). Refer to Figure 4-8 Format of Peripheral I/O Redirection Register (PIOR) in the RL78/G13 User's Manual.

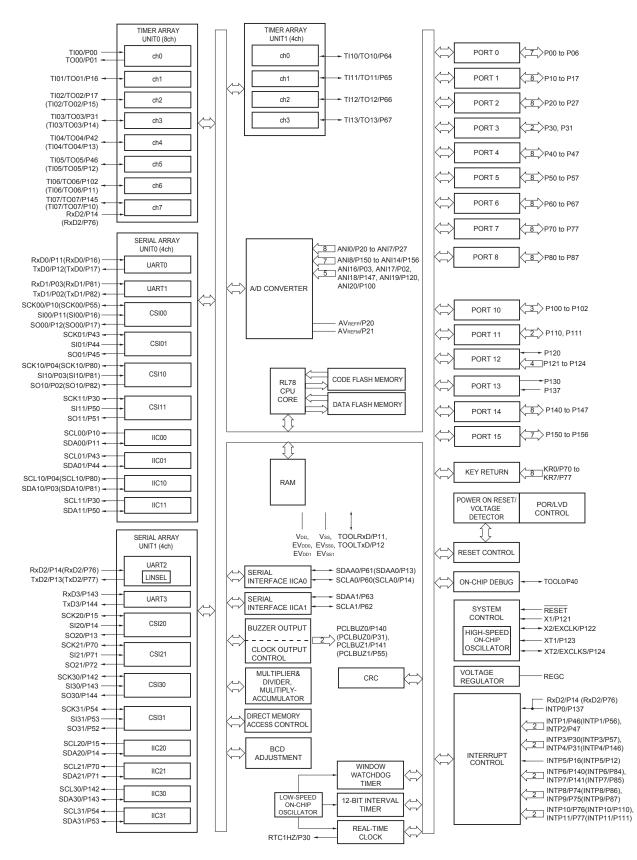


1.4 Pin Identification

ANI0 to ANI14,		REGC:	Regulator capacitance
ANI16 to ANI26:	Analog input	RESET:	Reset
AVREFM:	A/D converter reference	RTC1HZ:	Real-time clock correction clock
	potential (– side) input		(1 Hz) output
AVREFP:	A/D converter reference	RxD0 to RxD3:	Receive data
	potential (+ side) input	SCK00, SCK01, SCK10,	
EVDD0, EVDD1:	Power supply for port	SCK11, SCK20, SCK21,	
EVsso, EVss1:	Ground for port	SCLA0, SCLA1:	Serial clock input/output
EXCLK:	External clock input (Main	SCLA0, SCLA1, SCL00,	
	system clock)	SCL01, SCL10, SCL11,	
EXCLKS:	External clock input	SCL20,SCL21, SCL30,	
	(Subsystem clock)	SCL31:	Serial clock output
INTP0 to INTP11:	Interrupt request from	SDAA0, SDAA1, SDA00,	
	peripheral	SDA01,SDA10, SDA11,	
KR0 to KR7:	Key return	SDA20,SDA21, SDA30,	
P00 to P07:	Port 0	SDA31:	Serial data input/output
P10 to P17:	Port 1	SI00, SI01, SI10, SI11,	
P20 to P27:	Port 2	SI20, SI21, SI30, SI31:	Serial data input
P30 to P37:	Port 3	SO00, SO01, SO10,	
P40 to P47:	Port 4	SO11, SO20, SO21,	
P50 to P57:	Port 5	SO30, SO31:	Serial data output
P60 to P67:	Port 6	TI00 to TI07,	
P70 to P77:	Port 7	TI10 to TI17:	Timer input
P80 to P87:	Port 8	TO00 to TO07,	
P90 to P97:	Port 9	TO10 to TO17:	Timer output
P100 to P106:	Port 10	TOOL0:	Data input/output for tool
P110 to P117:	Port 11	TOOLRxD, TOOLTxD:	Data input/output for external device
P120 to P127:	Port 12	TxD0 to TxD3:	Transmit data
P130, P137:	Port 13	VDD:	Power supply
P140 to P147:	Port 14	Vss:	Ground
P150 to P156:	Port 15	X1, X2:	Crystal oscillator (main system clock)
PCLBUZ0, PCLBUZ1:	Programmable clock	XT1, XT2:	Crystal oscillator (subsystem clock)
	output/buzzer output		



1.5.13 100-pin products



Remark Functions in parentheses in the above figure can be assigned via settings in the peripheral I/O redirection register (PIOR). Refer to Figure 4-8 Format of Peripheral I/O Redirection Register (PIOR) in the RL78/G13 User's Manual.



Items	Symbol	Conditions		MIN.	TYP.	MAX.	Unit
Output voltage, high	Vон1	P00 to P07, P10 to P17, P30 to P37, P40 to P47, P50 to P57, P64	4.0 V \leq EV _{DD0} \leq 5.5 V, I _{OH1} = -10.0 mA	EV _{DD0} - 1.5			V
		to P67, P70 to P77, P80 to P87, P90 to P97, P100 to P106, P110 to	$\begin{array}{l} 4.0 \ V \leq EV_{\text{DD0}} \leq 5.5 \ V, \\ I_{\text{OH1}} = -3.0 \ \text{mA} \end{array}$	EV _{DD0} - 0.7			V
		P117, P120, P125 to P127, P130, P140 to P147	$2.7 \text{ V} \leq \text{EV}_{\text{DD0}} \leq 5.5 \text{ V},$ $I_{\text{OH1}} = -2.0 \text{ mA}$	EV _{DD0} - 0.6			V
			$1.8 \text{ V} \leq \text{EV}_{\text{DD0}} \leq 5.5 \text{ V},$ $I_{\text{OH1}} = -1.5 \text{ mA}$	EV _{DD0} - 0.5			V
			$eq:logical_lo$	EV _{DD0} – 0.5			V
	Vон2	P20 to P27, P150 to P156	$1.6 \text{ V} \le \text{V}_{\text{DD}} \le 5.5 \text{ V},$ Ioh2 = -100 μ A	V _{DD} - 0.5			V
Output voltage, low	P to P	P00 to P07, P10 to P17, P30 to P37, P40 to P47, P50 to P57, P64	$\begin{array}{l} 4.0 \ V \leq EV_{\text{DD0}} \leq 5.5 \ V, \\ I_{\text{OL1}} = 20 \ mA \end{array}$			1.3	V
		to P67, P70 to P77, P80 to P87, P90 to P97, P100 to P106, P110 to P117, P120, P125 to P127, P130, P140 to P147	$\begin{array}{l} 4.0 \ V \leq EV_{\text{DD0}} \leq 5.5 \ V, \\ I_{\text{OL1}} = 8.5 \ mA \end{array} \end{array} \label{eq:DD1}$			0.7	V
			$\begin{array}{l} 2.7 \ V \leq EV_{\text{DD0}} \leq 5.5 \ V, \\ I_{\text{OL1}} = 3.0 \ mA \end{array} \end{array} \label{eq:DD1}$			0.6	V
			$\begin{array}{l} 2.7 \ V \leq EV_{\text{DD0}} \leq 5.5 \ V, \\ I_{\text{OL1}} = 1.5 \ mA \end{array} \end{array} \label{eq:DD1}$			0.4	V
			$eq:local_$			0.4	V
			$eq:local_$			0.4	V
	Vol2	P20 to P27, P150 to P156	$1.6 \text{ V} \le \text{V}_{\text{DD}} \le 5.5 \text{ V},$ $\text{Iol2} = 400 \ \mu \text{ A}$			0.4	V
	Vol3	Vol3 P60 to P63	$\begin{array}{l} 4.0 \ V \leq EV_{\text{DD0}} \leq 5.5 \ V, \\ I_{\text{OL3}} = 15.0 \ \text{mA} \end{array}$			2.0	V
			$\begin{array}{l} 4.0 \ V \leq EV_{\text{DD0}} \leq 5.5 \ V, \\ I_{\text{OL3}} = 5.0 \ mA \end{array} \end{array} \label{eq:DD1}$			0.4	V
			$\begin{array}{l} 2.7 \ V \leq EV_{\text{DD0}} \leq 5.5 \ V, \\ I_{\text{OL3}} = 3.0 \ mA \end{array}$			0.4	V
			$\begin{array}{l} 1.8 \ V \leq EV_{\text{DD0}} \leq 5.5 \ V, \\ I_{\text{OL3}} = 2.0 \ mA \end{array}$			0.4	V
			$1.6 \text{ V} \le \text{EV}_{\text{DD0}} < 5.5 \text{ V},$ lol3 = 1.0 mA			0.4	V

Caution P00, P02 to P04, P10 to P15, P17, P43 to P45, P50, P52 to P55, P71, P74, P80 to P82, P96, and P142 to P144 do not output high level in N-ch open-drain mode.

Remark Unless specified otherwise, the characteristics of alternate-function pins are the same as those of the port pins.



(1) Flash ROM: 16 to 64 KB of 20- to 64-pin products

(TA = -40 to +85°C, 1.6 V \leq EVDD0 \leq VDD \leq 5.5 V, Vss = EVss0 = 0 V) (2/2)

Parameter	Symbol			Conditions		MIN.	TYP.	MAX.	Unit
Supply	IDD2	HALT	HS (high-	$f_{IH} = 32 \text{ MHz}^{Note 4}$	$V_{DD} = 5.0 V$		0.54	1.63	mA
Current	Note 2	mode	speed main) mode ^{Note 7}		$V_{DD} = 3.0 V$		0.54	1.63	mA
				fiH = 24 MHz ^{Note 4}	$V_{DD} = 5.0 V$		0.44	1.28	mA
					V _{DD} = 3.0 V		0.44	1.28	mA
				fin = 16 MHz ^{Note 4}	V _{DD} = 5.0 V		0.40	1.00	mA
					V _{DD} = 3.0 V		0.40	1.00	mA
			LS (low-	fin = 8 MHz ^{Note 4}	V _{DD} = 3.0 V		260	530	μA
			speed main) mode ^{Note 7}		V _{DD} = 2.0 V		260	530	μA
			LV (low-	fiH = 4 MHz ^{Note 4}	V _{DD} = 3.0 V		420	640	μA
			voltage main) mode		V _{DD} = 2.0 V		420	640	μA
			HS (high-	f _{MX} = 20 MHz ^{Note 3} ,	Square wave input		0.28	1.00	mA
		speed main) mode ^{Note 7}	$V_{DD} = 5.0 V$	Resonator connection		0.45	1.17	mA	
				f _{MX} = 20 MHz ^{Note 3} ,	Square wave input		0.28	1.00	mA
			$V_{DD} = 3.0 V$	Resonator connection		0.45	1.17	mA	
			$f_{MX} = 10 \text{ MHz}^{Note 3},$	Square wave input		0.19	0.60	mA	
			$V_{DD} = 5.0 V$	Resonator connection		0.26	0.67	mA	
			$f_{MX} = 10 \text{ MHz}^{Note 3}$,	Square wave input		0.19	0.60	mA	
			LS (low- speed main) mode ^{Note 7}	$V_{DD} = 3.0 V$	Resonator connection		0.26	0.67	mA
				$f_{MX} = 8 MHz^{Note 3}$,	Square wave input		95	330	μA
				$V_{DD} = 3.0 V$	Resonator connection		145	380	μA
				$f_{MX} = 8 MHz^{Note 3}$,	Square wave input		95	330	μA
				$V_{DD} = 2.0 V$	Resonator connection		145	380	μA
			Subsystem	fsub = 32.768 kHz ^{Note 5}	Square wave input		0.25	0.57	μA
			clock	$T_A = -40^{\circ}C$	Resonator connection		0.44	0.76	μA
			operation	$f_{SUB} = 32.768 \text{ kHz}^{Note 5}$	Square wave input		0.30	0.57	μA
				$T_A = +25^{\circ}C$	Resonator connection		0.49	0.76	μA
				fsub = 32.768 kHz ^{Note 5}	Square wave input		0.37	1.17	μA
				$T_A = +50^{\circ}C$	Resonator connection		0.56	1.36	μA
				fsuв = 32.768 kHz ^{Note 5}	Square wave input		0.53	1.97	μA
				$T_A = +70^{\circ}C$	Resonator connection		0.72	2.16	μA
				fsub = 32.768 kHz ^{Note 5}	Square wave input		0.82	3.37	μA
				T _A = +85°C	Resonator connection		1.01	3.56	μA
	DD3 ^{Note 6}	STOP	$T_A = -40^{\circ}C$				0.18	0.50	μA
		mode ^{Note 8}	$T_{A} = +25^{\circ}C$ $T_{A} = +50^{\circ}C$				0.23	0.50	μA
							0.30	1.10	μA
			$T_A = +70^{\circ}C$				0.46	1.90	μA
			T _A = +85°C				0.75	3.30	μA

(Notes and Remarks are listed on the next page.)

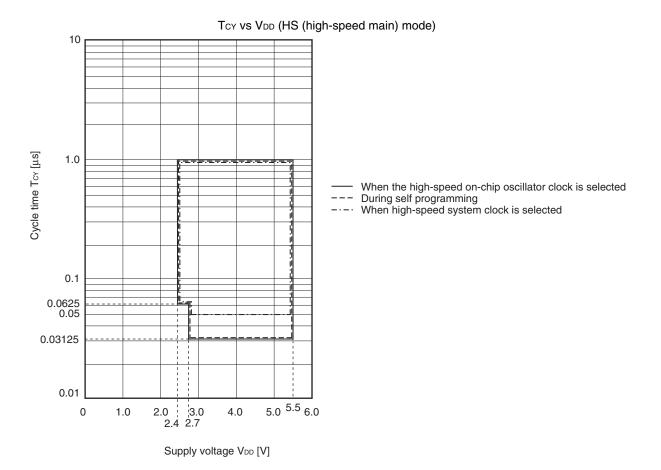


NoteThe following conditions are required for low voltage interface when $E_{VDD0} < V_{DD}$ $1.8 V \le EV_{DD0} < 2.7 V : MIN. 125 ns$ $1.6 V \le EV_{DD0} < 1.8 V : MIN. 250 ns$

 $\label{eq:rescaled} \textbf{Remark} \quad \text{f_{MCK}: Timer array unit operation clock frequency}$

(Operation clock to be set by the CKSmn0, CKSmn1 bits of timer mode register mn (TMRmn). m: Unit number (m = 0, 1), n: Channel number (n = 0 to 7))

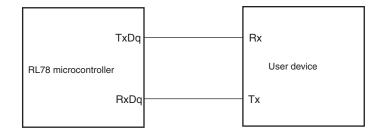
Minimum Instruction Execution Time during Main System Clock Operation



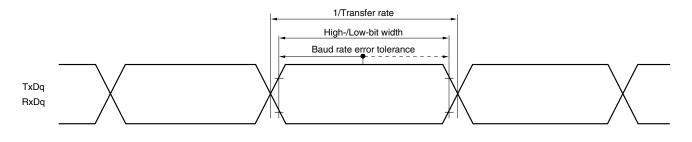
R01DS0131EJ0330 Rev.3.30 Mar 31, 2016



UART mode connection diagram (during communication at same potential)



UART mode bit width (during communication at same potential) (reference)



Remarks 1. q: UART number (q = 0 to 3), g: PIM and POM number (g = 0, 1, 8, 14)

fMCK: Serial array unit operation clock frequency
 (Operation clock to be set by the CKSmn bit of serial mode register mn (SMRmn). m: Unit number, n: Channel number (mn = 00 to 03, 10 to 13))



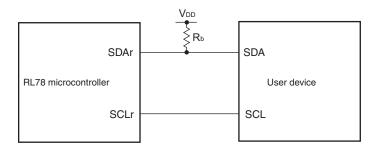
Parameter	Symbol	Conditions	、 U	h-speed Mode	``	/-speed Mode	``	-voltage Mode	Unit
			MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
SCLr clock frequency	fsc∟	$\begin{array}{l} 2.7 \ V \leq EV_{\text{DD0}} \leq 5.5 \ V, \\ C_{\text{b}} = 50 \ p\text{F}, \ R_{\text{b}} = 2.7 \ k\Omega \end{array}$		1000 Note 1		400 Note 1		400 Note 1	kHz
		$1.8 \text{ V} \leq \text{EV}_{\text{DD0}} \leq 5.5 \text{ V},$ $C_{\text{b}} = 100 \text{ pF}, \text{ R}_{\text{b}} = 3 \text{ k}\Omega$		400 Note 1		400 Note 1		400 Note 1	kHz
		$1.8 \text{ V} \leq \text{EV}_{\text{DD0}} < 2.7 \text{ V},$ $C_b = 100 \text{ pF}, \text{ R}_b = 5 \text{ k}\Omega$		300 Note 1		300 Note 1		300 Note 1	kHz
		$1.7 \text{ V} \leq \text{EV}_{\text{DD0}} < 1.8 \text{ V},$ $C_{\text{b}} = 100 \text{ pF}, \text{ R}_{\text{b}} = 5 \text{ k}\Omega$		250 Note 1		250 Note 1		250 Note 1	kHz
		$1.6 \text{ V} \le \text{EV}_{\text{DD0}} < 1.8 \text{ V},$ $C_{\text{b}} = 100 \text{ pF}, \text{ R}_{\text{b}} = 5 \text{ k}\Omega$		—		250 Note 1		250 Note 1	kHz
Hold time when SCLr = "L"	tLOW	$2.7 \text{ V} \le \text{EV}_{\text{DD0}} \le 5.5 \text{ V},$ $C_b = 50 \text{ pF}, \text{ R}_b = 2.7 \text{ k}\Omega$	475		1150		1150		ns
		$1.8 \text{ V} \leq \text{EV}_{\text{DD0}} \leq 5.5 \text{ V},$ $C_{\text{b}} = 100 \text{ pF}, \text{ R}_{\text{b}} = 3 \text{ k}\Omega$	1150		1150		1150		ns
		$1.8 \text{ V} \leq \text{EV}_{\text{DD0}} < 2.7 \text{ V},$ $C_{\text{b}} = 100 \text{ pF}, \text{ R}_{\text{b}} = 5 \text{ k}\Omega$	1550		1550		1550		ns
		$1.7 \text{ V} \le \text{EV}_{\text{DD0}} < 1.8 \text{ V},$ $C_b = 100 \text{ pF}, \text{R}_b = 5 \text{ k}\Omega$	1850		1850		1850		ns
		$1.6 \text{ V} \le \text{EV}_{\text{DD0}} < 1.8 \text{ V},$ $C_{\text{b}} = 100 \text{ pF}, \text{R}_{\text{b}} = 5 \text{ k}\Omega$			1850		1850		ns
Hold time when SCLr = "H"	tніgн	$2.7 \text{ V} \le \text{EV}_{\text{DD0}} \le 5.5 \text{ V},$ $C_b = 50 \text{ pF}, \text{ R}_b = 2.7 \text{ k}\Omega$	475		1150		1150		ns
		$1.8 \text{ V} \leq \text{EV}_{\text{DD0}} \leq 5.5 \text{ V},$ $C_{\text{b}} = 100 \text{ pF}, \text{ R}_{\text{b}} = 3 \text{ k}\Omega$	1150		1150		1150		ns
		$1.8 \text{ V} \le \text{EV}_{\text{DD0}} < 2.7 \text{ V},$ $C_b = 100 \text{ pF}, \text{R}_b = 5 \text{ k}\Omega$	1550		1550		1550		ns
		$1.7 \text{ V} \le \text{EV}_{\text{DD0}} < 1.8 \text{ V},$ $C_b = 100 \text{ pF}, \text{R}_b = 5 \text{ k}\Omega$	1850		1850		1850		ns
		$1.6 \text{ V} \le \text{EV}_{\text{DD0}} < 1.8 \text{ V},$ $C_b = 100 \text{ pF}, \text{R}_b = 5 \text{ k}\Omega$			1850		1850		ns

(5) During communication at same potential (simplified I²C mode) (1/2)

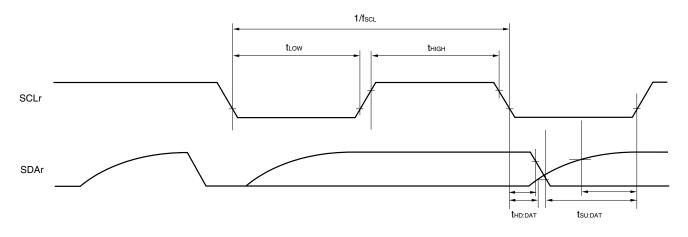
(Notes and Caution are listed on the next page, and Remarks are listed on the page after the next page.)



Simplified I²C mode mode connection diagram (during communication at same potential)



Simplified I²C mode serial transfer timing (during communication at same potential)



- **Remarks 1.** R_b[Ω]:Communication line (SDAr) pull-up resistance, C_b[F]: Communication line (SDAr, SCLr) load capacitance
 - r: IIC number (r = 00, 01, 10, 11, 20, 21, 30, 31), g: PIM number (g = 0, 1, 4, 5, 8, 14),
 h: POM number (g = 0, 1, 4, 5, 7 to 9, 14)
 - 3. fmck: Serial array unit operation clock frequency

(Operation clock to be set by the CKSmn bit of serial mode register mn (SMRmn). m: Unit number (m = 0, 1),

n: Channel number (n = 0 to 3), mn = 00 to 03, 10 to 13)



			$\sqrt{DD0} = EVDD1 \le VDD \le$						1.177	1	Lint
Parameter	Symbol		Conditions			high-		low-		low-	Unit
						main) ode	speed	main) ode		age Mode	
								1			
					MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
Transfer rate		Transmission	$4.0 \text{ V} \leq \text{EV}_{\text{DD0}} \leq 5.5 \text{ V},$			Note		Note		Note	bps
			$2.7 \text{ V} \leq V_b \leq 4.0 \text{ V}$			1		1		1	
				Theoretical		2.8		2.8		2.8	Mbps
				value of the		Note 2		Note 2		Note 2	
				maximum							
				transfer rate							
				C _b = 50 pF, R _b =							
				$1.4 \text{ k}\Omega, V_{\text{b}} = 2.7$							
				V							
			$2.7 \text{ V} \le \text{EV}_{\text{DD0}} < 4.0 \text{ V},$			Note		Note		Note	bps
			$2.3~V \leq V_b \leq 2.7~V$			3		3		3	
				Theoretical		1.2		1.2		1.2	Mbps
				value of the		Note 4		Note 4		Note 4	
				maximum							
				transfer rate							
				$C_b = 50 \text{ pF}, R_b =$							
				$2.7 \text{ k}\Omega$, V _b = 2.3							
				V							
			$1.8 \text{ V} \leq \text{EV}_{\text{DD0}} < 3.3 \text{ V},$			Notes		Notes		Notes	bps
			$1.6~V \leq V_b \leq 2.0~V$			5, 6		5, 6		5, 6	
				Theoretical		0.43		0.43		0.43	Mbps
				value of the		Note 7		Note 7		Note 7	
				maximum							
				transfer rate							
				$C_b = 50 \text{ pF}, R_b =$							
				$5.5 \text{ k}\Omega, \text{V}_{\text{b}} = 1.6$							
				V							

(6) Communication at different potential (1.8 V, 2.5 V, 3 V) (UART mode) (2/2) (TA = -40 to +85°C, 1.8 V \leq EVDD0 = EVDD1 \leq VDD \leq 5.5 V, Vss = EVss0 = EVss1 = 0 V)

Notes 1. The smaller maximum transfer rate derived by using fMck/6 or the following expression is the valid maximum transfer rate.

Expression for calculating the transfer rate when 4.0 V \leq EV $_{DD0} \leq$ 5.5 V and 2.7 V \leq V $_{b} \leq$ 4.0 V

Maximum transfer rate =
$$\frac{1}{\{-C_b \times R_b \times \ln (1 - \frac{2.2}{V_b})\} \times 3}$$
 [bps]

Baud rate error (theoretical value) =
$$\frac{\frac{1}{\text{Transfer rate} \times 2} - \{-C_b \times R_b \times \ln(1 - \frac{2.2}{V_b})\}}{(\frac{1}{\text{Transfer rate}}) \times \text{Number of transferred bits}} \times 100 [\%]$$

- * This value is the theoretical value of the relative difference between the transmission and reception sides.
- This value as an example is calculated when the conditions described in the "Conditions" column are met. Refer to Note 1 above to calculate the maximum transfer rate under conditions of the customer.



3. The smaller maximum transfer rate derived by using fMck/6 or the following expression is the valid maximum transfer rate.

Expression for calculating the transfer rate when 2.7 V \leq EV_{DD0} < 4.0 V and 2.3 V \leq V_b \leq 2.7 V

Maximum transfer rate =
$$\frac{1}{\{-C_b \times R_b \times \ln (1 - \frac{2.0}{V_b})\} \times 3}$$
 [bps]

Baud rate error (theoretical value) = $\frac{\frac{1}{|\text{Transfer rate} \times 2|} - \{-C_b \times R_b \times \ln(1 - \frac{2.0}{V_b})\}}{(\frac{1}{|\text{Transfer rate}|}) \times \text{Number of transferred bits}} \times 100 [\%]$

* This value is the theoretical value of the relative difference between the transmission and reception sides.

- **4.** This value as an example is calculated when the conditions described in the "Conditions" column are met. Refer to Note 3 above to calculate the maximum transfer rate under conditions of the customer.
- $\textbf{5.} \quad \textbf{Use it with } EV_{DD0} \geq V_{b}.$
- 6. The smaller maximum transfer rate derived by using fMCK/6 or the following expression is the valid maximum transfer rate.

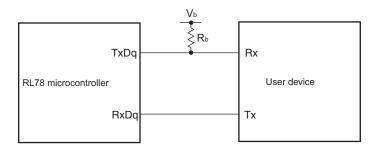
Expression for calculating the transfer rate when 1.8 V \leq EV_{DD0} < 3.3 V and 1.6 V \leq V_b \leq 2.0 V

Maximum transfer rate =
$$\frac{1}{\{-C_b \times R_b \times \ln (1 - \frac{1.5}{V_b})\} \times 3}$$
 [bps]

Baud rate error (theoretical value) = $\frac{\frac{1}{\text{Transfer rate} \times 2} - \{-C_b \times R_b \times \ln(1 - \frac{1.5}{V_b})\}}{(\frac{1}{\text{Transfer rate}}) \times \text{Number of transferred bits}} \times 100 [\%]$

- * This value is the theoretical value of the relative difference between the transmission and reception sides.
- **7.** This value as an example is calculated when the conditions described in the "Conditions" column are met. Refer to Note 6 above to calculate the maximum transfer rate under conditions of the customer.
- Caution Select the TTL input buffer for the RxDq pin and the N-ch open drain output (VDD tolerance (When 20- to 52-pin products)/EVDD tolerance (When 64- to 128-pin products)) mode for the TxDq pin by using port input mode register g (PIMg) and port output mode register g (POMg). For VIH and VIL, see the DC characteristics with TTL input buffer selected.

UART mode connection diagram (during communication at different potential)





Parameter	Symbol	Conditions	HS (higl main)		LS (low-speed main) Mode		LV (low-voltage main) Mode		Unit
			MIN.	MAX.	MIN.	MAX.	MIN.	MAX.	
Data setup time (reception)	tsu:dat		1/fмск + 135 ^{Note 3}		1/fмск + 190 _{Note 3}		1/fмск + 190 _{Note 3}		kHz
		$\label{eq:V} \begin{array}{l} 2.7 \ V \leq EV_{DD0} < 4.0 \ V, \\ 2.3 \ V \leq V_b \leq 2.7 \ V, \\ C_b = 50 \ pF, \ R_b = 2.7 \ k\Omega \end{array}$	1/fмск + 135 ^{Note 3}		1/fмск + 190 _{Note 3}		1/fмск + 190 _{Note 3}		kHz
			1/fмск + 190 ^{Note 3}		1/fмск + 190 _{Note 3}		1/fмск + 190 _{Note 3}		kHz
		$\label{eq:linear} \begin{array}{l} 2.7 \; V \leq EV_{DD0} < 4.0 \; V, \\ 2.3 \; V \leq V_b \leq 2.7 \; V, \\ C_b = 100 \; pF, \; R_b = 2.7 \; k\Omega \end{array}$	1/fмск + 190 ^{Note 3}		1/fмск + 190 _{Note 3}		1/fмск + 190 _{Note 3}		kHz
		$ \begin{split} & 1.8 \; V \leq EV_{DD0} < 3.3 \; V, \\ & 1.6 \; V \leq V_b \leq 2.0 \; V^{\text{Note 2}}, \\ & C_b = 100 \; pF, \; R_b = 5.5 \; k\Omega \end{split} $	1/f _{MCK} + 190 ^{Note 3}		1/fмск + 190 _{Note 3}		1/fмск + 190 _{Note 3}		kHz
Data hold time (transmission)	thd:dat	$\begin{array}{l} 4.0 \; V \leq EV_{DD0} \leq 5.5 \; V, \\ 2.7 \; V \leq V_b \leq 4.0 \; V, \\ C_b = 50 \; pF, \; R_b = 2.7 \; k\Omega \end{array}$	0	305	0	305	0	305	ns
		$\label{eq:2.7} \begin{array}{l} 2.7 \ V \leq EV_{DD0} < 4.0 \ V, \\ 2.3 \ V \leq V_b \leq 2.7 \ V, \\ C_b = 50 \ pF, \ R_b = 2.7 \ k\Omega \end{array}$	0	305	0	305	0	305	ns
			0	355	0	355	0	355	ns
		$\label{eq:linear} \begin{array}{l} 2.7 \; V \leq EV_{DD0} < 4.0 \; V, \\ 2.3 \; V \leq V_b \leq 2.7 \; V, \\ C_b = 100 \; pF, \; R_b = 2.7 \; k\Omega \end{array}$	0	355	0	355	0	355	ns
		$\label{eq:VDD} \begin{split} & 1.8 \ V \leq EV_{\text{DD0}} < 3.3 \ V, \\ & 1.6 \ V \leq V_{\text{b}} \leq 2.0 \ V^{\text{Note 2}}, \\ & C_{\text{b}} = 100 \ \text{pF}, \ R_{\text{b}} = 5.5 \ \text{k}\Omega \end{split}$	0	405	0	405	0	405	ns

(10) Communication at different potential (1.8 V, 2.5 V, 3 V) (simplified I²C mode) (2/2) (T_A = -40 to +85°C. 1.8 V \leq EV_{DD0} = EV_{DD1} \leq V_{DD} \leq 5.5 V. Vss = EV_{SS0} = EV_{SS1} = 0 V)

Notes 1. The value must also be equal to or less than f_MCK/4.

- **2.** Use it with $EV_{DD0} \ge V_b$.
- 3. Set the fmck value to keep the hold time of SCLr = "L" and SCLr = "H".
- Caution Select the TTL input buffer and the N-ch open drain output (V_{DD} tolerance (for the 20- to 52-pin products)/EV_{DD} tolerance (for the 64- to 128-pin products)) mode for the SDAr pin and the N-ch open drain output (V_{DD} tolerance (for the 20- to 52-pin products)/EV_{DD} tolerance (for the 64- to 128-pin products)) mode for the SCLr pin by using port input mode register g (PIMg) and port output mode register g (POMg). For V_{IH} and V_{IL}, see the DC characteristics with TTL input buffer selected.

(Remarks are listed on the next page.)



2.6 Analog Characteristics

2.6.1 A/D converter characteristics

Classification of A/D converter characteristics

		Reference Voltage	
	Reference voltage (+) = AVREFP	Reference voltage (+) = VDD	Reference voltage (+) = VBGR
Input channel	Reference voltage (-) = AVREFM	Reference voltage (-) = Vss	Reference voltage (-) = AVREFM
ANI0 to ANI14	Refer to 2.6.1 (1).	Refer to 2.6.1 (3).	Refer to 2.6.1 (4) .
ANI16 to ANI26	Refer to 2.6.1 (2) .		
Internal reference voltage	Refer to 2.6.1 (1) .		_
Temperature sensor output			
voltage			

(1) When reference voltage (+)= AV_{REFP}/ANI0 (ADREFP1 = 0, ADREFP0 = 1), reference voltage (-) = AV_{REFM}/ANI1 (ADREFM = 1), target pin : ANI2 to ANI14, internal reference voltage, and temperature sensor output voltage

(T_A = -40 to +85°C, 1.6 V \leq AV_{REFP} \leq V_{DD} \leq 5.5 V, V_{SS} = 0 V, Reference voltage (+) = AV_{REFP}, Reference voltage (-) = AV_{REFM} = 0 V)

Parameter	Symbol	Con	ditions	MIN.	TYP.	MAX.	Unit
Resolution	RES			8		10	bit
Overall error ^{Note 1}	AINL	10-bit resolution	$1.8~V \leq AV_{\text{REFP}} \leq 5.5~V$		1.2	±3.5	LSB
		$AV_{REFP} = V_{DD}{}^{Note 3}$	$1.6~V \leq AV_{\text{REFP}} \leq 5.5~V^{\text{Note 4}}$		1.2	±7.0	LSB
Conversion time	t CONV	10-bit resolution	$3.6~V \leq V \text{DD} \leq 5.5~V$	2.125		39	μS
		Target pin: ANI2 to	$2.7~V \leq V \text{DD} \leq 5.5~V$	3.1875		39	μS
		ANI14	$1.8~V \leq V \text{DD} \leq 5.5~V$	17		39	μS
			$1.6~V \leq V \text{DD} \leq 5.5~V$	57		95	μS
		10-bit resolution	$3.6~V \leq V \text{DD} \leq 5.5~V$	2.375		39	μS
		Target pin: Internal	$2.7~V \leq V \text{DD} \leq 5.5~V$	3.5625		39	μS
		reference voltage, and temperature sensor output voltage (HS (high-speed main) mode)	$2.4~V \leq V_{DD} \leq 5.5~V$	17		39	μs
Zero-scale error ^{Notes 1, 2}	Ezs	10-bit resolution	$1.8~V \leq AV_{\text{REFP}} \leq 5.5~V$			±0.25	%FSR
		$AV_{REFP} = V_{DD}^{Note 3}$	$1.6~V \leq AV_{\text{REFP}} \leq 5.5~V^{\text{Note 4}}$			±0.50	%FSR
Full-scale error ^{Notes 1, 2}	Ers	10-bit resolution	$1.8~V \leq AV_{\text{REFP}} \leq 5.5~V$			±0.25	%FSR
		$AV_{REFP} = V_{DD}^{Note 3}$	$1.6~V \leq AV_{\text{REFP}} \leq 5.5~V^{\text{Note 4}}$			±0.50	%FSR
Integral linearity error ^{Note 1}	ILE	10-bit resolution	$1.8~V \leq AV_{\text{REFP}} \leq 5.5~V$			±2.5	LSB
		$AV_{REFP} = V_{DD}{}^{Note 3}$	$1.6~V \leq AV_{\text{REFP}} \leq 5.5~V^{\text{Note 4}}$			±5.0	LSB
Differential linearity error Note 1	DLE	10-bit resolution	$1.8~V \leq AV_{\text{REFP}} \leq 5.5~V$			±1.5	LSB
		$AV_{REFP} = V_{DD}{}^{Note 3}$	$1.6~V \leq AV_{\text{REFP}} \leq 5.5~V^{\text{Note 4}}$			±2.0	LSB
Analog input voltage	VAIN	ANI2 to ANI14		0		AVREFP	V
	Internal reference voltage (2.4 V \leq VDD \leq 5.5 V, HS		(high-speed main) mode)		VBGR Note 5		V
		Temperature sensor outp (2.4 V \leq V _{DD} \leq 5.5 V, HS	0	١	TMPS25 Note	5	V

(Notes are listed on the next page.)



LVD Detection Voltage of Interrupt & Reset Mode

(TA = -40 to +85°C, VPDR \leq VDD \leq 5.5 V, Vss = 0 V)

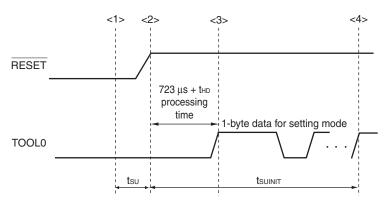
Parameter	Symbol	Conditions			MIN.	TYP.	MAX.	Unit
Interrupt and reset	VLVDA0	VPOC2,	$V_{POC1}, V_{POC0} = 0, 0, 0$, falling reset voltage	1.60	1.63	1.66	V
mode	VLVDA1		LVIS1, LVIS0 = 1, 0	Rising release reset voltage	1.74	1.77	1.81	V
				Falling interrupt voltage	1.70	1.73	1.77	V
	VLVDA2		LVIS1, LVIS0 = 0, 1	Rising release reset voltage	1.84	1.88	1.91	V
				Falling interrupt voltage	1.80	1.84	1.87	V
	VLVDA3		LVIS1, LVIS0 = 0, 0	Rising release reset voltage	2.86	2.92	2.97	V
				Falling interrupt voltage	2.80	2.86	2.91	V
	VLVDB0	VPOC2,	VPOC1, VPOC0 = 0, 0, 1	, falling reset voltage	1.80	1.84	1.87	V
	VLVDB1		LVIS1, LVIS0 = 1, 0	Rising release reset voltage	1.94	1.98	2.02	V
				Falling interrupt voltage	1.90	1.94	1.98	V
	VLVDB2		LVIS1, LVIS0 = 0, 1	Rising release reset voltage	2.05	2.09	2.13	V
				Falling interrupt voltage	2.00	2.04	2.08	V
	VLVDB3		LVIS1, LVIS0 = 0, 0	Rising release reset voltage	3.07	3.13	3.19	V
				Falling interrupt voltage	3.00	3.06	3.12	V
	VLVDC0	VPOC2,	VPOC1, VPOC0 = 0, 1, 0	, falling reset voltage	2.40	2.45	2.50	V
	VLVDC1		LVIS1, LVIS0 = 1, 0	Rising release reset voltage	2.56	2.61	2.66	V
				Falling interrupt voltage	2.50	2.55	2.60	V
	VLVDC2		LVIS1, LVIS0 = 0, 1	Rising release reset voltage	2.66	2.71	2.76	V
				Falling interrupt voltage	2.60	2.65	2.70	V
	VLVDC3		LVIS1, LVIS0 = 0, 0	Rising release reset voltage	3.68	3.75	3.82	V
				Falling interrupt voltage	3.60	3.67	3.74	V
	VLVDD0	VPOC2,	VPOC1, VPOC0 = 0, 1, 1	, falling reset voltage	2.70	2.75	2.81	V
	VLVDD1		LVIS1, LVIS0 = 1, 0	Rising release reset voltage	2.86	2.92	2.97	V
				Falling interrupt voltage	2.80	2.86	2.91	V
	VLVDD2		LVIS1, LVIS0 = 0, 1	Rising release reset voltage	2.96	3.02	3.08	V
				Falling interrupt voltage	2.90	2.96	3.02	V
	VLVDD3		LVIS1, LVIS0 = 0, 0	Rising release reset voltage	3.98	4.06	4.14	V
				Falling interrupt voltage	3.90	3.98	4.06	V



2.10 Timing of Entry to Flash Memory Programming Modes

$(T_{\text{A}} = -40 \text{ to } +85^{\circ}\text{C}, 1.8 \text{ V} \leq \text{EV}_{\text{DD}} = \text{EV}_{\text{DD}} \leq \text{V}_{\text{DD}} \leq 5.5 \text{ V}, \text{Vss} = \text{EV}_{\text{SS0}} = \text{EV}_{\text{SS1}} = 0 \text{ V})$

Parameter	Symbol	Conditions	MIN.	TYP.	MAX.	Unit
Time to complete the communication for the initial setting after the external reset is released	tsuinit	POR and LVD reset must be released before the external reset is released.			100	ms
Time to release the external reset after the TOOL0 pin is set to the low level	ts∪	POR and LVD reset must be released before the external reset is released.	10			μs
Time to hold the TOOL0 pin at the low level after the external reset is released (excluding the processing time of the firmware to control the flash memory)	tно	POR and LVD reset must be released before the external reset is released.	1			ms



- <1> The low level is input to the TOOL0 pin.
- <2> The external reset is released (POR and LVD reset must be released before the external reset is released.).
- <3> The TOOL0 pin is set to the high level.
- <4> Setting of the flash memory programming mode by UART reception and complete the baud rate setting.
- **Remark** tsuinit: Communication for the initial setting must be completed within 100 ms after the external reset is released during this period.
 - $t_{su:}$ Time to release the external reset after the TOOL0 pin is set to the low level
 - thd: Time to hold the TOOL0 pin at the low level after the external reset is released (excluding the processing time of the firmware to control the flash memory)



3.6.5 Power supply voltage rising slope characteristics

$(T_A = -40 \text{ to } +105^{\circ}\text{C}, \text{Vss} = 0 \text{ V})$

Parameter	Symbol	Conditions	MIN.	TYP.	MAX.	Unit
Power supply voltage rising slope	SVDD				54	V/ms

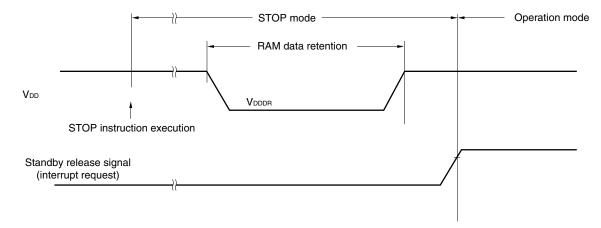
Caution Make sure to keep the internal reset state by the LVD circuit or an external reset until V_{DD} reaches the operating voltage range shown in 3.4 AC Characteristics.

3.7 RAM Data Retention Characteristics

$(T_A = -40 \text{ to } +105^{\circ}\text{C}, \text{Vss} = 0 \text{ V})$

Parameter	Symbol	Conditions	MIN.	TYP.	MAX.	Unit
Data retention supply voltage	VDDDR		1.44 ^{Note}		5.5	V

Note This depends on the POR detection voltage. For a falling voltage, data in RAM are retained until the voltage reaches the level that triggers a POR reset but not once it reaches the level at which a POR reset is generated.





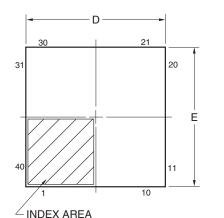
4.7 40-pin Products

R5F100EAANA, R5F100ECANA, R5F100EDANA, R5F100EEANA, R5F100EFANA, R5F100EGANA, R5F100EHANA R5F101EAANA, R5F101ECANA, R5F101EDANA, R5F101EEANA, R5F101EFANA, R5F101EGANA, R5F101EHANA R5F100EADNA, R5F100ECDNA, R5F100EDDNA, R5F100EEDNA, R5F100EFDNA, R5F100EGDNA, R5F100EHDNA

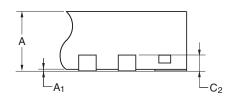
R5F101EADNA, R5F101ECDNA, R5F101EDDNA, R5F101EEDNA, R5F101EFDNA, R5F101EGDNA, R5F101EHDNA

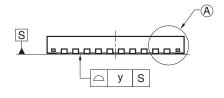
R5F100EAGNA, R5F100ECGNA, R5F100EDGNA, R5F100EEGNA, R5F100EFGNA, R5F100EGGNA, R5F100EHGNA

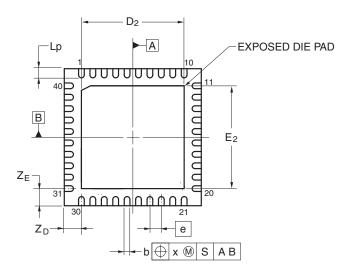
JEITA Package code	RENESAS code	Previous code	MASS (TYP.) [g]
P-HWQFN40-6x6-0.50	PWQN0040KC-A	P40K8-50-4B4-5	0.09



Detail of (A) Part







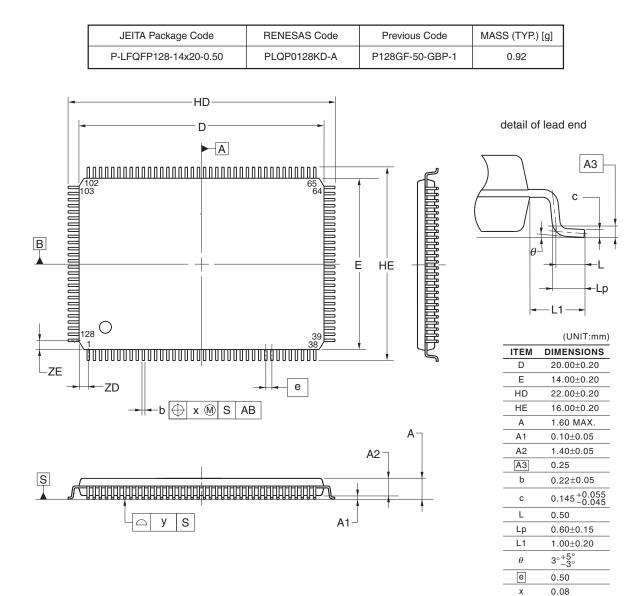
Referance	Dimension in Millimeters				
Symbol	Min	Nom	Max		
D	5.95	6.00	6.05		
E	5.95	6.00	6.05		
A			0.80		
A ₁	0.00				
b	0.18	0.25	0.30		
е		0.50			
Lp	0.30	0.40	0.50		
х			0.05		
у			0.05		
ZD		0.75	—		
Z _E		0.75	—		
C ₂	0.15	0.20	0.25		
D ₂		4.50			
E ₂		4.50			

©2013 Renesas Electronics Corporation. All rights reserved.



4.14 128-pin Products

R5F100SHAFB, R5F100SJAFB, R5F100SKAFB, R5F100SLAFB R5F101SHAFB, R5F101SJAFB, R5F101SKAFB, R5F101SLAFB R5F100SHDFB, R5F100SJDFB, R5F100SKDFB, R5F100SLDFB R5F101SHDFB, R5F101SJDFB, R5F101SKDFB, R5F101SLDFB



©2012 Renesas Electronics Corporation. All rights reserved.

х

y ZD

ZE

0.08

0.75

0.75



NOTES FOR CMOS DEVICES

- (1) VOLTAGE APPLICATION WAVEFORM AT INPUT PIN: Waveform distortion due to input noise or a reflected wave may cause malfunction. If the input of the CMOS device stays in the area between VIL (MAX) and VIH (MIN) due to noise, etc., the device may malfunction. Take care to prevent chattering noise from entering the device when the input level is fixed, and also in the transition period when the input level passes through the area between VIL (MAX) and VIH (MIN).
- (2) HANDLING OF UNUSED INPUT PINS: Unconnected CMOS device inputs can be cause of malfunction. If an input pin is unconnected, it is possible that an internal input level may be generated due to noise, etc., causing malfunction. CMOS devices behave differently than Bipolar or NMOS devices. Input levels of CMOS devices must be fixed high or low by using pull-up or pull-down circuitry. Each unused pin should be connected to VDD or GND via a resistor if there is a possibility that it will be an output pin. All handling related to unused pins must be judged separately for each device and according to related specifications governing the device.
- (3) PRECAUTION AGAINST ESD: A strong electric field, when exposed to a MOS device, can cause destruction of the gate oxide and ultimately degrade the device operation. Steps must be taken to stop generation of static electricity as much as possible, and quickly dissipate it when it has occurred. Environmental control must be adequate. When it is dry, a humidifier should be used. It is recommended to avoid using insulators that easily build up static electricity. Semiconductor devices must be stored and transported in an anti-static container, static shielding bag or conductive material. All test and measurement tools including work benches and floors should be grounded. The operator should be grounded using a wrist strap. Semiconductor devices must not be touched with bare hands. Similar precautions need to be taken for PW boards with mounted semiconductor devices.
- (4) STATUS BEFORE INITIALIZATION: Power-on does not necessarily define the initial status of a MOS device. Immediately after the power source is turned ON, devices with reset functions have not yet been initialized. Hence, power-on does not guarantee output pin levels, I/O settings or contents of registers. A device is not initialized until the reset signal is received. A reset operation must be executed immediately after power-on for devices with reset functions.
- (5) POWER ON/OFF SEQUENCE: In the case of a device that uses different power supplies for the internal operation and external interface, as a rule, switch on the external power supply after switching on the internal power supply. When switching the power supply off, as a rule, switch off the external power supply and then the internal power supply. Use of the reverse power on/off sequences may result in the application of an overvoltage to the internal elements of the device, causing malfunction and degradation of internal elements due to the passage of an abnormal current. The correct power on/off sequence must be judged separately for each device and according to related specifications governing the device.
- (6) INPUT OF SIGNAL DURING POWER OFF STATE : Do not input signals or an I/O pull-up power supply while the device is not powered. The current injection that results from input of such a signal or I/O pull-up power supply may cause malfunction and the abnormal current that passes in the device at this time may cause degradation of internal elements. Input of signals during the power off state must be judged separately for each device and according to related specifications governing the device.