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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	e200z0h
Core Size	32-Bit Single-Core
Speed	64MHz
Connectivity	CANbus, LINbus, SPI, UART/USART
Peripherals	DMA, POR, PWM, WDT
Number of I/O	37
Program Memory Size	256KB (256K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	20K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 3.6V
Data Converters	A/D 16x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LQFP
Supplier Device Package	64-LQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/spc560p40l1befby

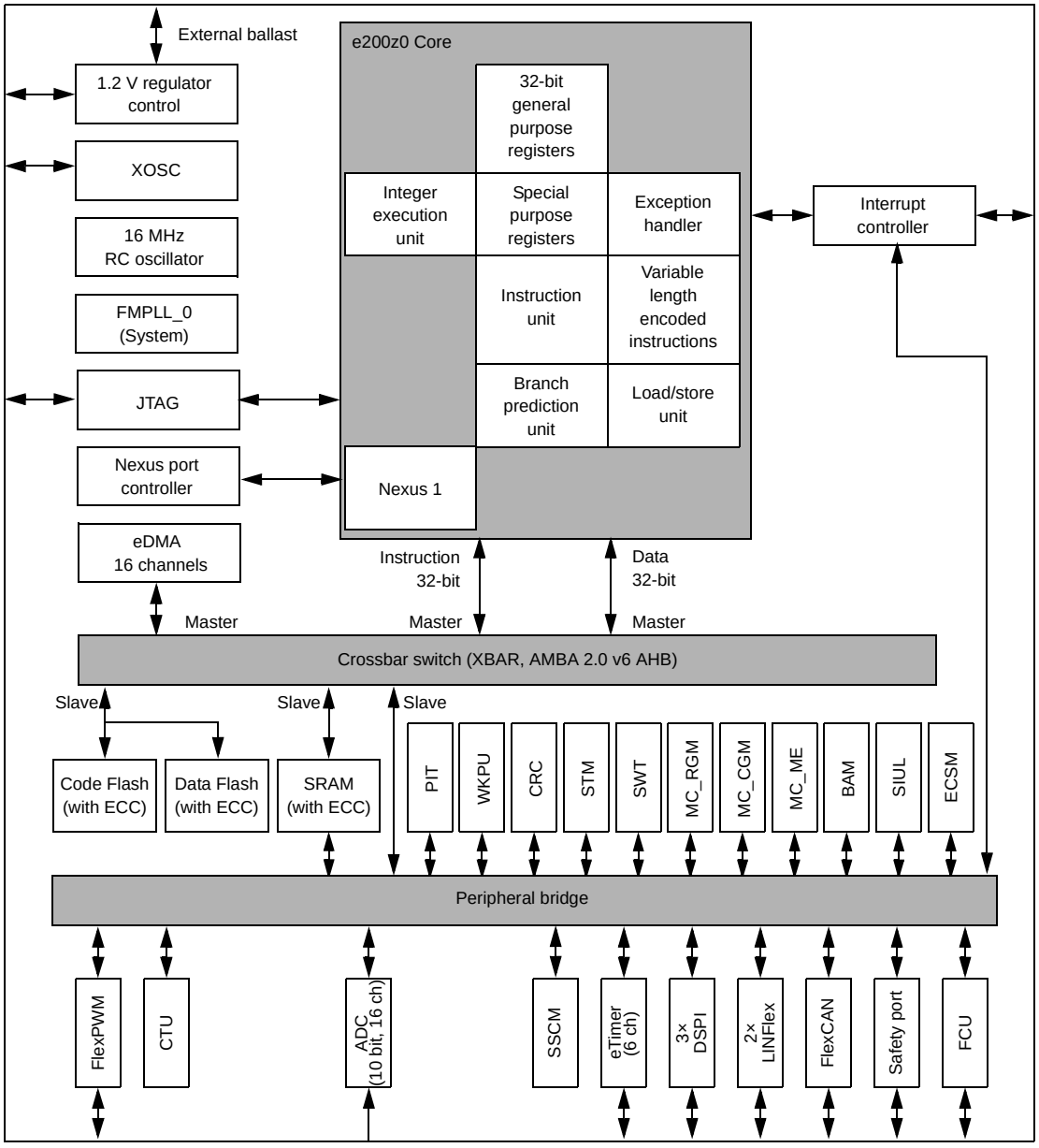
Table 2. SPC560P34/SPC560P40 device comparison (continued)

Feature		SPC560P34 Full-featured	SPC560P40 Full-featured
eDMA (enhanced direct memory access) channels		16	
FlexCAN (controller area network)		1(1)	2(1),(2)
Safety port		No	Yes (via second FlexCAN module)
FCU (fault collection unit)		Yes	
CTU (cross triggering unit)		Yes	Yes
eTimer		1 (16-bit, 6 channels)	
FlexPWM (pulse-width modulation) channels		8 (capture capabity not supported)	8 (capture capability not supported)
Analog-to-digital converter (ADC)		1 (10-bit, 16 channels)	
LINFlex		2 (1 × Master/Slave, 1 × Master only)	2 (1 × Master/Slave, 1 × Master only)
DSPI (deserial serial peripheral interface)		2	3
CRC (cyclic redundancy check) unit		Yes	
Junction temperature sensor		No	
JTAG controller		Yes	
Nexus port controller (NPC)		Yes (Nexus Class 1)	
Supply	Digital power supply(3)	3.3 V or 5 V single supply with external transistor	
	Analog power supply	3.3 V or 5 V	
	Internal RC oscillator	16 MHz	
	External crystal oscillator	4–40 MHz	
Packages		LQFP64 LQFP100	
Temperature	Standard ambient temperature	–40 to 125 °C	

1. Each FlexCAN module has 32 message buffers.
2. One FlexCAN module can act as a safety port with a bit rate as high as 8 Mbit/s at 64 MHz.
3. The different supply voltages vary according to the part number ordered.

SPC560P34/SPC560P40 is available in two configurations having different features: Full-featured and airbag. *Table 3* shows the main differences between the two versions of the SPC560P40 MCU.

Figure 1. Block diagram (SPC560P40 full-featured configuration)



Legend:

ADC	Analog-to-digital converter
BAM	Boot assist module
CRC	Cyclic redundancy check
CTU	Cross triggering unit
DSPI	Deserial serial peripheral interface
ECSM	Error correction status module
eDMA	Enhanced direct memory access
eTimer	Enhanced timer
FCU	Fault collection unit
Flash	Flash memory
FlexCAN	Controller area network
FlexPWM	Flexible pulse width modulation
FMPLL	Frequency-modulated phase-locked loop
INTC	Interrupt controller
JTAG	JTAG controller
LINFlex	Serial communication interface (LIN support)
MC_CGM	Clock generation module
MC_ME	Mode entry module
MC_PCU	Power control unit
MC_RGM	Reset generation module
PIT	Periodic interrupt timer
SIUL	System Integration unit Lite
SRAM	Static random-access memory
SSCM	System status and configuration module
STM	System timer module
SWT	Software watchdog timer
WKPU	Wakeup unit
XOSC	External oscillator
XBAR	Crossbar switch

Table 4. SPC560P34/SPC560P40 series block summary

Block	Function
Analog-to-digital converter (ADC)	Multi-channel, 10-bit analog-to-digital converter
Boot assist module (BAM)	Block of read-only memory containing VLE code which is executed according to the boot mode of the device
Clock generation module (MC_CGM)	Provides logic and control required for the generation of system and peripheral clocks
Controller area network (FlexCAN)	Supports the standard CAN communications protocol
Cross triggering unit (CTU)	Enables synchronization of ADC conversions with a timer event from the eMIOS or from the PIT
Crossbar switch (XBAR)	Supports simultaneous connections between two master ports and three slave ports; supports a 32-bit address bus width and a 32-bit data bus width
Cyclic redundancy check (CRC)	CRC checksum generator
Deserial serial peripheral interface (DSPI)	Provides a synchronous serial interface for communication with external devices
Enhanced direct memory access (eDMA)	Performs complex data transfers with minimal intervention from a host processor via "*n*" programmable channels
Enhanced timer (eTimer)	Provides enhanced programmable up/down modulo counting
Error correction status module (ECSM)	Provides a myriad of miscellaneous control functions for the device including program-visible information about configuration and revision levels, a reset status register, wakeup control for exiting sleep modes, and optional features such as information on memory errors reported by error-correcting codes
External oscillator (XOSC)	Provides an output clock used as input reference for FMPLL_0 or as reference clock for specific modules depending on system needs
Fault collection unit (FCU)	Provides functional safety to the device
Flash memory	Provides non-volatile storage for program code, constants and variables
Frequency-modulated phase-locked loop (FMPLL)	Generates high-speed system clocks and supports programmable frequency modulation
Interrupt controller (INTC)	Provides priority-based preemptive scheduling of interrupt requests
JTAG controller	Provides the means to test chip functionality and connectivity while remaining transparent to system logic when not in test mode
LINFlex controller	Manages a high number of LIN (Local Interconnect Network protocol) messages efficiently with a minimum of CPU load
Mode entry module (MC_ME)	Provides a mechanism for controlling the device operational mode and mode transition sequences in all functional states; also manages the power control unit, reset generation module and clock generation module, and holds the configuration, control and status registers accessible for applications
Periodic interrupt timer (PIT)	Produces periodic interrupts and triggers
Peripheral bridge (PBRIDGE)	Is the interface between the system bus and on-chip peripherals
Power control unit (MC_PCU)	Reduces the overall power consumption by disconnecting parts of the device from the power supply via a power switching device; device components are grouped into sections called "power domains" which are controlled by the PCU

The crossbar provides the following features:

- 3 master ports:
  - e200z0 core complex instruction port
  - e200z0 core complex Load/Store Data port
  - eDMA
- 3 slave ports:
  - Flash memory (Code and Data)
  - SRAM
  - Peripheral bridge
- 32-bit internal address, 32-bit internal data paths
- Fixed Priority Arbitration based on Port Master
- Temporary dynamic priority elevation of masters

1.5.3 Enhanced direct memory access (eDMA)

The enhanced direct memory access (eDMA) controller is a second-generation module capable of performing complex data movements via 16 programmable channels, with minimal intervention from the host processor. The hardware micro architecture includes a DMA engine which performs source and destination address calculations, and the actual data movement operations, along with an SRAM-based memory containing the transfer control descriptors (TCD) for the channels.

The eDMA module provides the following features:

- 16 channels support independent 8-, 16- or 32-bit single value or block transfers
- Supports variable-sized queues and circular queues
- Source and destination address registers are independently configured to either post-increment or to remain constant
- Each transfer is initiated by a peripheral, CPU, or eDMA channel request
- Each eDMA channel can optionally send an interrupt request to the CPU on completion of a single value or block transfer
- DMA transfers possible between system memories, DSPIs, ADC, FlexPWM, eTimer and CTU
- Programmable DMA channel multiplexer allows assignment of any DMA source to any available DMA channel with as many as 30 request sources
- eDMA abort operation through software

1.5.4 Flash memory

The SPC560P34/SPC560P40 provides 320 KB of programmable, non-volatile, flash memory. The non-volatile memory (NVM) can be used for instruction and/or data storage. The flash memory module is interfaced to the system bus by a dedicated flash memory controller. It supports a 32-bit data bus width at the system bus port, and a 128-bit read data interface to flash memory. The module contains four 128-bit wide prefetch buffers. Prefetch buffer hits allow no-wait responses. Normal flash memory array accesses are registered and are forwarded to the system bus on the following cycle, incurring two wait-states.

1.5.8 System clocks and clock generation

The following list summarizes the system clock and clock generation on the SPC560P34/SPC560P40:

- Lock detect circuitry continuously monitors lock status
- Loss of clock (LOC) detection for PLL outputs
- Programmable output clock divider (÷1, ÷2, ÷4, ÷8)
- FlexPWM module and eTimer module running at the same frequency as the e200z0h core
- Internal 16 MHz RC oscillator for rapid start-up and safe mode: supports frequency trimming by user application

1.5.9 Frequency-modulated phase-locked loop (FMPLL)

The FMPLL allows the user to generate high speed system clocks from a 4–40 MHz input clock. Further, the FMPLL supports programmable frequency modulation of the system clock. The PLL multiplication factor, output clock divider ratio are all software configurable.

The FMPLL has the following major features:

- Input clock frequency: 4–40 MHz
- Maximum output frequency: 64 MHz
- Voltage controlled oscillator (VCO)—frequency 256–512 MHz
- Reduced frequency divider (RFD) for reduced frequency operation without forcing the FMPLL to relock
- Frequency-modulated PLL
  - Modulation enabled/disabled through software
  - Triangle wave modulation
- Programmable modulation depth (±0.25% to ±4% deviation from center frequency): programmable modulation frequency dependent on reference frequency
- Self-clocked mode (SCM) operation

1.5.10 Main oscillator

The main oscillator provides these features:

- Input frequency range: 4–40 MHz
- Crystal input mode or oscillator input mode
- PLL reference

1.5.11 Internal RC oscillator

This device has an RC ladder phase-shift oscillator. The architecture uses constant current charging of a capacitor. The voltage at the capacitor is compared by the stable bandgap reference voltage.

1.5.23 Deserial serial peripheral interface (DSPI)

The deserial serial peripheral interface (DSPI) module provides a synchronous serial interface for communication between the SPC560P34/SPC560P40 MCU and external devices.

The DSPI modules provide these features:

- Full duplex, synchronous transfers
- Master or slave operation
- Programmable master bit rates
- Programmable clock polarity and phase
- End-of-transmission interrupt flag
- Programmable transfer baud rate
- Programmable data frames from 4 to 16 bits
- Up to 8 chip select lines available:
  - 8 on DSPI_0
  - 4 each on DSPI_1 and DSPI_2
- 8 clock and transfer attributes registers
- Chip select strobe available as alternate function on one of the chip select pins for deglitching
- FIFOs for buffering up to 4 transfers on the transmit and receive side
- Queueing operation possible through use of the I/O processor or eDMA
- General purpose I/O functionality on pins when not used for SPI

1.5.24 Pulse width modulator (FlexPWM)

The pulse width modulator module (PWM) contains four PWM submodules each of which is set up to control a single half-bridge power stage. There are also three fault channels.

This PWM is capable of controlling most motor types: AC induction motors (ACIM), permanent magnet AC motors (PMAC), both brushless (BLDC) and brush DC motors (BDC), switched (SRM) and variable reluctance motors (VRM), and stepper motors.

2 Package pinouts and signal descriptions

2.1 Package pinouts

The LQFP pinouts are shown in the following figures. For pin signal descriptions, please refer to *Table 7*.

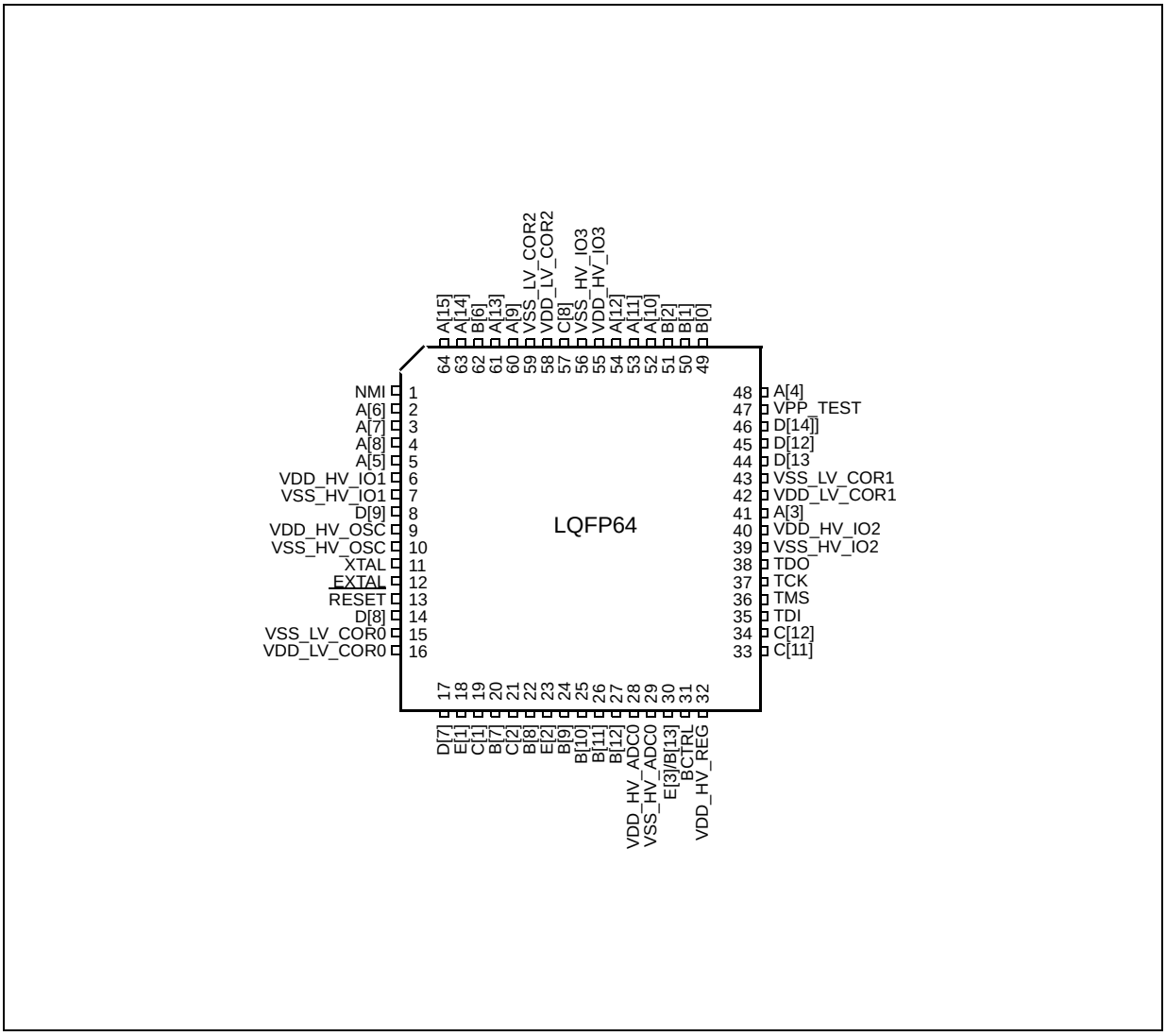


Figure 2. 64-pin LQFP pinout – Full featured configuration (top view)

Table 7. Pin muxing (continued)

Port pin	PCR register	Alternate function[1],[2]	Functions	Peripheral[3]	I/O direction[4]	Pad speed[5]		Pin	
						SRC = 0	SRC = 1	64-pin	100-pin
D[1]	PCR[49]	ALT0 ALT1 ALT2 ALT3	GPIO[49] — — EXT_TRG	SIUL — — CTU_0	I/O — — O	Slow	Medium	—	3
D[2]	PCR[50]	ALT0 ALT1 ALT2 ALT3	GPIO[50] — — X[3]	SIUL — — FlexPWM_0	I/O — — O	Slow	Medium	—	97
D[3]	PCR[51]	ALT0 ALT1 ALT2 ALT3	GPIO[51] — — A[3]	SIUL — — FlexPWM_0	I/O — — O	Slow	Medium	—	89
D[4]	PCR[52]	ALT0 ALT1 ALT2 ALT3	GPIO[52] — — B[3]	SIUL — — FlexPWM_0	I/O — — O	Slow	Medium	—	90
D[5]	PCR[53]	ALT0 ALT1 ALT2 ALT3	GPIO[53] CS3 F[0] —	SIUL DSPI_0 FCU_0 —	I/O O O —	Slow	Medium	—	22
D[6]	PCR[54]	ALT0 ALT1 ALT2 ALT3 —	GPIO[54] CS2 — — FAULT[1]	SIUL DSPI_0 — — FlexPWM_0	I/O O — — I	Slow	Medium	—	23
D[7]	PCR[55]	ALT0 ALT1 ALT2 ALT3	GPIO[55] CS3 F[1] CS4	SIUL DSPI_1 FCU_0 DSPI_0	I/O O O O	Slow	Medium	17	26
D[8]	PCR[56]	ALT0 ALT1 ALT2 ALT3	GPIO[56] CS2 — CS5	SIUL DSPI_1 — DSPI_0	I/O O — O	Slow	Medium	14	21
D[9]	PCR[57]	ALT0 ALT1 ALT2 ALT3	GPIO[57] X[0] TXD —	SIUL FlexPWM_0 LIN_1 —	I/O O O —	Slow	Medium	8	15
D[10]	PCR[58]	ALT0 ALT1 ALT2 ALT3	GPIO[58] A[0] — —	SIUL FlexPWM_0 — —	I/O O — —	Slow	Medium	—	53

Table 9. Absolute maximum ratings(1) (continued)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max(2)	
I_{INJSUM}	S R	Absolute sum of all input currents during overload condition	—	–50	50	mA
T_{STG}	S R	Storage temperature	—	–55	150	°C
T_J	S R	Junction temperature under bias	—	–40	150	°C

1. Functional operating conditions are given in the DC electrical characteristics. Absolute maximum ratings are stress ratings only, and functional operation at the maxima is not guaranteed. Stress beyond the listed maxima may affect device reliability or cause permanent damage to the device.
2. Absolute maximum voltages are currently maximum burn-in voltages.
3. The difference between each couple of voltage supplies must be less than 300 mV, $|V_{DD\_HV\_IOy} - V_{DD\_HV\_IOx}| < 300$ mV.
4. Guaranteed by device validation.
5. Minimum value of TV_{DD} must be guaranteed until $V_{DD\_HV\_REG}$ reaches 2.6 V (maximum value of V_{PORH})
6. Only when $V_{DD\_HV\_IOx} < 5.2$ V

Figure 6 shows the constraints of the different power supplies.

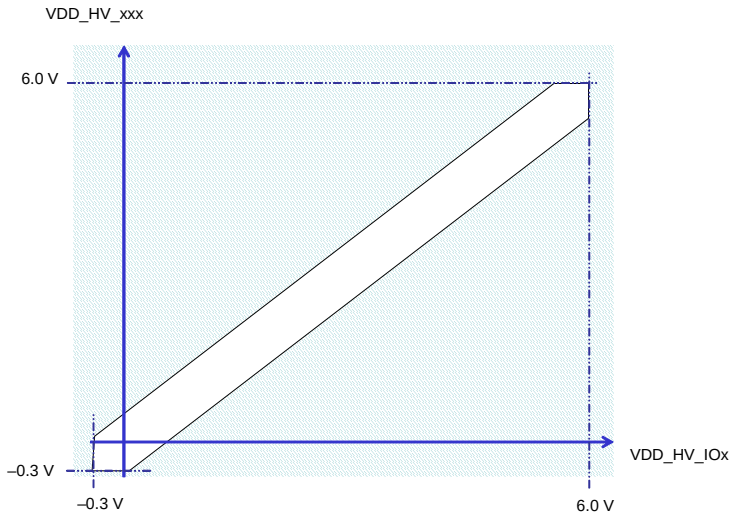
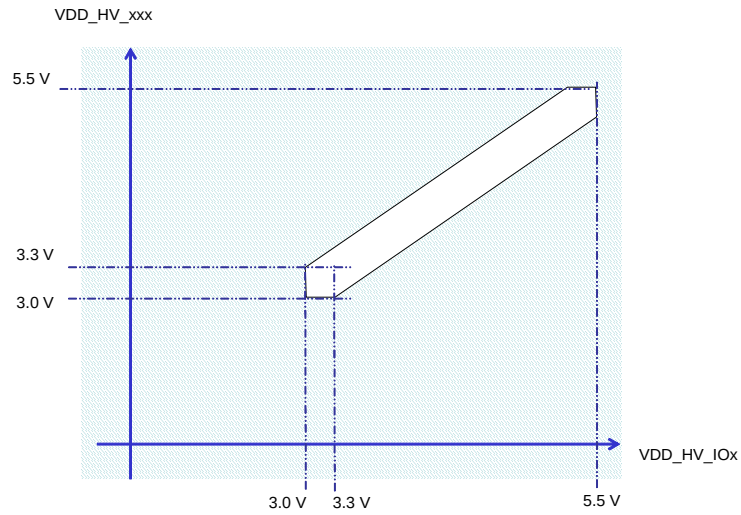


Figure 6. Power supplies constraints (–0.3 V ≤ $V_{DD\_HV\_IOx}$ ≤ 6.0 V)

The SPC560P34/SPC560P40 supply architecture allows the ADC supply to be managed independently from the standard V_{DD_HV} supply. *Figure 7* shows the constraints of the ADC power supply.

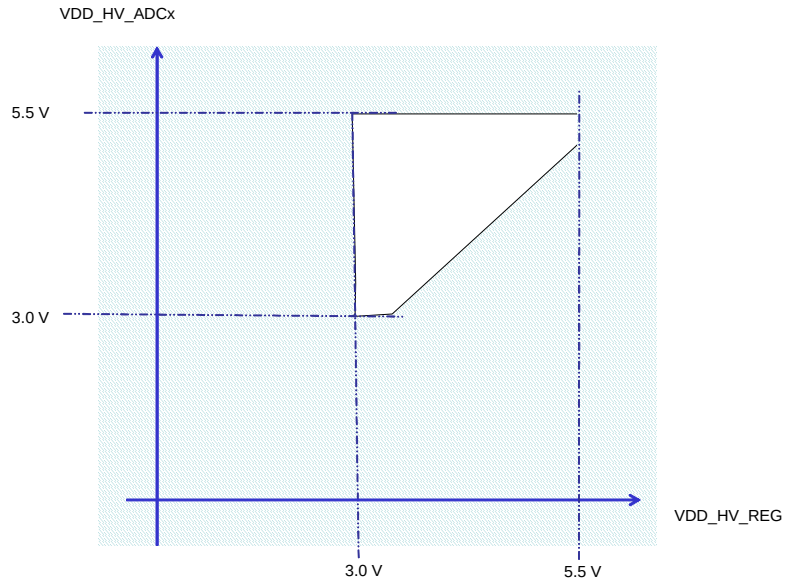
Figure 8. Power supplies constraints (3.0 V ≤ $V_{DD\_HV\_IOx}$ ≤ 5.5 V)



Note: IO AC and DC characteristics are guaranteed only in the range of 3.0–3.6 V when PAD3V5V is low, and in the range of 4.5–5.5 V when PAD3V5V is high.

The SPC560P34/SPC560P40 supply architecture allows the ADC supply to be managed independently from the standard V_{DD_HV} supply. *Figure 9* shows the constraints of the ADC power supply.

Figure 9. Independent ADC supply (3.0 V ≤ $V_{DD\_HV\_REG}$ ≤ 5.5 V)



3.8.2 Voltage monitor electrical characteristics

The device implements a power on reset module to ensure correct power-up initialization, as well as three low voltage detectors to monitor the V_{DD} and the V_{DD_LV} voltage while device is supplied:

- POR monitors V_{DD} during the power-up phase to ensure device is maintained in a safe reset state
- LVDHV3 monitors V_{DD} to ensure device reset below minimum functional supply
- LVDHV5 monitors V_{DD} when application uses device in the 5.0 V ± 10% range
- LVDLVCOR monitors low voltage digital power domain

Table 17. Low voltage monitor electrical characteristics

Symbol	C	Parameter	Conditions[1]	Value		Unit
				Min	Max	
V_{PORH}	T	Power-on reset threshold	—	1.5	2.7	V
V_{PORUP}	P	Supply for functional POR module	T_A = 25 °C	1.0	—	V
$V_{REGLVDMOK\_H}$	P	Regulator low voltage detector high threshold	—	—	2.95	V
$V_{REGLVDMOK\_L}$	P	Regulator low voltage detector low threshold	—	2.6	—	V
$V_{FLLVDMOK\_H}$	P	Flash low voltage detector high threshold	—	—	2.95	V
$V_{FLLVDMOK\_L}$	P	Flash low voltage detector low threshold	—	2.6	—	V
$V_{IOLVDMOK\_H}$	P	I/O low voltage detector high threshold	—	—	2.95	V
$V_{IOLVDMOK\_L}$	P	I/O low voltage detector low threshold	—	2.6	—	V
$V_{IOLVDM5OK\_H}$	P	I/O 5 V low voltage detector high threshold	—	—	4.4	V
$V_{IOLVDM5OK\_L}$	P	I/O 5 V low voltage detector low threshold	—	3.8	—	V
$V_{MLVDDOK\_H}$	P	Digital supply low voltage detector high	—	—	1.145	V
$V_{MLVDDOK\_L}$	P	Digital supply low voltage detector low	—	1.08	—	V

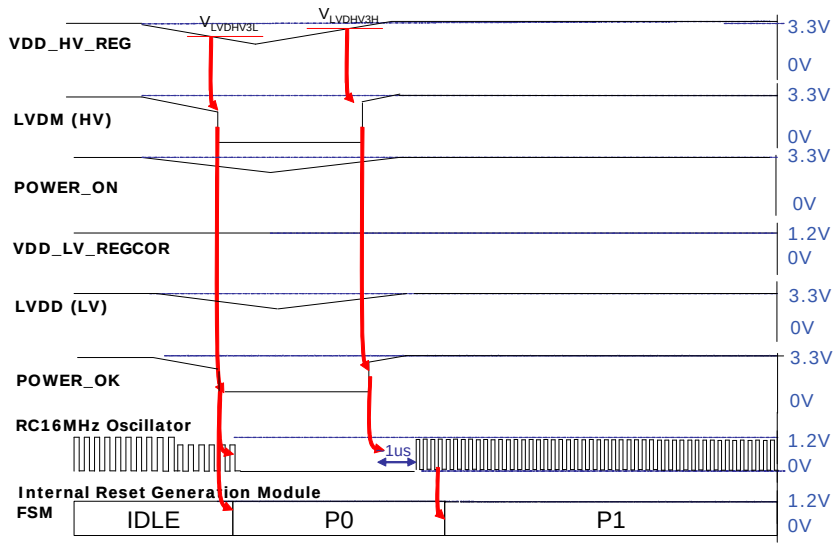
1. V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 °C to $T_{A\ MAX}$, unless otherwise specified

3.9 Power up/down sequencing

To prevent an overstress event or a malfunction within and outside the device, the SPC560P34/SPC560P40 implements the following sequence to ensure each module is started only when all conditions for switching it ON are available:

- A POWER_ON module working on voltage regulator supply controls the correct start-up of the regulator. This is a key module ensuring safe configuration for all voltage regulator functionality when supply is below 1.5 V. Associated POWER_ON (or POR) signal is active low.
- Several low voltage detectors, working on voltage regulator supply monitor the voltage of the critical modules (voltage regulator, I/Os, flash memory and low voltage domain). LVDs are gated low when POWER_ON is active.
- A POWER_OK signal is generated when all critical supplies monitored by the LVD are available. This signal is active high and released to all modules including I/Os, flash

Figure 13. Brown-out typical sequence



3.10 DC electrical characteristics

3.10.1 NVUSRO register

Portions of the device configuration, such as high voltage supply and watchdog enable/disable after reset are controlled via bit values in the non-volatile user options (NVUSRO) register.

For a detailed description of the NVUSRO register, please refer to the device reference manual.

NVUSRO[PAD3V5V] field description

The DC electrical characteristics are dependent on the PAD3V5V bit value. *Table 18* shows how NVUSRO[PAD3V5V] controls the device configuration.

Table 18. PAD3V5V field description

Value(1)	Description
0	High voltage supply is 5.0 V
1	High voltage supply is 3.3 V

1. Default manufacturing value before flash initialization is ‘1’ (3.3 V).

3.10.5 I/O pad current specification

The I/O pads are distributed across the I/O supply segment. Each I/O supply segment is associated to a V_{DD}/V_{SS} supply pair as described in *Table 23*.

Table 23. I/O supply segment

Package	Supply segment				
	1	2	3	4	5
LQFP100	pin15–pin26	pin27–pin46	pin51–pin61	pin64–pin86	pin89–pin10
LQFP64	pin8–pin17	pin18–pin30	pin33–pin38	pin41–pin54	pin57–pin5

Table 24. I/O consumption

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
I_{SWTSLW}[2]	CC	D	Dynamic I/O current for SLOW configuration	C_L = 25 pF	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	20	mA
					V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	16	
I_{SWTMED}[2]	CC	D	Dynamic I/O current for MEDIUM configuration	C_L = 25 pF	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	29	mA
					V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	17	
I_{SWTFST}[2]	CC	D	Dynamic I/O current for FAST configuration	C_L = 25 pF	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	110	mA
					V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	50	
I_{RMSSLW}	CC	D	Root medium square I/O current for SLOW configuration	C_L = 25 pF, 2 MHz	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	2.3	mA
				C_L = 25 pF, 4 MHz		—	—	3.2	
				C_L = 100 pF, 2 MHz		—	—	6.6	
				C_L = 25 pF, 2 MHz	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	1.6	
				C_L = 25 pF, 4 MHz		—	—	2.3	
				C_L = 100 pF, 2 MHz		—	—	4.7	
I_{RMSMED}	CC	D	Root medium square I/O current for MEDIUM configuration	C_L = 25 pF, 13 MHz	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	6.6	mA
				C_L = 25 pF, 40 MHz		—	—	13.4	
				C_L = 100 pF, 13 MHz		—	—	18.3	
				C_L = 25 pF, 13 MHz	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	5	
				C_L = 25 pF, 40 MHz		—	—	8.5	
				C_L = 100 pF, 13 MHz		—	—	11	

In particular two different transient periods can be distinguished:

- A first and quick charge transfer from the internal capacitance C_{P1} and C_{P2} to the sampling capacitance C_S occurs (C_S is supposed initially completely discharged): considering a worst case (since the time constant in reality would be faster) in which C_{P2} is reported in parallel to C_{P1} (call $C_P = C_{P1} + C_{P2}$), the two capacitances C_P and C_S are in series, and the time constant is

Equation 5

$$\tau_1 = (R_{SW} + R_{AD}) \bullet \frac{C_P \bullet C_S}{C_P + C_S}$$

Equation 5 can again be simplified considering only C_S as an additional worst condition. In reality, the transient is faster, but the A/D converter circuitry has been designed to be robust also in the very worst case: the sampling time T_S is always much longer than the internal time constant:

Equation 6

$$\tau_1 < (R_{SW} + R_{AD}) \bullet C_S \ll T_S$$

The charge of C_{P1} and C_{P2} is redistributed also on C_S, determining a new value of the voltage V_{A1} on the capacitance according to *Equation 7*:

Equation 7

$$V_{A1} \bullet (C_S + C_{P1} + C_{P2}) = V_A \bullet (C_{P1} + C_{P2})$$

- A second charge transfer involves also C_F (that is typically bigger than the on-chip capacitance) through the resistance R_L: again considering the worst case in which C_{P2} and C_S were in parallel to C_{P1} (since the time constant in reality would be faster), the time constant is:

Equation 8

$$\tau_2 < R_L \bullet (C_S + C_{P1} + C_{P2})$$

In this case, the time constant depends on the external circuit: in particular imposing that the transient is completed well before the end of sampling time T_S, a constraints on R_L sizing is obtained:

Equation 9

$$8.5 \bullet \tau_2 = 8.5 \bullet R_L \bullet (C_S + C_{P1} + C_{P2}) < T_S$$

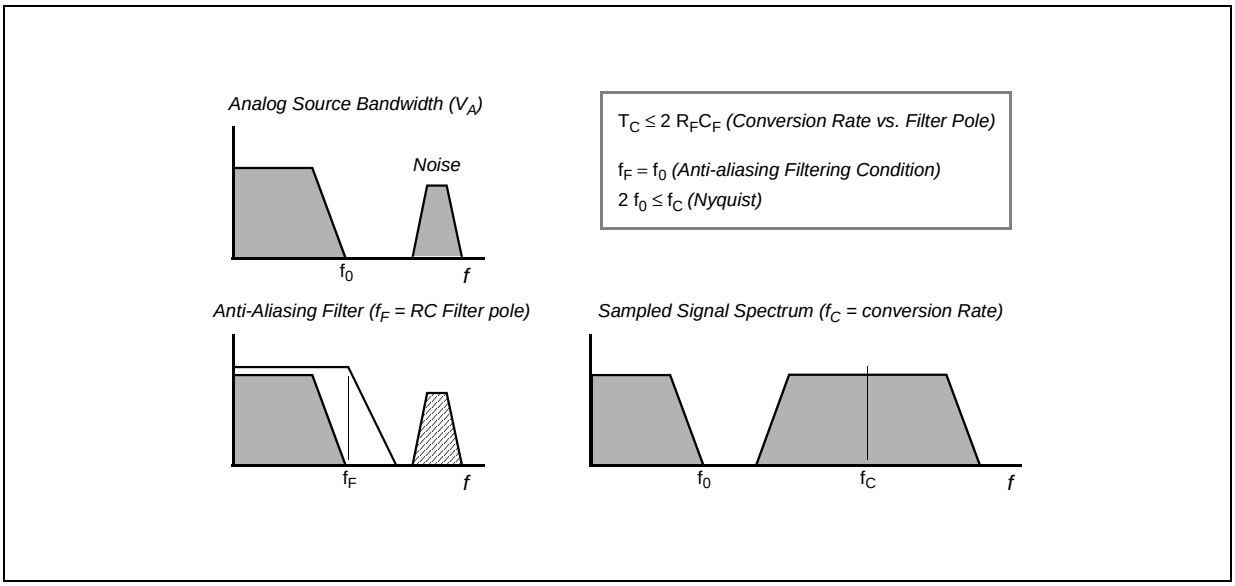
Of course, R_L shall be sized also according to the current limitation constraints, in combination with R_S (source impedance) and R_F (filter resistance). Being C_F definitively bigger than C_{P1}, C_{P2} and C_S, then the final voltage V_{A2} (at the end of the charge transfer transient) will be much higher than V_{A1}. *Equation 10* must be respected (charge balance assuming now C_S already charged at V_{A1}):

Equation 10

$$V_{A2} \bullet (C_S + C_{P1} + C_{P2} + C_F) = V_A \bullet C_F + V_{A1} \bullet (C_{P1} + C_{P2} + C_S)$$

The two transients above are not influenced by the voltage source that, due to the presence of the R_FC_F filter, is not able to provide the extra charge to compensate the voltage drop on C_S with respect to the ideal source V_A; the time constant R_FC_F of the filter is very high with respect to the sampling time (T_S). The filter is typically designed to act as anti-aliasing.

Figure 18. Spectral representation of input signal



Calling f_0 the bandwidth of the source signal (and as a consequence the cut-off frequency of the anti-aliasing filter, f_F), according to the Nyquist theorem the conversion rate f_C must be at least $2f_0$; it means that the constant time of the filter is greater than or at least equal to twice the conversion period (T_C). Again the conversion period T_C is longer than the sampling time T_S, which is just a portion of it, even when fixed channel continuous conversion mode is selected (fastest conversion rate at a specific channel): in conclusion it is evident that the time constant of the filter R_FC_F is definitively much higher than the sampling time T_S, so the charge level on C_S cannot be modified by the analog signal source during the time in which the sampling switch is closed.

The considerations above lead to impose new constraints on the external circuit, to reduce the accuracy error due to the voltage drop on C_S; from the two charge balance equations above, it is simple to derive *Equation 11* between the ideal and real sampled voltage on C_S:

Equation 11

$$\frac{V_A}{V_{A2}} = \frac{C_{P1} + C_{P2} + C_F}{C_{P1} + C_{P2} + C_F + C_S}$$

From this formula, in the worst case (when V_A is maximum, that is for instance 5 V), assuming to accept a maximum error of half a count, a constraint is evident on C_F value:

Equation 12

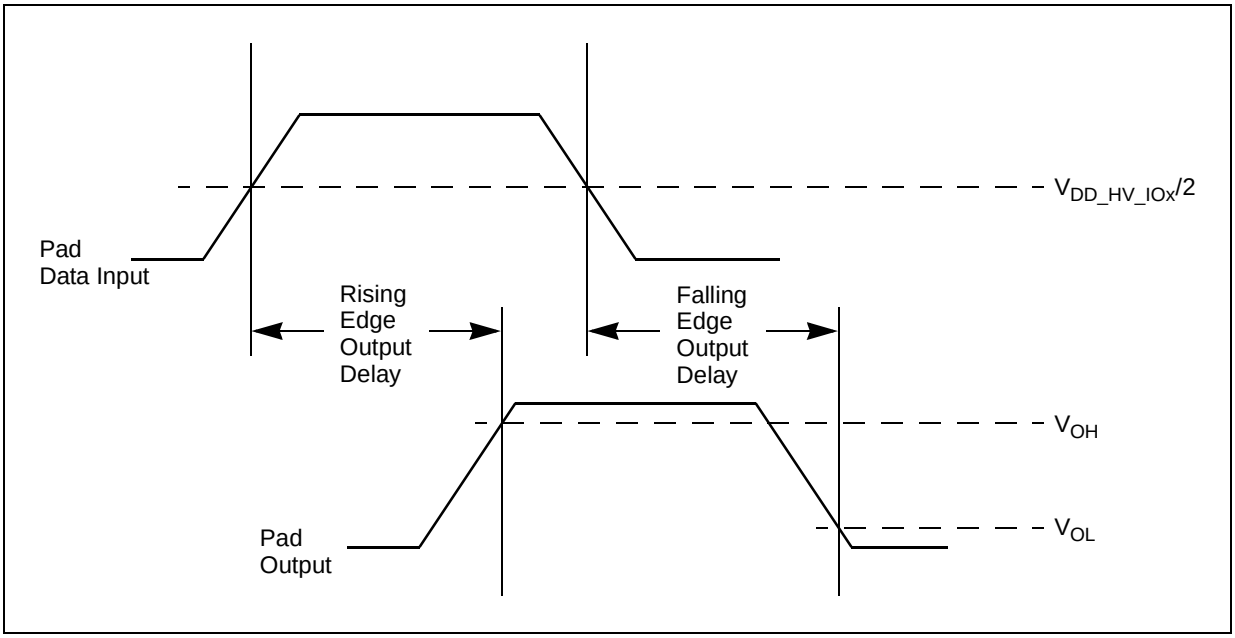
$$C_F > 2048 \bullet C_S$$

Table 36. Output pin transition times (continued)

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
t_{tr}	CC	D	Output transition time output pin[2] FAST configuration	C_L = 25 pF	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 SIUL.PCRx.SRC = 1	—	—	4	ns
				C_L = 50 pF		—	—	6	
				C_L = 100 pF		—	—	12	
				C_L = 25 pF	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 SIUL.PCRx.SRC = 1	—	—	4	
				C_L = 50 pF		—	—	7	
				C_L = 100 pF		—	—	12	
t_{SYM}[3]	CC	T	Symmetric transition time, same drive strength between N and P transistor	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0		—	—	4	ns
				V_{DD} = 3.3 V ± 10%, PAD3V5V = 1		—	—	5	

1. V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 °C to $T_{A\ MAX}$, unless otherwise specified.
2. C_L includes device and package capacitances (C_{PKG} < 5 pF).
3. Transition timing of both positive and negative slopes will differ maximum 50%.

Figure 19. Pad output delay



3.17 AC timing characteristics

3.17.1 $\overline{\text{RESET}}$ pin characteristics

The SPC560P34/SPC560P40 implements a dedicated bidirectional $\overline{\text{RESET}}$ pin.

Figure 20. Start-up reset requirements

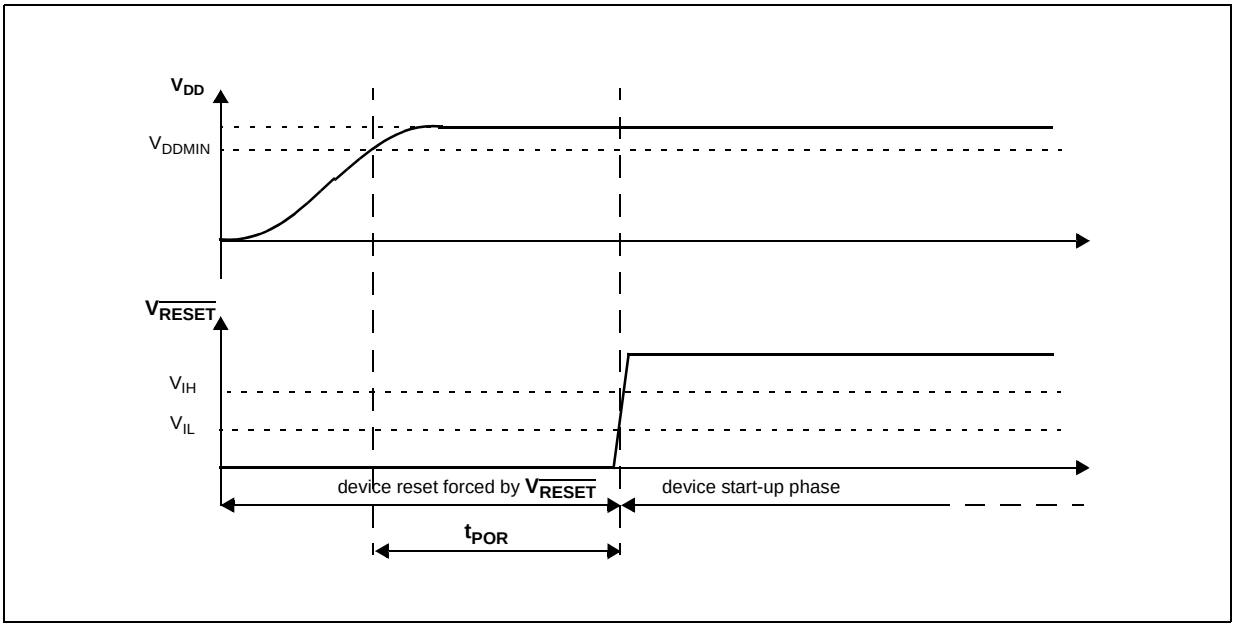


Figure 21. Noise filtering on reset signal

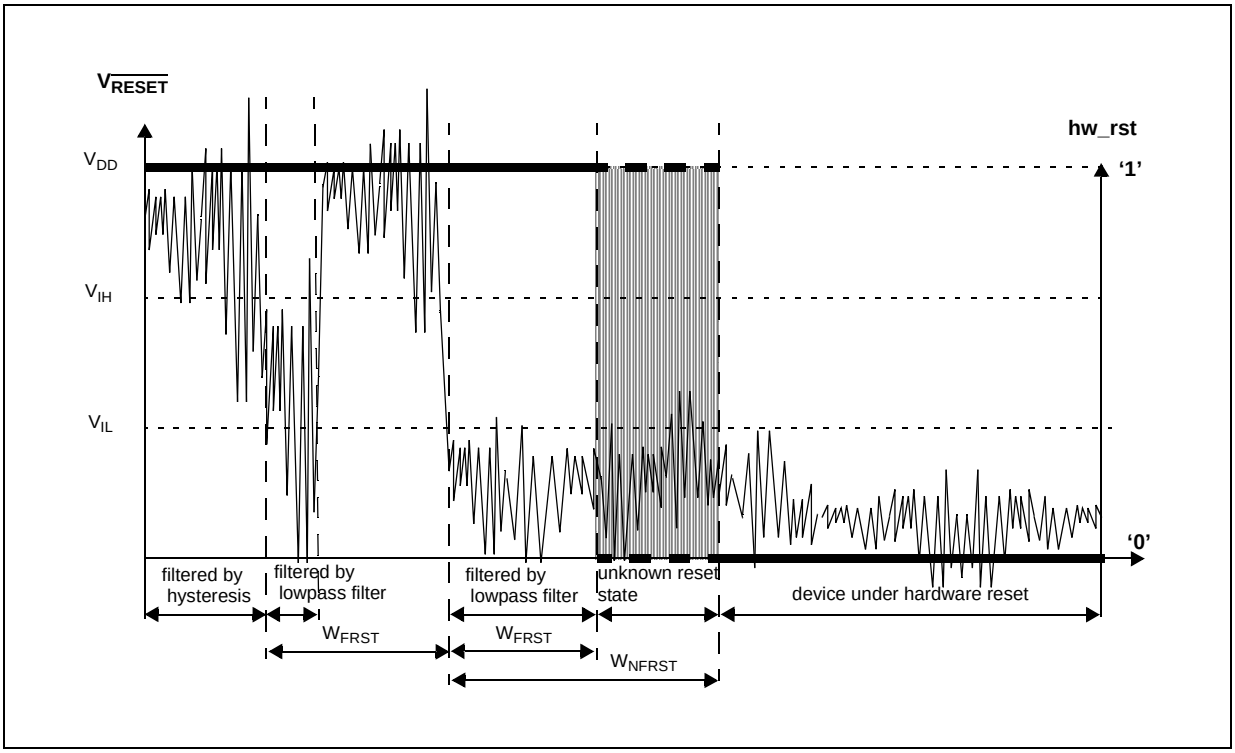
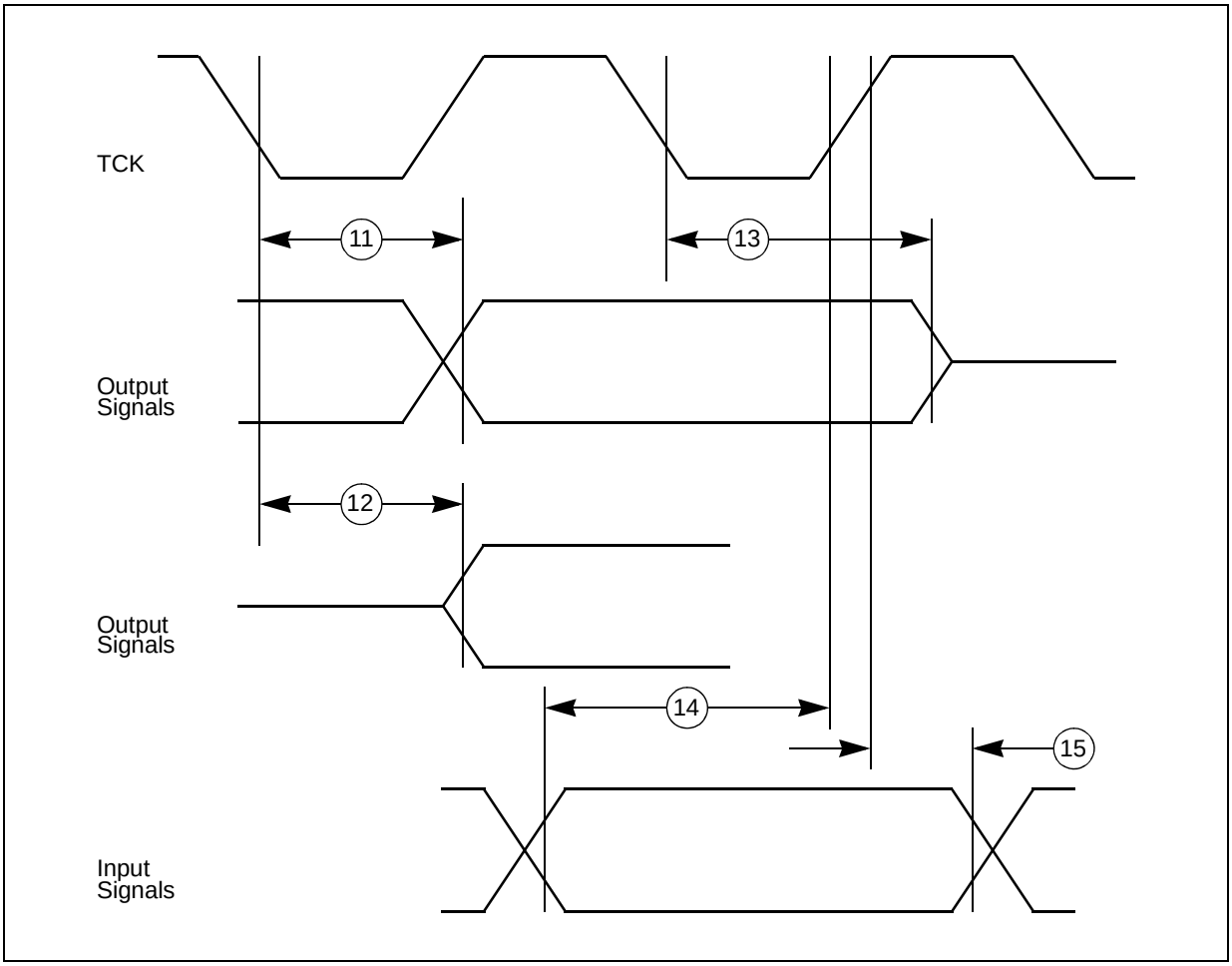


Table 37. $\overline{\text{RESET}}$ electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value[2]			Unit
					Min	Typ	Max	
V_{IH}	SR	P	Input high level CMOS (Schmitt Trigger)	—	$0.65V_{DD}$	—	V_{DD} + 0.4	V
V_{IL}	SR	P	Input low level CMOS (Schmitt Trigger)	—	–0.4	—	$0.35V_{DD}$	V
V_{HYS}	CC	C	Input hysteresis CMOS (Schmitt Trigger)	—	$0.1V_{DD}$	—	—	V
V_{OL}	CC	P	Output low level	Push Pull, I_{OL} = 2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	—	—	$0.1V_{DD}$	V
				Push Pull, I_{OL} = 1 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[3]	—	—	$0.1V_{DD}$	
				Push Pull, I_{OL} = 1 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	—	—	0.5	
t_{tr}	CC	D	Output transition time output pin[4] MEDIUM configuration	C_L = 25 pF, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	10	ns
				C_L = 50 pF, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	20	
				C_L = 100 pF, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	40	
				C_L = 25 pF, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	12	
				C_L = 50 pF, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	25	
				C_L = 100 pF, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	40	
W_{FRST}	SR	P	$\overline{\text{RESET}}$ input filtered pulse	—	—	—	40	ns
W_{NFRST}	SR	P	$\overline{\text{RESET}}$ input not filtered pulse	—	500	—	—	ns
t_{POR}	CC	D	Maximum delay before internal reset is released after all V_{DD_HV} reach nominal supply	Monotonic V_{DD_HV} supply ramp	—	—	1	ms
$\lvert I_{WPU} \rvert$	CC	P	Weak pull-up current absolute value	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	10	—	150	µA
				V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	10	—	150	
				V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[5]	10	—	250	

1. V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified
2. All values need to be confirmed during device validation.
3. This is a transient configuration during power-up, up to the end of reset PHASE2 (refer to RGM module section of device reference manual).

Figure 24. JTAG boundary scan timing



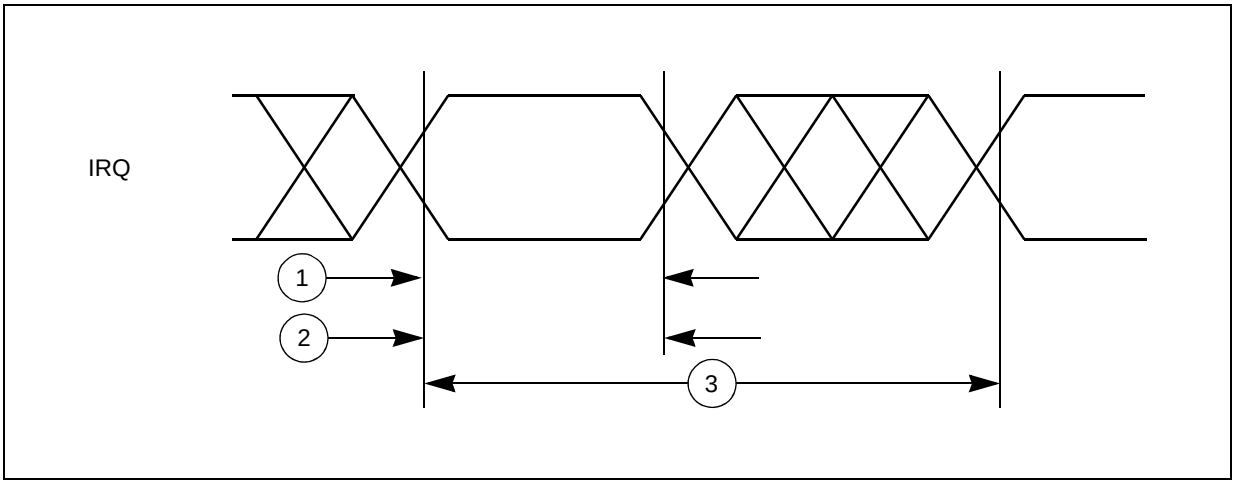
3.17.3 Nexus timing

Table 39. Nexus debug port timing(1)

No.	Symbol		C	Parameter	Value			Unit
					Min	Typ	Max	
1	t_{TCYC}	CC	D	TCK cycle time	4(2)	—	—	t_{CYC}
2	t_{NTDIS}	CC	D	TDI data setup time	5	—	—	ns
	t_{NTMSS}	CC	D	TMS data setup time	5	—	—	ns
3	t_{NTDIH}	CC	D	TDI data hold time	25	—	—	ns
	t_{NTMSH}	CC	D	TMS data hold time	25	—	—	ns
4	t_{TDOV}	CC	D	TCK low to TDO data valid	10	—	20	ns
5	t_{TDOI}	CC	D	TCK low to TDO data invalid	—	—	—	ns

1. All Nexus timing relative to MCKO is measured from 50% of MCKO and 50% of the respective signal.
2. Lower frequency is required to be fully compliant to standard.

Figure 28. External interrupt timing



3.17.5 DSPI timing

Table 41. DSPI timing(1)

No.	Symbol		C	Parameter	Conditions	Value		Unit
						Min	Max	
1	t_{SCK}	CC	D	DSPI cycle time	Master (MTFE = 0)	60	—	ns
					Slave (MTFE = 0)	60	—	
2	t_{CSC}	CC	D	CS to SCK delay	—	16	—	ns
3	t_{ASC}	CC	D	After SCK delay	—	26	—	ns
4	t_{SDC}	CC	D	SCK duty cycle	—	0.4 * t_{SCK}	0.6 * t_{SCK}	ns
5	t_A	CC	D	Slave access time	$\overline{SS}$ active to SOUT valid	—	30	ns
6	t_{DIS}	CC	D	Slave SOUT disable time	$\overline{SS}$ inactive to SOUT high impedance or invalid	—	16	ns
7	t_{PCSC}	CC	D	PCSx to $\overline{PCSS}$ time	—	13	—	ns
8	t_{PASC}	CC	D	$\overline{PCSS}$ to PCSx time	—	13	—	ns
9	t_{SUI}	CC	D	Data setup time for inputs	Master (MTFE = 0)	35	—	ns
					Slave	4	—	
					Master (MTFE = 1, CPHA = 0)	35	—	
					Master (MTFE = 1, CPHA = 1)	35	—	
10	t_{HI}	CC	D	Data hold time for inputs	Master (MTFE = 0)	–5	—	ns
					Slave	4	—	
					Master (MTFE = 1, CPHA = 0)	11	—	
					Master (MTFE = 1, CPHA = 1)	–5	—	