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Details

Product Status	Obsolete
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	38
Program Memory Size	60KB (60K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	48-VFQFN Exposed Pad
Supplier Device Package	48-QFN-EP (7x7)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mc9s08aw60mfde

Part Number	Package Description	Original (gold wire) package document number	Current (copper wire) package document number
MC68HC908JW32	48 QFN	98ARH99048A	98ASA00466D
MC9S08AC16			
MC9S908AC60			
MC9S08AC128			
MC9S08AW60			
MC9S08GB60A			
MC9S08GT16A			
MC9S08JM16			
MC9S08JM60			
MC9S08LL16			
MC9S08QE128			
MC9S08QE32			
MC9S08RG60			
MCF51CN128			
MC9RS08LA8	48 QFN	98ARL10606D	98ASA00466D
MC9S08GT16A	32 QFN	98ARH99035A	98ASA00473D
MC9S908QE32	32 QFN	98ARE10566D	98ASA00473D
MC9S908QE8	32 QFN	98ASA00071D	98ASA00736D
MC9S08JS16	24 QFN	98ARL10608D	98ASA00734D
MC9S08QB8			
MC9S08QG8	24 QFN	98ARL10605D	98ASA00474D
MC9S08SH8	24 QFN	98ARE10714D	98ASA00474D
MC9RS08KB12	24 QFN	98ASA00087D	98ASA00602D
MC9S08QG8	16 QFN	98ARE10614D	98ASA00671D
MC9RS08KB12	8 DFN	98ARL10557D	98ASA00672D
MC9S08QG8			
MC9RS08KA2	6 DFN	98ARL10602D	98ASA00735D



Revision History

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The following revision history table summarizes changes contained in this document. For your convenience, the page number designators have been linked to the appropriate location.

Revision Number	Revision Date	Description of Changes
1	1/2006	Initial external release.
2	12/2006	Includes KBI block changes; new V_{OL} / I_{OL} figures; R_{IDD} spec changes; SC part numbers with ICG trim modifications; addition of Temp Sensor to ADC. Resolved the stop IDD issues, added RTI figure, bandgap information, and incorporated electricals edits and any ProjectSync issues.

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- The output of the digitally-controlled oscillator (DCO) in the frequency-locked loop sub-module
- Control bits inside the ICG determine which source is connected.
- FFE is a control signal generated inside the ICG. If the frequency of ICGOUT $> 4 \times$ the frequency of ICGERCLK, this signal is a logic 1 and the fixed-frequency clock will be ICGERCLK/2. Otherwise the fixed-frequency clock will be BUSCLK.
- ICGLCLK — Development tools can select this internal self-clocked source (~ 8 MHz) to speed up BDC communications in systems where the bus clock is slow.
- ICGERCLK — External reference clock can be selected as the real-time interrupt clock source. Can also be used as the ALTCLK input to the ADC module.

- Illegal opcode detect
- Background debug forced reset
- The reset pin ($\overline{\text{RESET}}$)
- Clock generator loss of lock and loss of clock reset

Each of these sources, with the exception of the background debug forced reset, has an associated bit in the system reset status register. Whenever the MCU enters reset, the internal clock generator (ICG) module switches to self-clocked mode with the frequency of $f_{\text{Self_reset}}$ selected. The reset pin is driven low for 34 bus cycles where the internal bus frequency is half the ICG frequency. After the 34 bus cycles are completed, the pin is released and will be pulled up by the internal pullup resistor, unless it is held low externally. After the pin is released, it is sampled after another 38 bus cycles to determine whether the reset pin is the cause of the MCU reset.

5.4 Computer Operating Properly (COP) Watchdog

The COP watchdog is intended to force a system reset when the application software fails to execute as expected. To prevent a system reset from the COP timer (when it is enabled), application software must reset the COP timer periodically. If the application program gets lost and fails to reset the COP before it times out, a system reset is generated to force the system back to a known starting point. The COP watchdog is enabled by the COPE bit in SOPT (see [Section 5.9.4, “System Options Register \(SOPT\)”](#) for additional information). The COP timer is reset by writing any value to the address of SRS. This write does not affect the data in the read-only SRS. Instead, the act of writing to this address is decoded and sends a reset signal to the COP timer.

After any reset, the COP timer is enabled. This provides a reliable way to detect code that is not executing as intended. If the COP watchdog is not used in an application, it can be disabled by clearing the COPE bit in the write-once SOPT register. Also, the COPT bit can be used to choose one of two timeout periods (2^{18} or 2^{13} cycles of the bus rate clock). Even if the application will use the reset default settings in COPE and COPT, the user should write to write-once SOPT during reset initialization to lock in the settings. That way, they cannot be changed accidentally if the application program gets lost.

The write to SRS that services (clears) the COP timer should not be placed in an interrupt service routine (ISR) because the ISR could continue to be executed periodically even if the main application program fails.

When the MCU is in active background mode, the COP timer is temporarily disabled.

5.5 Interrupts

Interrupts provide a way to save the current CPU status and registers, execute an interrupt service routine (ISR), and then restore the CPU status so processing resumes where it left off before the interrupt. Other than the software interrupt (SWI), which is a program instruction, interrupts are caused by hardware events such as an edge on the IRQ pin or a timer-overflow event. The debug module can also generate an SWI under certain circumstances.

If an event occurs in an enabled interrupt source, an associated read-only status flag will become set. The CPU will not respond until and unless the local interrupt enable is a logic 1 to enable the interrupt. The

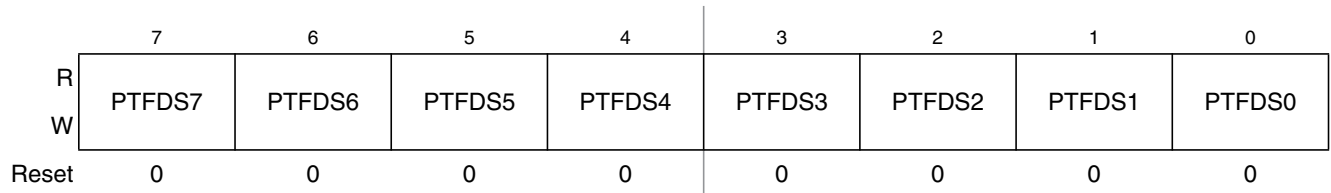


Figure 6-38. Output Drive Strength Selection for Port F (PTFDS)

Table 6-31. PTFDS Register Field Descriptions

Field	Description
7:0 PTFDS[7:0]	<p>Output Drive Strength Selection for Port F Bits — Each of these control bits selects between low and high output drive for the associated PTF pin.</p> <p>0 Low output drive enabled for port F bit n.</p> <p>1 High output drive enabled for port F bit n.</p>

6.7.14 Port G Pin Control Registers (PTGPE, PTGSE, PTGDS)

In addition to the I/O control, port G pins are controlled by the registers listed below.

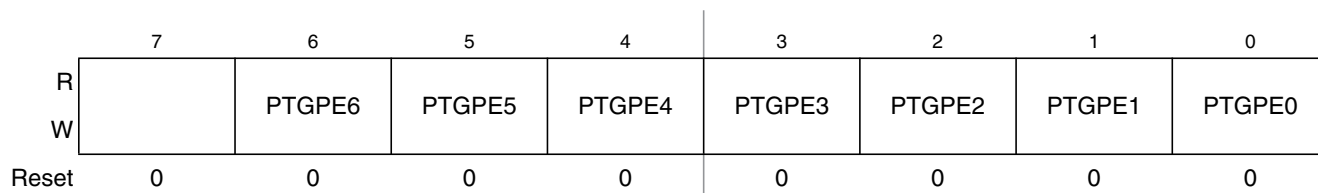


Figure 6-41. Internal Pullup Enable for Port G Bits (PTGPE)

Table 6-34. PTGPE Register Field Descriptions

Field	Description
6:0 PTGPE[6:0]	Internal Pullup Enable for Port G Bits — Each of these control bits determines if the internal pullup device is enabled for the associated PTG pin. For port G pins that are configured as outputs, these bits have no effect and the internal pullup devices are disabled. 0 Internal pullup device disabled for port G bit n. 1 Internal pullup device enabled for port G bit n.

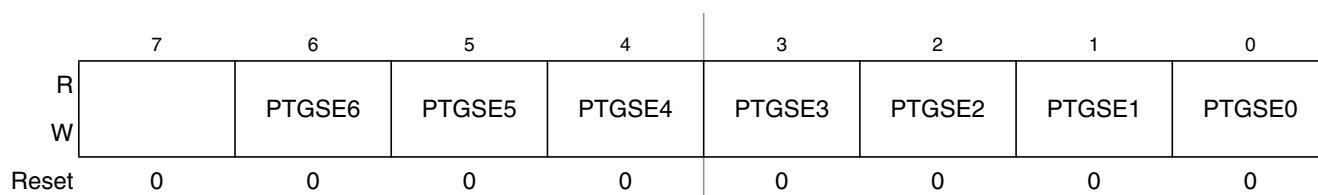


Figure 6-42. Output Slew Rate Control Enable for Port G Bits (PTGSE)

Table 6-35. PTGSE Register Field Descriptions

Field	Description
6:0 PTGSE[6:0]	Output Slew Rate Control Enable for Port G Bits — Each of these control bits determine whether output slew rate control is enabled for the associated PTG pin. For port G pins that are configured as inputs, these bits have no effect. 0 Output slew rate control disabled for port G bit n. 1 Output slew rate control enabled for port G bit n.

- 0 = Bit forced to 0
- 1 = Bit forced to 1
- = Bit set or cleared according to results of operation
- U = Undefined after the operation

Machine coding notation

- dd = Low-order 8 bits of a direct address 0x0000–0x00FF (high byte assumed to be 0x00)
- ee = Upper 8 bits of 16-bit offset
- ff = Lower 8 bits of 16-bit offset or 8-bit offset
- ii = One byte of immediate data
- jj = High-order byte of a 16-bit immediate data value
- kk = Low-order byte of a 16-bit immediate data value
- hh = High-order byte of 16-bit extended address
- ll = Low-order byte of 16-bit extended address
- rr = Relative offset

Source form

Everything in the source forms columns, *except expressions in italic characters*, is literal information that must appear in the assembly source file exactly as shown. The initial 3- to 5-letter mnemonic is always a literal expression. All commas, pound signs (#), parentheses, and plus signs (+) are literal characters.

- n* — Any label or expression that evaluates to a single integer in the range 0–7
- opr8i* — Any label or expression that evaluates to an 8-bit immediate value
- opr16i* — Any label or expression that evaluates to a 16-bit immediate value
- opr8a* — Any label or expression that evaluates to an 8-bit value. The instruction treats this 8-bit value as the low order 8 bits of an address in the direct page of the 64-Kbyte address space (0x00xx).
- opr16a* — Any label or expression that evaluates to a 16-bit value. The instruction treats this value as an address in the 64-Kbyte address space.
- opr8* — Any label or expression that evaluates to an unsigned 8-bit value, used for indexed addressing
- opr16* — Any label or expression that evaluates to a 16-bit value. Because the HCS08 has a 16-bit address bus, this can be either a signed or an unsigned value.
- rel* — Any label or expression that refers to an address that is within –128 to +127 locations from the next address after the last byte of object code for the current instruction. The assembler will calculate the 8-bit signed offset and include it in the object code for this instruction.

Address modes

- INH = Inherent (no operands)
- IMM = 8-bit or 16-bit immediate
- DIR = 8-bit direct
- EXT = 16-bit extended

10.5.3 Center-Aligned PWM Mode

This type of PWM output uses the up-/down-counting mode of the timer counter ($CPWMS = 1$). The output compare value in $TPMxCnVH:TPMxCnVL$ determines the pulse width (duty cycle) of the PWM signal and the period is determined by the value in $TPMxMODH:TPMxMODL$.

$TPMxMODH:TPMxMODL$ should be kept in the range of $0x0001$ to $0x7FFF$ because values outside this range can produce ambiguous results. $ELSnA$ will determine the polarity of the CPWM output.

$$\text{pulse width} = 2 \times (\text{TPMxCnVH:TPMxCnVL}) \quad \text{Eqn. 10-1}$$

$$\begin{aligned} \text{period} &= 2 \times (\text{TPMxMODH:TPMxMODL}); \\ \text{for } \text{TPMxMODH:TPMxMODL} &= 0x0001\text{--}0x7FFF \end{aligned} \quad \text{Eqn. 10-2}$$

If the channel value register $TPMxCnVH:TPMxCnVL$ is zero or negative (bit 15 set), the duty cycle will be 0%. If $TPMxCnVH:TPMxCnVL$ is a positive value (bit 15 clear) and is greater than the (nonzero) modulus setting, the duty cycle will be 100% because the duty cycle compare will never occur. This implies the usable range of periods set by the modulus register is $0x0001$ through $0x7FFE$ ($0x7FFF$ if generation of 100% duty cycle is not necessary). This is not a significant limitation because the resulting period is much longer than required for normal applications.

$TPMxMODH:TPMxMODL = 0x0000$ is a special case that should not be used with center-aligned PWM mode. When $CPWMS = 0$, this case corresponds to the counter running free from $0x0000$ through $0xFFFF$, but when $CPWMS = 1$ the counter needs a valid match to the modulus register somewhere other than at $0x0000$ in order to change directions from up-counting to down-counting.

Figure 10-12 shows the output compare value in the TPM channel registers (multiplied by 2), which determines the pulse width (duty cycle) of the CPWM signal. If $ELSnA = 0$, the compare match while counting up forces the CPWM output signal low and a compare match while counting down forces the output high. The counter counts up until it reaches the modulo setting in $TPMxMODH:TPMxMODL$, then counts down until it reaches zero. This sets the period equal to two times $TPMxMODH:TPMxMODL$.

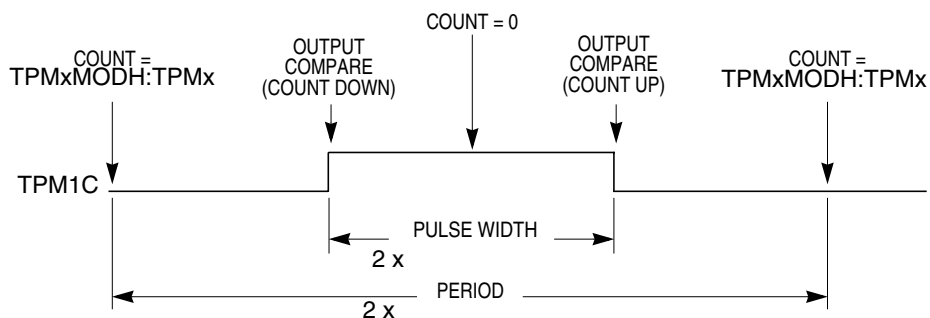


Figure 10-12. CPWM Period and Pulse Width ($ELSnA = 0$)

Center-aligned PWM outputs typically produce less noise than edge-aligned PWMs because fewer I/O pin transitions are lined up at the same system clock edge. This type of PWM is also required for some types of motor drives.

Because the HCS08 is a family of 8-bit MCUs, the settings in the timer channel registers are buffered to ensure coherent 16-bit updates and to avoid unexpected PWM pulse widths. Writes to any of the registers, $TPMxMODH$, $TPMxMODL$, $TPMxCnVH$, and $TPMxCnVL$, actually write to buffer registers. Values are

Table 11-5. SC1xS1 Register Field Descriptions (continued)

Field	Description
5 RDRF	<p>Receive Data Register Full Flag — RDRF becomes set when a character transfers from the receive shifter into the receive data register (SC1xD). In 8-bit mode, to clear RDRF, read SC1xS1 with RDRF = 1 and then read the SCI data register (SC1xD). In 9-bit mode, to clear RDRF, read SC1xS1 with RDRF = 1 and then read SC1xD and the SCI control 3 register (SC1xC3). SC1xD and SC1xC3 can be read in any order, but the flag is cleared only after both data registers are read.</p> <p>0 Receive data register empty. 1 Receive data register full.</p>
4 IDLE	<p>Idle Line Flag — IDLE is set when the SCI receive line becomes idle for a full character time after a period of activity. When ILT = 0, the receiver starts counting idle bit times after the start bit. So if the receive character is all 1s, these bit times and the stop bit time count toward the full character time of logic high (10 or 11 bit times depending on the M control bit) needed for the receiver to detect an idle line. When ILT = 1, the receiver doesn't start counting idle bit times until after the stop bit. So the stop bit and any logic high bit times at the end of the previous character do not count toward the full character time of logic high needed for the receiver to detect an idle line.</p> <p>To clear IDLE, read SC1xS1 with IDLE = 1 and then read the SCI data register (SC1xD). After IDLE has been cleared, it cannot become set again until after a new character has been received and RDRF has been set. IDLE will get set only once even if the receive line remains idle for an extended period.</p> <p>0 No idle line detected. 1 Idle line was detected.</p>
3 OR	<p>Receiver Overrun Flag — OR is set when a new serial character is ready to be transferred to the receive data register (buffer), but the previously received character has not been read from SC1xD yet. In this case, the new character (and all associated error information) is lost because there is no room to move it into SC1xD. To clear OR, read SC1xS1 with OR = 1 and then read the SCI data register (SC1xD).</p> <p>0 No overrun. 1 Receive overrun (new SCI data lost).</p>
2 NF	<p>Noise Flag — The advanced sampling technique used in the receiver takes seven samples during the start bit and three samples in each data bit and the stop bit. If any of these samples disagrees with the rest of the samples within any bit time in the frame, the flag NF will be set at the same time as the flag RDRF gets set for the character. To clear NF, read SC1xS1 and then read the SCI data register (SC1xD).</p> <p>0 No noise detected. 1 Noise detected in the received character in SC1xD.</p>
1 FE	<p>Framing Error Flag — FE is set at the same time as RDRF when the receiver detects a logic 0 where the stop bit was expected. This suggests the receiver was not properly aligned to a character frame. To clear FE, read SC1xS1 with FE = 1 and then read the SCI data register (SC1xD).</p> <p>0 No framing error detected. This does not guarantee the framing is correct. 1 Framing error.</p>
0 PF	<p>Parity Error Flag — PF is set at the same time as RDRF when parity is enabled (PE = 1) and the parity bit in the received character does not agree with the expected parity value. To clear PF, read SC1xS1 and then read the SCI data register (SC1xD).</p> <p>0 No parity error. 1 Parity error.</p>

13.7 Initialization/Application Information

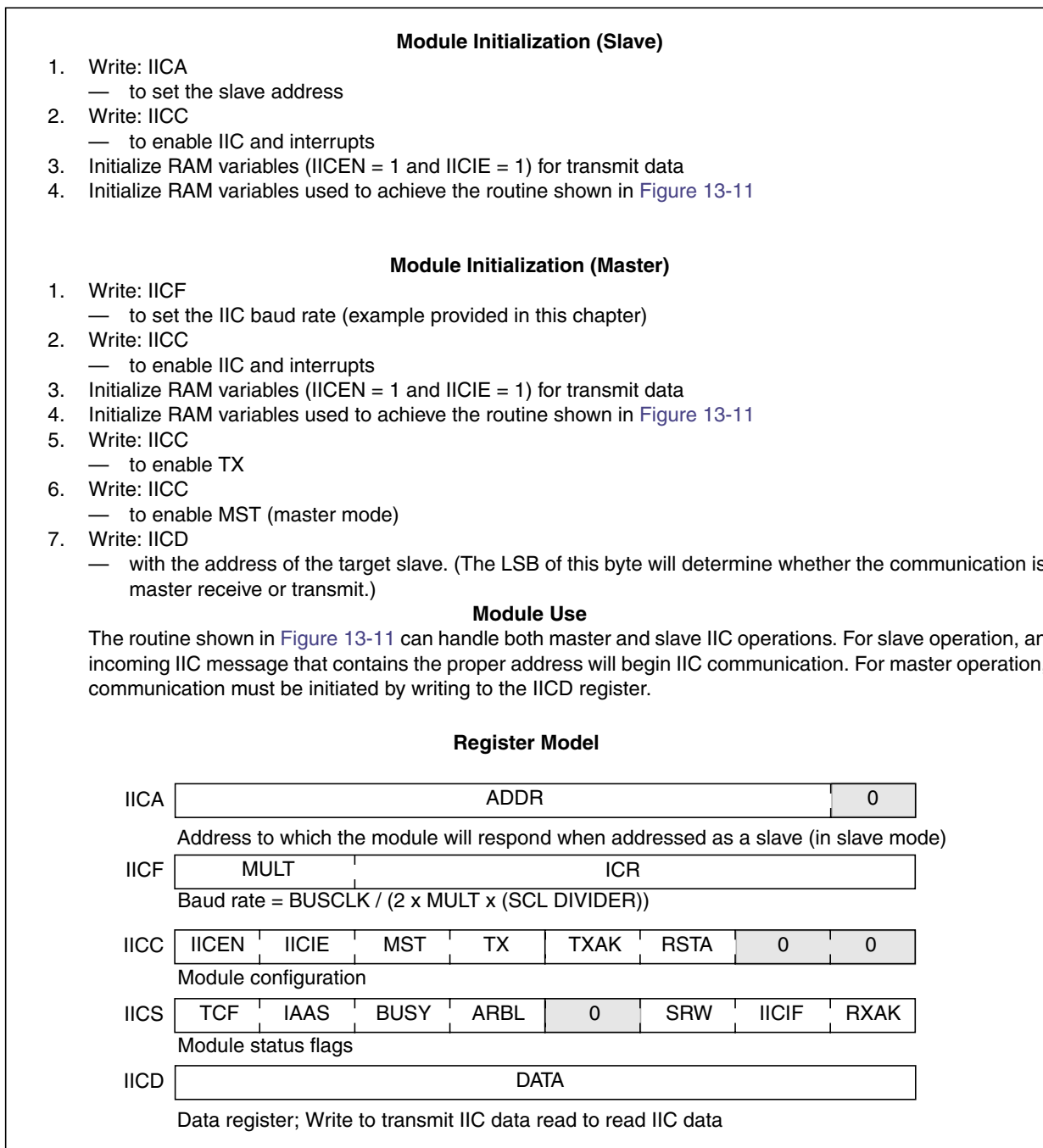


Figure 13-10. IIC Module Quick Start

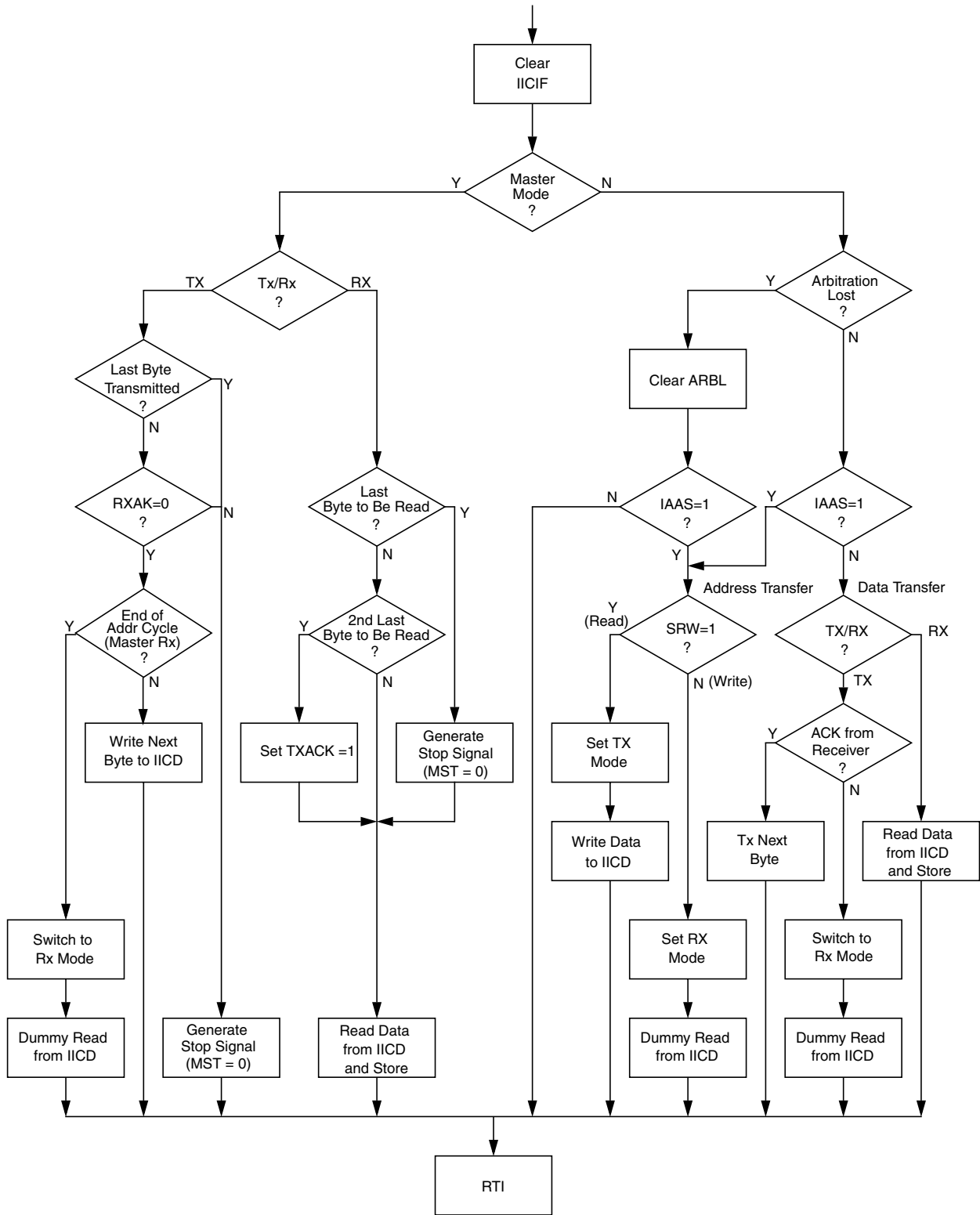


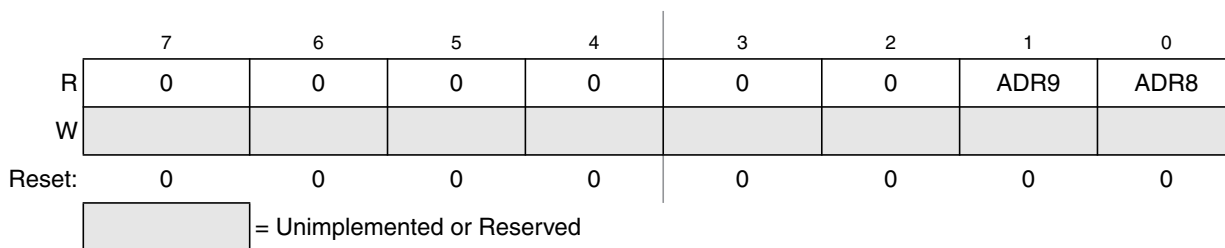
Figure 13-11. Typical IIC Interrupt Routine

Table 14-4. ADC1SC2 Register Field Descriptions (continued)

Field	Description
5 ACFE	Compare Function Enable — ACFE is used to enable the compare function. 0 Compare function disabled 1 Compare function enabled
4 ACFGT	Compare Function Greater Than Enable — ACFGT is used to configure the compare function to trigger when the result of the conversion of the input being monitored is greater than or equal to the compare value. The compare function defaults to triggering when the result of the compare of the input being monitored is less than the compare value. 0 Compare triggers when input is less than compare level 1 Compare triggers when input is greater than or equal to compare level

14.4.3 Data Result High Register (ADC1RH)

ADC1RH contains the upper two bits of the result of a 10-bit conversion. When configured for 8-bit conversions both ADR8 and ADR9 are equal to zero. ADC1RH is updated each time a conversion completes except when automatic compare is enabled and the compare condition is not met. In 10-bit MODE, reading ADC1RH prevents the ADC from transferring subsequent conversion results into the result registers until ADC1RL is read. If ADC1RL is not read until after the next conversion is completed, then the intermediate conversion result will be lost. In 8-bit mode there is no interlocking with ADC1RL. In the case that the MODE bits are changed, any data in ADC1RH becomes invalid.


Figure 14-6. Data Result High Register (ADC1RH)

14.4.4 Data Result Low Register (ADC1RL)

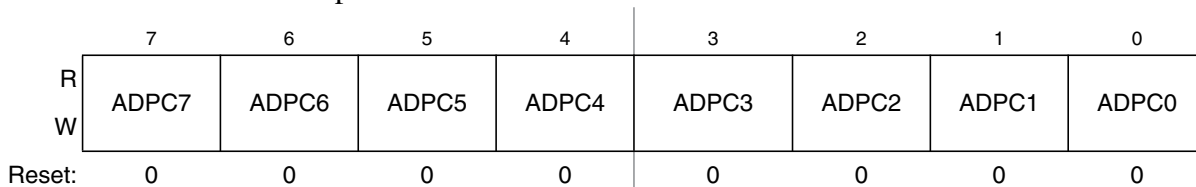
ADC1RL contains the lower eight bits of the result of a 10-bit conversion, and all eight bits of an 8-bit conversion. This register is updated each time a conversion completes except when automatic compare is enabled and the compare condition is not met. In 10-bit mode, reading ADC1RH prevents the ADC from transferring subsequent conversion results into the result registers until ADC1RL is read. If ADC1RL is not read until the after next conversion is completed, then the intermediate conversion results will be lost. In 8-bit mode, there is no interlocking with ADC1RH. In the case that the MODE bits are changed, any data in ADC1RL becomes invalid.

Table 14-8. Input Clock Select

ADICLK	Selected Clock Source
00	Bus clock
01	Bus clock divided by 2
10	Alternate clock (ALTCLK)
11	Asynchronous clock (ADACK)

14.4.8 Pin Control 1 Register (APCTL1)

The pin control registers are used to disable the I/O port control of MCU pins used as analog inputs. APCTL1 is used to control the pins associated with channels 0–7 of the ADC module.


Figure 14-11. Pin Control 1 Register (APCTL1)
Table 14-9. APCTL1 Register Field Descriptions

Field	Description
7 ADPC7	ADC Pin Control 7 — ADPC7 is used to control the pin associated with channel AD7. 0 AD7 pin I/O control enabled 1 AD7 pin I/O control disabled
6 ADPC6	ADC Pin Control 6 — ADPC6 is used to control the pin associated with channel AD6. 0 AD6 pin I/O control enabled 1 AD6 pin I/O control disabled
5 ADPC5	ADC Pin Control 5 — ADPC5 is used to control the pin associated with channel AD5. 0 AD5 pin I/O control enabled 1 AD5 pin I/O control disabled
4 ADPC4	ADC Pin Control 4 — ADPC4 is used to control the pin associated with channel AD4. 0 AD4 pin I/O control enabled 1 AD4 pin I/O control disabled
3 ADPC3	ADC Pin Control 3 — ADPC3 is used to control the pin associated with channel AD3. 0 AD3 pin I/O control enabled 1 AD3 pin I/O control disabled
2 ADPC2	ADC Pin Control 2 — ADPC2 is used to control the pin associated with channel AD2. 0 AD2 pin I/O control enabled 1 AD2 pin I/O control disabled

2. Update status and control register 2 (ADCSC2) to select the conversion trigger (hardware or software) and compare function options, if enabled.
3. Update status and control register 1 (ADCSC1) to select whether conversions will be continuous or completed only once, and to enable or disable conversion complete interrupts. The input channel on which conversions will be performed is also selected here.

14.6.1.2 Pseudo — Code Example

In this example, the ADC module will be set up with interrupts enabled to perform a single 10-bit conversion at low power with a long sample time on input channel 1, where the internal ADCK clock will be derived from the bus clock divided by 1.

ADCCFG = 0x98 (%10011000)

Bit 7	ADLPC	1	Configures for low power (lowers maximum clock speed)
Bit 6:5	ADIV	00	Sets the ADCK to the input clock ÷ 1
Bit 4	ADLSMP	1	Configures for long sample time
Bit 3:2	MODE	10	Sets mode at 10-bit conversions
Bit 1:0	ADICLK	00	Selects bus clock as input clock source

ADCSC2 = 0x00 (%00000000)

Bit 7	ADACT	0	Flag indicates if a conversion is in progress
Bit 6	ADTRG	0	Software trigger selected
Bit 5	ACFE	0	Compare function disabled
Bit 4	ACFGT	0	Not used in this example
Bit 3:2		00	Unimplemented or reserved, always reads zero
Bit 1:0		00	Reserved for Freescale's internal use; always write zero

ADCSC1 = 0x41 (%01000001)

Bit 7	COCO	0	Read-only flag which is set when a conversion completes
Bit 6	AIEN	1	Conversion complete interrupt enabled
Bit 5	ADCO	0	One conversion only (continuous conversions disabled)
Bit 4:0	ADCH	00001	Input channel 1 selected as ADC input channel

ADCRH/L = 0xxx

Holds results of conversion. Read high byte (ADCRH) before low byte (ADCRL) so that conversion data cannot be overwritten with data from the next conversion.

ADCCVH/L = 0xxx

Holds compare value when compare function enabled

APCTL1=0x02

AD1 pin I/O control disabled. All other AD pins remain general purpose I/O pins

APCTL2=0x00

All other AD pins remain general purpose I/O pins

- Average the result by converting the analog input many times in succession and dividing the sum of the results. Four samples are required to eliminate the effect of a 1LSB, one-time error.
- Reduce the effect of synchronous noise by operating off the asynchronous clock (ADACK) and averaging. Noise that is synchronous to ADCK cannot be averaged out.

14.7.2.4 Code Width and Quantization Error

The ADC quantizes the ideal straight-line transfer function into 1024 steps (in 10-bit mode). Each step ideally has the same height (1 code) and width. The width is defined as the delta between the transition points to one code and the next. The ideal code width for an N bit converter (in this case N can be 8 or 10), defined as 1LSB, is:

$$1\text{LSB} = (V_{\text{REFH}} - V_{\text{REFL}}) / 2^N \quad \text{Eqn. 14-2}$$

There is an inherent quantization error due to the digitization of the result. For 8-bit or 10-bit conversions the code will transition when the voltage is at the midpoint between the points where the straight line transfer function is exactly represented by the actual transfer function. Therefore, the quantization error will be $\pm 1/2\text{LSB}$ in 8- or 10-bit mode. As a consequence, however, the code width of the first (\$000) conversion is only $1/2\text{LSB}$ and the code width of the last (\$FF or \$3FF) is 1.5LSB .

14.7.2.5 Linearity Errors

The ADC may also exhibit non-linearity of several forms. Every effort has been made to reduce these errors but the system should be aware of them because they affect overall accuracy. These errors are:

- Zero-scale error (E_{ZS}) (sometimes called offset) — This error is defined as the difference between the actual code width of the first conversion and the ideal code width ($1/2\text{LSB}$). Note, if the first conversion is \$001, then the difference between the actual \$001 code width and its ideal (1LSB) is used.
- Full-scale error (E_{FS}) — This error is defined as the difference between the actual code width of the last conversion and the ideal code width (1.5LSB). Note, if the last conversion is \$3FE, then the difference between the actual \$3FE code width and its ideal (1LSB) is used.
- Differential non-linearity (DNL) — This error is defined as the worst-case difference between the actual code width and the ideal code width for all conversions.
- Integral non-linearity (INL) — This error is defined as the highest-value the (absolute value of the) running sum of DNL achieves. More simply, this is the worst-case difference of the actual transition voltage to a given code and its corresponding ideal transition voltage, for all codes.
- Total unadjusted error (TUE) — This error is defined as the difference between the actual transfer function and the ideal straight-line transfer function, and therefore includes all forms of error.

14.7.2.6 Code Jitter, Non-Monotonicity and Missing Codes

Analog-to-digital converters are susceptible to three special forms of error. These are code jitter, non-monotonicity, and missing codes.

Code jitter is when, at certain points, a given input voltage converts to one of two values when sampled repeatedly. Ideally, when the input voltage is infinitesimally smaller than the transition voltage, the

Figure 15-4 shows the host receiving a logic 0 from the target HCS08 MCU. Because the host is asynchronous to the target MCU, there is a 0-to-1 cycle delay from the host-generated falling edge on BKGD to the start of the bit time as perceived by the target MCU. The host initiates the bit time but the target HCS08 finishes it. Because the target wants the host to receive a logic 0, it drives the BKGD pin low for 13 BDC clock cycles, then briefly drives it high to speed up the rising edge. The host samples the bit level about 10 cycles after starting the bit time.

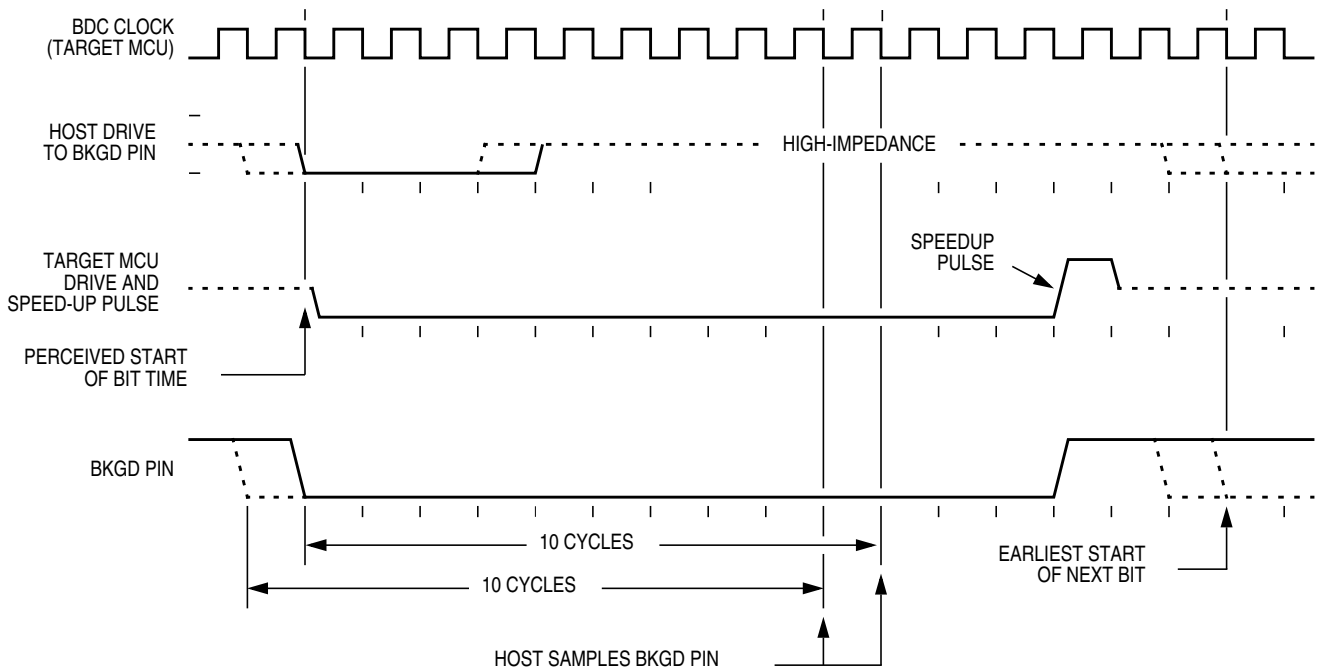


Figure 15-4. BDM Target-to-Host Serial Bit Timing (Logic 0)

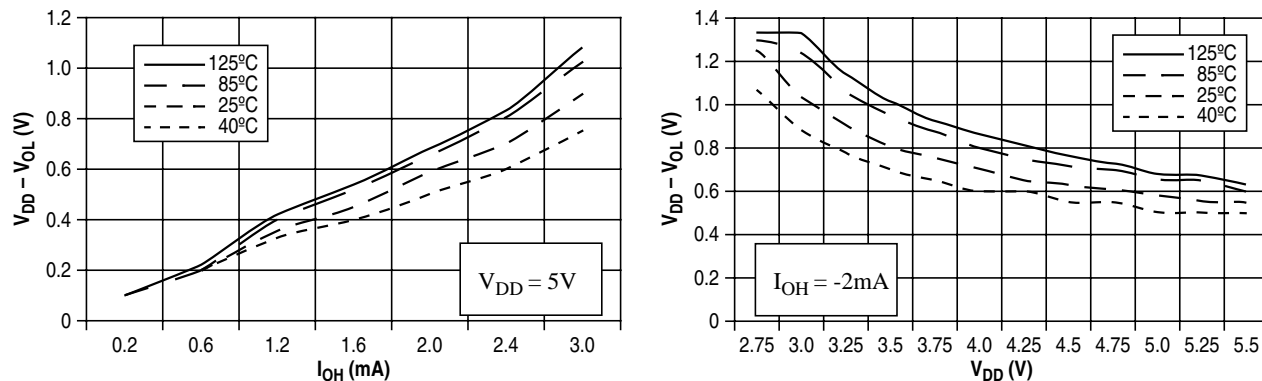


Figure A-3. Typical High-Side (Source) Characteristics — Low Drive (PTxDSn = 0)

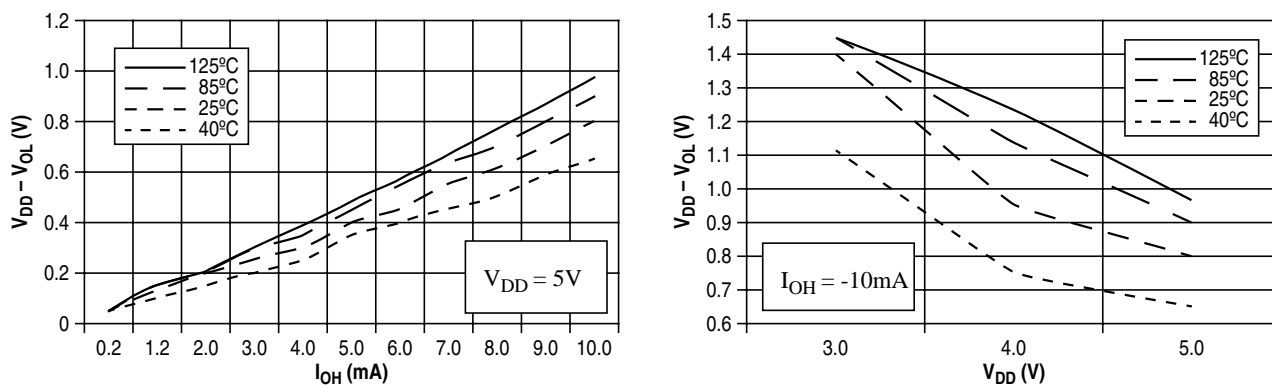
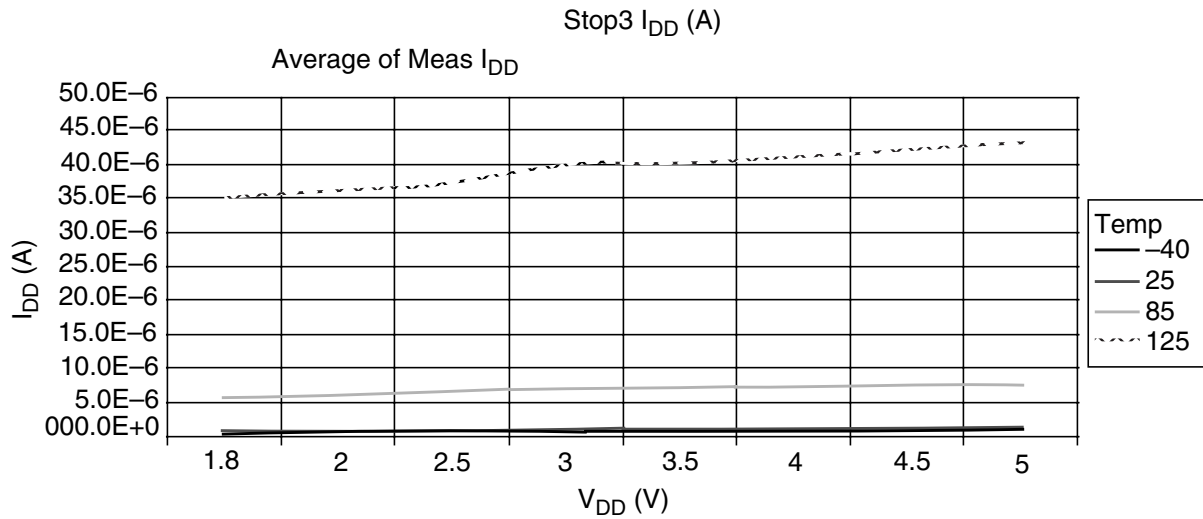


Figure A-4. Typical High-Side (Source) Characteristics — High Drive (PTxDSn = 1)


 Figure A-7. Typical Stop3 I_{DD}

A.8 ADC Characteristics

Table A-9. 5 Volt 10-bit ADC Operating Conditions

Characteristic	Conditions	Symb	Min	Typ ¹	Max	Unit
Supply voltage	Absolute	V_{DDAD}	2.7	—	5.5	V
	Delta to V_{DD} ($V_{DD}-V_{DDAD}$) ²	ΔV_{DDAD}	-100	0	+100	mV
Ground voltage	Delta to V_{SS} ($V_{SS}-V_{SSAD}$) ²	ΔV_{SSAD}	-100	0	+100	mV
Ref voltage high		V_{REFH}	2.7	V_{DDAD}	V_{DDAD}	V
Ref voltage low		V_{REFL}	V_{SSAD}	V_{SSAD}	V_{SSAD}	V
Input voltage		V_{ADIN}	V_{REFL}	—	V_{REFH}	V
Input capacitance		C_{ADIN}	—	4.5	5.5	pF
Input resistance		R_{ADIN}	—	3	5	k Ω
Analog source resistance External to MCU	10-bit mode $f_{ADCK} > 4\text{MHz}$ $f_{ADCK} < 4\text{MHz}$	R_{AS}	—	—	5	k Ω
	8-bit mode (all valid f_{ADCK})		—	—	10	
ADC conversion clock frequency	High speed (ADLPC = 0)	f_{ADCK}	0.4	—	8.0	MHz
	Low power (ADLPC = 1)		0.4	—	4.0	

¹ Typical values assume $V_{DDAD} = 5.0\text{ V}$, Temp = 25°C, $f_{ADCK} = 1.0\text{ MHz}$ unless otherwise stated. Typical values are for reference only and are not tested in production.

² dc potential difference.

Appendix B

Ordering Information and Mechanical Drawings

B.1 Ordering Information

This section contains ordering numbers for MC9S08AW60 Series devices. See below for an example of the device numbering system.

Table B-1. Consumer and Industrial Device Numbering System

Device Number ¹	Memory		Available Packages ²
	FLASH	RAM	Type
MC9S08AW60	63,280	2048	64-pin LQFP
MC9S08AW48	49,152		64-pin QFP
MC9S08AW32	32,768		48-pin QFN
MC9S08AW16	16,384	1024	44-pin LQFP

¹ See [Table 1-1](#) for a complete description of modules included on each device.

² See [Table B-3](#) for package information.

Table B-2. Automotive Device Numbering System

Device Number ¹	Memory		Available Packages ²
	FLASH	RAM	Type
S9S08AW60	63,280	2048	64-pin LQFP
S9S08AW48	49,152		48-pin QFN
S9S08AW32	32,768		44-pin LQFP
S9S08AW16	16,384	1024	48-pin QFN 44-pin LQFP

¹ See [Table 1-1](#) for a complete description of modules included on each device.

² See [Table B-3](#) for package information.