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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Active
Core Processor	8051
Core Size	8-Bit
Speed	67MHz
Connectivity	EBI/EMI, I ² C, LINbus, SPI, UART/USART, USB
Peripherals	CapSense, DMA, LCD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	64KB (64K x 8)
Program Memory Type	FLASH
EEPROM Size	2K x 8
RAM Size	8K x 8
Voltage - Supply (Vcc/Vdd)	1.71V ~ 5.5V
Data Converters	A/D 16x20b; D/A 4x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-BSSOP (0.295", 7.50mm Width)
Supplier Device Package	48-SSOP
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/cy8c3866pvi-021

Table 4-4. Boolean Instructions *(continued)*

Mnemonic	Description	Bytes	Cycles
ANL C, /bit	AND complement of direct bit to carry	2	2
ORL C, bit	OR direct bit to carry	2	2
ORL C, /bit	OR complement of direct bit to carry	2	2
MOV C, bit	Move direct bit to carry	2	2
MOV bit, C	Move carry to direct bit	2	3
JC rel	Jump if carry is set	2	3
JNC rel	Jump if no carry is set	2	3
JB bit, rel	Jump if direct bit is set	3	5
JNB bit, rel	Jump if direct bit is not set	3	5
JBC bit, rel	Jump if direct bit is set and clear bit	3	5

4.3.1.5 Program Branching Instructions

The 8051 supports a set of conditional and unconditional jump instructions that help to modify the program execution flow. [Table 4-5](#) shows the list of jump instructions.

Table 4-5. Jump Instructions

Mnemonic	Description	Bytes	Cycles
ACALL addr11	Absolute subroutine call	2	4
LCALL addr16	Long subroutine call	3	4
RET	Return from subroutine	1	4
RETI	Return from interrupt	1	4
AJMP addr11	Absolute jump	2	3
LJMP addr16	Long jump	3	4
SJMP rel	Short jump (relative address)	2	3
JMP @A + DPTR	Jump indirect relative to DPTR	1	5
JZ rel	Jump if accumulator is zero	2	4
JNZ rel	Jump if accumulator is nonzero	2	4
CJNE A, Direct, rel	Compare direct byte to accumulator and jump if not equal	3	5
CJNE A, #data, rel	Compare immediate data to accumulator and jump if not equal	3	4
CJNE Rn, #data, rel	Compare immediate data to register and jump if not equal	3	4
CJNE @Ri, #data, rel	Compare immediate data to indirect RAM and jump if not equal	3	5
DJNZ Rn, rel	Decrement register and jump if not zero	2	4
DJNZ Direct, rel	Decrement direct byte and jump if not zero	3	5
NOP	No operation	1	1

4.4 DMA and PHUB

The PHUB and the DMA controller are responsible for data transfer between the CPU and peripherals, and also data transfers between peripherals. The PHUB and DMA also control device configuration during boot. The PHUB consists of:

- A central hub that includes the DMA controller, arbiter, and router
- Multiple spokes that radiate outward from the hub to most peripherals

There are two PHUB masters: the CPU and the DMA controller. Both masters may initiate transactions on the bus. The DMA channels can handle peripheral communication without CPU intervention. The arbiter in the central hub determines which DMA channel is the highest priority if there are multiple requests.

4.4.1 PHUB Features

- CPU and DMA controller are both bus masters to the PHUB
- Eight multi-layer AHB bus parallel access paths (spokes) for peripheral access
- Simultaneous CPU and DMA access to peripherals located on different spokes
- Simultaneous DMA source and destination burst transactions on different spokes
- Supports 8-, 16-, 24-, and 32-bit addressing and data

Table 4-6. PHUB Spokes and Peripherals

PHUB Spokes	Peripherals
0	SRAM
1	IOs, PICU, EMIF
2	PHUB local configuration, Power manager, Clocks, IC, SWV, EEPROM, Flash programming interface
3	Analog interface and trim, Decimator
4	USB, CAN, I ² C, Timers, Counters, and PWMs
5	DFB
6	UDBs group 1
7	UDBs group 2

4.4.2 DMA Features

- 24 DMA channels
- Each channel has one or more transaction descriptors (TD) to configure channel behavior. Up to 128 total TDs can be defined
- TDs can be dynamically updated
- Eight levels of priority per channel
- Any digitally routable signal, the CPU, or another DMA channel, can trigger a transaction
- Each channel can generate up to two interrupts per transfer

- Transactions can be stalled or canceled
- Supports transaction size of infinite or 1 to 64 KB
- TDs may be nested and/or chained for complex transactions

4.4.3 Priority Levels

The CPU always has higher priority than the DMA controller when their accesses require the same bus resources. Due to the system architecture, the CPU can never starve the DMA. DMA channels of higher priority (lower priority number) may interrupt current DMA transfers. In the case of an interrupt, the current transfer is allowed to complete its current transaction. To ensure latency limits when multiple DMA accesses are requested simultaneously, a fairness algorithm guarantees an interleaved minimum percentage of bus bandwidth for priority levels 2 through 7. Priority levels 0 and 1 do not take part in the fairness algorithm and may use 100 percent of the bus bandwidth. If a tie occurs on two DMA requests of the same priority level, a simple round robin method is used to evenly share the allocated bandwidth. The round robin allocation can be disabled for each DMA channel, allowing it to always be at the head of the line. Priority levels 2 to 7 are guaranteed the minimum bus bandwidth shown in Table 4-7 after the CPU and DMA priority levels 0 and 1 have satisfied their requirements.

Table 4-7. Priority Levels

Priority Level	% Bus Bandwidth
0	100.0
1	100.0
2	50.0
3	25.0
4	12.5
5	6.2
6	3.1
7	1.5

When the fairness algorithm is disabled, DMA access is granted based solely on the priority level; no bus bandwidth guarantees are made.

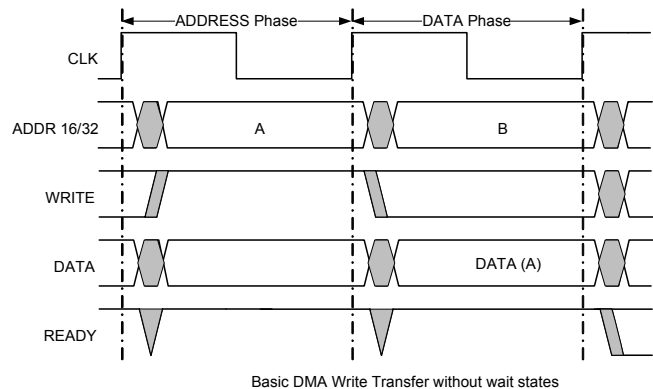
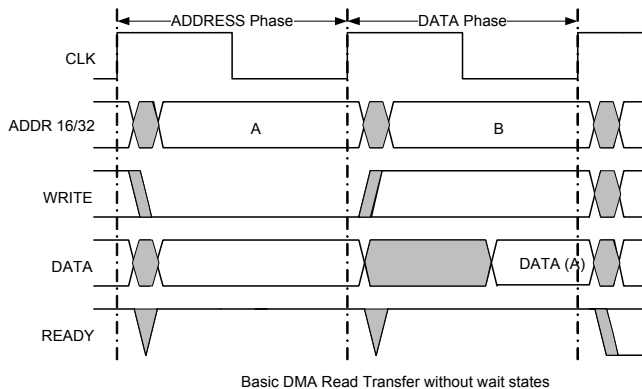
4.4.4 Transaction Modes Supported

The flexible configuration of each DMA channel and the ability to chain multiple channels allow the creation of both simple and complex use cases. General use cases include, but are not limited to:

4.4.4.1 Simple DMA

In a simple DMA case, a single TD transfers data between a source and sink (peripherals or memory location). The basic timing diagrams of DMA read and write cycles are shown in Figure 4-1. For more description on other transfer modes, refer to the Technical Reference Manual.

Figure 4-1. DMA Timing Diagram



4.4.4.2 Auto Repeat DMA

Auto repeat DMA is typically used when a static pattern is repetitively read from system memory and written to a peripheral. This is done with a single TD that chains to itself.

4.4.4.3 Ping Pong DMA

A ping pong DMA case uses double buffering to allow one buffer to be filled by one client while another client is consuming the data previously received in the other buffer. In its simplest form, this is done by chaining two TDs together so that each TD calls the opposite TD when complete.

4.4.4.4 Circular DMA

Circular DMA is similar to ping pong DMA except it contains more than two buffers. In this case there are multiple TDs; after the last TD is complete it chains back to the first TD.

4.4.4.5 Scatter Gather DMA

In the case of scatter gather DMA, there are multiple noncontiguous sources or destinations that are required to effectively carry out an overall DMA transaction. For example, a packet may need to be transmitted off of the device and the packet elements, including the header, payload, and trailer, exist in various noncontiguous locations in memory. Scatter gather DMA allows the segments to be concatenated together by using multiple TDs in a chain. The chain gathers the data from the multiple locations. A similar concept applies for the reception of data onto the device. Certain parts of the received data may need to be scattered to various locations in memory for software processing convenience. Each TD in the chain specifies the location for each discrete element in the chain.

4.4.4.6 Packet Queuing DMA

Packet queuing DMA is similar to scatter gather DMA but specifically refers to packet protocols. With these protocols, there may be separate configuration, data, and status phases associated with sending or receiving a packet.

For instance, to transmit a packet, a memory mapped configuration register can be written inside a peripheral, specifying the overall length of the ensuing data phase. The CPU can set up this configuration information anywhere in system memory and copy it with a simple TD to the peripheral. After the configuration phase, a data phase TD (or a series of data phase TDs) can begin (potentially using scatter gather). When the data

phase TD(s) finish, a status phase TD can be invoked that reads some memory mapped status information from the peripheral and copies it to a location in system memory specified by the CPU for later inspection. Multiple sets of configuration, data, and status phase 'subchains' can be strung together to create larger chains that transmit multiple packets in this way. A similar concept exists in the opposite direction to receive the packets.

4.4.4.7 Nested DMA

One TD may modify another TD, as the TD configuration space is memory mapped similar to any other peripheral. For example, a first TD loads a second TD's configuration and then calls the second TD. The second TD moves data as required by the application. When complete, the second TD calls the first TD, which again updates the second TD's configuration. This process repeats as often as necessary.

4.5 Interrupt Controller

The interrupt controller provides a mechanism for hardware resources to change program execution to a new address, independent of the current task being executed by the main code. The interrupt controller provides enhanced features not found on original 8051 interrupt controllers:

- Thirty-two interrupt vectors
- Jumps directly to ISR anywhere in code space with dynamic vector addresses
- Multiple sources for each vector
- Flexible interrupt to vector matching
- Each interrupt vector is independently enabled or disabled
- Each interrupt can be dynamically assigned one of eight priorities
- Eight level nestable interrupts
- Multiple I/O interrupt vectors
- Software can send interrupts
- Software can clear pending interrupts

Figure 4-2 on page 20 represents typical flow of events when an interrupt triggered. Figure 4-3 on page 21 shows the interrupt structure and priority polling.

Table 4-8. Interrupt Vector Table

#	Fixed Function	DMA	UDB
0	LVD	phub_termout0[0]	udb_intr[0]
1	Cache/ECC	phub_termout0[1]	udb_intr[1]
2	Reserved	phub_termout0[2]	udb_intr[2]
3	Sleep (Pwr Mgr)	phub_termout0[3]	udb_intr[3]
4	PICU[0]	phub_termout0[4]	udb_intr[4]
5	PICU[1]	phub_termout0[5]	udb_intr[5]
6	PICU[2]	phub_termout0[6]	udb_intr[6]
7	PICU[3]	phub_termout0[7]	udb_intr[7]
8	PICU[4]	phub_termout0[8]	udb_intr[8]
9	PICU[5]	phub_termout0[9]	udb_intr[9]
10	PICU[6]	phub_termout0[10]	udb_intr[10]
11	PICU[12]	phub_termout0[11]	udb_intr[11]
12	PICU[15]	phub_termout0[12]	udb_intr[12]
13	Comparators Combined	phub_termout0[13]	udb_intr[13]
14	Switched Caps Combined	phub_termout0[14]	udb_intr[14]
15	I ² C	phub_termout0[15]	udb_intr[15]
16	CAN	phub_termout1[0]	udb_intr[16]
17	Timer/Counter0	phub_termout1[1]	udb_intr[17]
18	Timer/Counter1	phub_termout1[2]	udb_intr[18]
19	Timer/Counter2	phub_termout1[3]	udb_intr[19]
20	Timer/Counter3	phub_termout1[4]	udb_intr[20]
21	USB SOF Int	phub_termout1[5]	udb_intr[21]
22	USB Arb Int	phub_termout1[6]	udb_intr[22]
23	USB Bus Int	phub_termout1[7]	udb_intr[23]
24	USB Endpoint[0]	phub_termout1[8]	udb_intr[24]
25	USB Endpoint Data	phub_termout1[9]	udb_intr[25]
26	Reserved	phub_termout1[10]	udb_intr[26]
27	LCD	phub_termout1[11]	udb_intr[27]
28	DFB Int	phub_termout1[12]	udb_intr[28]
29	Decimator Int	phub_termout1[13]	udb_intr[29]
30	PHUB Error Int	phub_termout1[14]	udb_intr[30]
31	EEPROM Fault Int	phub_termout1[15]	udb_intr[31]

6.1.1.1 Internal Oscillators

Figure 6-1 shows that there are two internal oscillators. They can be routed directly or divided. The direct routes may not have a 50% duty cycle. Divided clocks have a 50% duty cycle.

6.1.1.1.1 Internal Main Oscillator

In most designs the IMO is the only clock source required, due to its ± 1 -percent accuracy. The IMO operates with no external components and outputs a stable clock. A factory trim for each frequency range is stored in the device. With the factory trim, tolerance varies from ± 1 percent at 3 MHz, up to ± 7 percent at 62 MHz. The IMO, in conjunction with the PLL, allows generation of other clocks up to the device's maximum frequency (see PLL). The IMO provides clock outputs at 3, 6, 12, 24, 48, and 62 MHz.

6.1.1.1.2 Clock Doubler

The clock doubler outputs a clock at twice the frequency of the input clock. The doubler works at input frequency of 24 MHz, providing 48 MHz for the USB. It can be configured to use a clock from the IMO, MHzECO, or the DSI (external pin).

6.1.1.1.3 PLL

The PLL allows low-frequency, high-accuracy clocks to be multiplied to higher frequencies. This is a trade off between higher clock frequency and accuracy and, higher power consumption and increased startup time.

The PLL block provides a mechanism for generating clock frequencies based upon a variety of input sources. The PLL outputs clock frequencies in the range of 24 to 67 MHz. Its input and feedback dividers supply 4032 discrete ratios to create almost any desired clock frequency. The accuracy of the PLL output depends on the accuracy of the PLL input source. The most common PLL use is to multiply the IMO clock at 3 MHz, where it is most accurate, to generate the other clocks up to the device's maximum frequency.

The PLL achieves phase lock within 250 μ s (verified by bit setting). It can be configured to use a clock from the IMO, MHzECO or DSI (external pin). The PLL clock source can be used until lock is complete and signaled with a lock bit. The lock signal can be routed through the DSI to generate an interrupt. Disable the PLL before entering low-power modes.

6.1.1.1.4 Internal Low-Speed Oscillator

The ILO provides clock frequencies for low-power consumption, including the watchdog timer, and sleep timer. The ILO generates up to three different clocks: 1 kHz, 33 kHz, and 100 kHz. The 1-kHz clock (CLK1K) is typically used for a background 'heartbeat' timer. This clock inherently lends itself to low-power supervisory operations such as the watchdog timer and long sleep intervals using the central timewheel (CTW).

The central timewheel is a 1-kHz, free running, 13-bit counter clocked by the ILO. The central timewheel is always enabled, except in hibernate mode and when the CPU is stopped during debug on chip mode. It can be used to generate periodic interrupts for timing purposes or to wake the system from a low-power mode. Firmware can reset the central timewheel. Systems that require accurate timing should use the RTC capability instead of the central timewheel.

The 100-kHz clock (CLK100K) can be used as a low power master clock. It can also generate time intervals using the fast timewheel.

The fast timewheel is a 5-bit counter, clocked by the 100-kHz clock. It features programmable settings and automatically resets when the terminal count is reached. An optional interrupt can be generated each time the terminal count is reached. This enables flexible, periodic interrupts of the CPU at a higher rate than is allowed using the central timewheel.

The 33-kHz clock (CLK33K) comes from a divide-by-3 operation on CLK100K. This output can be used as a reduced accuracy version of the 32.768-kHz ECO clock with no need for a crystal.

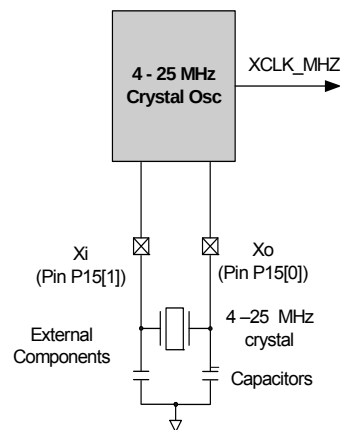
6.1.2 External Oscillators

Figure 6-1 shows that there are two external oscillators. They can be routed directly or divided. The direct routes may not have a 50% duty cycle. Divided clocks have a 50% duty cycle.

6.1.2.1 MHz External Crystal Oscillator

The MHzECO provides high frequency, high precision clocking using an external crystal (see Figure 6-2). It supports a wide variety of crystal types, in the range of 4 to 25 MHz. When used in conjunction with the PLL, it can generate other clocks up to the device's maximum frequency (see PLL). The GPIO pins connecting to the external crystal and capacitors are fixed. MHzECO accuracy depends on the crystal chosen.

Figure 6-2. MHzECO Block Diagram

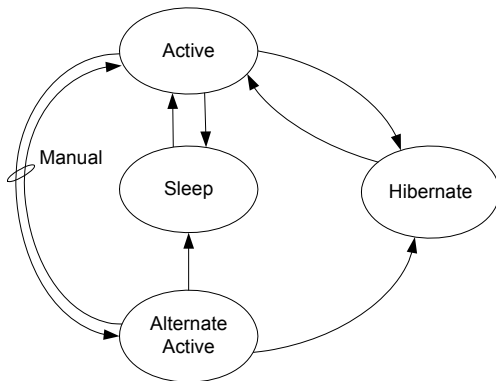


6.1.2.2 32.768-kHz ECO

The 32.768-kHz external crystal oscillator (32kHzECO) provides precision timing with minimal power consumption using an external 32.768-kHz watch crystal (see Figure 6-3). The 32kHzECO also connects directly to the sleep timer and provides the source for the RTC. The RTC uses a 1-second interrupt to implement the RTC functionality in firmware.

The oscillator works in two distinct power modes. This allows users to trade off power consumption with noise immunity from neighboring circuits. The GPIO pins connected to the external crystal and capacitors are fixed.

Figure 6-5. Power Mode Transitions



6.2.1.1 Active Mode

Active mode is the primary operating mode of the device. When in active mode, the active configuration template bits control which available resources are enabled or disabled. When a resource is disabled, the digital clocks are gated, analog bias currents are disabled, and leakage currents are reduced as appropriate. User firmware can dynamically control subsystem power by setting and clearing bits in the active configuration template. The CPU can disable itself, in which case the CPU is automatically reenabled at the next wakeup event.

When a wakeup event occurs, the global mode is always returned to active, and the CPU is automatically enabled, regardless of its template settings. Active mode is the default global power mode upon boot.

6.2.1.2 Alternate Active Mode

Alternate Active mode is very similar to Active mode. In alternate active mode, fewer subsystems are enabled, to reduce power consumption. One possible configuration is to turn off the CPU and flash, and run peripherals at full speed.

6.2.1.3 Sleep Mode

Sleep mode reduces power consumption when a resume time of 15 μ s is acceptable. The wake time is used to ensure that the regulator outputs are stable enough to directly enter active mode.

6.2.1.4 Hibernate Mode

In hibernate mode nearly all of the internal functions are disabled. Internal voltages are reduced to the minimal level to keep vital systems alive. Configuration state is preserved in hibernate mode and SRAM memory is retained. GPIOs configured as digital outputs maintain their previous values and external GPIO pin interrupt settings are preserved. The device can only return from hibernate mode in response to an external I/O interrupt. The resume time from hibernate mode is less than 100 μ s.

To achieve an extremely low current, the hibernate regulator has limited capacity. This limits the frequency of any signal present on the input pins - no GPIO should toggle at a rate greater than 10 kHz while in hibernate mode. If pins must be toggled at a high rate while in a low power mode, use sleep mode instead.

6.2.1.5 Wakeup Events

Wakeup events are configurable and can come from an interrupt or device reset. A wakeup event restores the system to active mode. Firmware enabled interrupt sources include internally generated interrupts, power supervisor, central timewheel, and I/O interrupts. Internal interrupt sources can come from a variety of peripherals, such as analog comparators and UDBs. The central timewheel provides periodic interrupts to allow the system to wake up, poll peripherals, or perform real-time functions. Reset event sources include the external reset I/O pin (XRES), WDT, and precision reset (PRES).

6.2.2 Boost Converter

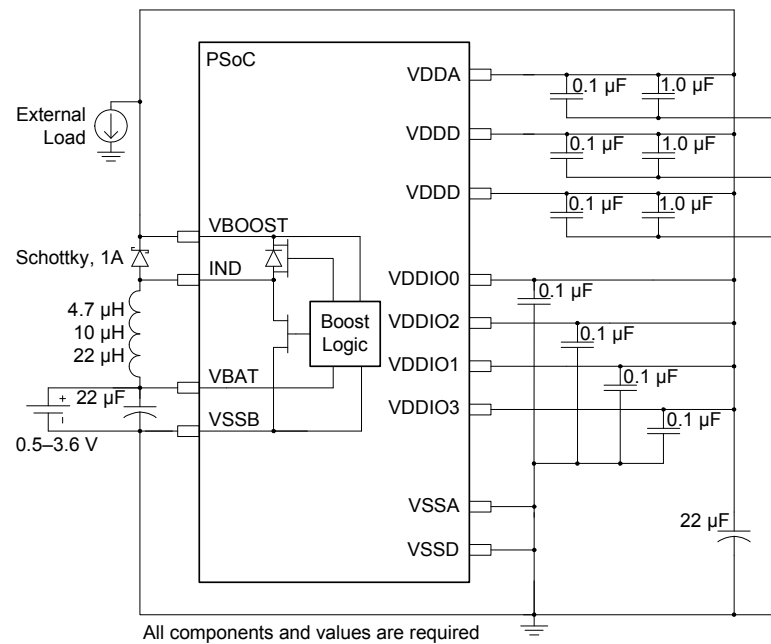
Applications that use a supply voltage of less than 1.71 V, such as solar panels or single cell battery supplies, may use the on-chip boost converter to generate a minimum of 1.8 V supply voltage. The boost converter may also be used in any system that requires a higher operating voltage than the supply provides such as driving 5.0 V LCD glass in a 3.3 V system. With the addition of an inductor, Schottky diode, and capacitors, it produces a selectable output voltage sourcing enough current to operate the PSoC and other on-board components.

The boost converter accepts an input voltage V_{BAT} from 0.5 V to 3.6 V, and can start up with V_{BAT} as low as 0.5 V. The converter provides a user configurable output voltage of 1.8 to 5.0 V (V_{OUT}) in 100 mV increments. V_{BAT} is typically less than V_{OUT} ; if V_{BAT} is greater than or equal to V_{OUT} , then V_{OUT} will be slightly less than V_{BAT} due to resistive losses in the boost converter. The block can deliver up to 50 mA (I_{BOOST}) depending on configuration to both the PSoC device and external components. The sum of all current sinks in the design including the PSoC device, PSoC I/O pin loads, and external component loads must be less than the I_{BOOST} specified maximum current.

Four pins are associated with the boost converter: VBAT, VSSB, VBOOST, and IND. The boosted output voltage is sensed at the VBOOST pin and must be connected directly to the chip's supply inputs; VDDA, VDDD, and VDDIO if used to power the PSoC device.

The boost converter requires four components in addition to those required in a non-boost design, as shown in Figure 6-6 on page 34. A 22 μ F capacitor (C_{BAT}) is required close to the VBAT pin to provide local bulk storage of the battery voltage and provide regulator stability. A diode between the battery and VBAT pin should not be used for reverse polarity protection because the diodes forward voltage drop reduces the V_{BAT} voltage. Between the VBAT and IND pins, an inductor of 4.7 μ H, 10 μ H, or 22 μ H is required. The inductor value can be optimized to increase the boost converter efficiency based on input voltage, output voltage, temperature, and current. Inductor size is determined by following the design guidance in this chapter and electrical specifications. The inductor must be placed within 1 cm of the VBAT and IND pins and have a minimum saturation current of 750 mA. Between the IND and VBOOST pins, place a Schottky diode within 1 cm of the pins. The Schottky diode shall have a forward current rating of at least 1.0 A and a reverse voltage of at least 20 V. Connect a 22- μ F bulk capacitor (C_{BOOST}) close to VBOOST to provide regulator output stability. It is important to sum the total capacitance connected to the VBOOST pin and ensure the maximum CBOOST specification is not exceeded. All capacitors must be rated for a minimum of 10 V to minimize capacitive losses due to voltage de-rating.

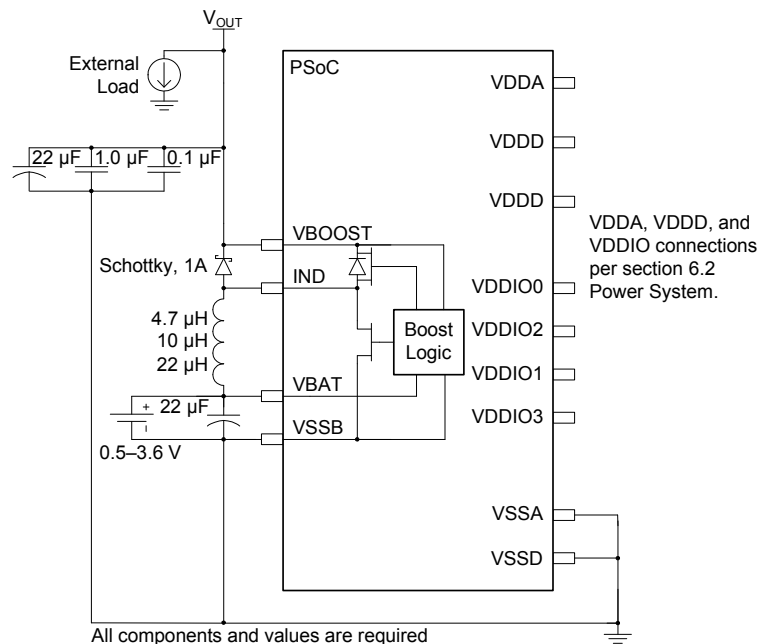
Figure 6-6. Application of Boost Converter powering PSoC device



The boost converter may also generate a supply that is not used directly by the PSoC device. An example of this use case is boosting a 1.8 V supply to 4.0 V to drive a white LED. If the boost converter is not supplying the PSoC devices V_{DDA} , V_{DDD} , and V_{DDIO} it must comply with the same design rules as supplying

the PSoC device, but with a change to the bulk capacitor requirements. A parallel arrangement 22 μ F, 1.0 μ F, and 0.1 μ F capacitors are all required on the Vout supply and must be placed within 1 cm of the VBOOST pin to ensure regulator stability.

Figure 6-7. Application of Boost Converter not powering PSoC device



The switching frequency is set to 400 kHz using an oscillator integrated into the boost converter. The boost converter can be operated in two different modes: active and standby. Active mode is the normal mode of operation where the boost regulator

actively generates a regulated output voltage. In standby mode, most boost functions are disabled, thus reducing power consumption of the boost circuit. Only minimal power is provided, typically $< 5 \mu\text{A}$ to power the PSoC device in Sleep mode. The

7. Digital Subsystem

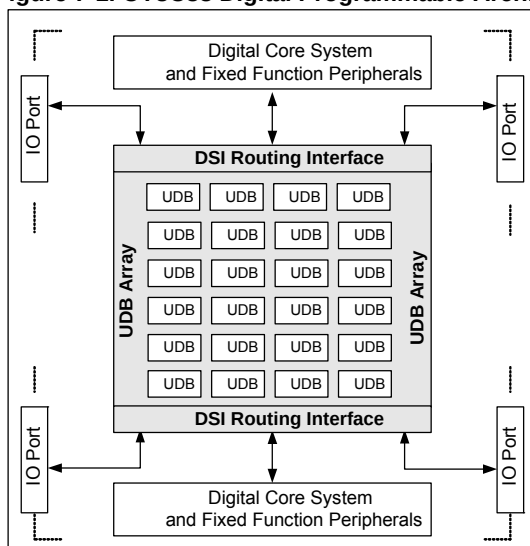
The digital programmable system creates application specific combinations of both standard and advanced digital peripherals and custom logic functions. These peripherals and logic are then interconnected to each other and to any pin on the device, providing a high level of design flexibility and IP security.

The features of the digital programmable system are outlined here to provide an overview of capabilities and architecture. You do not need to interact directly with the programmable digital system at the hardware and register level. PSoC Creator provides a high level schematic capture graphical interface to automatically place and route resources similar to PLDs.

The main components of the digital programmable system are:

- **UDB** – These form the core functionality of the digital programmable system. UDBs are a collection of uncommitted logic (PLD) and structural logic (Datapath) optimized to create all common embedded peripherals and customized functionality that are application or design specific.
- **Universal digital block array** – UDB blocks are arrayed within a matrix of programmable interconnect. The UDB array structure is homogeneous and allows for flexible mapping of digital functions onto the array. The array supports extensive and flexible routing interconnects between UDBs and the Digital System Interconnect.
- **Digital system interconnect (DSI)** – Digital signals from UDBs, fixed function peripherals, I/O pins, interrupts, DMA, and other system core signals are attached to the digital system interconnect to implement full featured device connectivity. The DSI allows any digital function to any pin or other feature routability when used with the universal digital block array.

Figure 7-1. CY8C38 Digital Programmable Architecture



7.1 Example Peripherals

The flexibility of the CY8C38 family's UDBs and analog blocks allow the user to create a wide range of components (peripherals). The most common peripherals were built and characterized by Cypress and are shown in the PSoC Creator component catalog, however, users may also create their own custom components using PSoC Creator. Using PSoC Creator, users may also create their own components for reuse within their organization, for example sensor interfaces, proprietary algorithms, and display interfaces.

The number of components available through PSoC Creator is too numerous to list in the data sheet, and the list is always growing. An example of a component available for use in CY8C38 family, but, not explicitly called out in this data sheet is the UART component.

7.1.1 Example Digital Components

The following is a sample of the digital components available in PSoC Creator for the CY8C38 family. The exact amount of hardware resources (UDBs, routing, RAM, flash) used by a component varies with the features selected in PSoC Creator for the component.

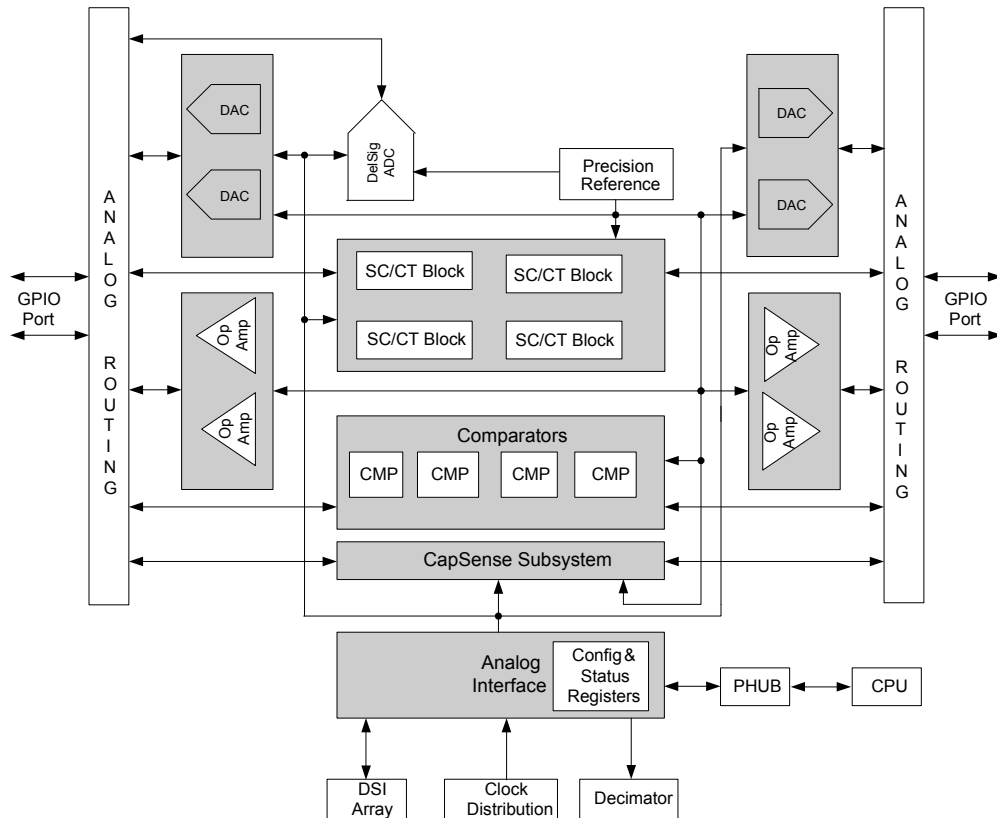
- **Communications**
 - I²C
 - UART
 - SPI
- **Functions**
 - EMIF
 - PWMs
 - Timers
 - Counters
- **Logic**
 - NOT
 - OR
 - XOR
 - AND

7.1.2 Example Analog Components

The following is a sample of the analog components available in PSoC Creator for the CY8C38 family. The exact amount of hardware resources (SC/CT blocks, routing, RAM, flash) used by a component varies with the features selected in PSoC Creator for the component.

- **Amplifiers**
 - TIA
 - PGA
 - opamp
- **ADC**
 - Delta-sigma
- **DACs**
 - Current
 - Voltage
 - PWM
- **Comparators**
- **Mixers**

Figure 8-1. Analog Subsystem Block Diagram



The PSoC Creator software program provides a user friendly interface to configure the analog connections between the GPIO and various analog resources and connections from one analog resource to another. PSoC Creator also provides component libraries that allow you to configure the various analog blocks to perform application specific functions (PGA, transimpedance amplifier, voltage DAC, current DAC, and so on). The tool also generates API interface libraries that allow you to write firmware that allows the communication between the analog peripheral and CPU/Memory.

8.1 Analog Routing

The CY8C38 family of devices has a flexible analog routing architecture that provides the capability to connect GPIOs and different analog blocks, and also route signals between different analog blocks. One of the strong points of this flexible routing architecture is that it allows dynamic routing of input and output connections to the different analog blocks.

For information on how to make pin selections for optimal analog routing, refer to the application note, [AN58304 - PSoC® 3 and PSoC® 5 - Pin Selection for Analog Designs](#).

8.1.1 Features

- Flexible, configurable analog routing architecture
- 16 analog globals (AG) and two analog mux buses (AMUXBUS) to connect GPIOs and the analog blocks
- Each GPIO is connected to one analog global and one analog mux bus
- Eight analog local buses (abus) to route signals between the different analog blocks
- Multiplexers and switches for input and output selection of the analog blocks

8.1.2 Functional Description

Analog globals (AGs) and analog mux buses (AMUXBUS) provide analog connectivity between GPIOs and the various analog blocks. There are 16 AGs in the CY8C38 family. The analog routing architecture is divided into four quadrants as shown in [Figure 8-2](#). Each quadrant has four analog globals (AGL[0..3], AGL[4..7], AGR[0..3], AGR[4..7]). Each GPIO is connected to the corresponding AG through an analog switch. The analog mux bus is a shared routing resource that connects to every GPIO through an analog switch. There are two AMUXBUS routes in CY8C38, one in the left half (AMUXBUSL) and one in the right half (AMUXBUSR), as shown in [Figure 8-2](#) on page 58.

8.2.2.3 Multi Sample

Multi sample mode is similar to continuous mode except that the ADC is reset between samples. This mode is useful when the input is switched between multiple signals. The decimator is re-primed between each sample so that previous samples do not affect the current conversion. Upon completion of a sample, the next sample is automatically initiated. The results can be transferred using either firmware polling, interrupt, or DMA.

8.2.2.4 Multi Sample (Turbo)

The multi sample (turbo) mode operates identical to the Multi-sample mode for resolutions of 8 to 16 bits. For resolutions of 17 to 20 bits, the performance is about four times faster than the multi sample mode, because the ADC is only reset once at the end of conversion.

More information on output formats is provided in the Technical Reference Manual.

8.2.3 Start of Conversion Input

The SoC signal is used to start an ADC conversion. A digital clock or UDB output can be used to drive this input. It can be used when the sampling period must be longer than the ADC conversion time or when the ADC must be synchronized to other hardware. This signal is optional and does not need to be connected if ADC is running in a continuous mode.

8.2.4 End of Conversion Output

The EoC signal goes high at the end of each ADC conversion. This signal may be used to trigger either an interrupt or DMA request.

8.3 Comparators

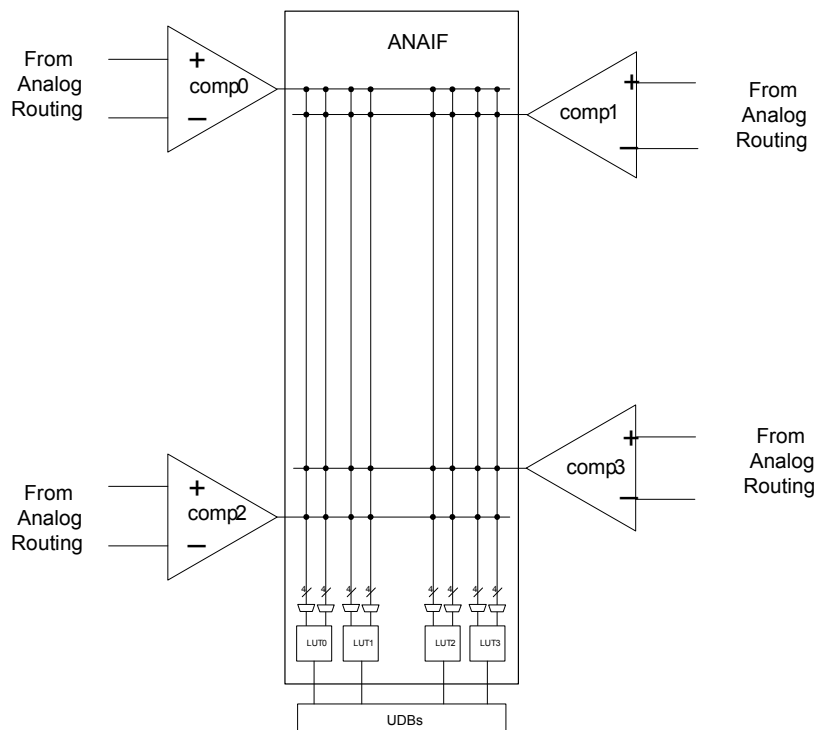
The CY8C38 family of devices contains four comparators in a device. Comparators have these features:

- Input offset factory trimmed to less than 5 mV
- Rail-to-rail common mode input range (VSSA to VDDA)
- Speed and power can be traded off by using one of three modes: fast, slow, or ultra low-power
- Comparator outputs can be routed to lookup tables to perform simple logic functions and then can also be routed to digital blocks
- The positive input of the comparators may be optionally passed through a low pass filter. Two filters are provided
- Comparator inputs can be connections to GPIO, DAC outputs and SC block outputs

8.3.1 Input and Output Interface

The positive and negative inputs to the comparators come from the analog global buses, the analog mux line, the analog local bus and precision reference through multiplexers. The output from each comparator could be routed to any of the two input LUTs. The output of that LUT is routed to the UDB DSI.

Figure 8-5. Analog Comparator



9.8 CSP Package Bootloader

A factory-installed bootloader program is included in all devices with CSP packages. The bootloader is compatible with PSoC Creator 3.0 bootloadable project files and has the following features:

- I²C-based
- SCLK and SDAT available at P1[6] and P1[7], respectively
- External pull-up resistors required
- I²C slave, address 4, data rate = 100 kbps
- Single application
- Wait two seconds for bootload command
- Other bootloader options are as set by the PSoC Creator 3.0 Bootloader Component default
- Occupies the bottom 9K of flash

For more information on this bootloader, see the following Cypress application notes:

- [AN89611](#) – PSoC® 3 AND PSoC 5LP - Getting Started With Chip Scale Packages (CSP)
- [AN73854](#) – PSoC 3 and PSoC 5 LP Introduction to Bootloaders
- [AN60317](#) – PSoC 3 and PSoC 5 LP I²C Bootloader

Note that a PSoC Creator bootloadable project must be associated with .hex and .elf files for a bootloader project that is configured for the target device. Bootloader .hex and .elf files can be found at www.cypress.com/go/PSoC3datasheet.

The factory-installed bootloader can be overwritten using JTAG or SWD programming.

11.4.2 SIO

Table 11-11. SIO DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
V _{inmax}	Maximum input voltage	All allowed values of V _{DDIO} and V _{DDD} , see Section 11.1	–	–	5.5	V
V _{inref}	Input voltage reference (Differential input mode)		0.5	–	0.52 × V _{DDIO}	V
V _{outref}	Output voltage reference (Regulated output mode)					
		V _{DDIO} > 3.7	1	–	V _{DDIO} – 1	V
		V _{DDIO} < 3.7	1	–	V _{DDIO} – 0.5	V
V _{IH}	Input voltage high threshold					
	GPIO mode	CMOS input	0.7 × V _{DDIO}	–	–	V
	Differential input mode ^[42]	Hysteresis disabled	SIO_ref + 0.2	–	–	V
V _{IL}	Input voltage low threshold					
	GPIO mode	CMOS input	–	–	0.3 × V _{DDIO}	V
	Differential input mode ^[42]	Hysteresis disabled	–	–	SIO_ref – 0.2	V
V _{OH}	Output voltage high					
	Unregulated mode	I _{OH} = 4 mA, V _{DDIO} = 3.3 V	V _{DDIO} – 0.4	–	–	V
	Regulated mode ^[42]	I _{OH} = 1 mA	SIO_ref – 0.65	–	SIO_ref + 0.2	V
	Regulated mode ^[42]	I _{OH} = 0.1 mA	SIO_ref – 0.3	–	SIO_ref + 0.2	V
V _{OL}	Output voltage low					
		V _{DDIO} = 3.30 V, I _{OL} = 25 mA	–	–	0.8	V
		V _{DDIO} = 3.30 V, I _{OL} = 20 mA	–	–	0.4	V
		V _{DDIO} = 1.80 V, I _{OL} = 4 mA	–	–	0.4	V
R _{pullup}	Pull-up resistor		3.5	5.6	8.5	kΩ
R _{pulldown}	Pull-down resistor		3.5	5.6	8.5	kΩ
I _{IL}	Input leakage current (Absolute value) ^[43]					
	V _{IH} ≤ V _{ddsio}	25 °C, V _{ddsio} = 3.0 V, V _{IH} = 3.0 V	–	–	14	nA
	V _{IH} > V _{ddsio}	25 °C, V _{ddsio} = 0 V, V _{IH} = 3.0 V	–	–	10	μA
C _{IN}	Input Capacitance ^[43]		–	–	7	pF
V _H	Input voltage hysteresis (Schmitt-Trigger) ^[43]					
		Single ended mode (GPIO mode)	–	40	–	mV
		Differential mode	–	35	–	mV
I _{diode}	Current through protection diode to V _{SSIO}		–	–	100	μA

Notes

42. See [Figure 6-10](#) on page 39 and [Figure 6-13](#) on page 43 for more information on SIO reference.

43. Based on device characterization (Not production tested).

Figure 11-22. USBIO Output High Voltage and Current, GPIO Mode

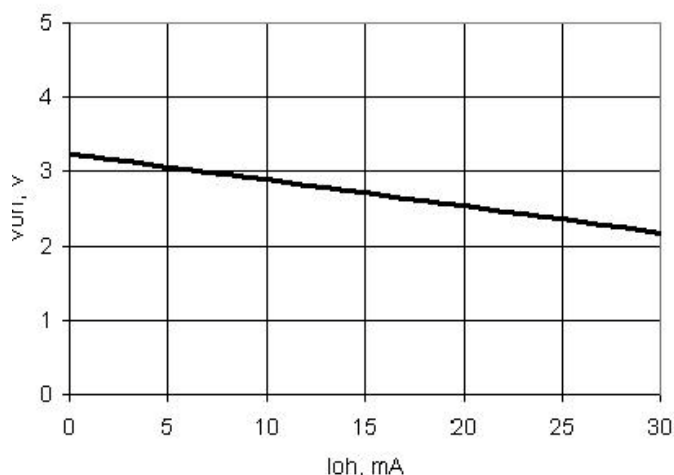


Figure 11-23. USBIO Output Low Voltage and Current, GPIO Mode

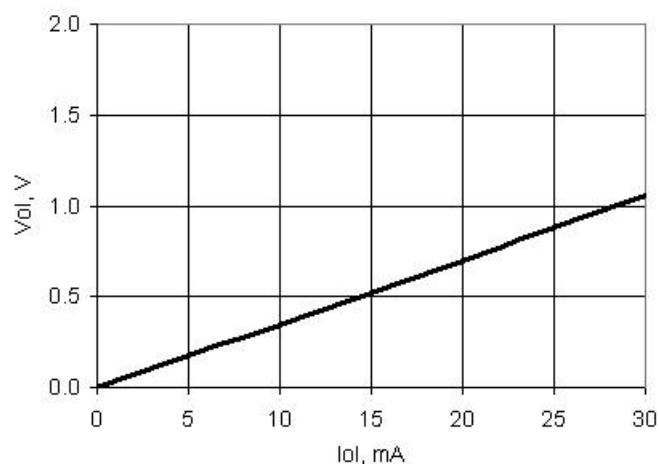


Table 11-15. USBIO AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
Tdrate	Full-speed data rate average bit rate		12 – 0.25%	12	12 + 0.25%	MHz
Tjr1	Receiver data jitter tolerance to next transition		–8	–	8	ns
Tjr2	Receiver data jitter tolerance to pair transition		–5	–	5	ns
Tdj1	Driver differential jitter to next transition		–3.5	–	3.5	ns
Tdj2	Driver differential jitter to pair transition		–4	–	4	ns
Tfdeop	Source jitter for differential transition to SE0 transition		–2	–	5	ns
Tfeopt	Source SE0 interval of EOP		160	–	175	ns
Tfeopr	Receiver SE0 interval of EOP		82	–	–	ns
Tfst	Width of SE0 interval during differential transition		–	–	14	ns
Fgpio_out	GPIO mode output operating frequency	3 V ≤ V _{DDD} ≤ 5.5 V	–	–	20	MHz
		V _{DDD} = 1.71 V	–	–	6	MHz
Tr_gpio	Rise time, GPIO mode, 10%/90% V _{DDD}	V _{DDD} > 3 V, 25 pF load	–	–	12	ns
		V _{DDD} = 1.71 V, 25 pF load	–	–	40	ns
Tf_gpio	Fall time, GPIO mode, 90%/10% V _{DDD}	V _{DDD} > 3 V, 25 pF load	–	–	12	ns
		V _{DDD} = 1.71 V, 25 pF load	–	–	40	ns

Figure 11-53. IDAC PSRR vs Frequency

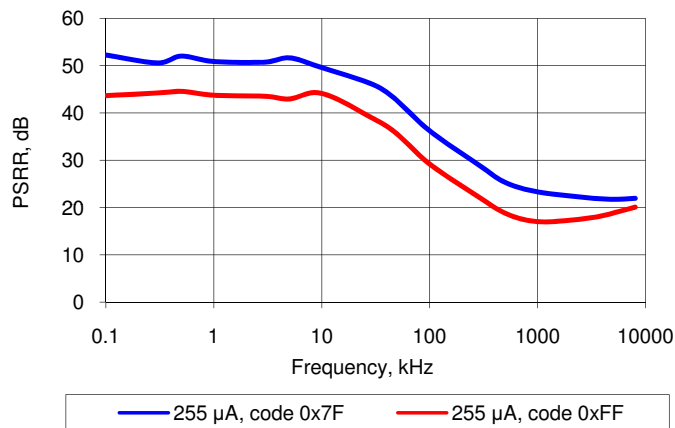
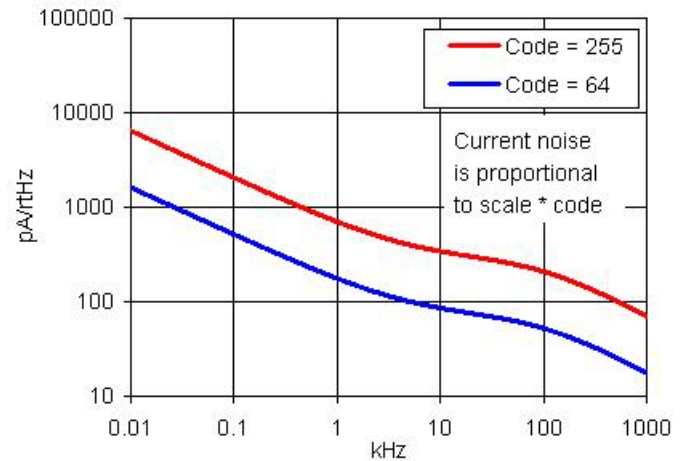


Figure 11-54. IDAC Current Noise, 255 µA Mode, Source Mode, High speed mode, $V_{DDA} = 5$ V



11.5.7 Voltage Digital to Analog Converter (VDAC)

See the VDAC component datasheet in PSoC Creator for full electrical specifications and APIs.

Unless otherwise specified, all charts and graphs show typical values.

Table 11-34. VDAC DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
	Resolution		–	8	–	bits
V_{OUT}	Output voltage range, code = 255	1 V scale	–	1.02	–	V
		4 V scale, $V_{DDA} = 5$ V	–	4.08	–	V
INL1	Integral nonlinearity	1 V scale	–	±2.1	±2.5	LSB
INL4	Integral nonlinearity ^[57]	4 V scale	–	±2.1	±2.5	LSB
DNL1	Differential nonlinearity	1 V scale	–	±0.3	±1	LSB
DNL4	Differential nonlinearity ^[57]	4 V scale	–	±0.3	±1	LSB
Rout	Output resistance	1 V scale	–	4	–	kΩ
		4 V scale	–	16	–	kΩ
	Monotonicity		–	–	Yes	–
V_{OS}	Zero scale error		–	0	±0.9	LSB
Eg	Gain error	1 V scale,	–	–	±2.5	%
		4 V scale	–	–	±2.5	%
TC_Eg	Temperature coefficient, gain error	1 V scale,	–	–	0.03	%FSR / °C
		4 V scale	–	–	0.03	%FSR / °C
I_{DD}	Operating current	Low speed mode	–	–	100	µA
		High speed mode	–	–	500	µA

Note

57. Based on device characterization (Not production tested).

Figure 11-61. VDAC Operating Current vs Temperature, 1 V Mode, Low speed mode

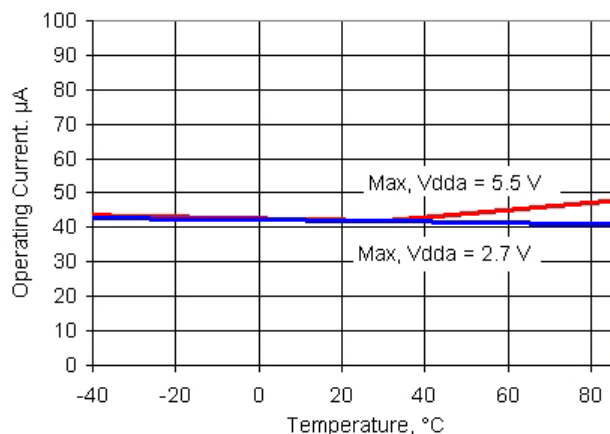


Figure 11-62. VDAC Operating Current vs Temperature, 1 V Mode, High speed mode

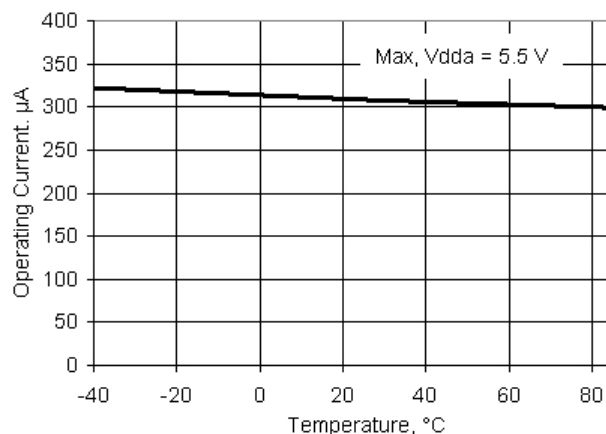


Table 11-35. VDAC AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
F_{DAC}	Update rate	1 V scale	–	–	1000	ksps
		4 V scale	–	–	250	ksps
$T_{settleP}$	Settling time to 0.1%, step 25% to 75%	1 V scale, Cload = 15 pF	–	0.45	1	µs
		4 V scale, Cload = 15 pF	–	0.8	3.2	µs
$T_{settleN}$	Settling time to 0.1%, step 75% to 25%	1 V scale, Cload = 15 pF	–	0.45	1	µs
		4 V scale, Cload = 15 pF	–	0.7	3	µs
	Voltage noise	Range = 1 V, High speed mode, $V_{DDA} = 5$ V, 10 kHz	–	750	–	nV/sqrtHz

Figure 11-63. VDAC Step Response, Codes 0x40 - 0xC0, 1 V Mode, High speed mode, $V_{DDA} = 5$ V

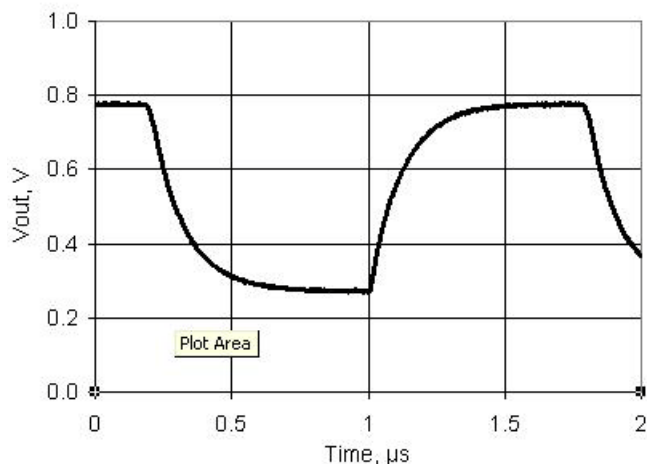
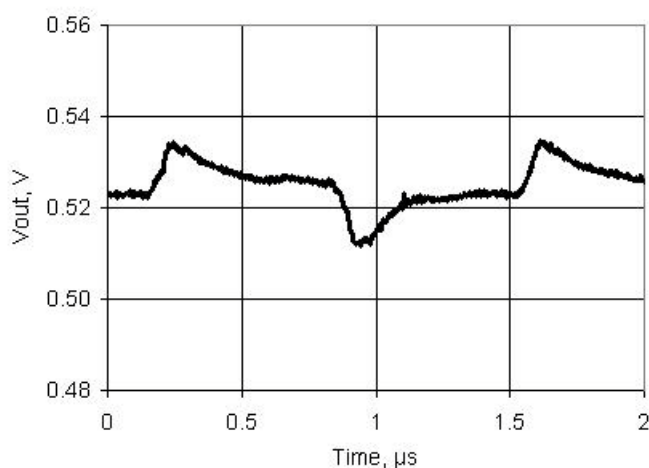


Figure 11-64. VDAC Glitch Response, Codes 0x7F - 0x80, 1 V Mode, High speed mode, $V_{DDA} = 5$ V



11.5.9 Transimpedance Amplifier

The TIA is created using a SC/CT analog block; see the TIA component data sheet in PSoC Creator for full electrical specifications and APIs.

Table 11-38. Transimpedance Amplifier (TIA) DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
V _{IOFF}	Input offset voltage		–	–	10	mV
R _{conv}	Conversion resistance ^[58]	R = 20K; 40 pF load	–25	–	+35	%
		R = 30K; 40 pF load	–25	–	+35	%
		R = 40K; 40 pF load	–25	–	+35	%
		R = 80K; 40 pF load	–25	–	+35	%
		R = 120K; 40 pF load	–25	–	+35	%
		R = 250K; 40 pF load	–25	–	+35	%
		R = 500K; 40 pF load	–25	–	+35	%
		R = 1M; 40 pF load	–25	–	+35	%
	Quiescent current		–	1.1	2	mA

Table 11-39. Transimpedance Amplifier (TIA) AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
BW	Input bandwidth (–3 dB)	R = 20K; –40 pF load	1500	–	–	kHz
		R = 120K; –40 pF load	240	–	–	kHz
		R = 1M; –40 pF load	25	–	–	kHz

Note

58. Conversion resistance values are not calibrated. Calibrated values and details about calibration are provided in PSoC Creator component data sheets. External precision resistors can also be used.

12. Ordering Information

In addition to the features listed in [Table 12-1](#), every CY8C38 device includes: a precision on-chip voltage reference, precision oscillators, flash, ECC, DMA, a fixed function I²C, 4 KB trace RAM, JTAG/SWD programming and debug, external memory interface, and more. In addition to these features, the flexible UDBs and analog subsection support a wide range of peripherals. To assist you in selecting the ideal part, PSoC Creator makes a part recommendation after you choose the components required by your application. All CY8C38 derivatives incorporate device and flash security in user-selectable security levels; see the TRM for details.

Table 12-1. CY8C38 Family with Single Cycle 8051

Part Number	MCU Core				Analog								Digital			I/O ^[83]				Package	JTAG ID ^[84]	
	CPU Speed (MHz)	Flash (KB)	SRAM (KB)	EEPROM (KB)	LCD Segment Drive	ADC	DAC	Comparator	SC/CT Analog Blocks ^[81]	Opamps	DFB	CapSense	UDBs ^[82]	16-bit Timer/PWM	FS USB	CAN 2.0b	Total I/O	GPIO	SIO			USBIO
32 KB Flash																						
CY8C3865AXI-019	67	32	4	1	✓	20-bit Del-Sig	4	4	4	4	✓	✓	20	4	✓	–	72	62	8	2	100-pin TQFP	0x1E013069
CY8C3865LTI-014	67	32	4	1	✓	20-bit Del-Sig	4	4	4	4	✓	✓	20	4	✓	–	48	38	8	2	68-pin QFN	0x1E00E069
CY8C3865AXI-204	67	32	8	1	✓	20-bit Del-Sig	2	0	0	0	–	✓	16	4	✓	–	72	62	8	2	100-pin TQFP	0x1E0CC069
CY8C3865LTI-205	67	32	8	1	✓	20-bit Del-Sig	2	0	0	0	–	✓	16	4	✓	–	48	38	8	2	68-pin QFN	0x1E0CD069
64 KB Flash																						
CY8C3866LTI-067	67	64	8	2	✓	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	✓	–	31	25	4	2	48-pin QFN	0x1E043069
CY8C3866PVI-021	67	64	8	2	✓	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	✓	–	31	25	4	2	48-pin SSOP	0x1E015069
CY8C3866AXI-035	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	–	✓	70	62	8	0	100-pin TQFP	0x1E023069
CY8C3866AXI-039	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	✓	–	72	62	8	2	100-pin TQFP	0x1E027069
CY8C3866LTI-030	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	✓	–	48	38	8	2	68-pin QFN	0x1E01E069
CY8C3866LTI-068	67	64	8	2	✓	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	✓	✓	31	25	4	2	48-pin QFN	0x1E044069
CY8C3866AXI-040	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	✓	✓	72	62	8	2	100-pin TQFP	0x1E028069
CY8C3866PVI-070	67	64	8	2	✓	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	–	✓	29	25	4	0	48-pin SSOP	0x1E046069
CY8C3866AXI-206	67	64	8	2	✓	20-bit Del-Sig	2	2	0	2	✓	✓	20	4	✓	–	72	62	8	2	100-pin TQFP	0x1E0CE069
CY8C3866LTI-207	67	64	8	2	✓	20-bit Del-Sig	2	2	0	2	✓	✓	20	4	✓	–	48	38	8	2	68-pin QFN	0x1E0CF069
CY8C3866AXI-208	67	64	8	2	✓	20-bit Del-Sig	2	2	2	2	✓	✓	24	4	✓	✓	72	62	8	2	100-pin TQFP	0x1E0D0069
CY8C3866LTI-209	67	64	8	2	✓	20-bit Del-Sig	2	2	2	2	✓	✓	24	4	✓	✓	48	38	8	2	68-pin QFN	0x1E0D1069
CY8C3866FNI-210	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	✓	✓	48	38	8	2	72 WLCSP	0x1E0D2069

Notes

81. Analog blocks support a variety of functionality including TIA, PGA, and mixers. See the [Example Peripherals](#) on page 44 for more information on how analog blocks can be used.
82. UDBs support a variety of functionality including SPI, LIN, UART, timer, counter, PWM, PRS, and others. Individual functions may use a fraction of a UDB or multiple UDBs. Multiple functions can share a single UDB. See the [Example Peripherals](#) on page 44 for more information on how UDBs can be used.
83. The I/O Count includes all types of digital I/O: GPIO, SIO, and the two USB I/O. See the [I/O System and Routing](#) on page 37 for details on the functionality of each of these types of I/O.
84. The JTAG ID has three major fields. The most significant nibble (left digit) is the version, followed by a 2 byte part number and a 3 nibble manufacturer ID.

17. Revision History

Description Title: PSoC® 3: CY8C38 Family Datasheet Programmable System-on-Chip (PSoC®) Document Number: 001-11729				
Revision	ECN	Submission Date	Orig. of Change	Description of Change
**	571504	See ECN	HMT	New data sheet for new device Part Number family.
*A	754416	See ECN	HMT	Prepare Preliminary for PR1.
*B	2253366	See ECN	DSG	Prepare Preliminary2 for PR3--total rewrite.
*C	2350209	See ECN	DSG	Minor change: Added "Confidential" watermark. Corrected typo on 68QFN pinout: pin 13 XREF to XRES.
*D	2481747	See ECN	SFV	Changed part numbers and data sheet title.
*E	2521877	See ECN	DSG	Prelim3 release--extensive spec, writing, and formatting changes
*F	2660161	02/16/09	GDK	Reorganized content to be consistent with the TRM. Added Xdata Space Access SFRs and DAC sections. Updated Boost Converter section and Conversion Signals section. Classified Ordering Information according to CPU speed; added information on security features and ROHS compliance. Added a section on XRES Specifications under Electrical Specification. Updated Analog Subsystem and CY8C35/55 Architecture block diagrams. Updated Electrical Specifications. Renamed CyDesigner as PSoC Creator
*G	2712468	05/29/09	MKEA	Updates to Electrical Specifications. Added Analog Routing section Updates to Ordering Information table
*H	2758970	09/02/09	MKEA	Updated Part Numbering Conventions. Added Section 11.7.5 (EMIF Figures and Tables). Updated GPIO and SIO AC specifications. Updated XRES Pin Description and Xdata Address Map specifications. Updated DFB and Comparator specifications. Updated PHUB features section and RTC in sleep mode. Updated IDAC and VDAC DC and Analog Global specifications. Updated USBIO AC and Delta Sigma ADC specifications. Updated PPOR and Voltage Monitors DC specifications. Updated Drive Mode diagram. Added 48-QFN Information. Updated other electrical specifications
*I	2824546	12/09/09	MKEA	Updated I2C section to reflect 1 Mbps. Updated Table 11-6 and 11-7 (Boost AC and DC specs); also added Schottky Diode specs. Changed current for sleep/hibernate mode to include SIO; Added footnote to analog global specs. Updated Figures 1-1, 6-2, 7-14, and 8-1. Updated Table 6-2 and Table 6-3 (Hibernate and Sleep rows) and Power Modes section. Updated GPIO and SIO AC specifications. Updated Gain error in IDAC and VDAC specifications. Updated description of V_{DDA} spec in Table 11-1 and removed GPIO Clamp Current parameter. Updated number of UDBs on page 1. Moved FILO from ILO DC to AC table. Added PCB Layout and PCB Schematic diagrams. Updated Fgpioout spec (Table 11-9). Added duty cycle frequency in PLL AC spec table. Added note for Sleep and Hibernate modes and Active Mode specs in Table 11-2. Linked URL in Section 10.3 to PSoC Creator site. Updated J_a and J_c values in Table 13-1. Updated Single Sample Mode and Fast FIR Mode sections. Updated Input Resistance specification in Del-Sig ADC table. Added T_{io_init} parameter. Updated PGA and UGB AC Specs. Removed SPC ADC. Updated Boost Converter section. Added section 'SIO as Comparator'; updated Hysteresis spec (differential mode) in Table 11-10. Updated V_{BAT} condition and deleted V_{start} parameter in Table 11-6. Added 'Bytes' column for Tables 4-1 to 4-5.
*J	2873322	02/04/10	MKEA	Changed maximum value of PPOR_TR to '1'. Updated V_{BIAS} specification. Updated PCB Schematic. Updated Figure 8-1 and Figure 6-3. Updated Interrupt Vector table, Updated Sales links. Updated JTAG and SWD specifications. Removed J_{p-p} and J_{period} from ECO AC Spec table. Added note on sleep timer in Table 11-2. Updated ILO AC and DC specifications. Added Resolution parameter in VDAC and IDAC tables. Updated I_{OUT} typical and maximum values. Changed Temperature Sensor range to -40 °C to +85 °C. Removed Latchup specification from Table 11-1.

Description Title: PSoC® 3: CY8C38 Family Datasheet Programmable System-on-Chip (PSoC®) (continued)
Document Number: 001-11729

Revision	ECN	Submission Date	Orig. of Change	Description of Change
*L	2938381	05/27/10	MKEA	Replaced V_{DDIO} with V_{DDD} in USBIO diagram and specification tables, added text in USBIO section of Electrical Specifications. Added Table 13-2 (Package MSL) Modified Tstorag condition and changed max spec to 100 Added bullet (Pass) under ALU (section 7.2.2.2) Added figures for kHzECO and MHzECO in the External Oscillator section Updated Figure 6-1(Clocking Subsystem diagram) Removed CPUCLK_DIV in table 5-2, Deleted Clock Divider SFR subsection Updated PSoC Creator Framework image Updated SIO DC Specifications (V_{IH} and V_{IL} parameters) Updated bullets in Clocking System and Clocking Distribution sections Updated Figure 8-2 Updated PCB Layout and Schematic, updated as per MTRB review comments Updated Table 6-3 (power changed to current) In 32kHz EC DC Specifications table, changed I_{CC} Max to 0.25 In IMO DC Specifications table, updated Supply Current values Updated GPIO DC Specs table
*M	2958674	06/22/10	SHEA	Minor ECN to post data sheet to external website
*N	2989685	08/04/10	MKEA	INL max is changed from 16 to 32 in Table 11-20, 20-bit Delta-sigma ADC AC Specifications. Added to Table 6-6 a footnote and references to same. Added sentences to the resistive pullup and pull-down description bullets. Added sentence to Section 6.4.11, Adjustable Output Level. Updated section 5.5 External Memory Interface Updated Table 11-73 JTAG Interface AC Specifications Updated Table 11-74 SWD Interface AC Specifications Updated style changes as per the new template.
*O	3078568	11/04/10	MKEA	Added 48-SSOP pin and package details. Removed PLL output duty cycle spec. Updated "Current Digital-to-analog Converter (IDAC)" on page 97 Updated "Voltage Digital to Analog Converter (VDAC)" on page 101 Updated Table 11-2, "DC Specifications," on page 72 Updated Table 11-28, "Voltage Reference Specifications," on page 95

Description Title: PSoC® 3: CY8C38 Family Datasheet Programmable System-on-Chip (PSoC®) (continued)
Document Number: 001-11729

Revision	ECN	Submission Date	Orig. of Change	Description of Change
*P	3107314	12/10/2010	MKEA	Updated delta-sigma tables and graphs. Formatted table 11.2. Updated interrupt controller table Updated transimpedance amplifier section Updated SIO DC specs table Updated Voltage Monitors DC Specifications table Updated LCD Direct Drive DC specs table Replaced the Discrete Time Mixer and Continuous Time Mixer tables with Mixer DC and AC specs tables Updated ESD _{HBM} value. Updated IDAC and VDAC sections Removed ESO parts from ordering information Changed USBIO pins from NC to DNU and removed redundant USBIO pin description notes Updated POR with brown out DC and AC specs Updated PGA AC specs Updated 32 kHz External Crystal DC Specifications Updated opamp AC specs Updated XRES IO specs Updated Inductive boost regulator section Delta sigma ADC spec updates Updated comparator section Removed buzz mode from Power Mode Transition diagram Updated opamp DC and AC spec tables Updated PGA DC table
*Q	3179219	22/02/2011	MKEA	Updated conditions for flash data retention time Updated 100-pin TQFP package spec. Updated EEPROM AC specifications.
*R	3200146	03/28/2011	MKEA	Removed Preliminary status from the data sheet. Updated JTAG ID Deleted Cin_G1, ADC input capacitance from Delta-Sigma ADC DC spec table Updated JTAG Interface AC Specifications and SWD Interface Specifications tables Updated USBIO DC specs Added 0.01 to max speed Updated Features on page 1 Added Section 5.5, Nonvolatile Latches Updated Flash AC specs Added CAN DC specs Updated delta-sigma graphs, noise histogram figures and RMS Noise spec tables Add reference to application note AN58304 in section 8.1 Updated 100-pin TQFP package spec Added oscillator, I/O, VDAC, regulator graphs Updated JTAG/SWD timing diagrams Updated GPIO and SIO AC specs Updated POR with Brown Out AC spec table Updated IDAC graphs Added DMA timing diagram, interrupt timing and interrupt vector, I2C timing diagrams Updated opamp graphs and PGA graphs Added full chip performance graphs Changed MHzECO range. Added "Solder Reflow Peak Temperature" table.