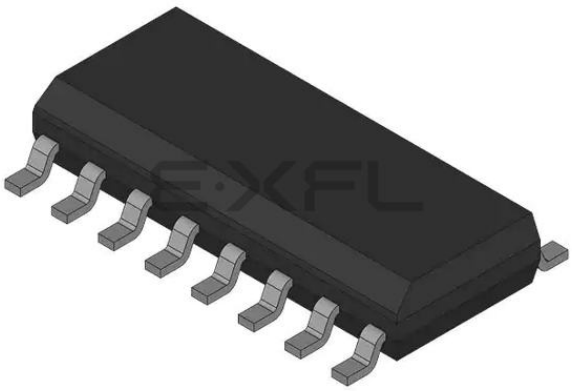

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Program Memory Size	8KB (8K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 4x10b
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Enhanced Serial Communications Interface (ESCI) Module

Chapter 14
System Integration Module (SIM)

Chapter 15
Serial Peripheral Interface (SPI) Module

2.6.3 FLASH Mass Erase Operation

Use the following procedure to erase the entire FLASH memory to read as a 1:

1. Set both the ERASE bit and the MASS bit in the FLASH control register.
2. Read the FLASH block protect register.
3. Write any data to any FLASH address[1] within the FLASH memory address range.
4. Wait for a time, t_{NVS}.
5. Set the HVEN bit.
6. Wait for a time, t_{MErase}.
7. Clear the ERASE and MASS bits.

NOTE

Mass erase is disabled whenever any block is protected (FLBPR does not equal $FF).

8. Wait for a time, t_{NVHL}.
9. Clear the HVEN bit.
10. After time, t_{RCV}, the memory can be accessed in read mode again.

NOTE

Programming and erasing of FLASH locations cannot be performed by code being executed from the FLASH memory. While these operations must be performed in the order as shown, other unrelated operations may occur between the steps.

CAUTION

A mass erase will erase the internal oscillator trim value at $FFC0.

2.6.4 FLASH Program Operation

Programming of the FLASH memory is done on a row basis. A row consists of 32 consecutive bytes starting from addresses $XX00, $XX20, $XX40, $XX60, $XX80, $XXA0, $XXC0, or $XXE0. Use the following step-by-step procedure to program a row of FLASH memory

Figure 2-4 shows a flowchart of the programming algorithm.

NOTE

Do not program any byte in the FLASH more than once after a successful erase operation. Reprogramming bits to a byte which is already programmed is not allowed without first erasing the page in which the byte resides or mass erasing the entire FLASH memory. Programming without first erasing may disturb data stored in the FLASH.

1. Set the PGM bit. This configures the memory for program operation and enables the latching of address and data for programming.
2. Read the FLASH block protect register.
3. Write any data to any FLASH location within the address range desired.
4. Wait for a time, t_{NVS}.
5. Set the HVEN bit.

1. When in monitor mode, with security sequence failed (see 17.3.2 Security), write to the FLASH block protect register instead of any FLASH address.

break, the bit cannot change during the break state as long as BCFE is cleared. After the break, doing the second step clears the status bit.

3.7 I/O Signals

The ADC10 module shares its pins with general-purpose input/output (I/O) port pins. See Figure 3-1 for port location of these shared pins. The ADC10 on this MCU uses V_{DD} and V_{SS} as its supply and reference pins. This MCU does not have an external trigger source.

3.7.1 ADC10 Analog Power Pin (V_{DDA})

The ADC10 analog portion uses V_{DDA} as its power pin. In some packages, V_{DDA} is connected internally to V_{DD}. If externally available, connect the V_{DDA} pin to the same voltage potential as V_{DD}. External filtering may be necessary to ensure clean V_{DDA} for good results.

NOTE

If externally available, route V_{DDA} carefully for maximum noise immunity and place bypass capacitors as near as possible to the package.

3.7.2 ADC10 Analog Ground Pin (V_{SSA})

The ADC10 analog portion uses V_{SSA} as its ground pin. In some packages, V_{SSA} is connected internally to V_{SS}. If externally available, connect the V_{SSA} pin to the same voltage potential as V_{SS}.

In cases where separate power supplies are used for analog and digital power, the ground connection between these supplies should be at the V_{SSA} pin. This should be the only ground connection between these supplies if possible. The V_{SSA} pin makes a good single point ground location.

3.7.3 ADC10 Voltage Reference High Pin (V_{REFH})

V_{REFH} is the power supply for setting the high-reference voltage for the converter. In some packages, V_{REFH} is connected internally to V_{DDA}. If externally available, V_{REFH} may be connected to the same potential as V_{DDA}, or may be driven by an external source that is between the minimum V_{DDA} spec and the V_{DDA} potential (V_{REFH} must never exceed V_{DDA}).

NOTE

Route V_{REFH} carefully for maximum noise immunity and place bypass capacitors as near as possible to the package.

AC current in the form of current spikes required to supply charge to the capacitor array at each successive approximation step is drawn through the V_{REFH} and V_{REFL} loop. The best external component to meet this current demand is a 0.1 μF capacitor with good high frequency characteristics. This capacitor is connected between V_{REFH} and V_{REFL} and must be placed as close as possible to the package pins. Resistance in the path is not recommended because the current will cause a voltage drop which could result in conversion errors. Inductance in this path must be minimum (parasitic only).

3.7.4 ADC10 Voltage Reference Low Pin (V_{REFL})

V_{REFL} is the power supply for setting the low-reference voltage for the converter. In some packages, V_{REFL} is connected internally to V_{SSA}. If externally available, connect the V_{REFL} pin to the same voltage potential as V_{SSA}. There will be a brief current associated with V_{REFL} when the sampling capacitor is

4.3 Functional Description

The function of the auto wakeup logic is to generate periodic wakeup requests to bring the microcontroller unit (MCU) out of stop mode. The wakeup requests are treated as regular keyboard interrupt requests, with the difference that instead of a pin, the interrupt signal is generated by an internal logic.

Entering stop mode will enable the auto wakeup generation logic. Writing the AWUIE bit in the keyboard interrupt enable register enables or disables the auto wakeup interrupt input (see Figure 4-1). A 1 applied to the AWUIREQ input with auto wakeup interrupt request enabled, latches an auto wakeup interrupt request.

Auto wakeup latch, AWUL, can be read directly from the bit 6 position of port A data register (PTA). This is a read-only bit which is occupying an empty bit position on PTA. No PTA associated registers, such as PTA6 data direction or PTA6 pullup exist for this bit.

There are two clock sources for the AWU. An internal RC oscillator (INTRCOSC, exclusive for the auto wakeup feature) drives the wakeup request generator provided the OSCENINSTOP bit in the CONFIG2 register Figure 4-1 is cleared. More accurate wakeup periods are possible using the BUSCLKX2 signal (from the oscillator module) which is selected by setting OSCENINSTOP.

Once the overflow count is reached in the generator counter, a wakeup request, AWUIREQ, is latched and sent to the KBI logic. See Figure 4-1.

Wakeup interrupt requests will only be serviced if the associated interrupt enable bit, AWUIE, in KBIER is set. The AWU shares the keyboard interrupt vector.

The overflow count can be selected from two options defined by the COPRS bit in CONFIG1. This bit was “borrowed” from the computer operating properly (COP) using the fact that the COP feature is idle (no MCU clock available) in stop mode. COPRS = 1 selects the short wakeup period while COPRS = 0 selects the long wakeup period.

The auto wakeup RC oscillator is highly dependent on operating voltage and temperature. This feature is not recommended for use as a time-keeping function.

The wakeup request is latched to allow the interrupt source identification. The latched value, AWUL, can be read directly from the bit 6 position of PTA data register. This is a read-only bit which is occupying an empty bit position on PTA. No PTA associated registers, such as PTA6 data, PTA6 direction, and PTA6 pullup exist for this bit. The latch can be cleared by writing to the ACKK bit in the KBSCR register. Reset also clears the latch. AWUIE bit in KBI interrupt enable register (see Figure 4-1) has no effect on AWUL reading.

The AWU oscillator and counters are inactive in normal operating mode and become active only upon entering stop mode.

4.4 Interrupts

The AWU can generate an interrupt requests:.

AWU Latch (AWUL) — The AWUL bit is set when the AWU counter overflows. The auto wakeup interrupt mask bit, AWUIE, is used to enable or disable AWU interrupt requests.

The AWU shares its interrupt with the KBI vector.

4.5 Low-Power Modes

The WAIT and STOP instructions put the MCU in low power-consumption standby modes.

4.5.1 Wait Mode

The AWU module remains inactive in wait mode.

4.5.2 Stop Mode

When the AWU module is enabled (AWUIE = 1 in the keyboard interrupt enable register) it is activated automatically upon entering stop mode. Clearing the IMASKK bit in the keyboard status and control register enables keyboard interrupt requests to bring the MCU out of stop mode. The AWU counters start from 0 each time stop mode is entered.

4.6 Registers

The AWU shares registers with the keyboard interrupt (KBI) module, the port A I/O module and configuration register 2. The following I/O registers control and monitor operation of the AWU:

- Port A data register (PTA)
- Keyboard interrupt status and control register (KBSCR)
- Keyboard interrupt enable register (KBIER)
- Configuration register 1 (CONFIG1)
- Configuration register 2 (CONFIG2)

4.6.1 Port A I/O Register

The port A data register (PTA) contains a data latch for the state of the AWU interrupt request, in addition to the data latches for port A.

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	0	AWUL	PTA5	PTA4	PTA3	PTA2	PTA1	PTA0
Write:								
Reset:	0	0	Unaffected by reset					

= Unimplemented

Figure 4-2. Port A Data Register (PTA)

AWUL — Auto Wakeup Latch

This is a read-only bit which has the value of the auto wakeup interrupt request latch. The wakeup request signal is generated internally. There is no PTA6 port or any of the associated bits such as PTA6 data direction or pullup bits.

1 = Auto wakeup interrupt request is pending
0 = Auto wakeup interrupt request is not pending

NOTE

PTA5–PTA0 bits are not used in conjuction with the auto wakeup feature. To see a description of these bits, see 12.2.1 Port A Data Register.

COPRS (In Stop Mode) — Auto Wakeup Period Selection Bit, depends on OSCSTOPEN in CONFIG2 and bus clock source (BUSCLKX2).

1 = Auto wakeup short cycle = 512 × (INTRCOSC or BUSCLKX2)
0 = Auto wakeup long cycle = 16,384 × (INTRCOSC or BUSCLKX2)

SSREC — Short Stop Recovery Bit

SSREC enables the CPU to exit stop mode with a delay of 32 BUSCLKX4 cycles instead of a 4096 BUSCLKX4 cycle delay.

1 = Stop mode recovery after 32 BUSCLKX4 cycles
0 = Stop mode recovery after 4096 BUSCLKX4 cycles

NOTE

LVISTOP, LVIRST, LVIPWRD, LVITRIP and COPD bits are not used in conjuction with the auto wakeup feature. To see a description of these bits, see Chapter 5 Configuration Register (CONFIG)

Table 7-1. Instruction Set Summary (Sheet 2 of 6)

Source Form	Operation	Description	Effect on CCR						Address Mode	Opcode	Operand	Cycles
			V	H	I	N	Z	C				
BHS *rel*	Branch if Higher or Same (Same as BCC)	PC ← (PC) + 2 + *rel* ? (C) = 0	–	–	–	–	–	–	REL	24	rr	3
BIH *rel*	Branch if $\overline{\text{IRQ}}$ Pin High	PC ← (PC) + 2 + *rel* ? $\overline{\text{IRQ}}$ = 1	–	–	–	–	–	–	REL	2F	rr	3
BIL *rel*	Branch if $\overline{\text{IRQ}}$ Pin Low	PC ← (PC) + 2 + *rel* ? $\overline{\text{IRQ}}$ = 0	–	–	–	–	–	–	REL	2E	rr	3
BIT #*opr* BIT *opr* BIT *opr* BIT *opr*,X BIT *opr*,X BIT ,X BIT *opr*,SP BIT *opr*,SP	Bit Test	(A) & (M)	0	–	–	↕	↕	–	IMM DIR EXT IX2 IX1 IX SP1 SP2	A5 B5 C5 D5 E5 F5 9EE5 9ED5	ii dd hh ll ee ff ff ff ee ff	2 3 4 4 3 2 4 5
BLE *opr*	Branch if Less Than or Equal To (Signed Operands)	PC ← (PC) + 2 + *rel* ? (Z) ∣ (N ⊕ V) = 1	–	–	–	–	–	–	REL	93	rr	3
BLO *rel*	Branch if Lower (Same as BCS)	PC ← (PC) + 2 + *rel* ? (C) = 1	–	–	–	–	–	–	REL	25	rr	3
BLS *rel*	Branch if Lower or Same	PC ← (PC) + 2 + *rel* ? (C) ∣ (Z) = 1	–	–	–	–	–	–	REL	23	rr	3
BLT *opr*	Branch if Less Than (Signed Operands)	PC ← (PC) + 2 + *rel* ? (N ⊕ V) =1	–	–	–	–	–	–	REL	91	rr	3
BMC *rel*	Branch if Interrupt Mask Clear	PC ← (PC) + 2 + *rel* ? (I) = 0	–	–	–	–	–	–	REL	2C	rr	3
BMI *rel*	Branch if Minus	PC ← (PC) + 2 + *rel* ? (N) = 1	–	–	–	–	–	–	REL	2B	rr	3
BMS *rel*	Branch if Interrupt Mask Set	PC ← (PC) + 2 + *rel* ? (I) = 1	–	–	–	–	–	–	REL	2D	rr	3
BNE *rel*	Branch if Not Equal	PC ← (PC) + 2 + *rel* ? (Z) = 0	–	–	–	–	–	–	REL	26	rr	3
BPL *rel*	Branch if Plus	PC ← (PC) + 2 + *rel* ? (N) = 0	–	–	–	–	–	–	REL	2A	rr	3
BRA *rel*	Branch Always	PC ← (PC) + 2 + *rel*	–	–	–	–	–	–	REL	20	rr	3
BRCLR *n*,*opr*,*rel*	Branch if Bit *n* in M Clear	PC ← (PC) + 3 + *rel* ? (Mn) = 0	–	–	–	–	–	↕	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	01 03 05 07 09 0B 0D 0F	dd rr dd rr dd rr dd rr dd rr dd rr dd rr dd rr	5 5 5 5 5 5 5 5
BRN *rel*	Branch Never	PC ← (PC) + 2	–	–	–	–	–	–	REL	21	rr	3
BRSET *n*,*opr*,*rel*	Branch if Bit *n* in M Set	PC ← (PC) + 3 + *rel* ? (Mn) = 1	–	–	–	–	–	↕	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	00 02 04 06 08 0A 0C 0E	dd rr dd rr dd rr dd rr dd rr dd rr dd rr dd rr	5 5 5 5 5 5 5 5
BSET *n*,*opr*	Set Bit *n* in M	Mn ← 1	–	–	–	–	–	–	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	10 12 14 16 18 1A 1C 1E	dd dd dd dd dd dd dd dd	4 4 4 4 4 4 4 4
BSR *rel*	Branch to Subroutine	PC ← (PC) + 2; push (PCL) SP ← (SP) – 1; push (PCH) SP ← (SP) – 1 PC ← (PC) + *rel*	–	–	–	–	–	–	REL	AD	rr	4
CBEQ *opr*,*rel* CBEQA #*opr*,*rel* CBEQX #*opr*,*rel* CBEQ *opr*,X+,*rel* CBEQ X+,*rel* CBEQ *opr*,SP,*rel*	Compare and Branch if Equal	PC ← (PC) + 3 + rel ? (A) – (M) = $00 PC ← (PC) + 3 + rel ? (A) – (M) = $00 PC ← (PC) + 3 + rel ? (X) – (M) = $00 PC ← (PC) + 3 + rel ? (A) – (M) = $00 PC ← (PC) + 2 + rel ? (A) – (M) = $00 PC ← (PC) + 4 + rel ? (A) – (M) = $00	–	–	–	–	–	–	DIR IMM IMM IX1+ IX+ SP1	31 41 51 61 71 9E61	dd rr ii rr ii rr ff rr rr ff rr	5 4 4 5 4 6
CLC	Clear Carry Bit	C ← 0	–	–	–	–	–	0	INH	98		1
CLI	Clear Interrupt Mask	I ← 0	–	–	0	–	–	–	INH	9A		2

11.3.4 XTAL Oscillator

The XTAL oscillator circuit is designed for use with an external crystal or ceramic resonator to provide an accurate clock source. In this configuration, the OSC2 pin is dedicated to the external crystal circuit. The OSC2EN bit has no effect when this clock mode is selected.

In its typical configuration, the XTAL oscillator is connected in a Pierce oscillator configuration, as shown in Figure 11-2. This figure shows only the logical representation of the internal components and may not represent actual circuitry.

The oscillator configuration uses five components:

- Crystal, X_1
- Fixed capacitor, C_1
- Tuning capacitor, C_2 (can also be a fixed capacitor)
- Feedback resistor, R_B
- Series resistor, R_S (optional)

NOTE

The series resistor (R_S) is included in the diagram to follow strict Pierce oscillator guidelines and may not be required for all ranges of operation, especially with high frequency crystals. Refer to the oscillator characteristics table in the Electricals section for more information.

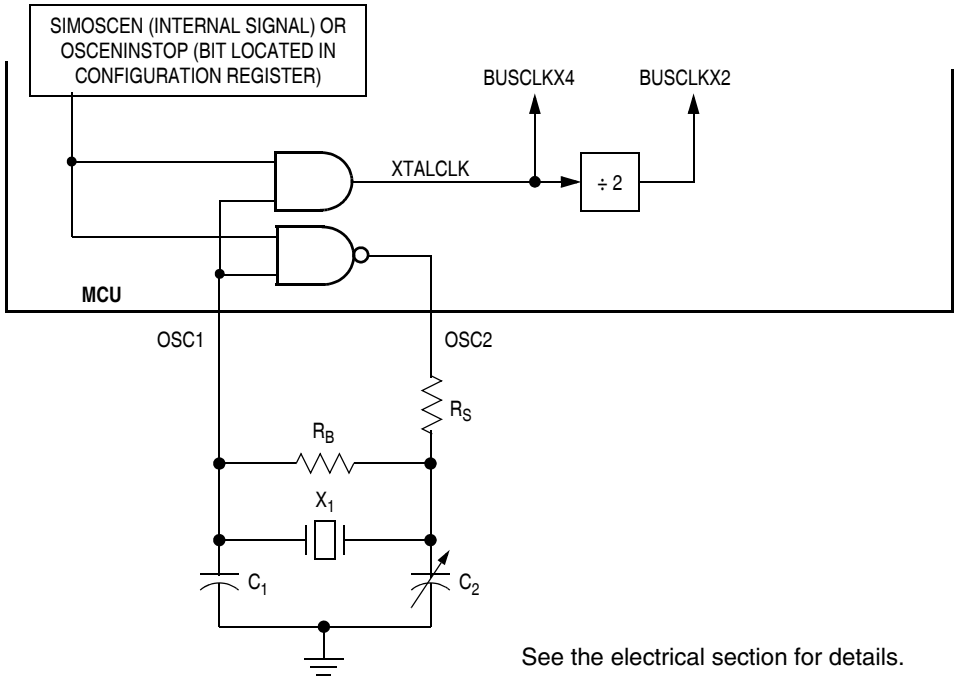

Figure 11-2. XTAL Oscillator External Connections

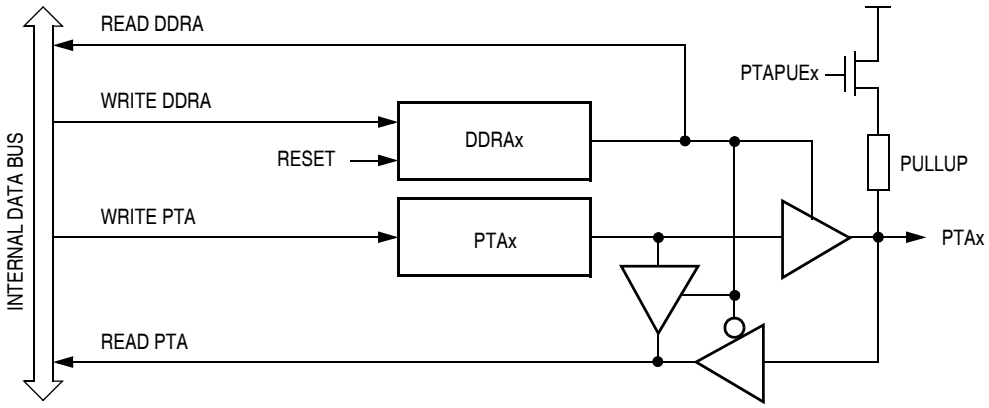

Figure 12-3. Port A I/O Circuit

NOTE

Figure 12-3 does not apply to PTA2

When DDRAx is a 1, reading PTA reads the PTAx data latch. When DDRAx is a 0, reading PTA reads the logic level on the PTAx pin. The data latch can always be written, regardless of the state of its data direction bit.

12.2.3 Port A Input Pullup Enable Register

The port A input pullup enable register (PTAPUE) contains a software configurable pullup device for each of the port A pins. Each bit is individually configurable and requires the corresponding data direction register, DDRAx, to be configured as input. Each pullup device is automatically and dynamically disabled when its corresponding DDRAx bit is configured as output.

	Bit 7	6	5	4	3	2	1	Bit 0
Read: Write:	OSC2EN		PTAPUE5	PTAPUE4	PTAPUE3	PTAPUE2	PTAPUE1	PTAPUE0
Reset:	0	0	0	0	0	0	0	0

= Unimplemented

Figure 12-4. Port A Input Pullup Enable Register (PTAPUE)

OSC2EN — Enable PTA4 on OSC2 Pin

This read/write bit configures the OSC2 pin function when internal oscillator or RC oscillator option is selected. This bit has no effect for the XTAL or external oscillator options.

1 = OSC2 pin outputs the internal or RC oscillator clock (BUSCLKX4)

0 = OSC2 pin configured for PTA4 I/O, having all the interrupt and pullup functions

PTAPUE[5:0] — Port A Input Pullup Enable Bits

These read/write bits are software programmable to enable pullup devices on port A pins.

1 = Corresponding port A pin configured to have internal pullup if its DDRA bit is set to 0

0 = Pullup device is disconnected on the corresponding port A pin regardless of the state of its DDRA bit

12.3.3 Port B Input Pullup Enable Register

The port B input pullup enable register (PTBPUE) contains a software configurable pullup device for each of the eight port B pins. Each bit is individually configurable and requires the corresponding data direction register, DDRBx, be configured as input. Each pullup device is automatically and dynamically disabled when its corresponding DDRBx bit is configured as output.

	Bit 7	6	5	4	3	2	1	Bit 0
Read: Write:	PTBPUE7	PTBPUE6	PTBPUE5	PTBPUE4	PTBPUE3	PTBPUE2	PTBPUE2	PTBPUE0
Reset:	0	0	0	0	0	0	0	0

Figure 12-8. Port B Input Pullup Enable Register (PTBPUE)

PTBPUE[7:0] — Port B Input Pullup Enable Bits

These read/write bits are software programmable to enable pullup devices on port B pins

1 = Corresponding port B pin configured to have internal pull if its DDRB bit is set to 0

0 = Pullup device is disconnected on the corresponding port B pin regardless of the state of its DDRB bit.

12.3.4 Port B Summary Table

Table 12-2 summarizes the operation of the port A pins when used as a general-purpose input/output pins.

Table 12-2. Port B Pin Functions

DDRB Bit	PTB Bit	I/O Pin Mode	Accesses to DDRB	Accesses to PTB	
			Read/Write	Read	Write
0	X(1)	Input, Hi-Z(2)	DDRB7–DDRB0	Pin	PTB7–PTB0(3)
1	X	Output	DDRB7–DDRB0	Pin	PTB7–PTB0

1. X = don't care
2. Hi-Z = high impedance
3. Writing affects data register, but does not affect the input.

Slow Data Tolerance

Figure 13-7 shows how much a slow received character can be misaligned without causing a noise error or a framing error. The slow stop bit begins at RT8 instead of RT1 but arrives in time for the stop bit data samples at RT8, RT9, and RT10.

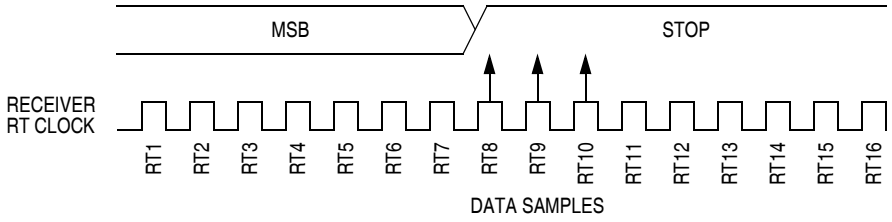

Figure 13-7. Slow Data

For an 8-bit character, data sampling of the stop bit takes the receiver 9 bit times × 16 RT cycles + 10 RT cycles = 154 RT cycles.

With the misaligned character shown in Figure 13-7, the receiver counts 154 RT cycles at the point when the count of the transmitting device is 9 bit times × 16 RT cycles + 3 RT cycles = 147 RT cycles.

The maximum percent difference between the receiver count and the transmitter count of a slow 8-bit character with no errors is:

$$\left|\frac{154 - 147}{154}\right| \times 100 = 4.54\%$$

For a 9-bit character, data sampling of the stop bit takes the receiver 10 bit times × 16 RT cycles + 10 RT cycles = 170 RT cycles.

With the misaligned character shown in Figure 13-7, the receiver counts 170 RT cycles at the point when the count of the transmitting device is 10 bit times × 16 RT cycles + 3 RT cycles = 163 RT cycles.

The maximum percent difference between the receiver count and the transmitter count of a slow 9-bit character with no errors is:

$$\left|\frac{170 - 163}{170}\right| \times 100 = 4.12\%$$

Fast Data Tolerance

Figure 13-8 shows how much a fast received character can be misaligned without causing a noise error or a framing error. The fast stop bit ends at RT10 instead of RT16 but is still there for the stop bit data samples at RT8, RT9, and RT10.

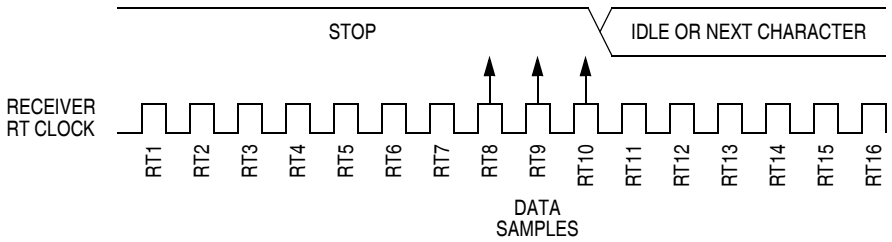

Figure 13-8. Fast Data

15.3.5 Resetting the SPI

Any system reset completely resets the SPI. Partial resets occur whenever the SPI enable bit (SPE) is 0. Whenever SPE is 0, the following occurs:

- The SPTE flag is set.
- Any transmission currently in progress is aborted.
- The shift register is cleared.
- The SPI state counter is cleared, making it ready for a new complete transmission.
- All the SPI pins revert back to being general-purpose I/O.

These items are reset only by a system reset:

- All control bits in the SPCR register
- All control bits in the SPSCR register (MODFEN, ERRIE, SPR1, and SPR0)
- The status flags SPRF, OVRF, and MODF

By not resetting the control bits when SPE is low, the user can clear SPE between transmissions without having to set all control bits again when SPE is set high for the next transmission.

By not resetting the SPRF, OVRF, and MODF flags, the user can still service these interrupts after the SPI has been disabled. The user can disable the SPI by writing 0 to the SPE bit. The SPI can also be disabled by a mode fault occurring in an SPI that was configured as a master with the MODFEN bit set.

15.3.6 Error Conditions

The following flags signal SPI error conditions:

- Overflow (OVRF) — Failing to read the SPI data register before the next full byte enters the shift register sets the OVRF bit. The new byte does not transfer to the receive data register, and the unread byte still can be read. OVRF is in the SPI status and control register.
- Mode fault error (MODF) — The MODF bit indicates that the voltage on the slave select pin ($\overline{SS}$) is inconsistent with the mode of the SPI. MODF is in the SPI status and control register.

15.3.6.1 Overflow Error

The overflow flag (OVRF) becomes set if the receive data register still has unread data from a previous transmission when the capture strobe of bit 1 of the next transmission occurs. The bit 1 capture strobe occurs in the middle of SPSCK cycle 7 (see Figure 15-4 and Figure 15-6.) If an overflow occurs, all data received after the overflow and before the OVRF bit is cleared does not transfer to the receive data register and does not set the SPI receiver full bit (SPRF). The unread data that transferred to the receive data register before the overflow occurred can still be read. Therefore, an overflow error always indicates the loss of data. Clear the overflow flag by reading the SPI status and control register and then reading the SPI data register.

OVRF generates a receiver/error interrupt request if the error interrupt enable bit (ERRIE) is also set. The SPRF, MODF, and OVRF interrupts share the same interrupt vector (see Figure 15-11.) It is not possible to enable MODF or OVRF individually to generate a receiver/error interrupt request. However, leaving MODFEN low prevents MODF from being set.

If the SPRF interrupt is enabled and the OVRF interrupt is not, watch for an overflow condition. Figure 15-9 shows how it is possible to miss an overflow. The first part of Figure 15-9 shows how it is possible to read the SPSCR and SPDR to clear the SPRF without problems. However, as illustrated by the second transmission example, the OVRF bit can be set in between the time that SPSCR and SPDR are read.

MODF — Mode Fault Bit

This clearable, read-only flag is set in a slave SPI if the $\overline{SS}$ pin goes high during a transmission with MODFEN set. In a master SPI, the MODF flag is set if the $\overline{SS}$ pin goes low at any time with the MODFEN bit set. Clear MODF by reading the SPI status and control register (SPSCR) with MODF set and then writing to the SPI control register (SPCR).

1 = $\overline{SS}$ pin at inappropriate logic level
0 = $\overline{SS}$ pin at appropriate logic level

SPTE — SPI Transmitter Empty Bit

This clearable, read-only flag is set each time the transmit data register transfers a byte into the shift register. SPTE generates an SPTE interrupt request if the SPTIE bit in the SPI control register is also set.

NOTE

Do not write to the SPI data register unless SPTE is high.

During an SPTE interrupt, user software can clear SPTE by writing to the transmit data register.

1 = Transmit data register empty
0 = Transmit data register not empty

MODFEN — Mode Fault Enable Bit

This read/write bit, when set, allows the MODF flag to be set. If the MODF flag is set, clearing MODFEN does not clear the MODF flag. If the SPI is enabled as a master and the MODFEN bit is 0, then the $\overline{SS}$ pin is available as a general-purpose I/O.

If the MODFEN bit is 1, then this pin is not available as a general-purpose I/O. When the SPI is enabled as a slave, the $\overline{SS}$ pin is not available as a general-purpose I/O regardless of the value of MODFEN. See 15.7.4 SS (Slave Select).

If the MODFEN bit is 0, the level of the $\overline{SS}$ pin does not affect the operation of an enabled SPI configured as a master. For an enabled SPI configured as a slave, having MODFEN low only prevents the MODF flag from being set. It does not affect any other part of SPI operation. See 15.3.6.2 Mode Fault Error.

SPR1 and SPR0 — SPI Baud Rate Select Bits

In master mode, these read/write bits select one of four baud rates as shown in Table 15-3. SPR1 and SPR0 have no effect in slave mode.

Table 15-3. SPI Master Baud Rate Selection

SPR1 and SPR0	Baud Rate Divisor (BD)
00	2
01	8
10	32
11	128

Use this formula to calculate the SPI baud rate:

$$\text{Baud rate} = \frac{\text{BUSCLK}}{\text{BD}}$$

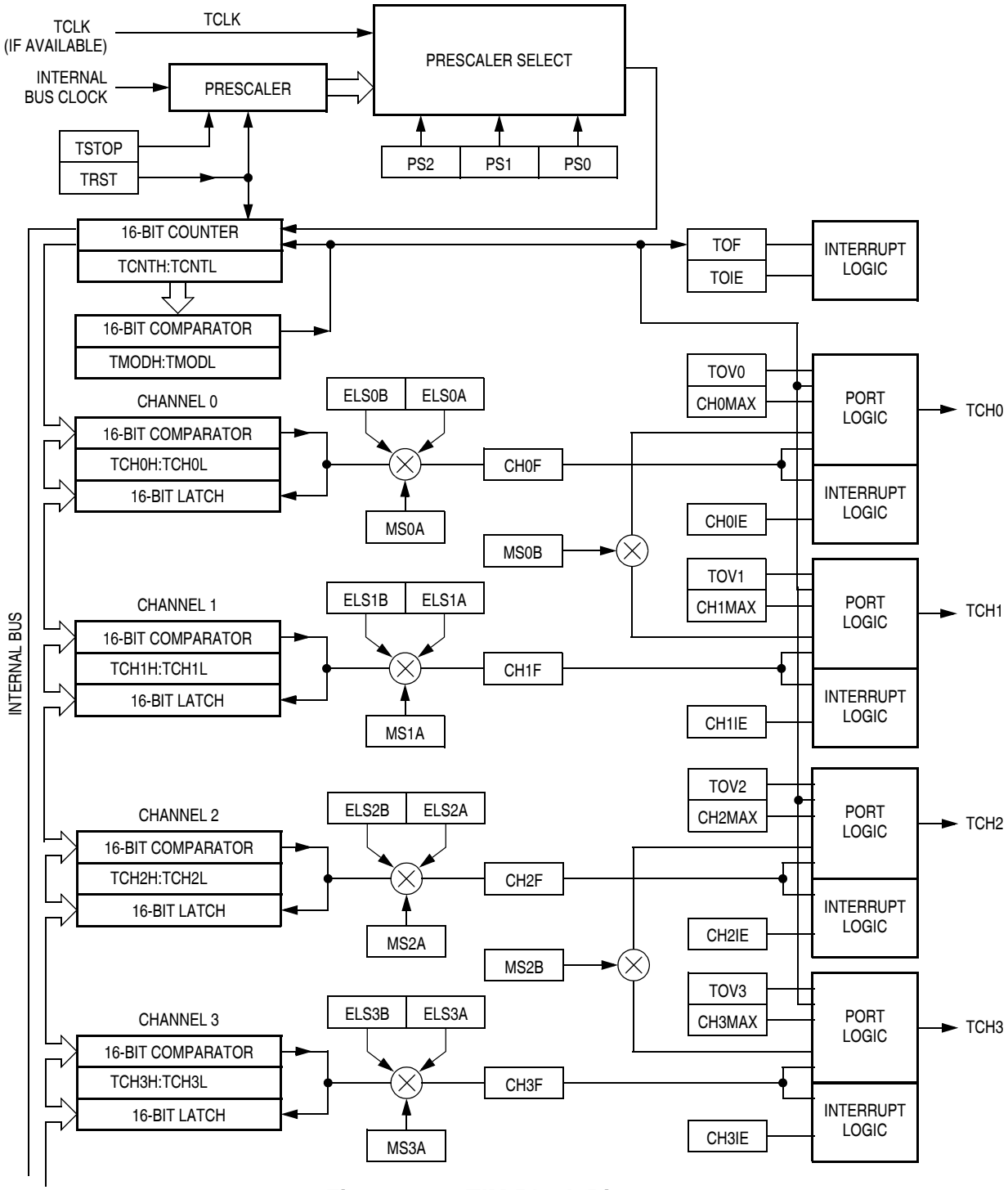

Figure 16-2. TIM Block Diagram

16.3.3 Output Compare

With the output compare function, the TIM can generate a periodic pulse with a programmable polarity, duration, and frequency. When the counter reaches the value in the registers of an output compare

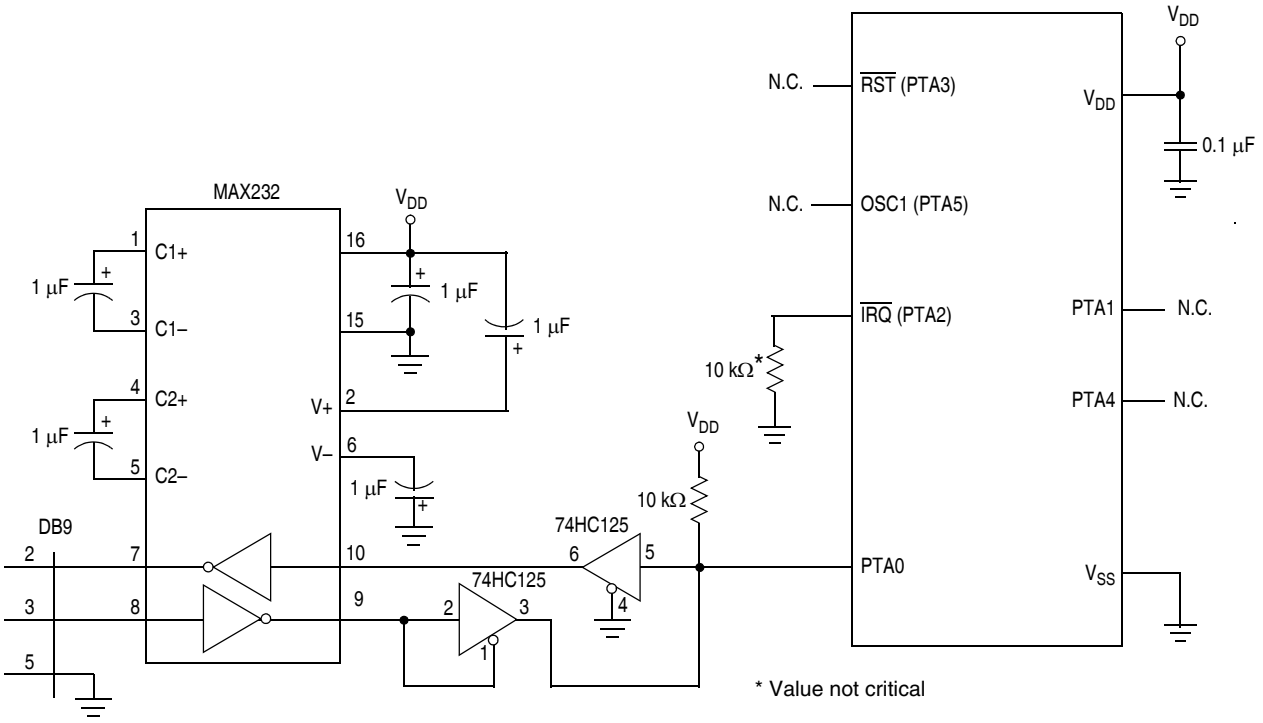

Figure 17-12. Monitor Mode Circuit (Internal Clock, No High Voltage)

The monitor code has been updated from previous versions of the monitor code to allow enabling the internal oscillator to generate the internal clock. This addition, which is enabled when $\overline{IRQ}$ is held low out of reset, is intended to support serial communication/programming at 9600 baud in monitor mode by using the internal oscillator, and the internal oscillator user trim value OSCTRIM (FLASH location $FFC0, if programmed) to generate the desired internal frequency (3.2 MHz). Since this feature is enabled only when $\overline{IRQ}$ is held low out of reset, it cannot be used when the reset vector is programmed (i.e., the value is not $FFFF) because entry into monitor mode in this case requires V_{TST} on $\overline{IRQ}$. The $\overline{IRQ}$ pin must remain low during this monitor session in order to maintain communication.

Table 17-1 shows the pin conditions for entering monitor mode. As specified in the table, monitor mode may be entered after a power-on reset (POR) and will allow communication at 9600 baud provided one of the following sets of conditions is met:

- If $FFFE and $FFFF do not contain $FF (programmed state):
  - The external clock is 9.8304 MHz
  - $\overline{IRQ}$ = V_{TST}
- If $FFFE and $FFFF contain $FF (erased state):
  - The external clock is 9.8304 MHz
  - $\overline{IRQ}$ = V_{DD} (this can be implemented through the internal $\overline{IRQ}$ pullup)
- If $FFFE and $FFFF contain $FF (erased state):
  - $\overline{IRQ}$ = V_{SS} (internal oscillator is selected, no external clock required)

The rising edge of the internal $\overline{RST}$ signal latches the monitor mode. Once monitor mode is latched, the values on PTA1 and PTA4 pins can be changed.

Once out of reset, the MCU waits for the host to send eight security bytes (see 17.3.2 Security). After the security bytes, the MCU sends a break signal (10 consecutive 0s) to the host, indicating that it is ready to receive a command.

17.3.1.2 Forced Monitor Mode

If entering monitor mode without high voltage on $\overline{IRQ}$, then startup port pin requirements and conditions, (PTA1/PTA4) are not in effect. This is to reduce circuit requirements when performing in-circuit programming.

NOTE

If the reset vector is blank and monitor mode is entered, the chip will see an additional reset cycle after the initial power-on reset (POR). Once the reset vector has been programmed, the traditional method of applying a voltage, V_{TST}, to $\overline{IRQ}$ must be used to enter monitor mode.

If monitor mode was entered as a result of the reset vector being blank, the COP is always disabled regardless of the state of $\overline{IRQ}$.

If the voltage applied to the $\overline{IRQ}$ is less than V_{TST}, the MCU will come out of reset in user mode. Internal circuitry monitors the reset vector fetches and will assert an internal reset if it detects that the reset vectors are erased ($FF). When the MCU comes out of reset, it is forced into monitor mode without requiring high voltage on the $\overline{IRQ}$ pin. Once out of reset, the monitor code is initially executing with the internal clock at its default frequency.

If $\overline{IRQ}$ is held high, all pins will default to regular input port functions except for PTA0 and PTA5 which will operate as a serial communication port and OSC1 input respectively (refer to Figure 17-11). That will allow the clock to be driven from an external source through OSC1 pin.

If $\overline{IRQ}$ is held low, all pins will default to regular input port function except for PTA0 which will operate as serial communication port. Refer to Figure 17-12.

Regardless of the state of the $\overline{IRQ}$ pin, it will not function as a port input pin in monitor mode. Bit 2 of the Port A data register will always read 0. The BIH and BIL instructions will behave as if the $\overline{IRQ}$ pin is enabled, regardless of the settings in the configuration register. See Chapter 5 Configuration Register (CONFIG).

The COP module is disabled in forced monitor mode. Any reset other than a power-on reset (POR) will automatically force the MCU to come back to the forced monitor mode.

17.3.1.3 Monitor Vectors

In monitor mode, the MCU uses different vectors for reset, SWI (software interrupt), and break interrupt than those for user mode. The alternate vectors are in the $FE page instead of the $FF page and allow code execution from the internal monitor firmware instead of user code.

NOTE

Exiting monitor mode after it has been initiated by having a blank reset vector requires a power-on reset (POR). Pulling $\overline{RST}$ (when $\overline{RST}$ pin available) low will not exit monitor mode in this situation.

Table 17-2 summarizes the differences between user mode and monitor mode regarding vectors.

Table 17-2. Mode Difference

Modes	Functions					
	Reset Vector High	Reset Vector Low	Break Vector High	Break Vector Low	SWI Vector High	SWI Vector Low
User	$FFFE	$FFFF	$FFFC	$FFFD	$FFFC	$FFFD
Monitor	$FEFE	$FEFF	$FEFC	$FEFD	$FEFC	$FEFD

Characteristic	Conditions	Symbol	Min	Typ(1)	Max	Unit	Comment
Integral non-linearity	10-bit mode	INL	0	±0.5	—	LSB	
	8-bit mode		0	±0.3	—		
Zero-scale error	10-bit mode	E_{ZS}	0	±0.5	—	LSB	$V_{ADIN} = V_{SS}$
	8-bit mode		0	±0.3	—		
Full-scale error	10-bit mode	E_{FS}	0	±0.5	—	LSB	$V_{ADIN} = V_{DD}$
	8-bit mode		0	±0.3	—		
Quantization error	10-bit mode	E_Q	—	—	±0.5	LSB	8-bit mode is not truncated
	8-bit mode		—	—	±0.5		
Input leakage error	10-bit mode	E_{IL}	0	±0.2	±5	LSB	Pad leakage(5) * R_{AS}
	8-bit mode		0	±0.1	±1.2		
Bandgap voltage (3)(6)		V_{BG}	1.17	1.245	1.32	V	

1. Typical values assume V_{DD} = 5.0 V, temperature = 25•C, f_{ADCK} = 1.0 MHz unless otherwise stated. Typical values are for reference only and are not tested in production.
2. Incremental I_{DD} added to MCU mode current.
3. Values are based on characterization results, not tested in production.
4. Reference the ADC module specification for more information on calculating conversion times.
5. Based on typical input pad leakage current.
6. LVI must be enabled, (LVIPWRD = 0, in CONFIG1). Voltage input to ADCH4:0 = $1A, an ADC conversion on this channel allows user to determine supply voltage.

18.16 Timer Interface Module Characteristics

Characteristic	Symbol	Min	Max	Unit
Timer input capture pulse width(1)	t_{TH}, t_{TL}	2	—	t_{cyc}
Timer input capture period	t_{TLTL}	Note(2)	—	t_{cyc}
Timer input clock pulse width(1)	t_{TCL}, t_{TCH}	t_{cyc} + 5	—	ns

1. Values are based on characterization results, not tested in production.
2. The minimum period is the number of cycles it takes to execute the interrupt service routine plus 1 t_{cyc}.

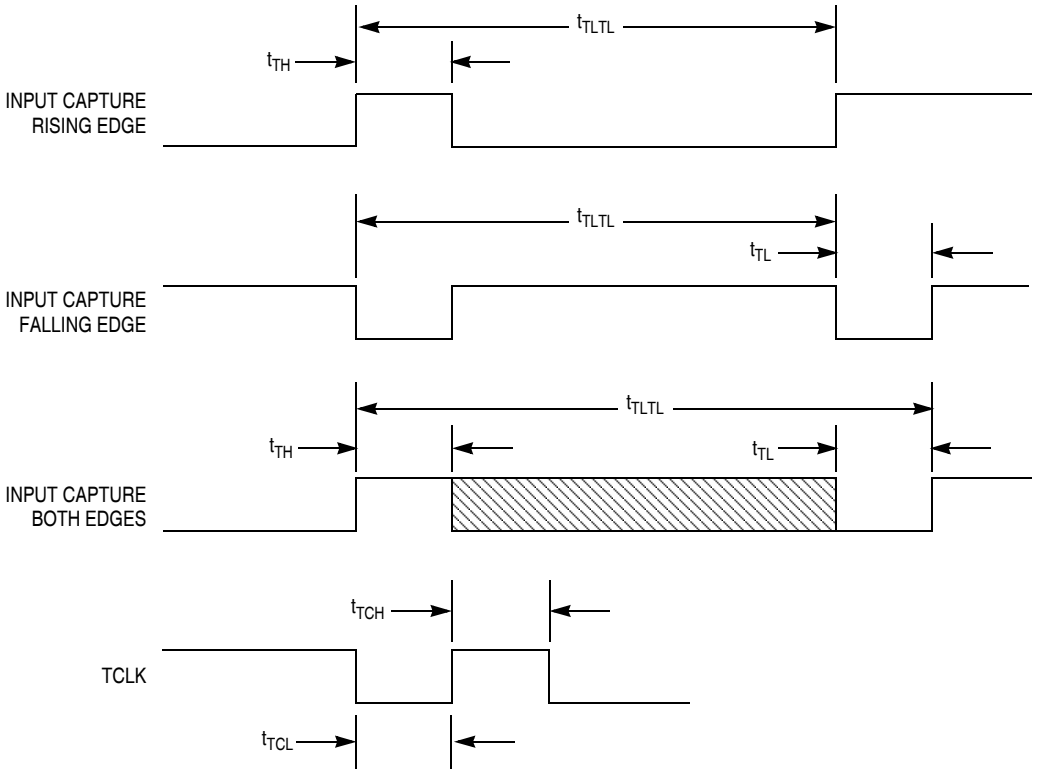

Figure 18-13. Timer Input Timing