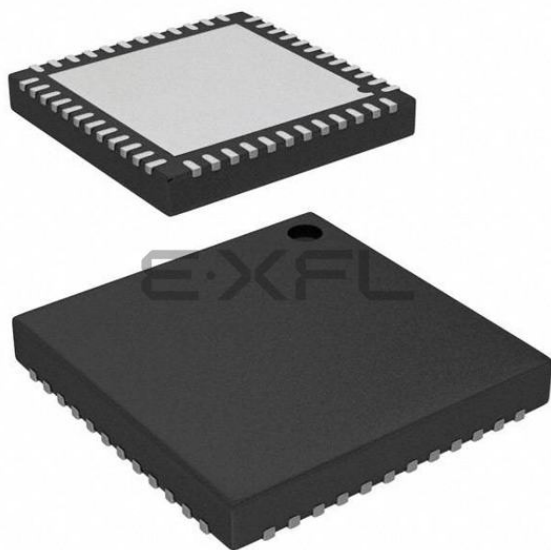




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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Active
Core Processor	8051
Core Size	8-Bit
Speed	67MHz
Connectivity	EBI/EMI, I ² C, LINbus, SPI, UART/USART, USB
Peripherals	CapSense, DMA, LCD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	64KB (64K x 8)
Program Memory Type	FLASH
EEPROM Size	2K x 8
RAM Size	8K x 8
Voltage - Supply (Vcc/Vdd)	1.71V ~ 5.5V
Data Converters	A/D 16x20b; D/A 4x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-VFQFN Exposed Pad
Supplier Device Package	48-QFN (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/cy8c3866lti-067t

4.3.1.2 Logical Instructions

The logical instructions perform Boolean operations such as AND, OR, XOR on bytes, rotate of accumulator contents, and swap of nibbles in an accumulator. The Boolean operations on the bytes are performed on the bit-by-bit basis. [Table 4-2](#) shows the list of logical instructions and their description.

Table 4-2. Logical Instructions

Mnemonic	Description	Bytes	Cycles
ANL A,Rn	AND register to accumulator	1	1
ANL A,Direct	AND direct byte to accumulator	2	2
ANL A,@Ri	AND indirect RAM to accumulator	1	2
ANL A,#data	AND immediate data to accumulator	2	2
ANL Direct, A	AND accumulator to direct byte	2	3
ANL Direct, #data	AND immediate data to direct byte	3	3
ORL A,Rn	OR register to accumulator	1	1
ORL A,Direct	OR direct byte to accumulator	2	2
ORL A,@Ri	OR indirect RAM to accumulator	1	2
ORL A,#data	OR immediate data to accumulator	2	2
ORL Direct, A	OR accumulator to direct byte	2	3
ORL Direct, #data	OR immediate data to direct byte	3	3
XRL A,Rn	XOR register to accumulator	1	1
XRL A,Direct	XOR direct byte to accumulator	2	2
XRL A,@Ri	XOR indirect RAM to accumulator	1	2
XRL A,#data	XOR immediate data to accumulator	2	2
XRL Direct, A	XOR accumulator to direct byte	2	3
XRL Direct, #data	XOR immediate data to direct byte	3	3
CLR A	Clear accumulator	1	1
CPL A	Complement accumulator	1	1
RL A	Rotate accumulator left	1	1
RLC A	Rotate accumulator left through carry	1	1
RR A	Rotate accumulator right	1	1
RRC A	Rotate accumulator right though carry	1	1
SWAP A	Swap nibbles within accumulator	1	1

4.3.1.3 Data Transfer Instructions

The data transfer instructions are of three types: the core RAM, xdata RAM, and the lookup tables. The core RAM transfer includes transfer between any two core RAM locations or SFRs. These instructions can use direct, indirect, register, and immediate addressing. The xdata RAM transfer includes only the transfer between the accumulator and the xdata RAM location. It can use only indirect addressing. The lookup tables involve nothing but the read of program memory using the Indexed

addressing mode. [Table 4-3](#) lists the various data transfer instructions available.

4.3.1.4 Boolean Instructions

The 8051 core has a separate bit-addressable memory location. It has 128 bits of bit addressable RAM and a set of SFRs that are bit addressable. The instruction set includes the whole menu of bit operations such as move, set, clear, toggle, OR, and AND instructions and the conditional jump instructions. [Table 4-4](#) lists the available Boolean instructions.

Table 4-4. Boolean Instructions *(continued)*

Mnemonic	Description	Bytes	Cycles
ANL C, /bit	AND complement of direct bit to carry	2	2
ORL C, bit	OR direct bit to carry	2	2
ORL C, /bit	OR complement of direct bit to carry	2	2
MOV C, bit	Move direct bit to carry	2	2
MOV bit, C	Move carry to direct bit	2	3
JC rel	Jump if carry is set	2	3
JNC rel	Jump if no carry is set	2	3
JB bit, rel	Jump if direct bit is set	3	5
JNB bit, rel	Jump if direct bit is not set	3	5
JBC bit, rel	Jump if direct bit is set and clear bit	3	5

4.3.1.5 Program Branching Instructions

The 8051 supports a set of conditional and unconditional jump instructions that help to modify the program execution flow. [Table 4-5](#) shows the list of jump instructions.

Table 4-5. Jump Instructions

Mnemonic	Description	Bytes	Cycles
ACALL addr11	Absolute subroutine call	2	4
LCALL addr16	Long subroutine call	3	4
RET	Return from subroutine	1	4
RETI	Return from interrupt	1	4
AJMP addr11	Absolute jump	2	3
LJMP addr16	Long jump	3	4
SJMP rel	Short jump (relative address)	2	3
JMP @A + DPTR	Jump indirect relative to DPTR	1	5
JZ rel	Jump if accumulator is zero	2	4
JNZ rel	Jump if accumulator is nonzero	2	4
CJNE A, Direct, rel	Compare direct byte to accumulator and jump if not equal	3	5
CJNE A, #data, rel	Compare immediate data to accumulator and jump if not equal	3	4
CJNE Rn, #data, rel	Compare immediate data to register and jump if not equal	3	4
CJNE @Ri, #data, rel	Compare immediate data to indirect RAM and jump if not equal	3	5
DJNZ Rn, rel	Decrement register and jump if not zero	2	4
DJNZ Direct, rel	Decrement direct byte and jump if not zero	3	5
NOP	No operation	1	1

Table 4-8. Interrupt Vector Table

#	Fixed Function	DMA	UDB
0	LVD	phub_termout0[0]	udb_intr[0]
1	Cache/ECC	phub_termout0[1]	udb_intr[1]
2	Reserved	phub_termout0[2]	udb_intr[2]
3	Sleep (Pwr Mgr)	phub_termout0[3]	udb_intr[3]
4	PICU[0]	phub_termout0[4]	udb_intr[4]
5	PICU[1]	phub_termout0[5]	udb_intr[5]
6	PICU[2]	phub_termout0[6]	udb_intr[6]
7	PICU[3]	phub_termout0[7]	udb_intr[7]
8	PICU[4]	phub_termout0[8]	udb_intr[8]
9	PICU[5]	phub_termout0[9]	udb_intr[9]
10	PICU[6]	phub_termout0[10]	udb_intr[10]
11	PICU[12]	phub_termout0[11]	udb_intr[11]
12	PICU[15]	phub_termout0[12]	udb_intr[12]
13	Comparators Combined	phub_termout0[13]	udb_intr[13]
14	Switched Caps Combined	phub_termout0[14]	udb_intr[14]
15	I ² C	phub_termout0[15]	udb_intr[15]
16	CAN	phub_termout1[0]	udb_intr[16]
17	Timer/Counter0	phub_termout1[1]	udb_intr[17]
18	Timer/Counter1	phub_termout1[2]	udb_intr[18]
19	Timer/Counter2	phub_termout1[3]	udb_intr[19]
20	Timer/Counter3	phub_termout1[4]	udb_intr[20]
21	USB SOF Int	phub_termout1[5]	udb_intr[21]
22	USB Arb Int	phub_termout1[6]	udb_intr[22]
23	USB Bus Int	phub_termout1[7]	udb_intr[23]
24	USB Endpoint[0]	phub_termout1[8]	udb_intr[24]
25	USB Endpoint Data	phub_termout1[9]	udb_intr[25]
26	Reserved	phub_termout1[10]	udb_intr[26]
27	LCD	phub_termout1[11]	udb_intr[27]
28	DFB Int	phub_termout1[12]	udb_intr[28]
29	Decimator Int	phub_termout1[13]	udb_intr[29]
30	PHUB Error Int	phub_termout1[14]	udb_intr[30]
31	EEPROM Fault Int	phub_termout1[15]	udb_intr[31]

5.5 Nonvolatile Latches (NVLs)

PSoC has a 4-byte array of nonvolatile latches (NVLs) that are used to configure the device at reset. The NVL register map is shown in [Table 5-2](#).

Table 5-2. Device Configuration NVL Register Map

Register Address	7	6	5	4	3	2	1	0
0x00	PRT3RDM[1:0]		PRT2RDM[1:0]		PRT1RDM[1:0]		PRT0RDM[1:0]	
0x01	PRT12RDM[1:0]		PRT6RDM[1:0]		PRT5RDM[1:0]		PRT4RDM[1:0]	
0x02	XRESMEN	DBGEN					PRT15RDM[1:0]	
0x03	DIG_PHS_DLY[3:0]				ECCEN	DPS[1:0]		CFGSPD

The details for individual fields and their factory default settings are shown in [Table 5-3](#).

Table 5-3. Fields and Factory Default Settings

Field	Description	Settings
PRTxRDM[1:0]	Controls reset drive mode of the corresponding IO port. See “Reset Configuration” on page 43. All pins of the port are set to the same mode.	00b (default) - high impedance analog 01b - high impedance digital 10b - resistive pull up 11b - resistive pull down
XRESMEN	Controls whether pin P1[2] is used as a GPIO or as an external reset. See “Pin Descriptions” on page 12, XRES description.	0 (default for 68-pin, 72-pin, and 100-pin parts) - GPIO 1 (default for 48-pin parts) - external reset
DBGEN	Debug Enable allows access to the debug system, for third-party programmers.	0 - access disabled 1 (default) - access enabled
CFGSPD	Controls the speed of the IMO-based clock during the device boot process, for faster boot or low-power operation	0 (default) - 12 MHz IMO 1 - 48 MHz IMO
DPS[1:0]	Controls the usage of various P1 pins as a debug port. See “Programming, Debug Interfaces, Resources” on page 65.	00b - 5-wire JTAG 01b (default) - 4-wire JTAG 10b - SWD 11b - debug ports disabled
ECCEN	Controls whether ECC flash is used for ECC or for general configuration and data storage. See “Flash Program Memory” on page 23.	0 - ECC disabled 1 (default) - ECC enabled
DIG_PHS_DLY[3:0]	Selects the digital clock phase delay.	See the TRM for details.

Although PSoC Creator provides support for modifying the device configuration NVLs, the number of NVL erase / write cycles is limited – see [“Nonvolatile Latches \(NVL\)”](#) on page 113.

Independent of the ALU operation, these functions are available:

- Shift left
- Shift right
- Nibble swap
- Bitwise OR mask

7.2.2.3 Conditionals

Each datapath has two compares, with bit masking options. Compare operands include the two accumulators and the two data registers in a variety of configurations. Other conditions include zero detect, all ones detect, and overflow. These conditions are the primary datapath outputs, a selection of which can be driven out to the UDB routing matrix. Conditional computation can use the built in chaining to neighboring UDBs to operate on wider data widths without the need to use routing resources.

7.2.2.4 Variable MSB

The most significant bit of an arithmetic and shift function can be programmatically specified. This supports variable width CRC and PRS functions, and in conjunction with ALU output masking, can implement arbitrary width timers, counters and shift blocks.

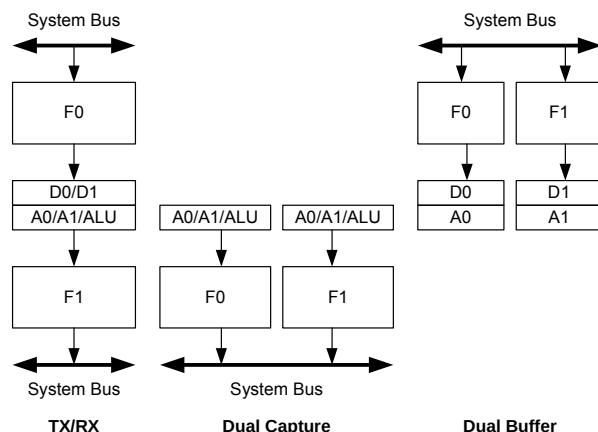
7.2.2.5 Built in CRC/PRS

The datapath has built-in support for single cycle CRC computation and PRS generation of arbitrary width and arbitrary polynomial. CRC/PRS functions longer than 8 bits may be implemented in conjunction with PLD logic, or built in chaining may be used to extend the function into neighboring UDBs.

7.2.2.6 Input/Output FIFOs

Each datapath contains two four-byte deep FIFOs, which can be independently configured as an input buffer (system bus writes to the FIFO, datapath internal reads the FIFO), or an output buffer (datapath internal writes to the FIFO, the system bus reads from the FIFO). The FIFOs generate status that are selectable as datapath outputs and can therefore be driven to the routing, to interact with sequencers, interrupts, or DMA.

Figure 7-5. Example FIFO Configurations



7.2.2.7 Chaining

The datapath can be configured to chain conditions and signals such as carries and shift data with neighboring datapaths to create higher precision arithmetic, shift, CRC/PRS functions.

7.2.2.8 Time Multiplexing

In applications that are over sampled, or do not need high clock rates, the single ALU block in the datapath can be efficiently shared with two sets of registers and condition generators. Carry and shift out data from the ALU are registered and can be selected as inputs in subsequent cycles. This provides support for 16-bit functions in one (8-bit) datapath.

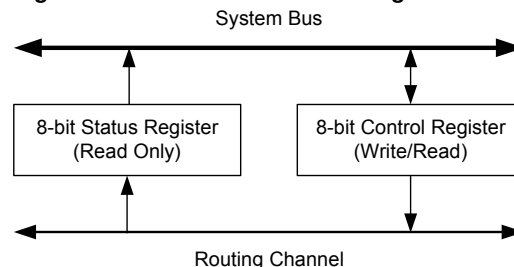
7.2.2.9 Datapath I/O

There are six inputs and six outputs that connect the datapath to the routing matrix. Inputs from the routing provide the configuration for the datapath operation to perform in each cycle, and the serial data inputs. Inputs can be routed from other UDB blocks, other device peripherals, device I/O pins, and so on. The outputs to the routing can be selected from the generated conditions, and the serial data outputs. Outputs can be routed to other UDB blocks, device peripherals, interrupt and DMA controller, I/O pins, and so on.

7.2.3 Status and Control Module

The primary purpose of this circuitry is to coordinate CPU firmware interaction with internal UDB operation.

Figure 7-6. Status and Control Registers



The bits of the control register, which may be written to by the system bus, are used to drive into the routing matrix, and thus provide firmware with the opportunity to control the state of UDB processing. The status register is read-only and it allows internal UDB state to be read out onto the system bus directly from internal routing. This allows firmware to monitor the state of UDB processing. Each bit of these registers has programmable connections to the routing matrix and routing connections are made depending on the requirements of the application.

7.2.3.1 Usage Examples

As an example of control input, a bit in the control register can be allocated as a function enable bit. There are multiple ways to enable a function. In one method the control bit output would be routed to the clock control block in one or more UDBs and serve as a clock enable for the selected UDB blocks. A status example is a case where a PLD or datapath block generated a condition, such as a “compare true” condition that is captured and latched by the status register and then read (and cleared) by CPU firmware.

The opamp and resistor array is programmable to perform various analog functions including

- Naked operational amplifier – Continuous mode
- Unity-gain buffer – Continuous mode
- PGA – Continuous mode
- Transimpedance amplifier (TIA) – Continuous mode
- Up/down mixer – Continuous mode
- Sample and hold mixer (NRZ S/H) – Switched cap mode
- First order analog to digital modulator – Switched cap mode

8.5.1 Naked Opamp

The Naked Opamp presents both inputs and the output for connection to internal or external signals. The opamp has a unity gain bandwidth greater than 6.0 MHz and output drive current up to 650 μ A. This is sufficient for buffering internal signals (such as DAC outputs) and driving external loads greater than 7.5 kohms.

8.5.2 Unity Gain

The Unity Gain buffer is a Naked Opamp with the output directly connected to the inverting input for a gain of 1.00. It has a –3 dB bandwidth greater than 6.0 MHz.

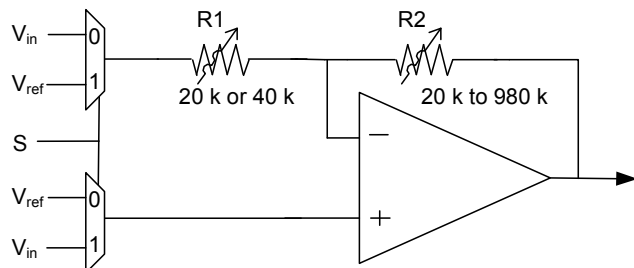
8.5.3 PGA

The PGA amplifies an external or internal signal. The PGA can be configured to operate in inverting mode or noninverting mode. The PGA function may be configured for both positive and negative gains as high as 50 and 49 respectively. The gain is adjusted by changing the values of R1 and R2 as illustrated in Figure 8-8. The schematic in Figure 8-8 shows the configuration and possible resistor settings for the PGA. The gain is switched from inverting and non inverting by changing the shared select value of the both the input muxes. The bandwidth for each gain case is listed in Table 8-3.

Table 8-3. Bandwidth

Gain	Bandwidth
1	6.0 MHz
24	340 kHz
48	220 kHz
50	215 kHz

Figure 8-8. PGA Resistor Settings



The PGA is used in applications where the input signal may not be large enough to achieve the desired resolution in the ADC, or dynamic range of another SC/CT block such as a mixer. The gain is adjustable at runtime, including changing the gain of the PGA prior to each ADC sample.

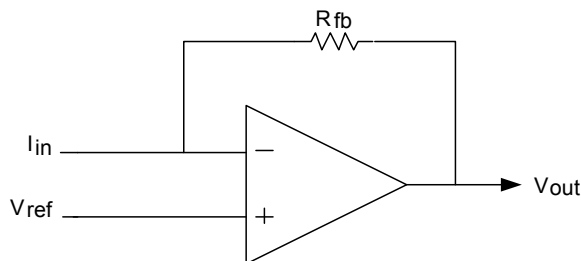
8.5.4 TIA

The Transimpedance Amplifier (TIA) converts an internal or external current to an output voltage. The TIA uses an internal feedback resistor in a continuous time configuration to convert input current to output voltage. For an input current I_{in} , the output voltage is $V_{REF} - I_{in} \times R_{fb}$, where V_{REF} is the value placed on the non inverting input. The feedback resistor R_{fb} is programmable between 20 K Ω and 1 M Ω through a configuration register. Table 8-4 shows the possible values of R_{fb} and associated configuration settings.

Table 8-4. Feedback Resistor Settings

Configuration Word	Nominal R_{fb} (K Ω)
000b	20
001b	30
010b	40
011b	60
100b	120
101b	250
110b	500
111b	1000

Figure 8-9. Continuous Time TIA Schematic

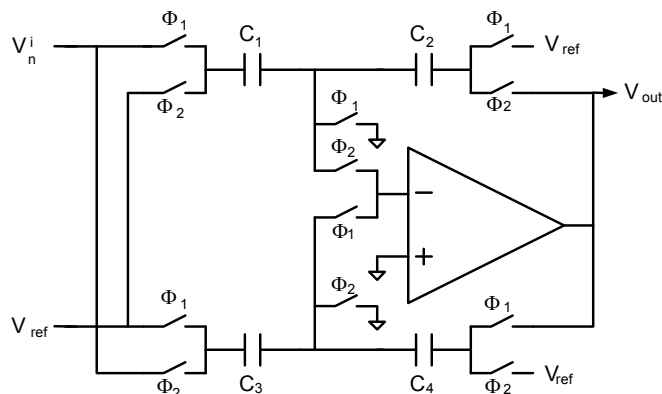


The TIA configuration is used for applications where an external sensor's output is current as a function of some type of stimulus such as temperature, light, magnetic flux etc. In a common application, the voltage DAC output can be connected to the V_{REF} TIA input to allow calibration of the external sensor bias current by adjusting the voltage DAC output voltage.

8.6 LCD Direct Drive

The PSoC LCD driver system is a highly configurable peripheral designed to allow PSoC to directly drive a broad range of LCD glass. All voltages are generated on chip, eliminating the need for external components. With a high multiplex ratio of up to 1/16, the CY8C38 family LCD driver system can drive a maximum of 736 segments. The PSoC LCD driver module was also designed with the conservative power budget of portable devices in mind, enabling different LCD drive modes and power down modes to conserve power.

Figure 8-13. Sample and Hold Topology
($\Phi 1$ and $\Phi 2$ are opposite phases of a clock)



8.11.1 Down Mixer

The SC/CT block can be used as a mixer to down convert an input signal. This circuit is a high bandwidth passive sample network that can sample input signals up to 14 MHz. This sampled value is then held using the opamp with a maximum clock rate of 4 MHz. The output frequency is at the difference between the input frequency and the highest integer multiple of the Local Oscillator that is less than the input.

8.11.2 First Order Modulator – SC Mode

A first order modulator is constructed by placing the SC/CT block in an integrator mode and using a comparator to provide a 1-bit feedback to the input. Depending on this bit, a reference voltage is either subtracted or added to the input signal. The block output is the output of the comparator and not the integrator in the modulator case. The signal is downshifted and buffered and then processed by a decimator to make a delta-sigma converter or a counter to make an incremental converter. The accuracy of the sampled data from the first-order modulator is determined from several factors.

The main application for this modulator is for a low-frequency ADC with high accuracy. Applications include strain gauges, thermocouples, precision voltage, and current measurement.

9. Programming, Debug Interfaces, Resources

PSoC devices include extensive support for programming, testing, debugging, and tracing both hardware and firmware. Three interfaces are available: JTAG, SWD, and SWV. JTAG and SWD support all programming and debug features of the device. JTAG also supports standard JTAG scan chains for board level test and chaining multiple JTAG devices to a single JTAG connection.

For more information on PSoC 3 Programming, refer to the [PSoC® 3 Device Programming Specifications](#).

Complete Debug on Chip (DoC) functionality enables full device debugging in the final system using the standard production

device. It does not require special interfaces, debugging pods, simulators, or emulators. Only the standard programming connections are required to fully support debug.

The PSoC Creator IDE software provides fully integrated programming and debug support for PSoC devices. The low cost MiniProg3 programmer and debugger is designed to provide full programming and debug support of PSoC devices in conjunction with the PSoC Creator IDE. PSoC JTAG, SWD, and SWV interfaces are fully compatible with industry standard third party tools.

All DOC circuits are disabled by default and can only be enabled in firmware. If not enabled, the only way to reenables them is to erase the entire device, clear flash protection, and reprogram the device with new firmware that enables DOC. Disabling DOC features, robust flash protection, and hiding custom analog and digital functionality inside the PSoC device provide a level of security not possible with multichip application solutions. Additionally, all device interfaces can be permanently disabled (Device Security) for applications concerned about phishing attacks due to a maliciously reprogrammed device. Permanently disabling interfaces is not recommended in most applications because you cannot access the device later. Because all programming, debug, and test interfaces are disabled when device security is enabled, PSoCs with Device Security enabled may not be returned for failure analysis.

Table 9-1. Debug Configurations

Debug and Trace Configuration	GPIO Pins Used
All debug and trace disabled	0
JTAG	4 or 5
SWD	2
SWV	1
SWD + SWV	3

9.1 JTAG Interface

The IEEE 1149.1 compliant JTAG interface exists on four or five pins (the nTRST pin is optional). The JTAG interface is used for programming the flash memory, debugging, I/O scan chains, and JTAG device chaining.

PSoC 3 has certain timing requirements to be met for entering programming mode through the JTAG interface. Due to these timing requirements, not all standard JTAG programmers, or standard JTAG file formats such as SVF or STAPL, can support PSoC 3 programming. The list of programmers that support PSoC 3 programming is available at <http://www.cypress.com/go/programming>.

The JTAG clock frequency can be up to 14 MHz, or 1/3 of the CPU clock frequency for 8 and 16-bit transfers, or 1/5 of the CPU clock frequency for 32-bit transfers. By default, the JTAG pins are enabled on new devices but the JTAG interface can be disabled, allowing these pins to be used as GPIO instead.

9.3 Debug Features

Using the JTAG or SWD interface, the CY8C38 supports the following debug features:

- Halt and single-step the CPU
- View and change CPU and peripheral registers, and RAM addresses
- Eight program address breakpoints
- One memory access breakpoint—break on reading or writing any memory address and data value
- Break on a sequence of breakpoints (non recursive)
- Debugging at the full speed of the CPU
- Compatible with PSoC Creator and MiniProg3 programmer and debugger
- Standard JTAG programming and debugging interfaces make CY8C38 compatible with other popular third-party tools (for example, ARM / Keil)

9.4 Trace Features

The CY8C38 supports the following trace features when using JTAG or SWD:

- Trace the 8051 program counter (PC), accumulator register (ACC), and one SFR / 8051 core RAM register
- Trace depth up to 1000 instructions if all registers are traced, or 2000 instructions if only the PC is traced (on devices that include trace memory)
- Program address trigger to start tracing
- Trace windowing, that is, only trace when the PC is within a given range
- Two modes for handling trace buffer full: continuous (overwriting the oldest trace data) or break when trace buffer is full

9.5 Single Wire Viewer Interface

The SWV interface is closely associated with SWD but can also be used independently. SWV data is output on the JTAG interface's TDO pin. If using SWV, you must configure the device for SWD, not JTAG. SWV is not supported with the JTAG interface.

SWV is ideal for application debug where it is helpful for the firmware to output data similar to 'printf' debugging on PCs. The SWV is ideal for data monitoring, because it requires only a single pin and can output data in standard UART format or Manchester encoded format. For example, it can be used to tune a PID control loop in which the output and graphing of the three error terms greatly simplifies coefficient tuning.

The following features are supported in SWV:

- 32 virtual channels, each 32 bits long
- Simple, efficient packing and serializing protocol
- Supports standard UART format (N81)

9.6 Programming Features

The JTAG and SWD interfaces provide full programming support. The entire device can be erased, programmed, and verified. You can increase flash protection levels to protect firmware IP. Flash protection can only be reset after a full device

erase. Individual flash blocks can be erased, programmed, and verified, if block security settings permit.

9.7 Device Security

PSoC 3 offers an advanced security feature called device security, which permanently disables all test, programming, and debug ports, protecting your application from external access. The device security is activated by programming a 32-bit key (0x50536F43) to a Write Once Latch (WOL).

The Write Once Latch is a type of nonvolatile latch (NVL). The cell itself is an NVL with additional logic wrapped around it. Each WOL device contains four bytes (32 bits) of data. The wrapper outputs a '1' if a super-majority (28 of 32) of its bits match a pre-determined pattern (0x50536F43); it outputs a '0' if this majority is not reached. When the output is 1, the Write Once NV latch locks the part out of Debug and Test modes; it also permanently gates off the ability to erase or alter the contents of the latch. Matching all bits is intentionally not required, so that single (or few) bit failures do not deassert the WOL output. The state of the NVL bits after wafer processing is truly random with no tendency toward 1 or 0.

The WOL only locks the part after the correct 32-bit key (0x50536F43) is loaded into the NVL's volatile memory, programmed into the NVL's nonvolatile cells, and the part is reset. The output of the WOL is only sampled on reset and used to disable the access. This precaution prevents anyone from reading, erasing, or altering the contents of the internal memory.

The user can write the key into the WOL to lock out external access only if no flash protection is set (see "Flash Security" on page 23). However, after setting the values in the WOL, a user still has access to the part until it is reset. Therefore, a user can write the key into the WOL, program the flash protection data, and then reset the part to lock it.

If the device is protected with a WOL setting, Cypress cannot perform failure analysis and, therefore, cannot accept RMAs from customers. The WOL can be read out through the SWD port to electrically identify protected parts. The user can write the key in WOL to lock out external access only if no flash protection is set. For more information on how to take full advantage of the security features in PSoC see the PSoC 3 TRM.

Disclaimer

Note the following details of the flash code protection features on Cypress devices.

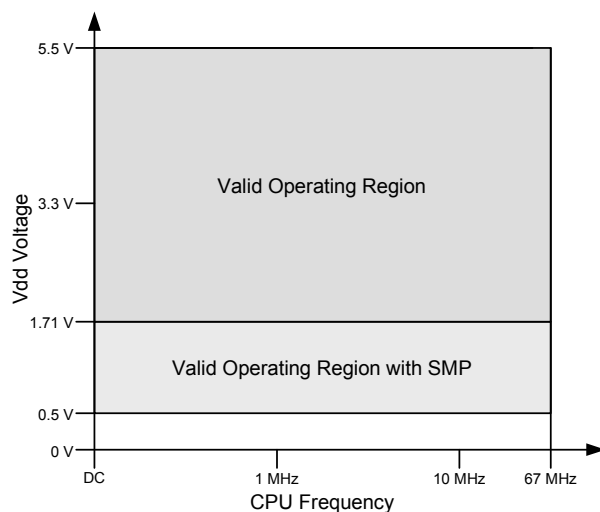
Cypress products meet the specifications contained in their particular Cypress data sheets. Cypress believes that its family of products is one of the most secure families of its kind on the market today, regardless of how they are used. There may be methods, unknown to Cypress, that can breach the code protection features. Any of these methods, to our knowledge, would be dishonest and possibly illegal. Neither Cypress nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as "unbreakable."

Cypress is willing to work with the customer who is concerned about the integrity of their code. Code protection is constantly evolving. We at Cypress are committed to continuously improving the code protection features of our products.

Table 11-3. AC Specifications^[33]

Parameter	Description	Conditions	Min	Typ	Max	Units
F _{CPU}	CPU frequency	$1.71\text{ V} \leq V_{DD} \leq 5.5\text{ V}$	DC	–	67.01	MHz
F _{BUSCLK}	Bus frequency	$1.71\text{ V} \leq V_{DD} \leq 5.5\text{ V}$	DC	–	67.01	MHz
Svdd	V _{DD} ramp rate		–	–	0.066	V/μs
T _{IO_INIT}	Time from V _{DDD} /V _{DDA} /V _{CCD} /V _{CCA} ≥ IPOR to I/O ports set to their reset states		–	–	10	μs
T _{STARTUP}	Time from V _{DDD} /V _{DDA} /V _{CCD} /V _{CCA} ≥ PRES to CPU executing code at reset vector	V _{CCA} /V _{DDA} = regulated from V _{DDA} /V _{DDD} , no PLL used, fast IMO boot mode (48 MHz typ.)	–	–	40	μs
		V _{CCA} /V _{CCD} = regulated from V _{DDA} /V _{DDD} , no PLL used, slow IMO boot mode (12 MHz typ.)	–	–	74	μs
T _{SLEEP}	Wakeup from sleep mode – Application of non-LVD interrupt to beginning of execution of next CPU instruction		–	–	15	μs
T _{HIBERNATE}	Wakeup from hibernate mode – Application of external interrupt to beginning of execution of next CPU instruction		–	–	100	μs

Figure 11-4. F_{CPU} vs. V_{DD}



Note

33. Based on device characterization (Not production tested).

11.3.3 Inductive Boost Regulator

Unless otherwise specified, operating conditions are: $V_{BAT} = 0.5\text{ V} - 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V} - 5.0\text{ V}$, $I_{OUT} = 0\text{ mA} - 50\text{ mA}$, $L_{BOOST} = 4.7\text{ }\mu\text{H} - 22\text{ }\mu\text{H}$, $C_{BOOST} = 22\text{ }\mu\text{F} \parallel 3 \times 1.0\text{ }\mu\text{F} \parallel 3 \times 0.1\text{ }\mu\text{F}$, $C_{BAT} = 22\text{ }\mu\text{F}$, $I_F = 1.0\text{ A}$. Unless otherwise specified, all charts and graphs show typical values.

Table 11-6. Inductive Boost Regulator DC Specifications

Parameter	Description	Conditions		Min	Typ	Max	Units
V _{OUT}	Boost output voltage ^[34]	vsel = 1.8 V in register BOOST_CR0		1.71	1.8	1.89	V
		vsel = 1.9 V in register BOOST_CR0		1.81	1.90	2.00	V
		vsel = 2.0 V in register BOOST_CR0		1.90	2.00	2.10	V
		vsel = 2.4 V in register BOOST_CR0		2.16	2.40	2.64	V
		vsel = 2.7 V in register BOOST_CR0		2.43	2.70	2.97	V
		vsel = 3.0 V in register BOOST_CR0		2.70	3.00	3.30	V
		vsel = 3.3 V in register BOOST_CR0		2.97	3.30	3.63	V
		vsel = 3.6 V in register BOOST_CR0		3.24	3.60	3.96	V
		vsel = 5.0 V in register BOOST_CR0		4.50	5.00	5.50	V
V _{BAT}	Input voltage to boost ^[35]	I _{OUT} = 0 mA–5 mA	vsel = 1.8 V–2.0 V, T _A = 0 °C–70 °C	0.5	–	0.8	V
		I _{OUT} = 0 mA–15 mA	vsel = 1.8 V–5.0 V ^[36] , T _A = –10 °C–85 °C	1.6	–	3.6	V
		I _{OUT} = 0 mA–25 mA	vsel = 1.8 V–2.7 V, T _A = –10 °C–85 °C	0.8	–	1.6	V
		I _{OUT} = 0 mA–50 mA	vsel = 1.8 V–3.3 V ^[36] , T _A = –40 °C–85 °C	1.8	–	2.5	V
			vsel = 1.8 V–3.3 V ^[36] , T _A = –10 °C–85 °C	1.3	–	2.5	V
			vsel = 2.5 V–5.0 V ^[36] , T _A = –10 °C–85 °C	2.5	–	3.6	V
		I _{OUT}	Output current	T _A = 0 °C–70 °C	V _{BAT} = 0.5 V–0.8 V	0	–
T _A = –10 °C–85 °C	V _{BAT} = 1.6 V–3.6 V			0	–	15	mA
	V _{BAT} = 0.8 V–1.6 V			0	–	25	mA
	V _{BAT} = 1.3 V–2.5 V			0	–	50	mA
	V _{BAT} = 2.5 V–3.6 V			0	–	50	mA
T _A = –40 °C–85 °C	V _{BAT} = 1.8 V–2.5 V			0	–	50	mA
I _{LPK}	Inductor peak current					–	–
I _Q	Quiescent current	Boost active mode		–	250	–	μA
		Boost sleep mode, I _{OUT} < 1 μA		–	25	–	μA
Reg _{LOAD}	Load regulation			–	–	10	%
Reg _{LINE}	Line regulation			–	–	10	%

Notes

34. Listed v_{sel} options are characterized. Additional v_{sel} options are valid and guaranteed by design.

35. The boost will start at all valid V_{BAT} conditions including down to $V_{BAT} = 0.5\text{ V}$.

36. If V_{BAT} is greater than or equal to V_{OUT} boost setting, then V_{OUT} will be less than V_{BAT} due to resistive losses in the boost circuit.

Figure 11-17. SIO Output High Voltage and Current, Unregulated Mode

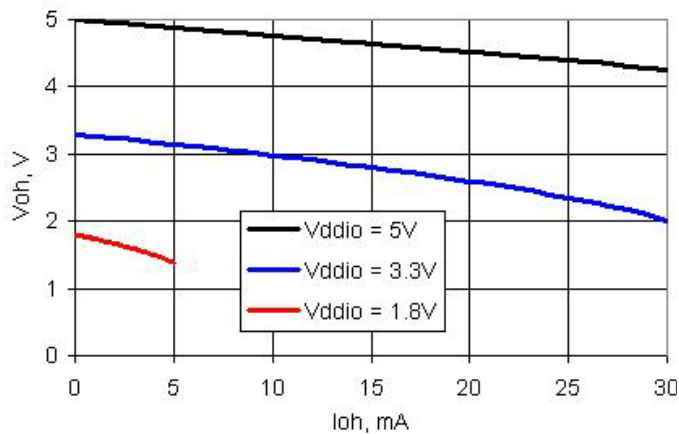


Figure 11-18. SIO Output Low Voltage and Current, Unregulated Mode

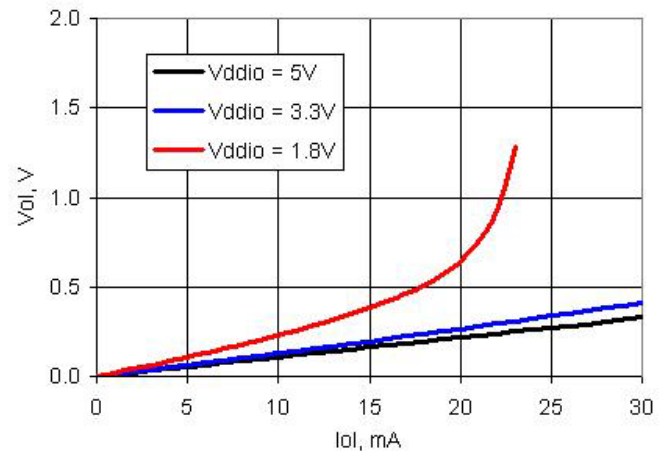


Figure 11-19. SIO Output High Voltage and Current, Regulated Mode

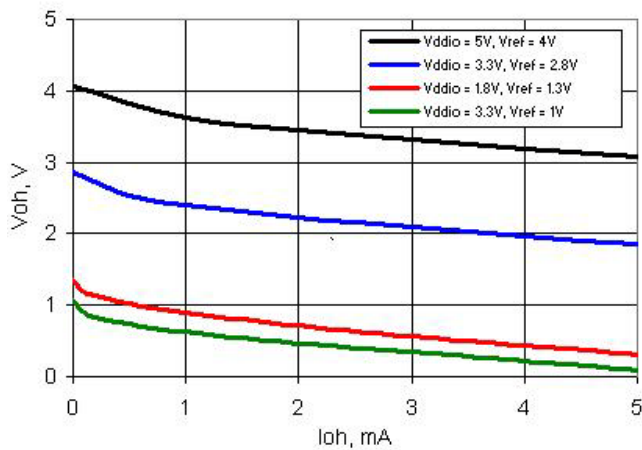


Table 11-12. SIO AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
TriseF	Rise time in fast strong mode (90/10%) ^[44]	Cload = 25 pF, VDDIO = 3.3 V	–	–	12	ns
TfallF	Fall time in fast strong mode (90/10%) ^[44]	Cload = 25 pF, VDDIO = 3.3 V	–	–	12	ns
TriseS	Rise time in slow strong mode (90/10%) ^[44]	Cload = 25 pF, VDDIO = 3.0 V	–	–	75	ns
TfallS	Fall time in slow strong mode (90/10%) ^[44]	Cload = 25 pF, VDDIO = 3.0 V	–	–	60	ns

Note

44. Based on device characterization (Not production tested).

11.4.4 XRES

Table 11-17. XRES DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
V _{IH}	Input voltage high threshold		$0.7 \times V_{DDIO}$	–	–	V
V _{IL}	Input voltage low threshold		–	–	$0.3 \times V_{DDIO}$	V
R _{pullup}	Pull-up resistor		3.5	5.6	8.5	kΩ
C _{IN}	Input capacitance ^[46]		–	3	–	pF
V _H	Input voltage hysteresis (Schmitt-Trigger) ^[46]		–	100	–	mV
I _{diode}	Current through protection diode to V _{DDIO} and V _{SSIO}		–	–	100	μA

Table 11-18. XRES AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
T _{RESET}	Reset pulse width		1	–	–	μs

11.5 Analog Peripherals

Specifications are valid for $-40\text{ °C} \leq T_A \leq 85\text{ °C}$ and $T_J \leq 100\text{ °C}$, except where noted. Specifications are valid for 1.71 V to 5.5 V, except where noted.

11.5.1 Opamp

Table 11-19. Opamp DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
V _I	Input voltage range		V _{SSA}	–	V _{DDA}	V
V _{OS}	Input offset voltage		–	–	2.5	mV
		Operating temperature -40 °C to 70 °C	–	–	2	mV
TCV _{OS}	Input offset voltage drift with temperature	Power mode = high	–	–	±30	μV/°C
Ge1	Gain error, unity gain buffer mode	Rload = 1 kΩ	–	–	±0.1	%
C _{IN}	Input capacitance	Routing from pin	–	–	18	pF
V _O	Output voltage range	1 mA, source or sink, power mode = high	V _{SSA} + 0.05	–	V _{DDA} – 0.05	V
I _{OUT}	Output current capability, source or sink	V _{SSA} + 500 mV ≤ V _{out} ≤ V _{DDA} –500 mV, V _{DDA} > 2.7 V	25	–	–	mA
		V _{SSA} + 500 mV ≤ V _{out} ≤ V _{DDA} –500 mV, 1.7 V = V _{DDA} ≤ 2.7 V	16	–	–	mA
I _{DD}	Quiescent current	Power mode = min	–	250	400	μA
		Power mode = low	–	250	400	μA
		Power mode = med	–	330	950	μA
		Power mode = high	–	1000	2500	μA
CMRR	Common mode rejection ratio		80	–	–	dB
PSRR	Power supply rejection ratio	V _{DDA} ≥ 2.7 V	85	–	–	dB
		V _{DDA} < 2.7 V	70	–	–	dB
I _{IB}	Input bias current ^[46]	25 °C	–	10	–	pA

Note

46. Based on device characterization (Not production tested).

Table 11-20. Opamp AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
GBW	Gain-bandwidth product	Power mode = minimum, 15 pF load	1	–	–	MHz
		Power mode = low, 15 pF load	2	–	–	MHz
		Power mode = medium, 200 pF load	1	–	–	MHz
		Power mode = high, 200 pF load	3	–	–	MHz
SR	Slew rate, 20% - 80%	Power mode = low, 15 pF load	1.1	–	–	V/μs
		Power mode = medium, 200 pF load	0.9	–	–	V/μs
		Power mode = high, 200 pF load	3	–	–	V/μs
e_n	Input noise density	Power mode = high, $V_{DDA} = 5$ V, at 100 kHz	–	45	–	nV/sqrtHz

Figure 11-30. Opamp Noise vs Frequency, Power Mode = High, $V_{DDA} = 5$ V

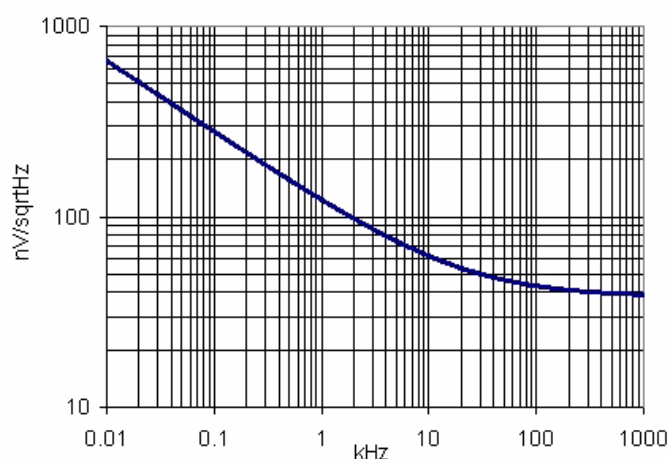


Figure 11-31. Opamp Step Response, Rising

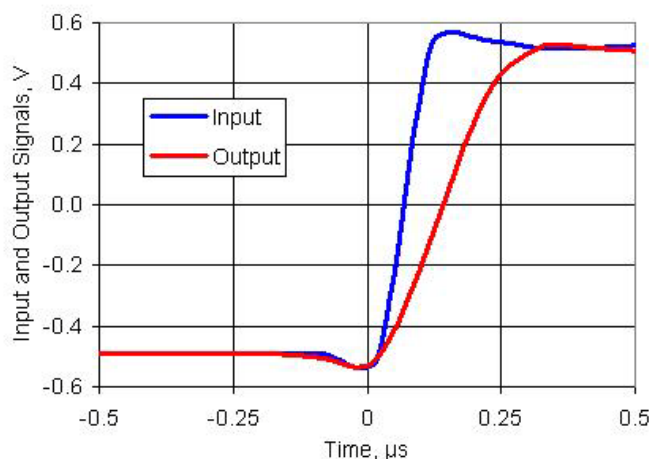


Figure 11-32. Opamp Step Response, Falling

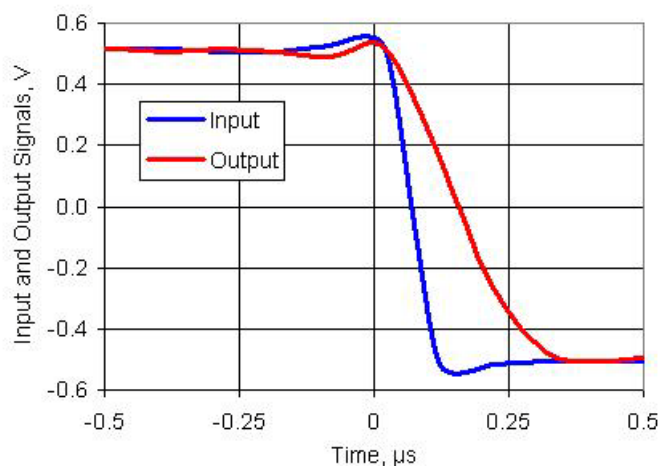


Figure 11-72. Synchronous Write and Read Cycle Timing, No Wait States

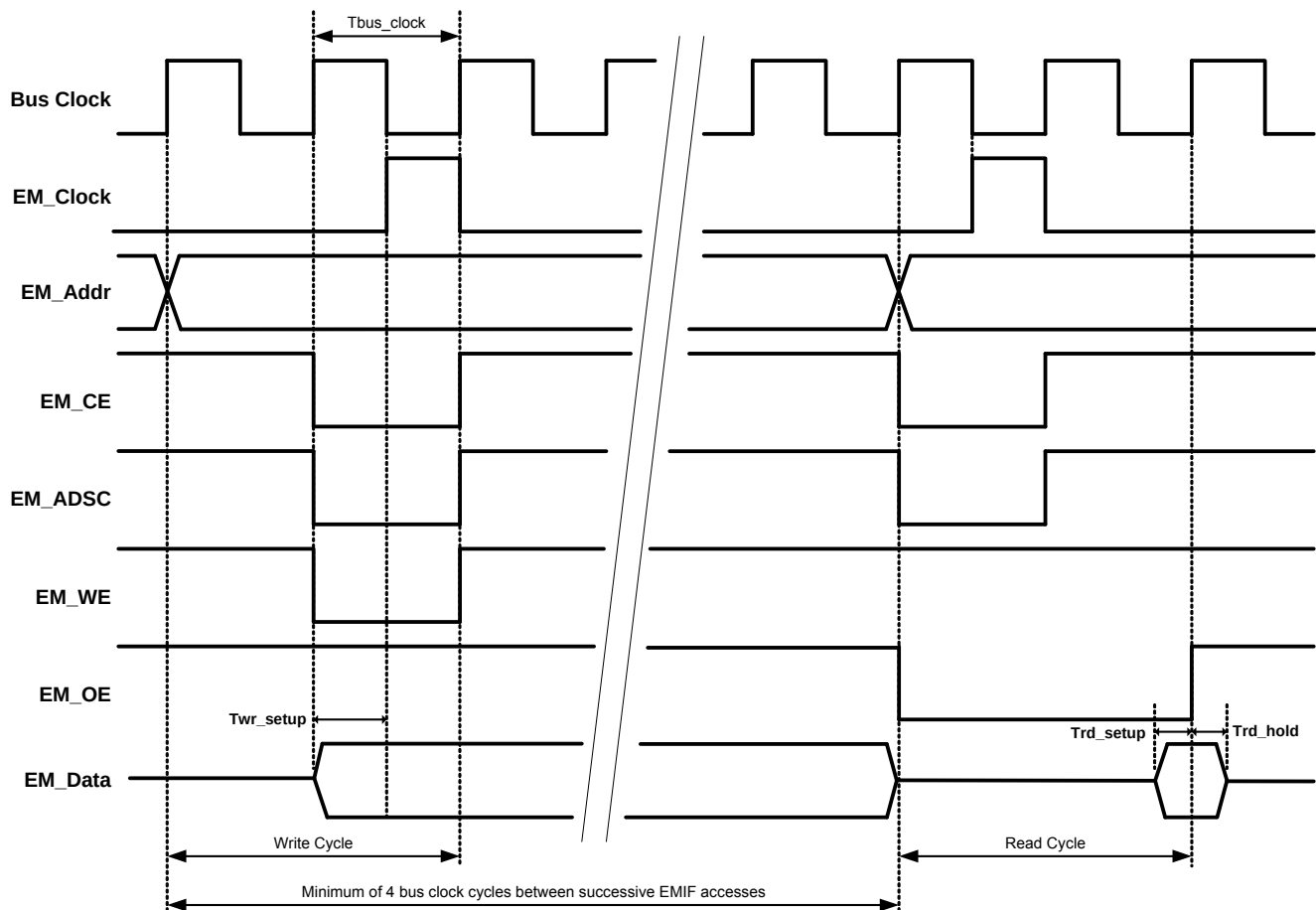


Table 11-68. Synchronous Write and Read Timing Specifications^[65]

Parameter	Description	Conditions	Min	Typ	Max	Units
Fbus_clock	Bus clock frequency ^[66]		–	–	33	MHz
Tbus_clock	Bus clock period ^[67]		30.3	–	–	ns
Twr_Setup	Time from EM_data valid to rising edge of EM_Clock		$T_{bus_clock} - 10$	–	–	ns
Trd_setup	Time that EM_data must be valid before rising edge of EM_OE		5	–	–	ns
Trd_hold	Time that EM_data must be valid after rising edge of EM_OE		5	–	–	ns

Notes

65. Based on device characterization (Not production tested).

66. EMIF signal timings are limited by GPIO frequency limitations. See “GPIO” section on page 80.

67. EMIF output signals are generally synchronized to bus clock, so EMIF signal timings are dependent on bus clock frequency.

11.8.3 Interrupt Controller

Table 11-73. Interrupt Controller AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
	Delay from interrupt signal input to ISR code execution from ISR code	Includes worse case completion of longest instruction DIV with 6 cycles	–	–	25	Tcy CPU

11.8.4 JTAG Interface

Figure 11-73. JTAG Interface Timing

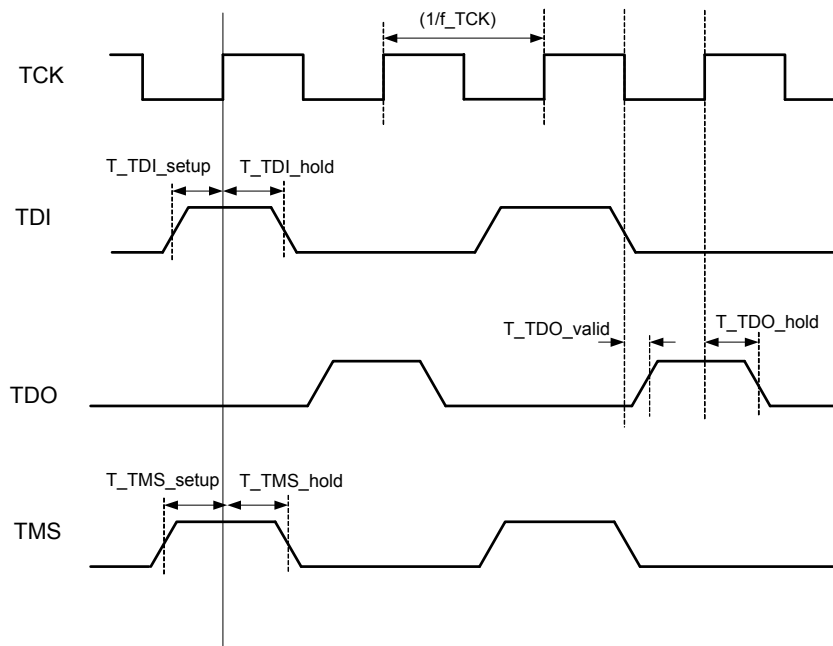


Table 11-74. JTAG Interface AC Specifications^[69]

Parameter	Description	Conditions	Min	Typ	Max	Units
f_TCK	TCK frequency	$3.3\text{ V} \leq V_{DD} \leq 5\text{ V}$	–	–	14 ^[70]	MHz
		$1.71\text{ V} \leq V_{DD} < 3.3\text{ V}$	–	–	7 ^[70]	MHz
T_TDI_setup	TDI setup before TCK high		$(T/10) - 5$	–	–	ns
T_TMS_setup	TMS setup before TCK high		T/4	–	–	
T_TDI_hold	TDI, TMS hold after TCK high	$T = 1/f_TCK\text{ max}$	T/4	–	–	
T_TDO_valid	TCK low to TDO valid	$T = 1/f_TCK\text{ max}$	–	–	2T/5	
T_TDO_hold	TDO hold after TCK high	$T = 1/f_TCK\text{ max}$	T/4	–	–	

Notes

69. Based on device characterization (Not production tested).
70. f_TCK must also be no more than 1/3 CPU clock frequency.

12. Ordering Information

In addition to the features listed in [Table 12-1](#), every CY8C38 device includes: a precision on-chip voltage reference, precision oscillators, flash, ECC, DMA, a fixed function I²C, 4 KB trace RAM, JTAG/SWD programming and debug, external memory interface, and more. In addition to these features, the flexible UDBs and analog subsection support a wide range of peripherals. To assist you in selecting the ideal part, PSoC Creator makes a part recommendation after you choose the components required by your application. All CY8C38 derivatives incorporate device and flash security in user-selectable security levels; see the TRM for details.

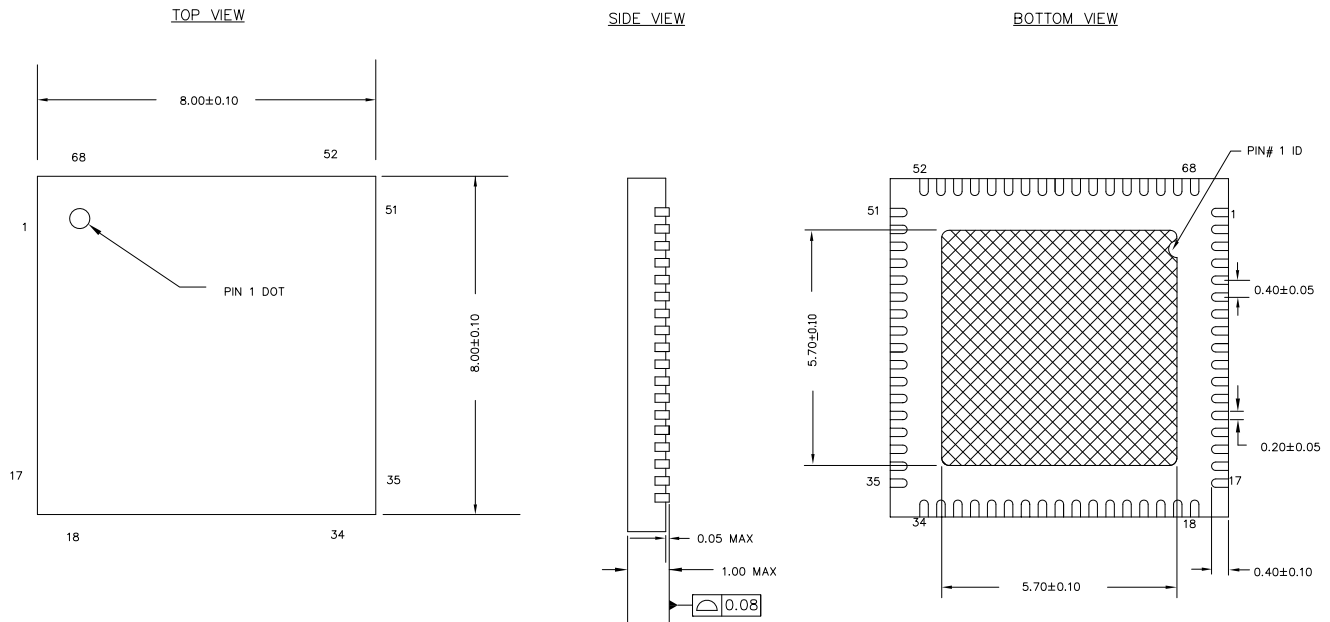
Table 12-1. CY8C38 Family with Single Cycle 8051

Part Number	MCU Core				Analog								Digital			I/O ^[83]				Package	JTAG ID ^[84]	
	CPU Speed (MHz)	Flash (KB)	SRAM (KB)	EEPROM (KB)	LCD Segment Drive	ADC	DAC	Comparator	SC/CT Analog Blocks ^[81]	Opamps	DFB	CapSense	UDBs ^[82]	16-bit Timer/PWM	FS USB	CAN 2.0b	Total I/O	GPIO	SIO			USBIO
32 KB Flash																						
CY8C3865AXI-019	67	32	4	1	✓	20-bit Del-Sig	4	4	4	4	✓	✓	20	4	✓	–	72	62	8	2	100-pin TQFP	0x1E013069
CY8C3865LTI-014	67	32	4	1	✓	20-bit Del-Sig	4	4	4	4	✓	✓	20	4	✓	–	48	38	8	2	68-pin QFN	0x1E00E069
CY8C3865AXI-204	67	32	8	1	✓	20-bit Del-Sig	2	0	0	0	–	✓	16	4	✓	–	72	62	8	2	100-pin TQFP	0x1E0CC069
CY8C3865LTI-205	67	32	8	1	✓	20-bit Del-Sig	2	0	0	0	–	✓	16	4	✓	–	48	38	8	2	68-pin QFN	0x1E0CD069
64 KB Flash																						
CY8C3866LTI-067	67	64	8	2	✓	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	✓	–	31	25	4	2	48-pin QFN	0x1E043069
CY8C3866PVI-021	67	64	8	2	✓	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	✓	–	31	25	4	2	48-pin SSOP	0x1E015069
CY8C3866AXI-035	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	–	✓	70	62	8	0	100-pin TQFP	0x1E023069
CY8C3866AXI-039	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	✓	–	72	62	8	2	100-pin TQFP	0x1E027069
CY8C3866LTI-030	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	✓	–	48	38	8	2	68-pin QFN	0x1E01E069
CY8C3866LTI-068	67	64	8	2	✓	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	✓	✓	31	25	4	2	48-pin QFN	0x1E044069
CY8C3866AXI-040	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	✓	✓	72	62	8	2	100-pin TQFP	0x1E028069
CY8C3866PVI-070	67	64	8	2	✓	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	–	✓	29	25	4	0	48-pin SSOP	0x1E046069
CY8C3866AXI-206	67	64	8	2	✓	20-bit Del-Sig	2	2	0	2	✓	✓	20	4	✓	–	72	62	8	2	100-pin TQFP	0x1E0CE069
CY8C3866LTI-207	67	64	8	2	✓	20-bit Del-Sig	2	2	0	2	✓	✓	20	4	✓	–	48	38	8	2	68-pin QFN	0x1E0CF069
CY8C3866AXI-208	67	64	8	2	✓	20-bit Del-Sig	2	2	2	2	✓	✓	24	4	✓	✓	72	62	8	2	100-pin TQFP	0x1E0D0069
CY8C3866LTI-209	67	64	8	2	✓	20-bit Del-Sig	2	2	2	2	✓	✓	24	4	✓	✓	48	38	8	2	68-pin QFN	0x1E0D1069
CY8C3866FNI-210	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	✓	✓	48	38	8	2	72 WLCSP	0x1E0D2069

Notes

81. Analog blocks support a variety of functionality including TIA, PGA, and mixers. See the [Example Peripherals](#) on page 44 for more information on how analog blocks can be used.
82. UDBs support a variety of functionality including SPI, LIN, UART, timer, counter, PWM, PRS, and others. Individual functions may use a fraction of a UDB or multiple UDBs. Multiple functions can share a single UDB. See the [Example Peripherals](#) on page 44 for more information on how UDBs can be used.
83. The I/O Count includes all types of digital I/O: GPIO, SIO, and the two USB I/O. See the [I/O System and Routing](#) on page 37 for details on the functionality of each of these types of I/O.
84. The JTAG ID has three major fields. The most significant nibble (left digit) is the version, followed by a 2 byte part number and a 3 nibble manufacturer ID.

Figure 13-3. 68-pin QFN 8×8 with 0.4 mm Pitch Package Outline (Sawn Version)

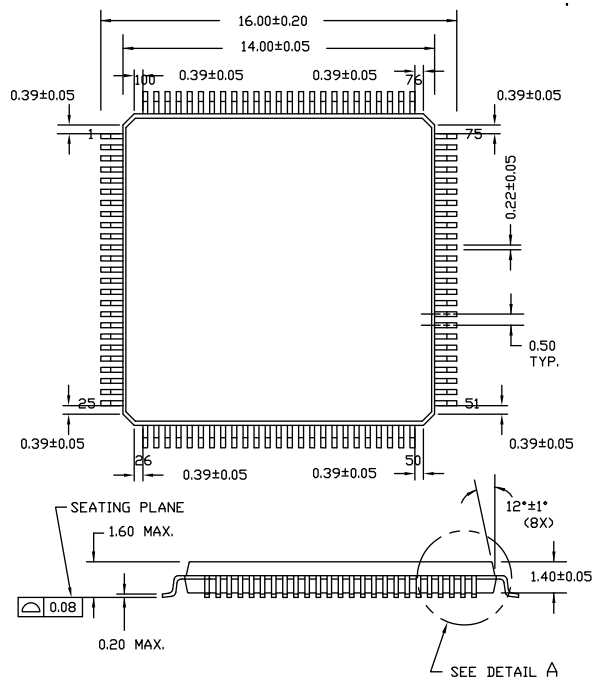


NOTES:

1. HATCH AREA IS SOLDERABLE EXPOSED METAL.
2. REFERENCE JEDEC#: MO-220
3. PACKAGE WEIGHT: 17 ± 2mg
4. ALL DIMENSIONS ARE IN MILLIMETERS

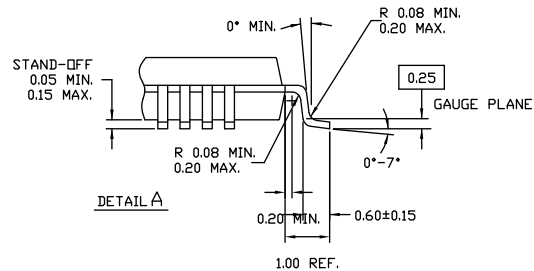
001-09618 *E

Figure 13-4. 100-pin TQFP (14 × 14 × 1.4 mm) Package Outline

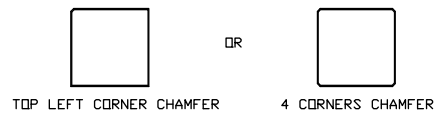


NOTE:

1. JEDEC STD REF MS-026
2. BODY LENGTH DIMENSION DOES NOT INCLUDE MOLD PROTRUSION/END FLASH
MOLD PROTRUSION/END FLASH SHALL NOT EXCEED 0.0098 in (0.25 mm) PER SIDE
BODY LENGTH DIMENSIONS ARE MAX PLASTIC BODY SIZE INCLUDING MOLD MISMATCH
3. DIMENSIONS IN MILLIMETERS



NOTE: PKG. CAN HAVE



51-85048 *J

Table 14-1. Acronyms Used in this Document *(continued)*

Acronym	Description
PHUB	peripheral hub
PHY	physical layer
PICU	port interrupt control unit
PLA	programmable logic array
PLD	programmable logic device, see also PAL
PLL	phase-locked loop
PMDD	package material declaration data sheet
POR	power-on reset
PRES	precise low-voltage reset
PRS	pseudo random sequence
PS	port read data register
PSoC®	Programmable System-on-Chip™
PSRR	power supply rejection ratio
PWM	pulse-width modulator
RAM	random-access memory
RISC	reduced-instruction-set computing
RMS	root-mean-square
RTC	real-time clock
RTL	register transfer language
RTR	remote transmission request
RX	receive
SAR	successive approximation register
SC/CT	switched capacitor/continuous time
SCL	I ² C serial clock
SDA	I ² C serial data
S/H	sample and hold
SINAD	signal to noise and distortion ratio
SIO	special input/output, GPIO with advanced features. See GPIO.
SOC	start of conversion

Table 14-1. Acronyms Used in this Document *(continued)*

Acronym	Description
SOF	start of frame
SPI	Serial Peripheral Interface, a communications protocol
SR	slew rate
SRAM	static random access memory
SRES	software reset
SWD	serial wire debug, a test protocol
SWV	single-wire viewer
TD	transaction descriptor, see also DMA
THD	total harmonic distortion
TIA	transimpedance amplifier
TRM	technical reference manual
TTL	transistor-transistor logic
TX	transmit
UART	Universal Asynchronous Transmitter Receiver, a communications protocol
UDB	universal digital block
USB	Universal Serial Bus
USBIO	USB input/output, PSoC pins used to connect to a USB port
VDAC	voltage DAC, see also DAC, IDAC
WDT	watchdog timer
WOL	write once latch, see also NVL
WRES	watchdog timer reset
XRES	external reset I/O pin
XTAL	crystal

15. Reference Documents

[PSoC® 3, PSoC® 5 Architecture TRM](#)

[PSoC® 3 Registers TRM](#)

Description Title: PSoC® 3: CY8C38 Family Datasheet Programmable System-on-Chip (PSoC®) (continued)
Document Number: 001-11729

Revision	ECN	Submission Date	Orig. of Change	Description of Change
*S	3259185	05/17/2011	MKEA	Added JTAG and SWD interface connection diagrams Updated T_{JA} and T_{JC} values in Table 13-1 Changed typ and max values for the TCV_{OS} parameter in Opamp DC specifications table. Updated Clocking subsystem diagram. Changed V_{SSD} to V_{SSB} in the PSoC Power System diagram Updated Ordering information.
*T	3464258	12/14/2011	MKEA	Updated Analog Global specs Updated IDAC range Updated TIA section Modified VDDIO description in Section 3 Added note on Sleep and Hibernation modes in the Power Modes section Updated Boost Converter section Updated conditions for Inductive boost AC specs Added VDAC/IDAC noise graphs and specs Added pin capacitance specs for ECO pins Removed C_L from 32 kHz External Crystal DC Specs table. Added reference to AN54439 in Section 6.1.2.2 Deleted T_{SWDO_hold} row from the SWD Interface AC Specifications table Removed Pin 46 connections in "Example Schematic for 100-pin TQFP Part with Power Connections" Updated Active Mode I_{DD} description in Table 11-2. Added I_{DDDR} and I_{DDAR} specs in Table 11-2. Replaced "total device program time" with T_{PROG} in Flash AC specs table Added I_{GPIO}, I_{SIO} and I_{USBIO} specs in Absolute Maximum Ratings Added conditions to I_{CC} spec in 32 kHz External Crystal DC Specs table. Updated TCV_{OS} value Removed Boost Efficiency vs V_{OUT} graph Updated boost graphs Updated min value of GPIO input edge rate Removed 3.4 Mbps in UDBs from I2C section Updated USBIO Block diagram; added USBIO drive mode description Updated Analog Interconnect diagram Changed max IMO startup time to 12 μs Added note for I_{IL} spec in USBIO DC specs table Updated GPIO Block diagram Updated voltage reference specs Added text explaining power supply ramp up in Section 11-4.

18. Sales, Solutions, and Legal Information

Worldwide Sales and Design Support

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